

Power Supply Design Seminar

Phase-shifted full-bridge
converter fundamentals

Authors

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Agenda

- Why a phase-shifted full bridge (PSFB)?
 - Power trends and the need for soft-switching converters
 - PSFB target applications
- PSFB operation and design considerations
 - How does a PSFB work, and how does it achieve soft switching?
 - Types of output rectifiers for PSFB
 - Voltage spikes on the output rectifier and clamping options
 - Voltage-mode and peak current-mode control
 - Synchronous rectifier (SR) modes of operation
 - Light load management
- PSFB design example
- Summary

Why a PSFB?

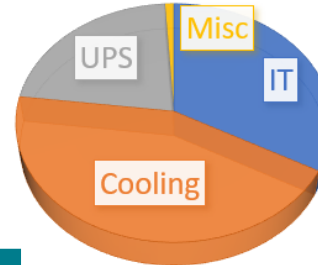
Power trends

- Demands for smaller size, lighter and higher-efficiency systems
- 80 Plus certification and Open Compute Project (OCP) Modular Hardware System-Common Redundant Power Supply (M-CRPS) specifications drive high efficiency

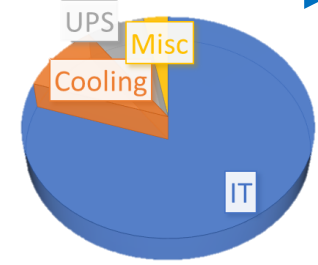
Efficiency at	10% load	20% load	50% load	100% load	Note
80 Plus Titanium	90%	94%	96%	91%	At 230-V _{AC} input
M-CRPS (<2,500 W)	90%	94%	96%	92%	At 240-V _{AC} input
M-CRPS (≥2,500 W)	90%	94%	96%	94%	At 240-V _{AC} input



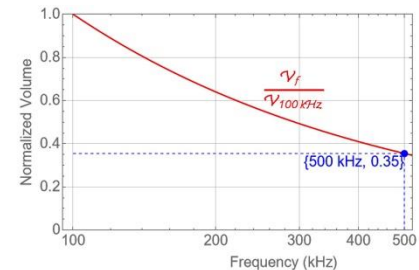
$$PUE = \frac{\text{Total Datacenter Power}}{\text{Actual IT Power}}$$



Power-use efficiency (PUE): 3

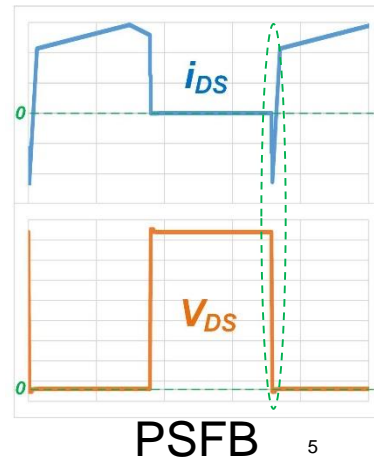
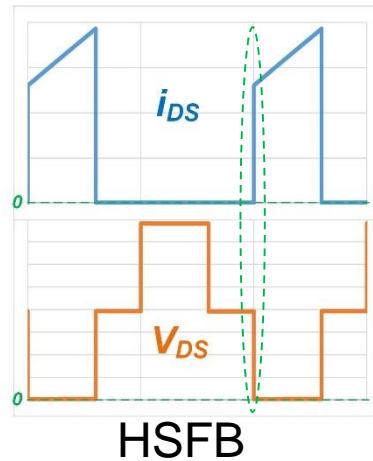
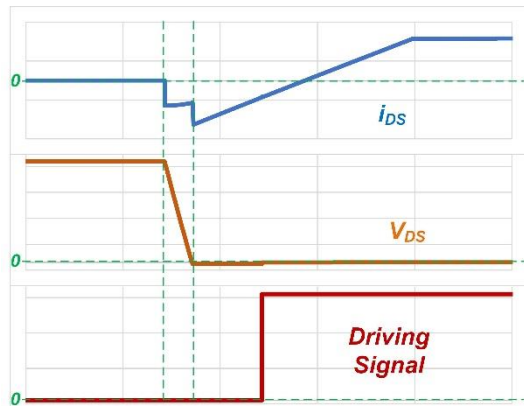
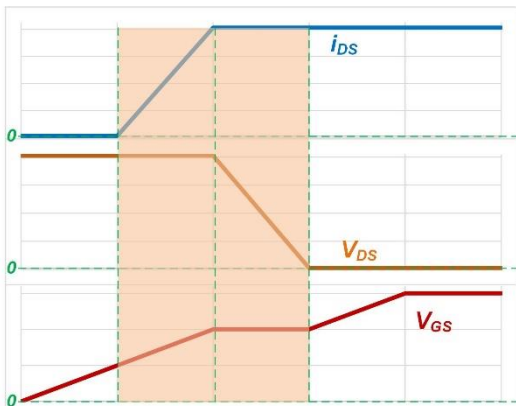
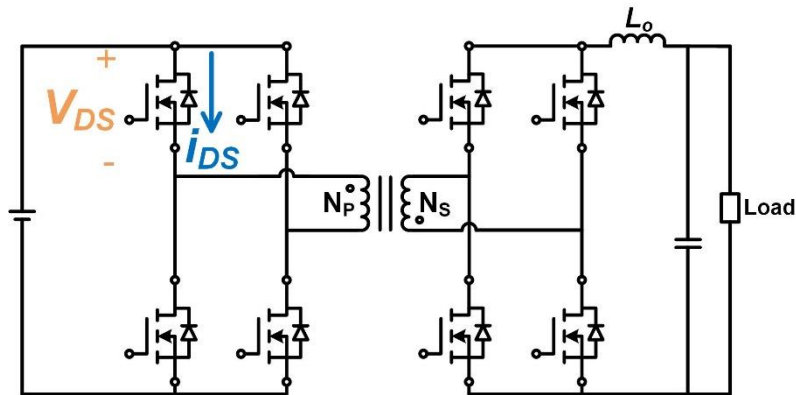


PUE: 1.25

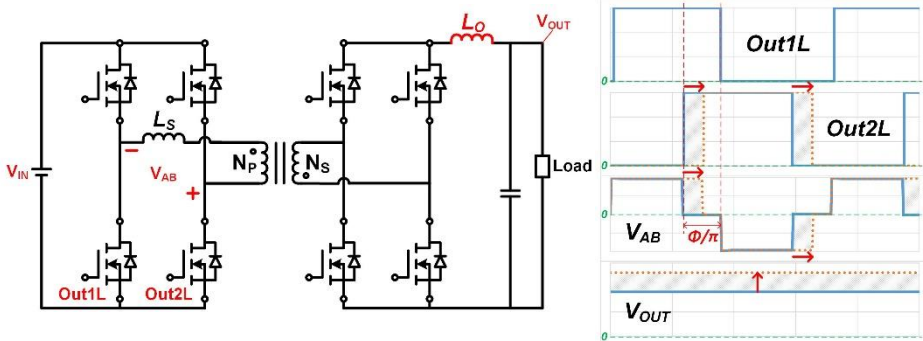


Motivation for soft switching

- Hard-switched turnon
 - MOSFET current/voltage overlap at turnon
 - Trade-off of turnon speed (t_{on} losses) vs. switch-node transient voltages (dV/dt) (EMI noise)
 - Ex: Hard-switched full-bridge converter (HSFB)
- Soft-switched turnon
 - MOSFET V_{DS} discharged before V_{GS} applied
 - Turnon speed not critical for high efficiency
 - Allows higher F_{sw}

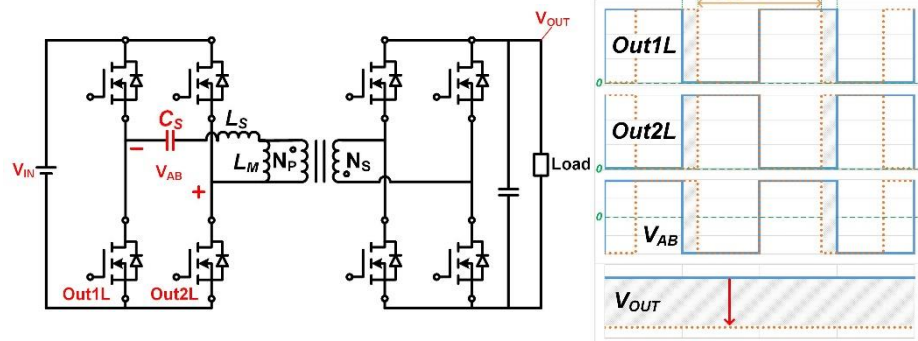


Topology comparison: PSFB vs. LLC



PSFB converter

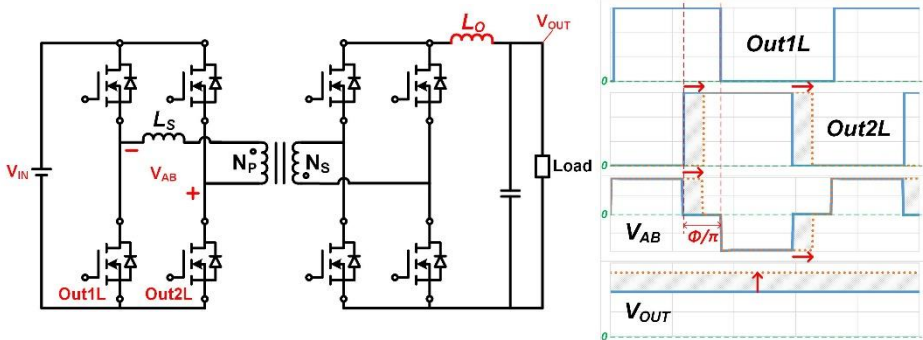
- Uses energy stored in L_S and L_O for zero voltage switching (ZVS)
- **Change phase angle for different V_{OUT}/V_{IN} gain**
- Advantages
 - **Faster transient response**
 - **Capable of wide voltage operating range**
 - Fixed frequency, easy SR control
- Disadvantages
 - Rectifier reverse recovery
 - Lost of ZVS on primary leg(s) at light loads
 - Higher rectifier voltage stress



LLC series resonant converter

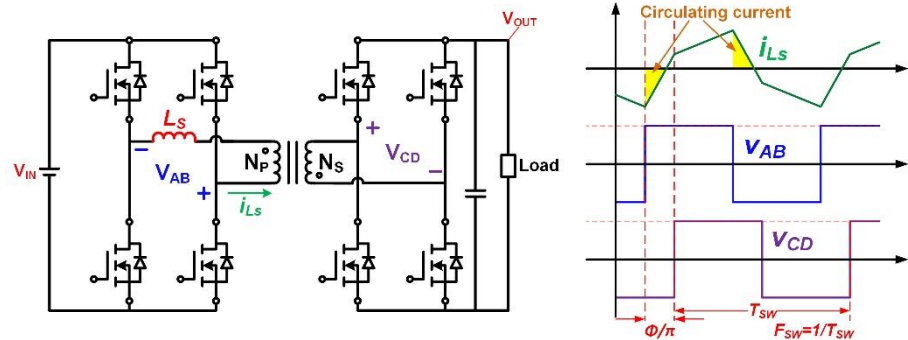
- Uses energy stored in L_M for ZVS
 - **Change F_{sw} for different V_{OUT}/V_{IN} gain**
 - Advantages
 - Soft switching on output rectifier at $F_{sw} \leq f_r$
 - Higher efficiency when $f_{sw} \approx f_r$
 - Disadvantages
 - Larger secondary RMS currents
 - SR control more challenging
 - Limited operating range
- Series resonant f \uparrow

Topology comparison: PSFB vs. DAB



PSFB converter

- Uses energy stored in L_S and L_O for ZVS
- V_{OUT}/V_{IN} gain: determined by phase angle between the **primary half bridges**
- Lower output ripple current than a dual active bridge (DAB)
- Higher output rectifier stress

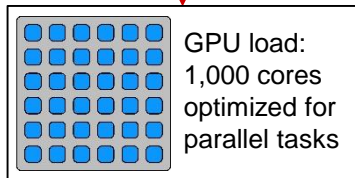
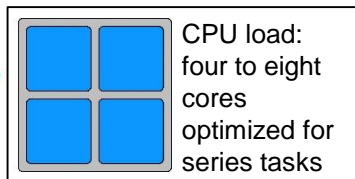
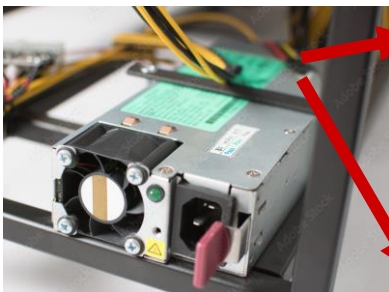


DAB converter

- Uses energy stored in L_S for ZVS
- V_{OUT}/V_{IN} gain: determined by phase angle between **transformer windings**
- Large circulating current with single-phase shift control
- Applying multiple phase-shift control will lower the circulating current
- Higher output ripple current than a PSFB

PSFB target applications

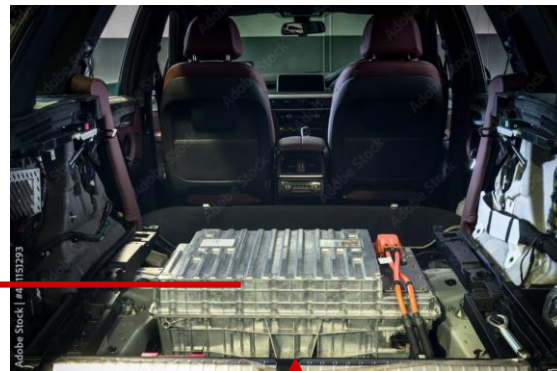
- PSFB can be a good fit for applications with:
 - Fast transient requirements
 - $\geq 10 \text{ A}/\mu\text{s}$ in a server power-supply unit (PSU) 12-V bus with a graphics processing unit (GPU) load
 - Wide input/output ranges
 - High-power battery applications



Electric
vehicle
battery:
250 V to
450 V



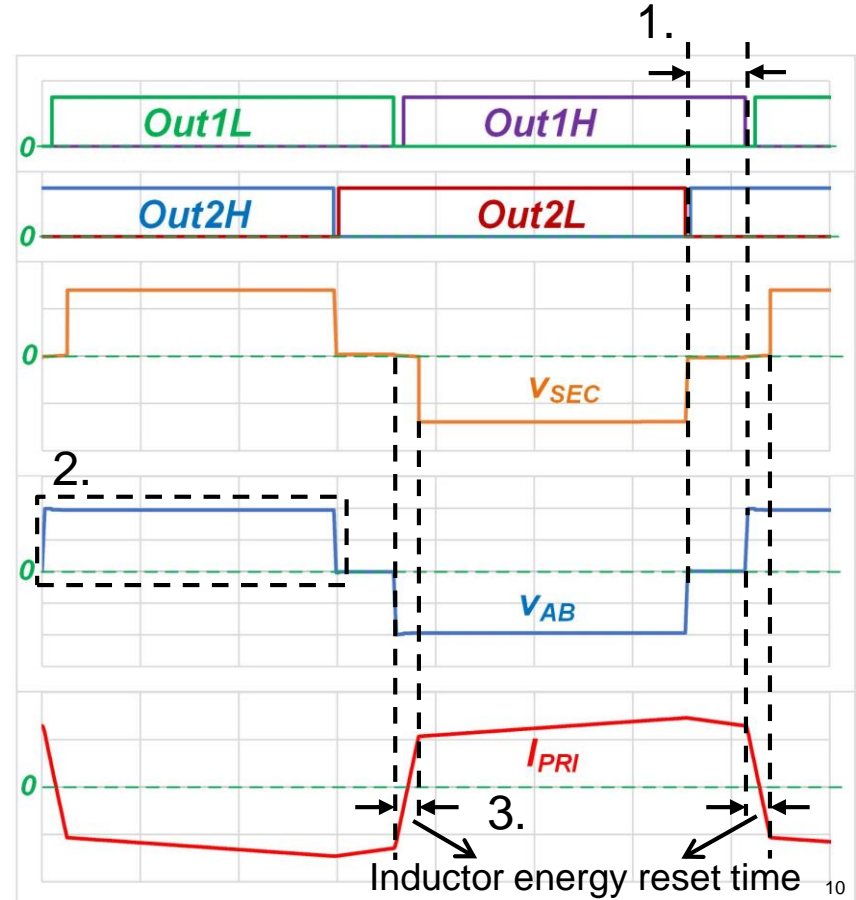
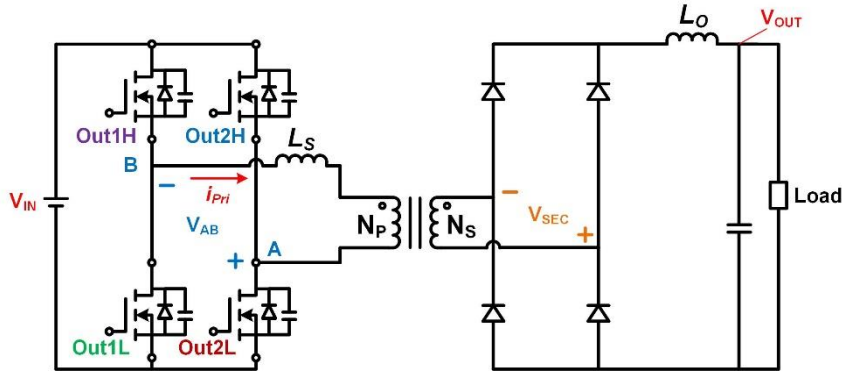
Car battery:
9 V to 16 V
up to 300 A



Soft switching and operation

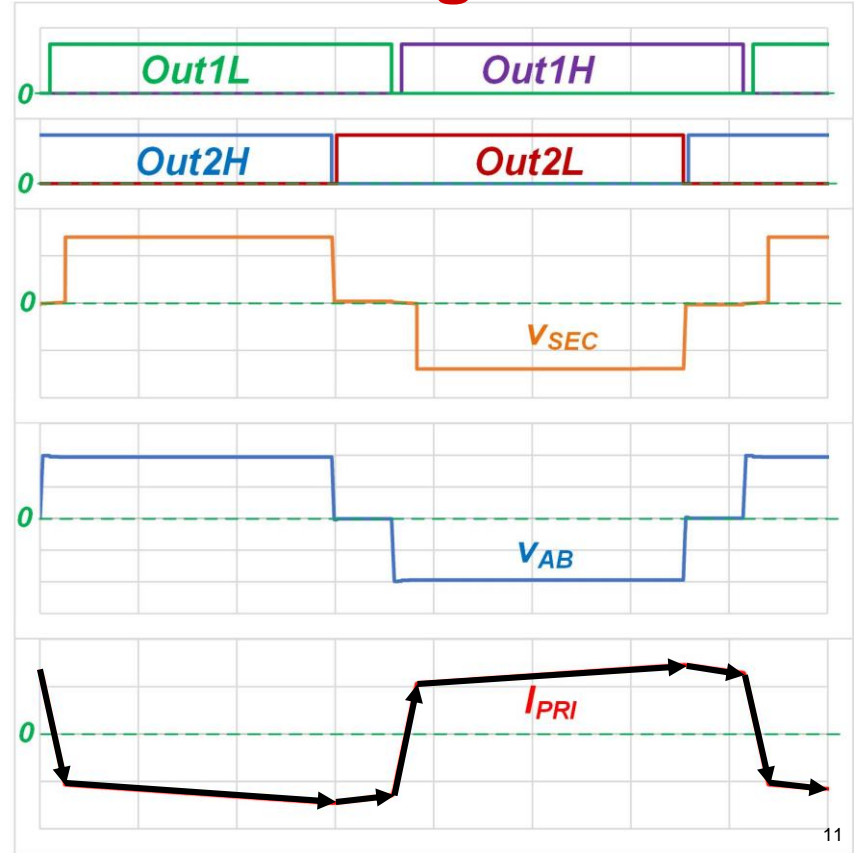
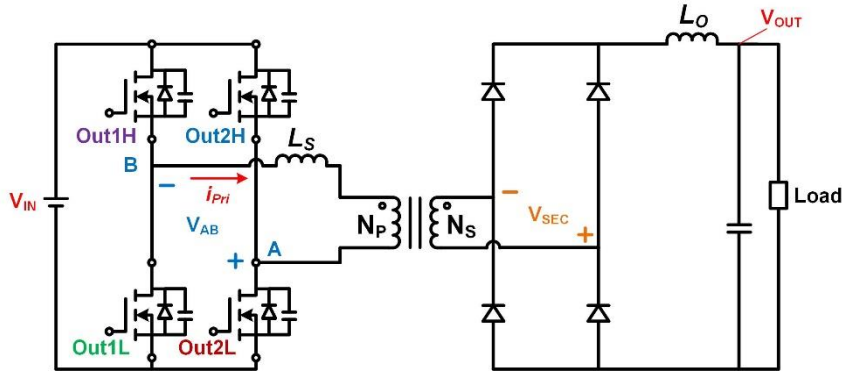
How does a PSFB work?

1. Phase shifts to control nonzero voltage transformer pulse width
2. Buck operation on output stage for regulation
3. Duty-cycle loss limits effective duty cycle



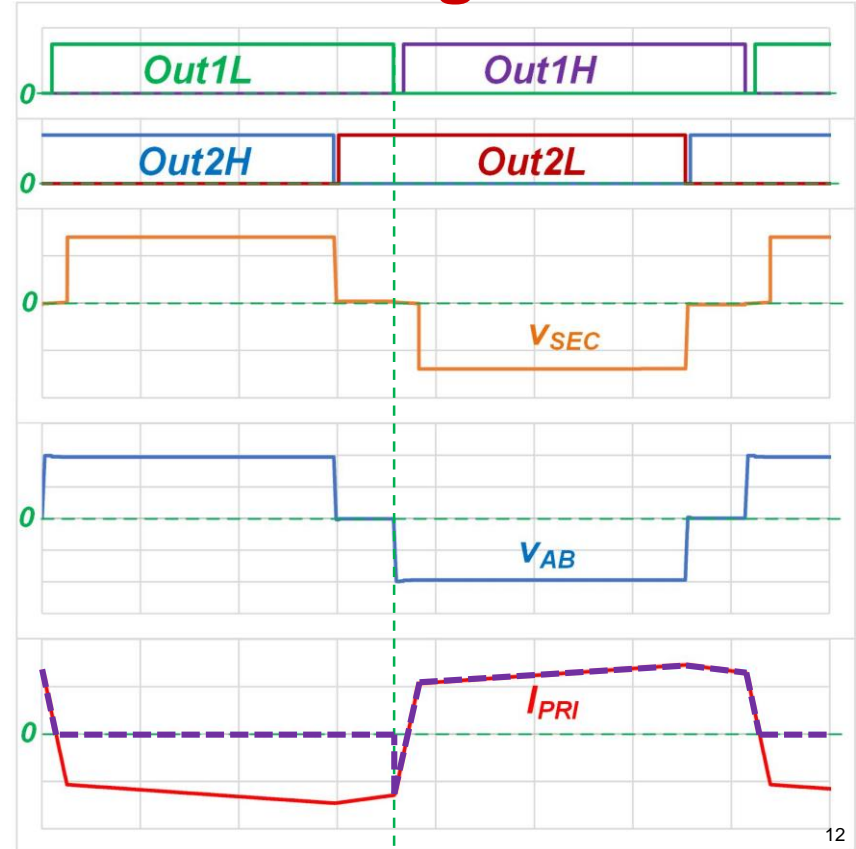
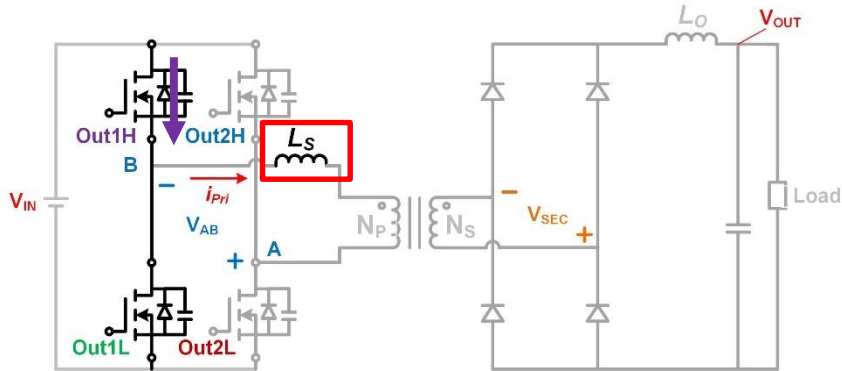
How does a PSFB achieve soft switching?

- Allows freewheeling current
 - L_S and L_O energy can be used for ZVS



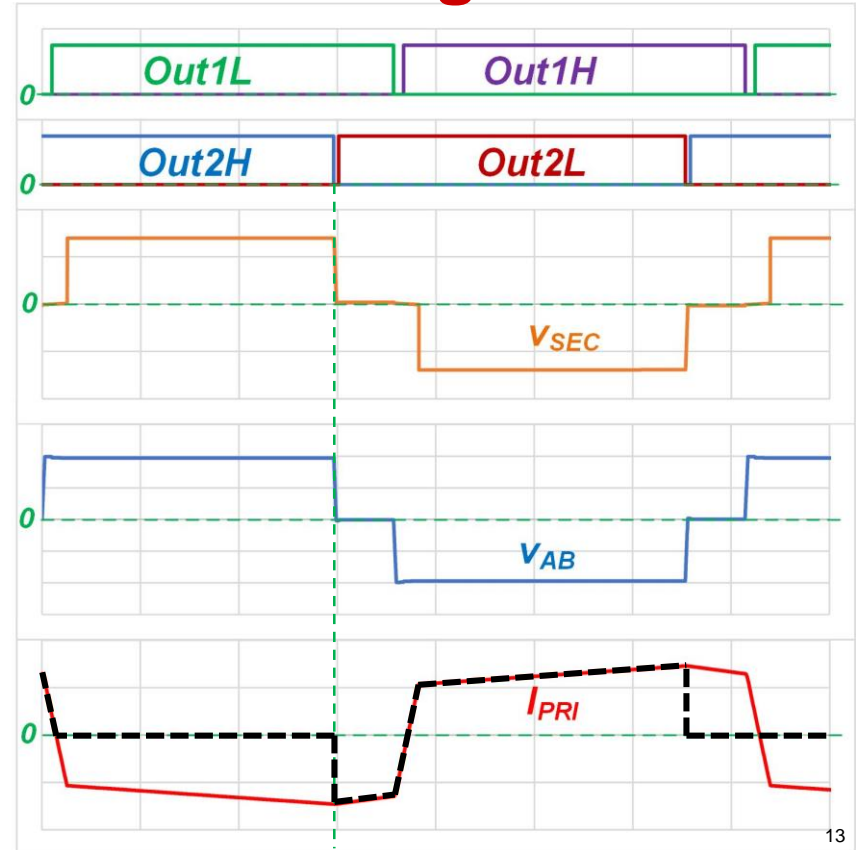
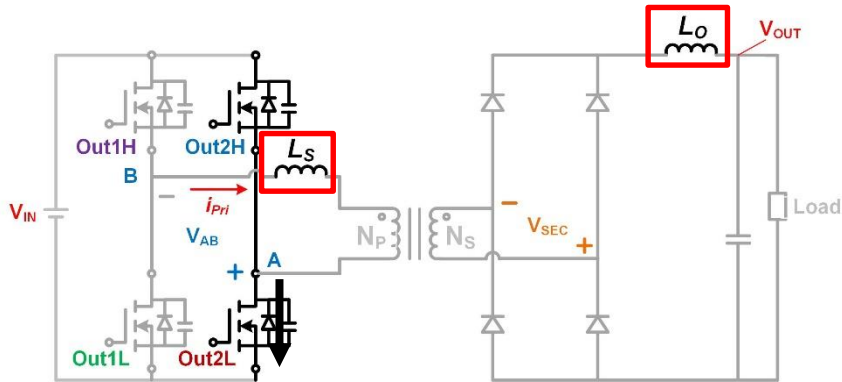
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- Leg 1 ZVS: mainly relies on L_S



How does a PSFB achieve soft switching?

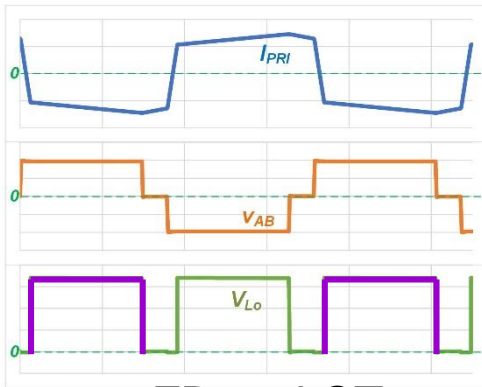
- Allows freewheeling current
 - L_S and L_O energy can be used for ZVS
- Leg 1 ZVS: mainly relies on L_S
 - Easy to **lose** ZVS at light loads
- Leg 2 ZVS: counts on both L_S and L_O
 - Easy to **keep** ZVS at light loads



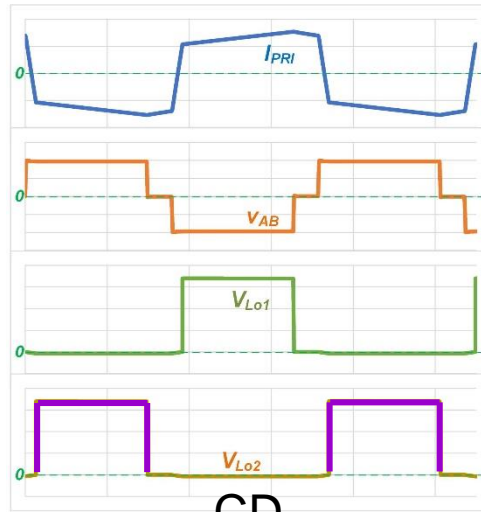
Rectifier and clamp

Output rectifiers for a PSFB

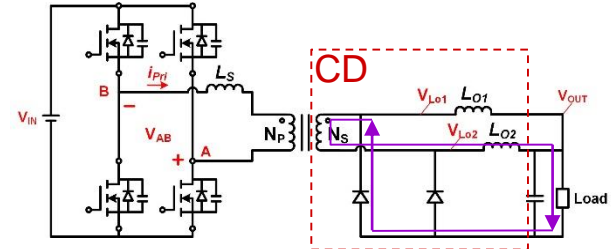
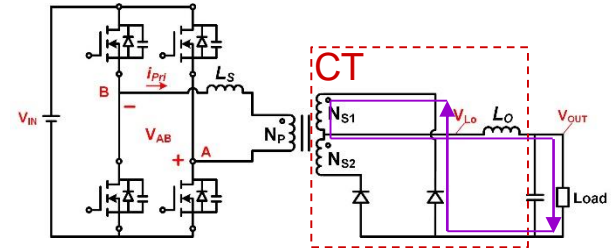
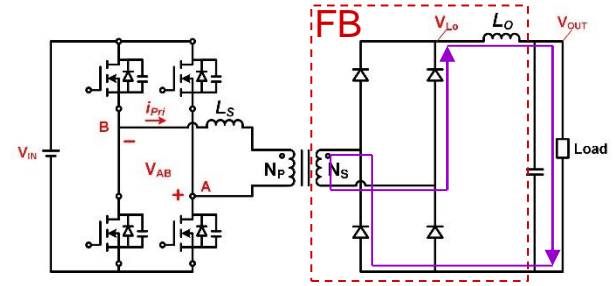
- There are three types of rectifiers:
 - Full-bridge (FB) rectifier
 - Center-tapped (CT) rectifier
 - Current-doubler (CD) rectifier



FB and CT

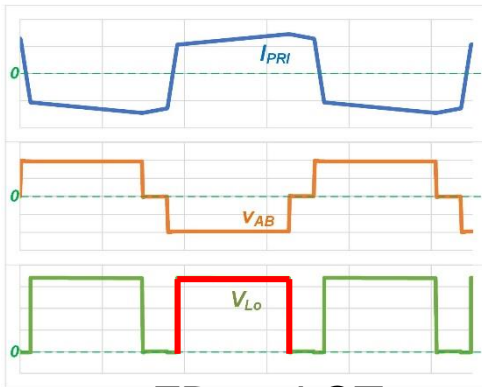


CD

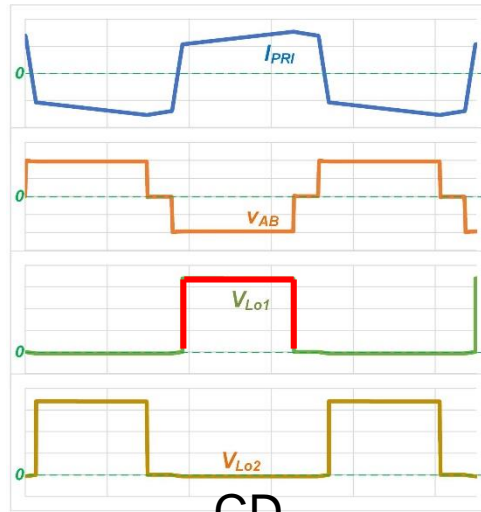


Output rectifiers for a PSFB

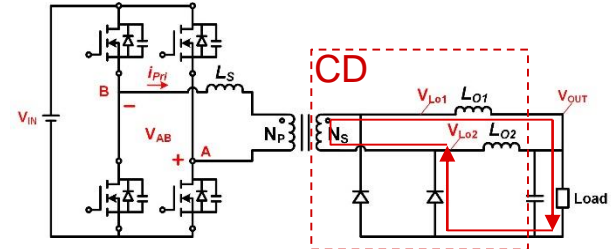
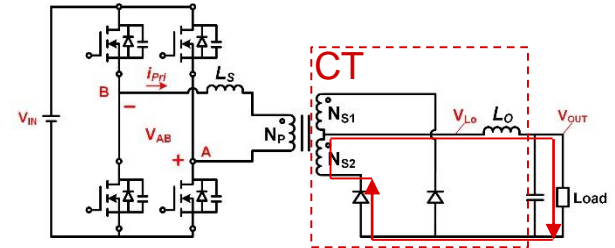
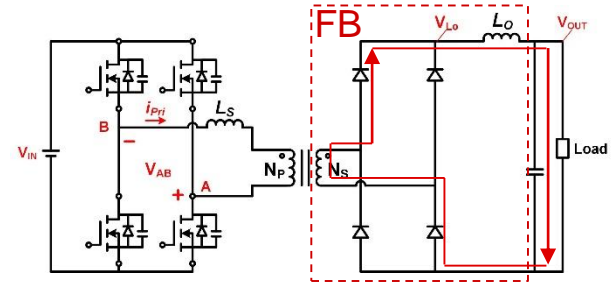
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FB and CT

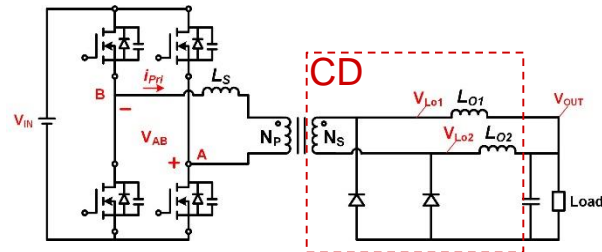
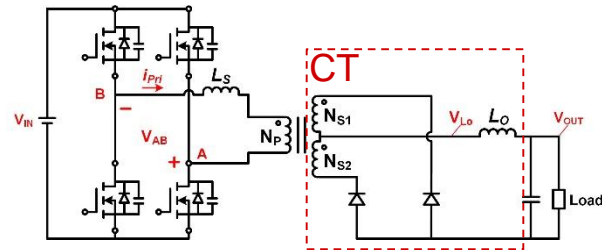
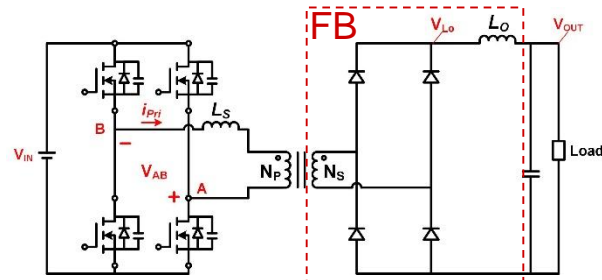
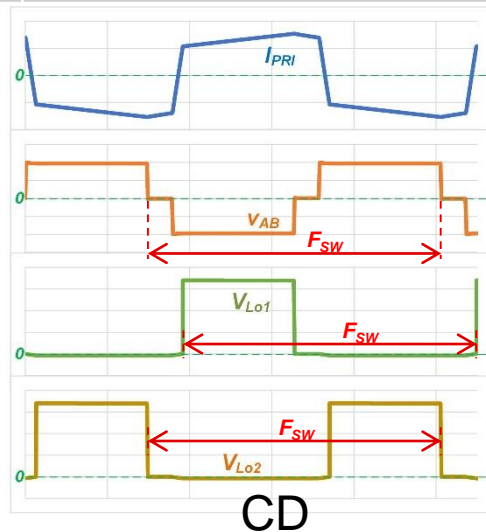
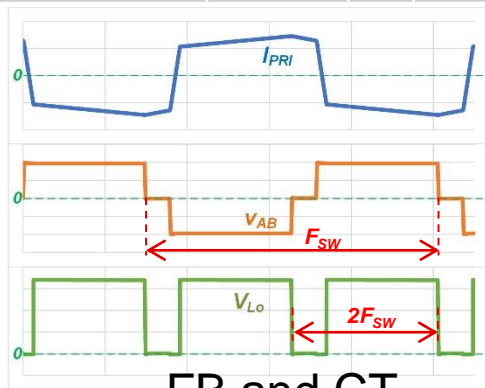


CD



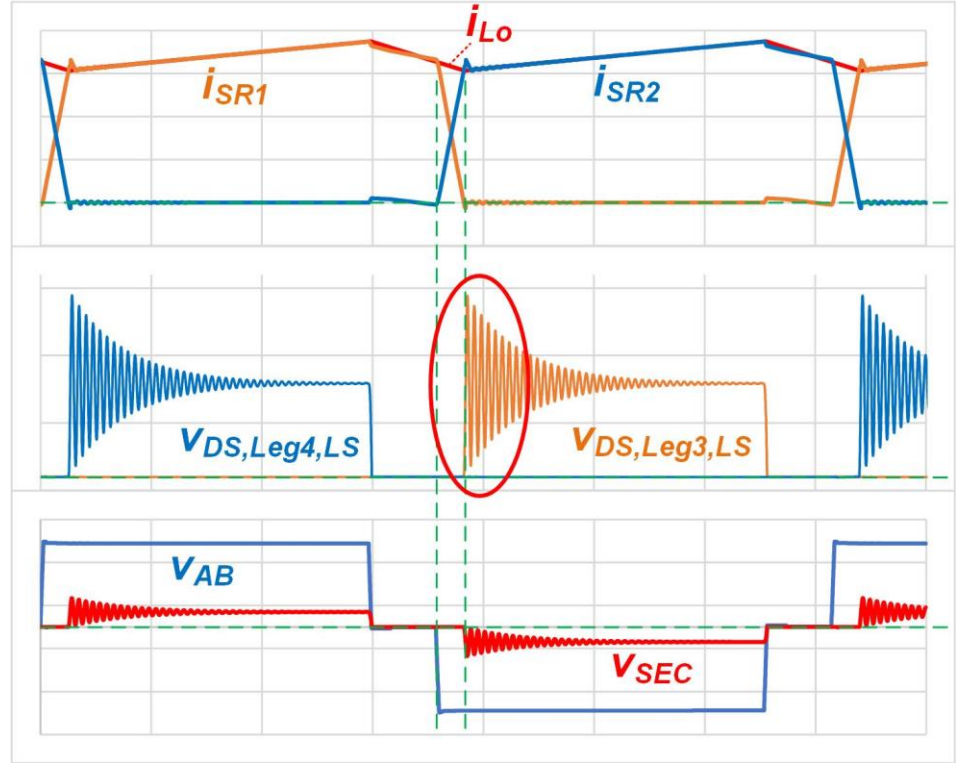
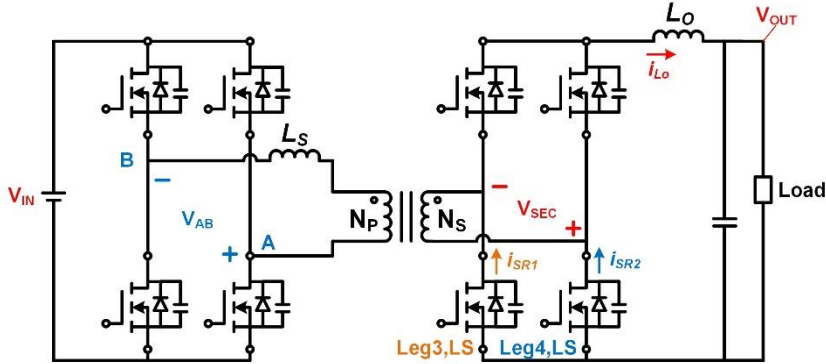
Output rectifiers for a PSFB

Type	Inductor F_{SW}	D range	N_s	Rec	L_o	Characteristics
FB	$2 \times F_{SW,PRI}$	0-100%	1	4	1	Better transformer utilization rate, good for high V_o
CT	$2 \times F_{SW,PRI}$	0-100%	2	2	1	Lowest component count, lower transformer utilization rate
CD	$1 \times F_{SW,PRI}$	0-50%	1	2	2	Better transformer utilization rate, half i_{L_o}



Voltage spikes on the output rectifier

- Transformer winding series inductor resonant with rectifier C_{oss}
- Voltage peak could be as high as $2V_{IN} \frac{N_S}{N_P}$



Clamp rectifier voltage spikes: Passive

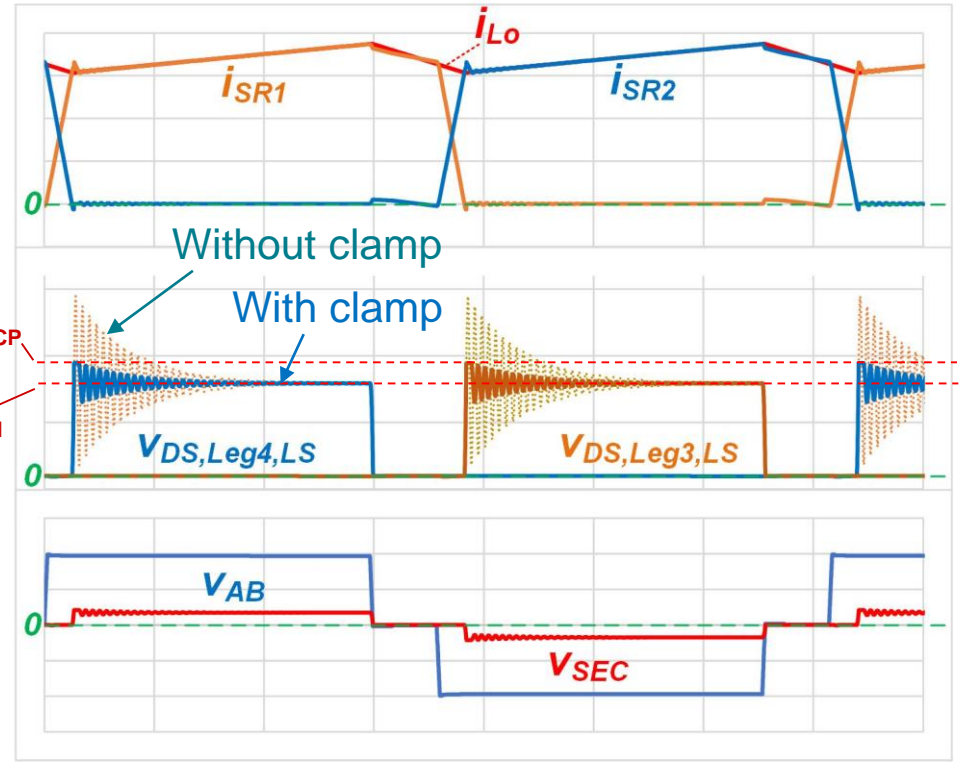
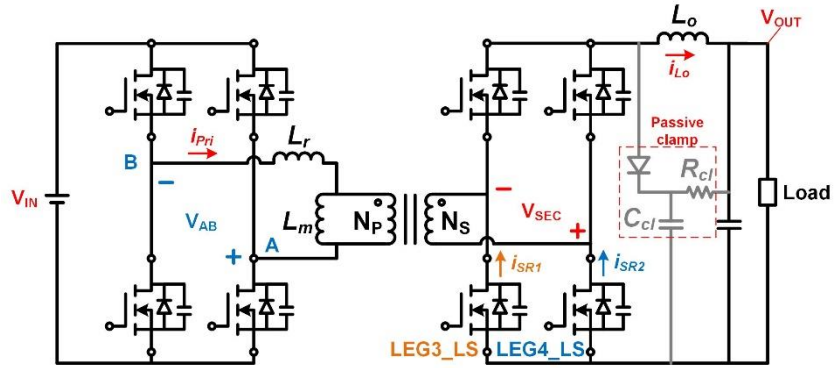
- C_{clamp} as an ideal voltage source

$$R_{\text{cl}} = \frac{(V_{\text{CP}} - V_{\text{OUT}})(V_{\text{CP}} - V_d)}{C_{\text{cl}} V_{\text{CP}} (2V_d - V_{\text{CP}}) F_{\text{sw}}}$$

- Clamp resistor dissipates

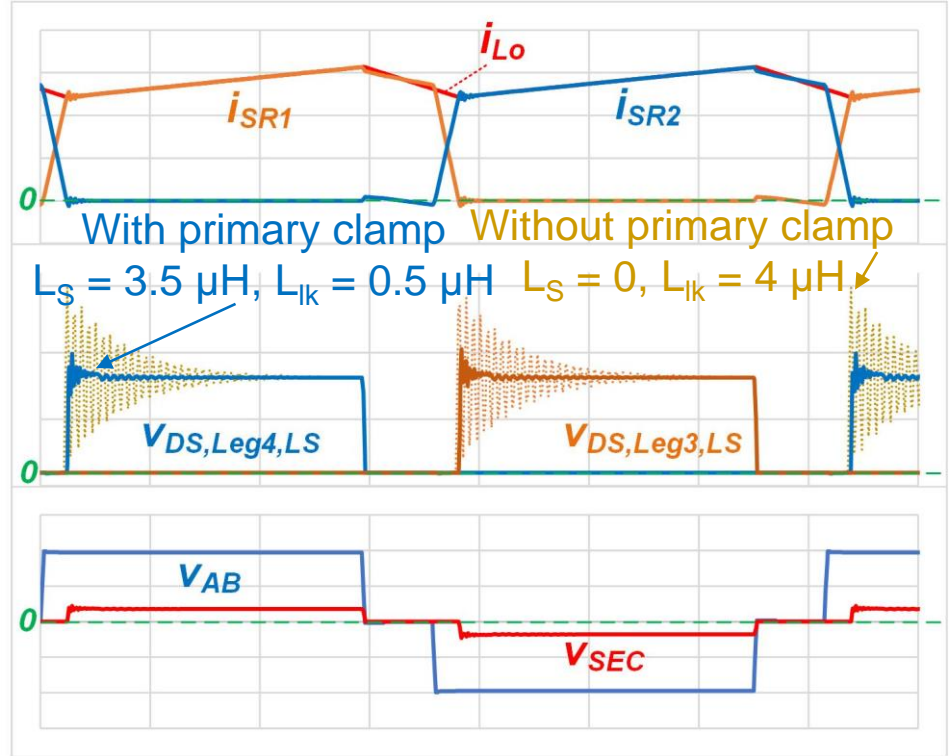
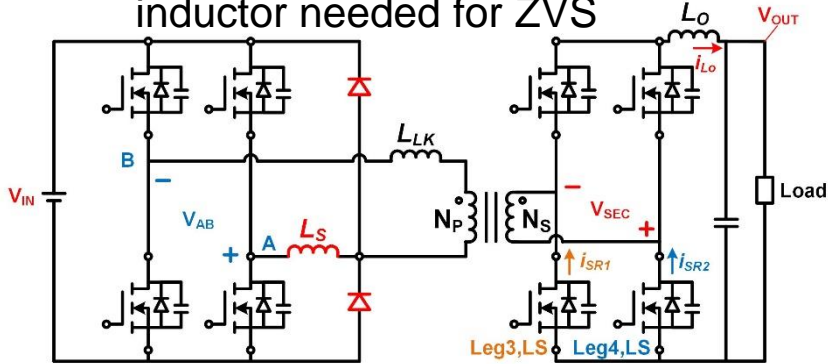
$$P_{R_{\text{cl}}} = \frac{(V_{\text{CP}} - V_{\text{OUT}})^2}{R_{\text{cl}}}$$

\Rightarrow Smaller $R_{\text{cl}} \Rightarrow$ Lower V_{CP}
 \Rightarrow Larger $P_{R_{\text{cl}}}$



Clamp rectifier voltage spikes: Primary clamp

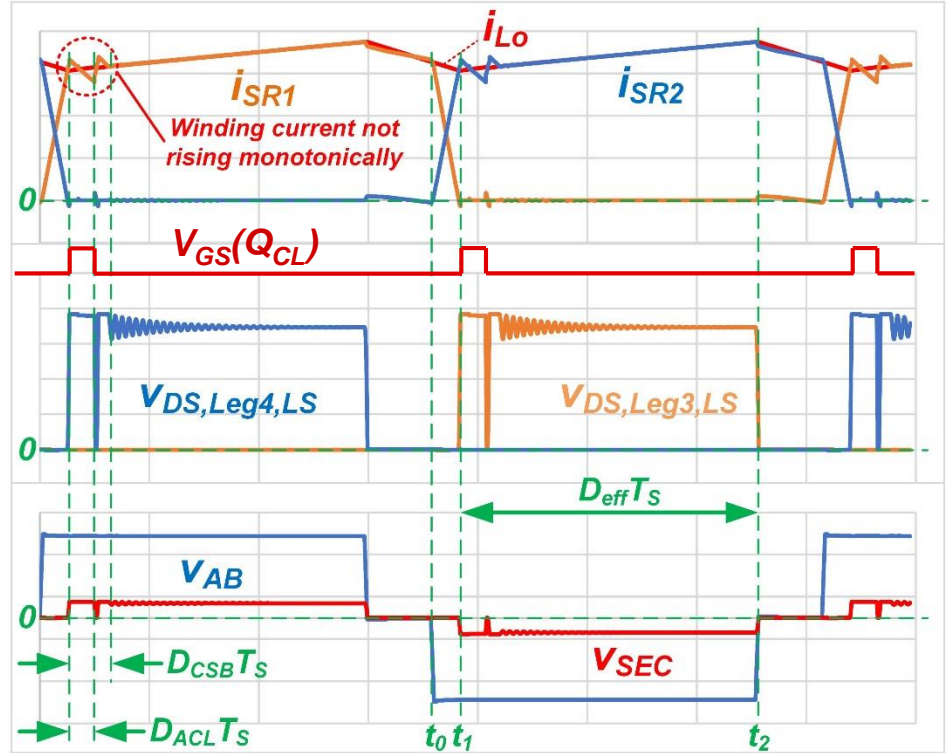
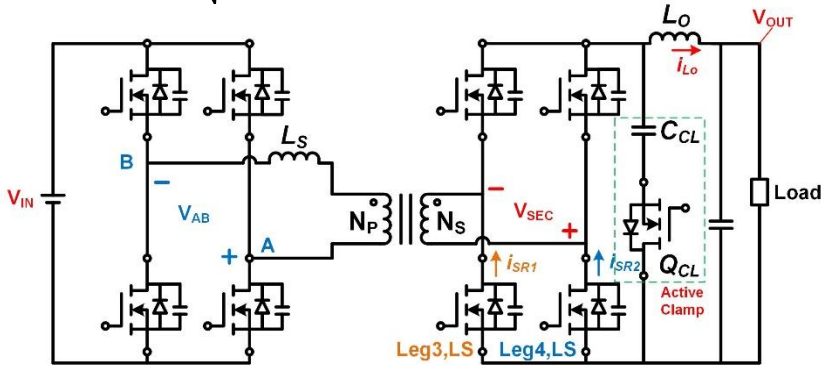
- Pros:
 - Allows L_S energy recycling
 - Allows well-coupled transformer windings for lower rectifier voltage stress
- Cons:
 - Additional diodes and discrete series inductor needed for ZVS



Clamp rectifier voltage spikes: Active clamp

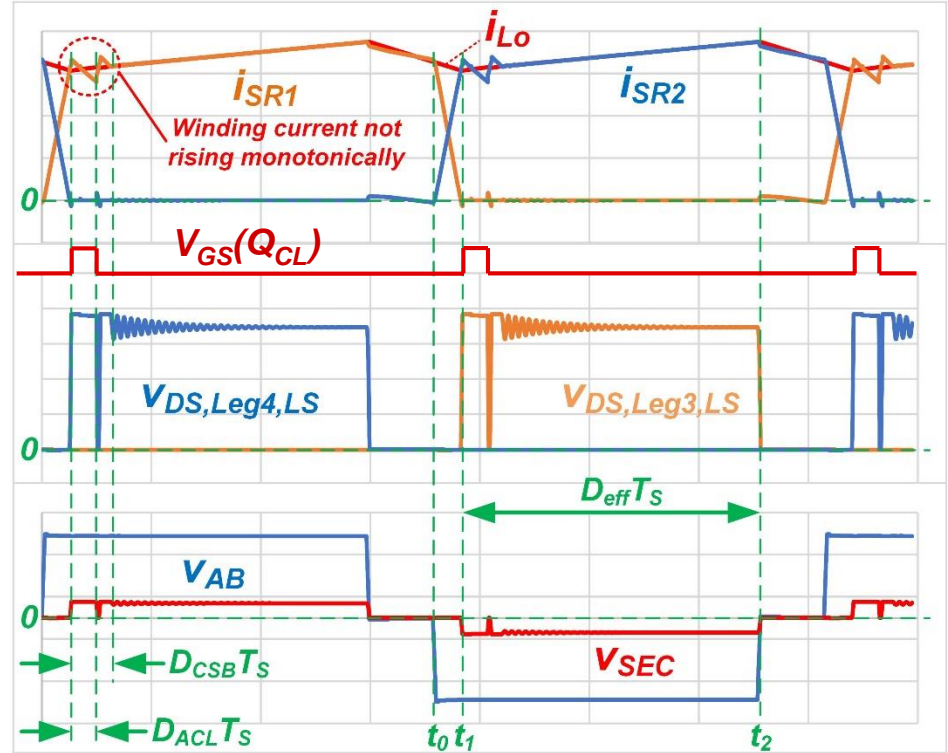
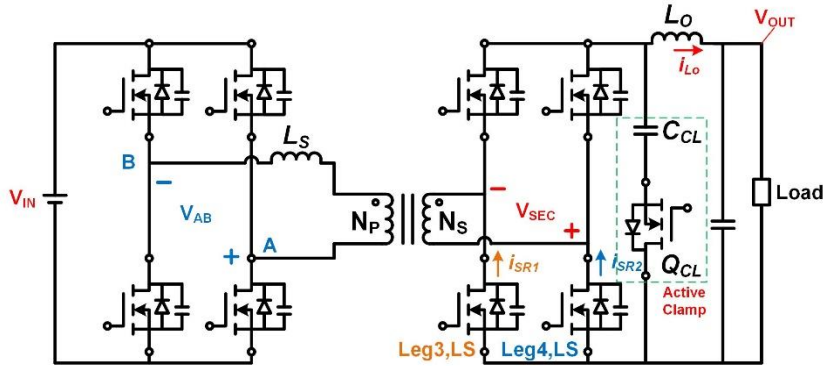
- Turn on Q_{CL} after t_1 to allow rectifier voltage to clamp to $V_{C_{CL}}$
- Size C_{CL} capacitance for relatively low ripple voltage:

$$2\pi \sqrt{\left(\frac{N_S}{N_P}\right)^2 \times L_S \times C_{CL}} \gg T_S$$



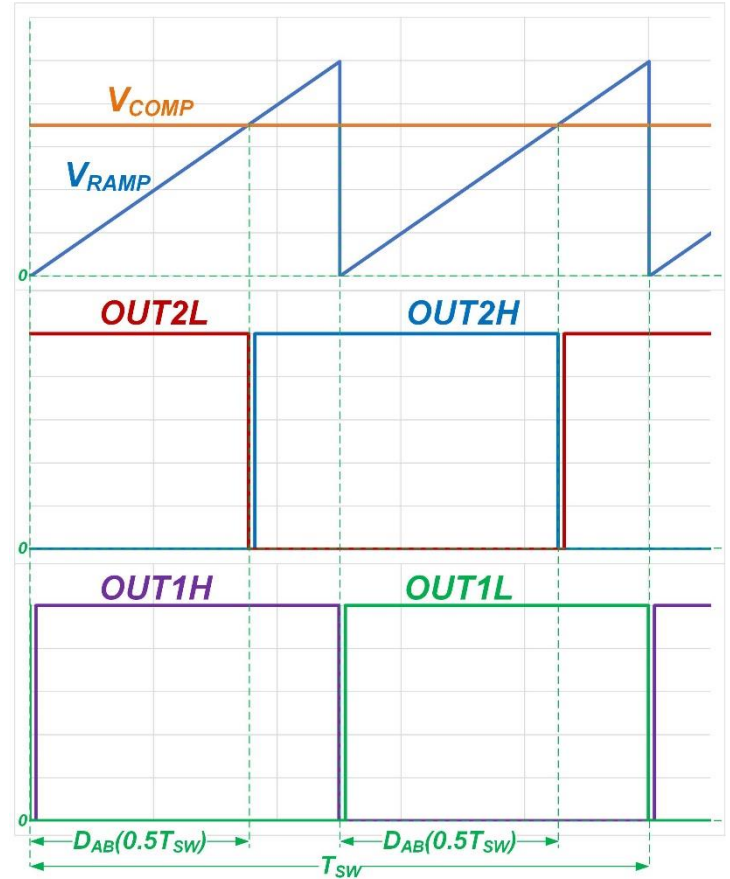
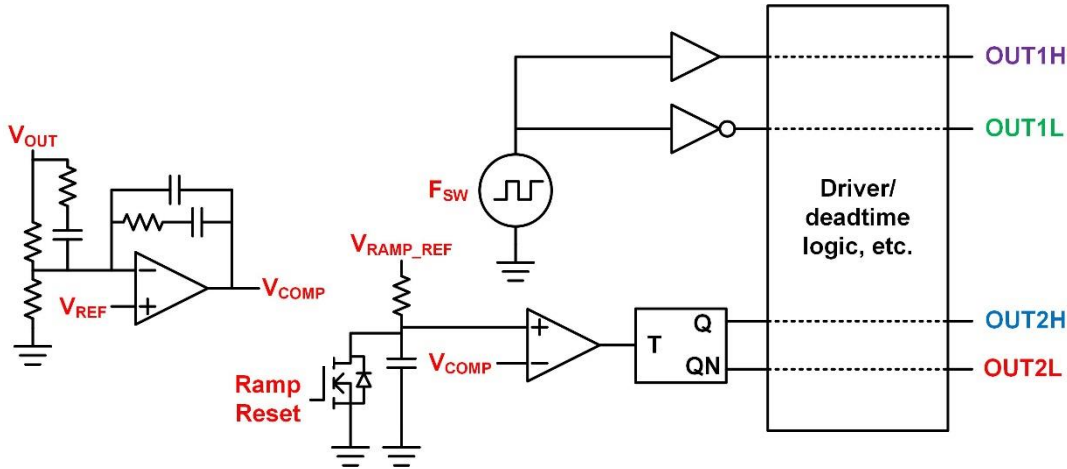
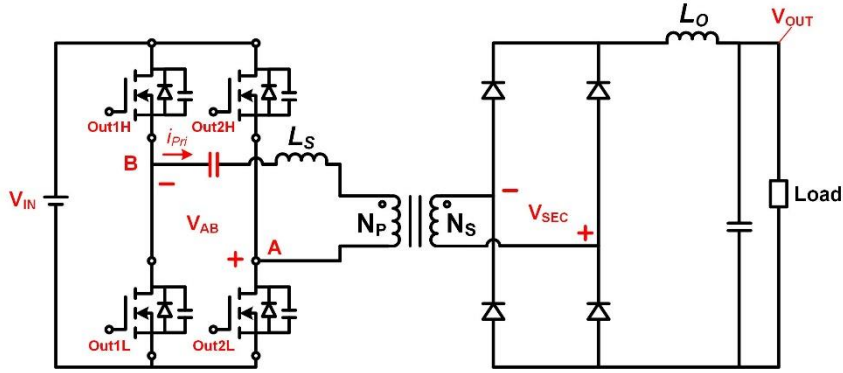
Clamp rectifier voltage spikes: Active clamp

- Creates current distortion on i_{PRI} and i_{SR}
 - Makes peak current-mode control difficult
- Q_{CL} only needs to conduct for a very short period to clamp
 - Longer $D_{ACL}T_S \Rightarrow$ larger i_{CL} and wider nonmonotonic current duration

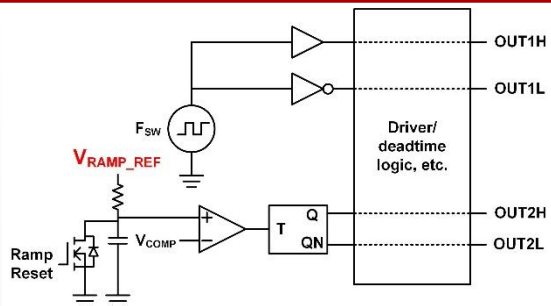


Control

PSFB control

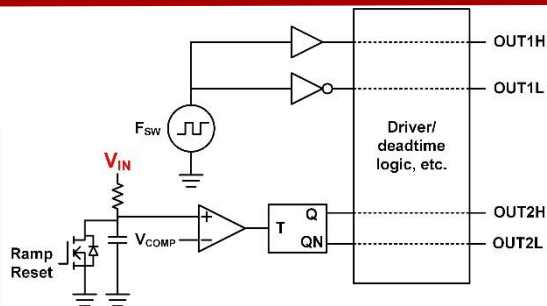


Modes of PSFB control



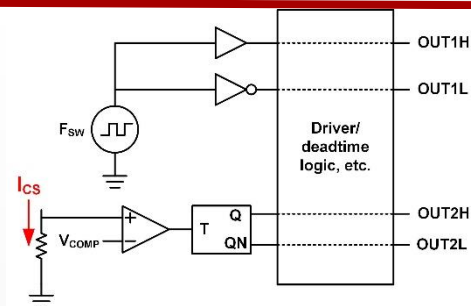
Voltage-mode control

- One high-side/low-side pair directly controlled by clock at the switching frequency
- Second high-side/low-side pair controlled by T flip-flop



Voltage-mode control with feedforward

- Voltage ramp reference is fed from V_{IN} or a voltage proportional to V_{IN}
- Immediate response to changes in input voltage

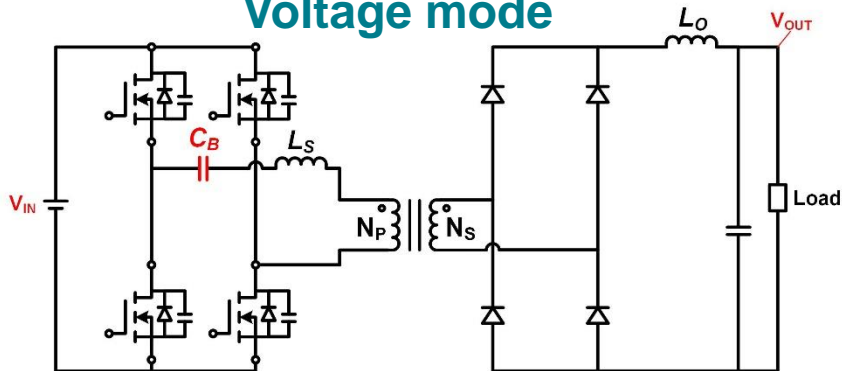


Peak current-mode control

- Current information from power stage replaces ramp signal
- Current-sense (CS) resistor plus amplifier or CS transformer common methods

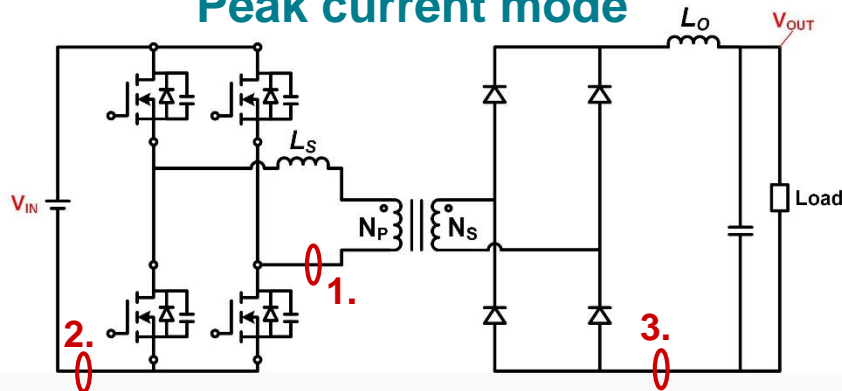
Voltage mode vs. peak current mode

Voltage mode



- DC blocking capacitor, C_B , is required to prevent transformer saturation
- Blocking capacitor will increase PSU footprint and voltage stress on SRs

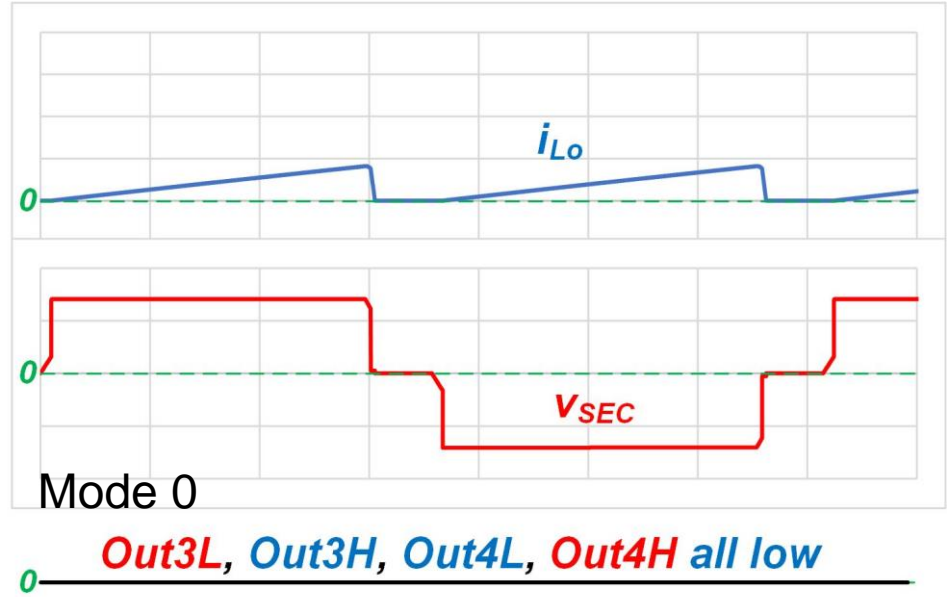
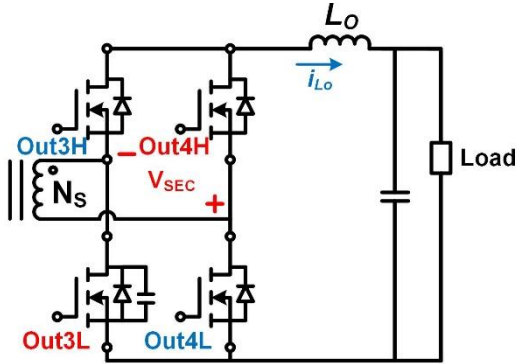
Peak current mode



1.
 - CS transformer core automatically resets
 - Contains no DC current information
2.
 - CS transformer reset more challenging
 - Higher parasitic loop inductance
3.
 - Resistor plus CS amplifier is common
 - Missing magnetizing current information

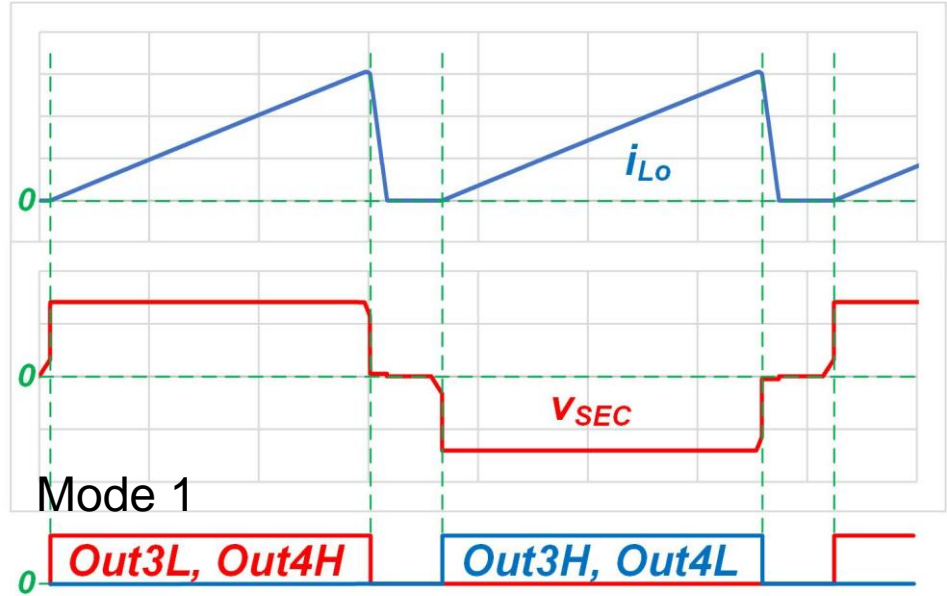
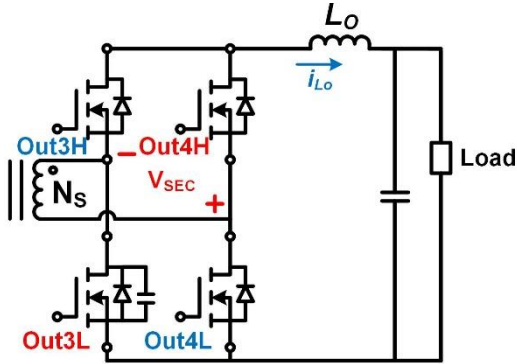
Different modes of SRs in a PSFB

- Mode 0: diode conduction (discontinuous conduction mode [DCM])



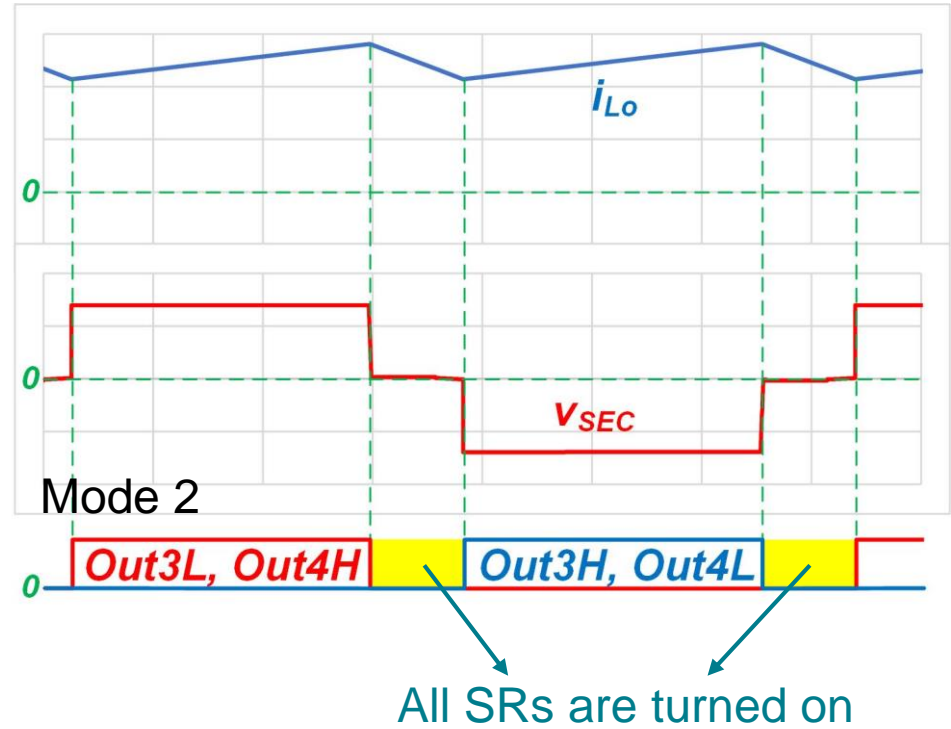
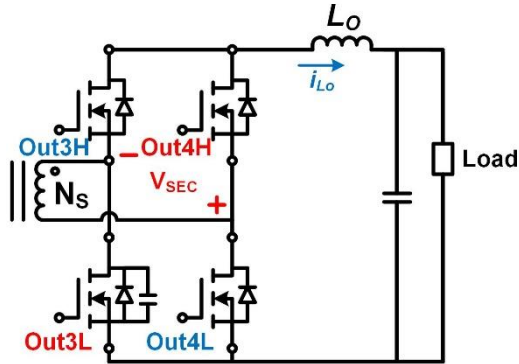
Different modes of SRs in a PSFB

- Mode 1: SR channel conduction only during inductor charging period
 - Avoids reverse current conduction



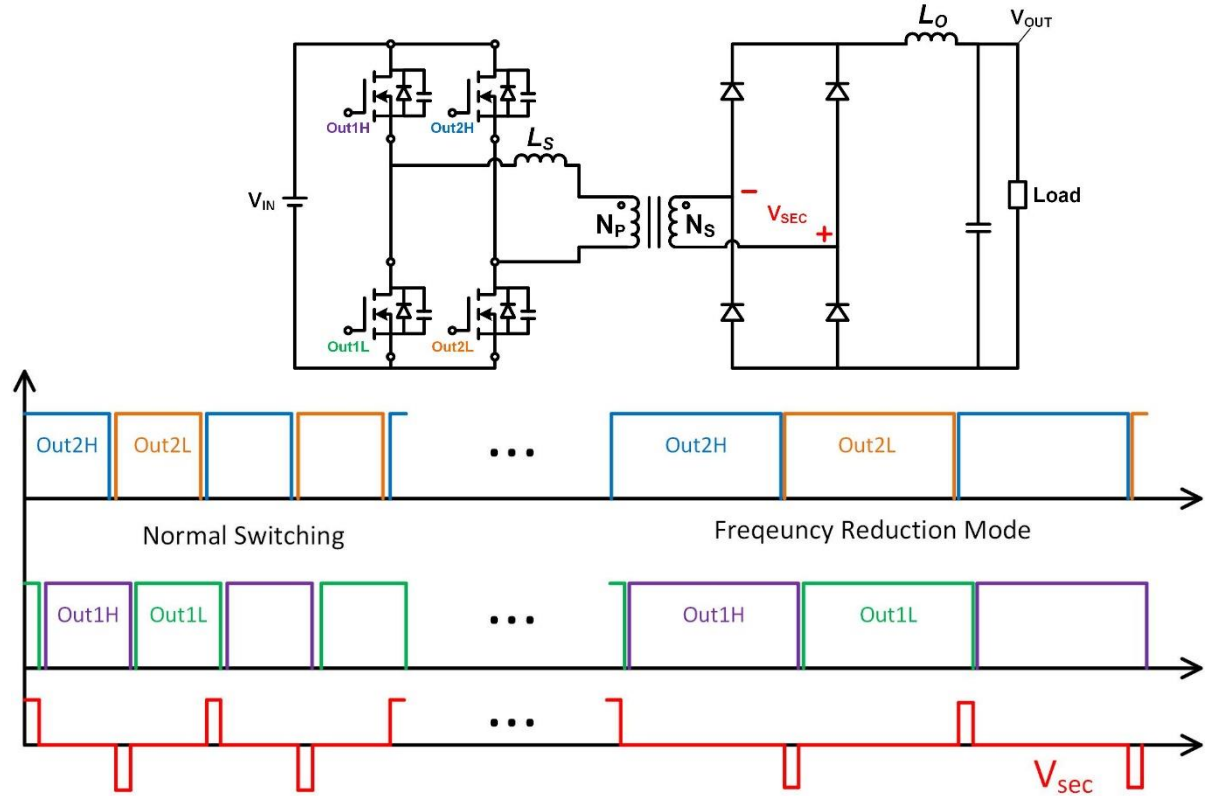
Different modes of SRs in a PSFB

- Mode 2: turn on all rectifiers/FETs during freewheeling period
 - Lower conduction losses at heavy loads



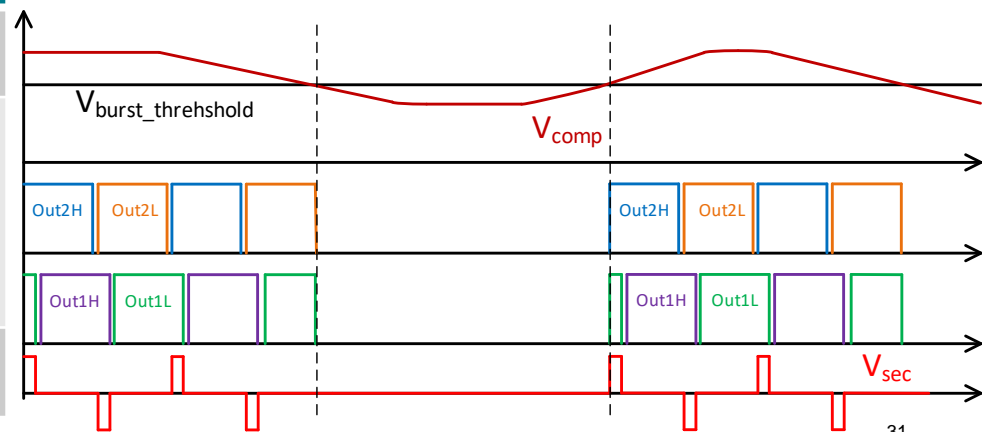
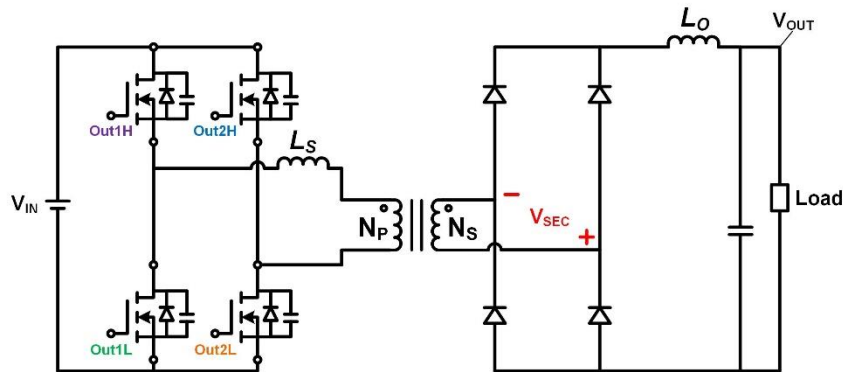
Light-load management: Frequency reduction mode

- The PSFB continuously switches
- Able to reduce effective duty cycle while maintaining a minimum on-time
- Gate-drive transformer not recommended (saturation)



Light-load management: Hysteretic burst mode

- When V_{comp} becomes $< V_{burst_threshold}$, the PSFB stops switching
- When V_{comp} becomes $> V_{burst_threshold}$, the PSFB resumes switching

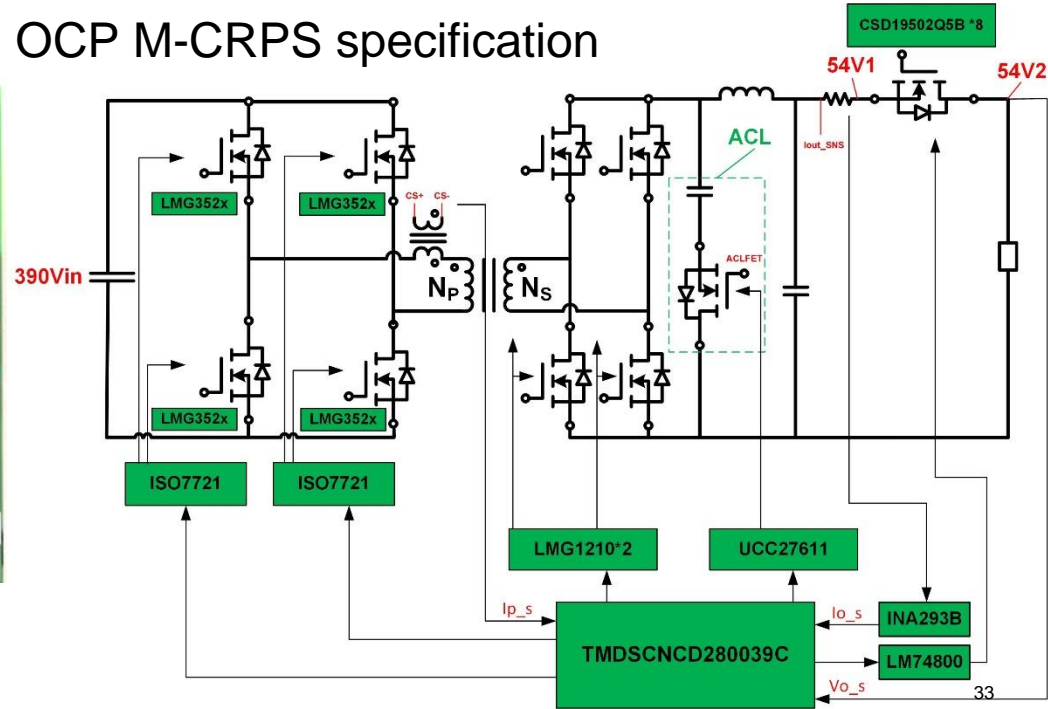
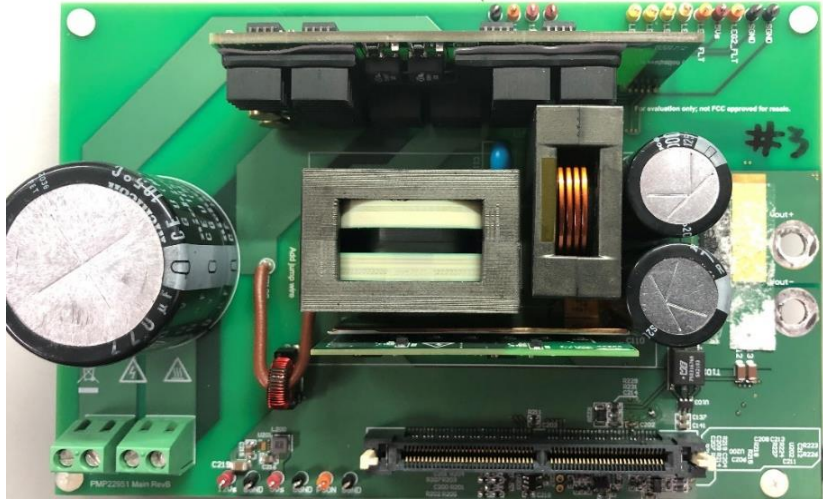


31

PSFB design example

PSFB design example: PMP22951

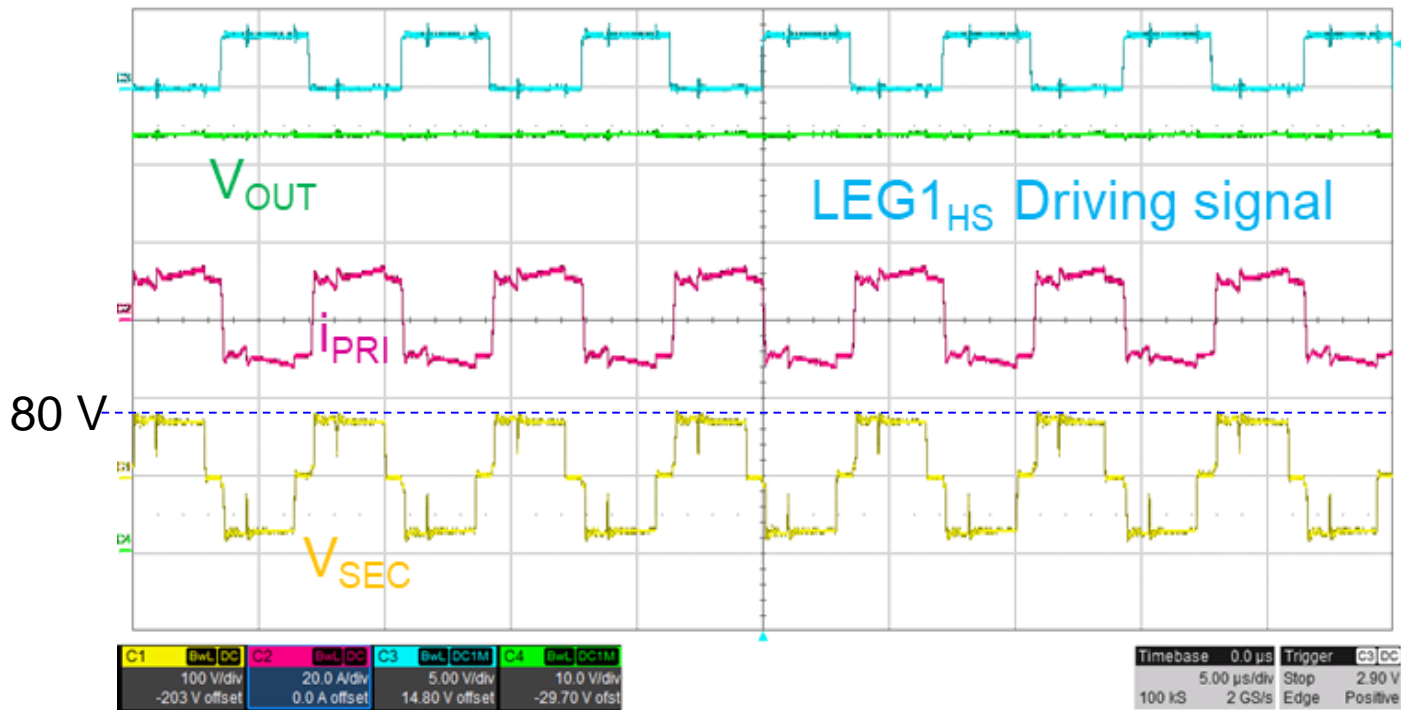
- Input voltage: 390 V_{nom}, 340 V_{min}
- Output: 54-V/3-kW max, targeting OCP M-CRPS specification



54-V, 3-kW Phase-Shifted Full Bridge with Active Clamp Reference Design

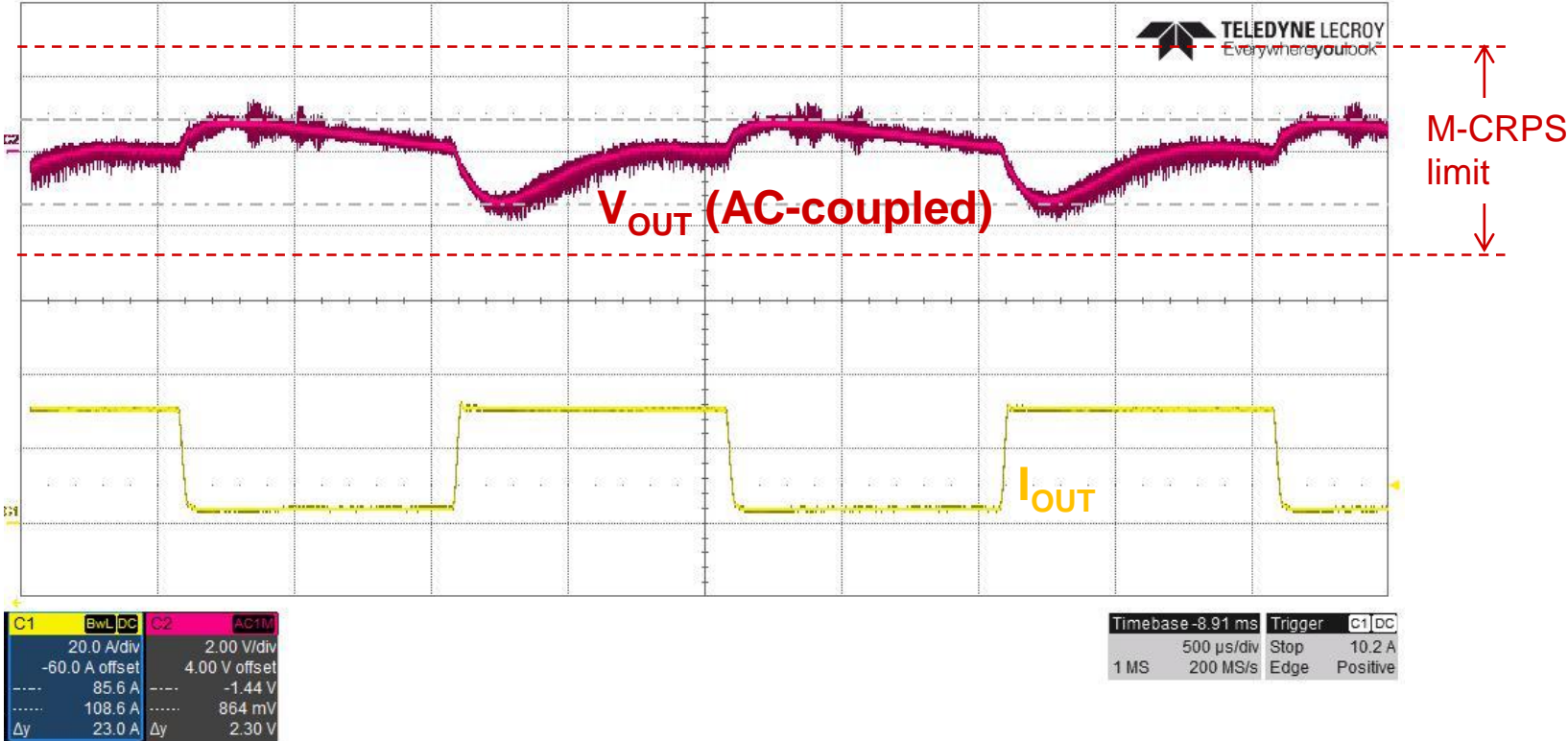
3-kW steady-state waveforms

- 140-kHz operation

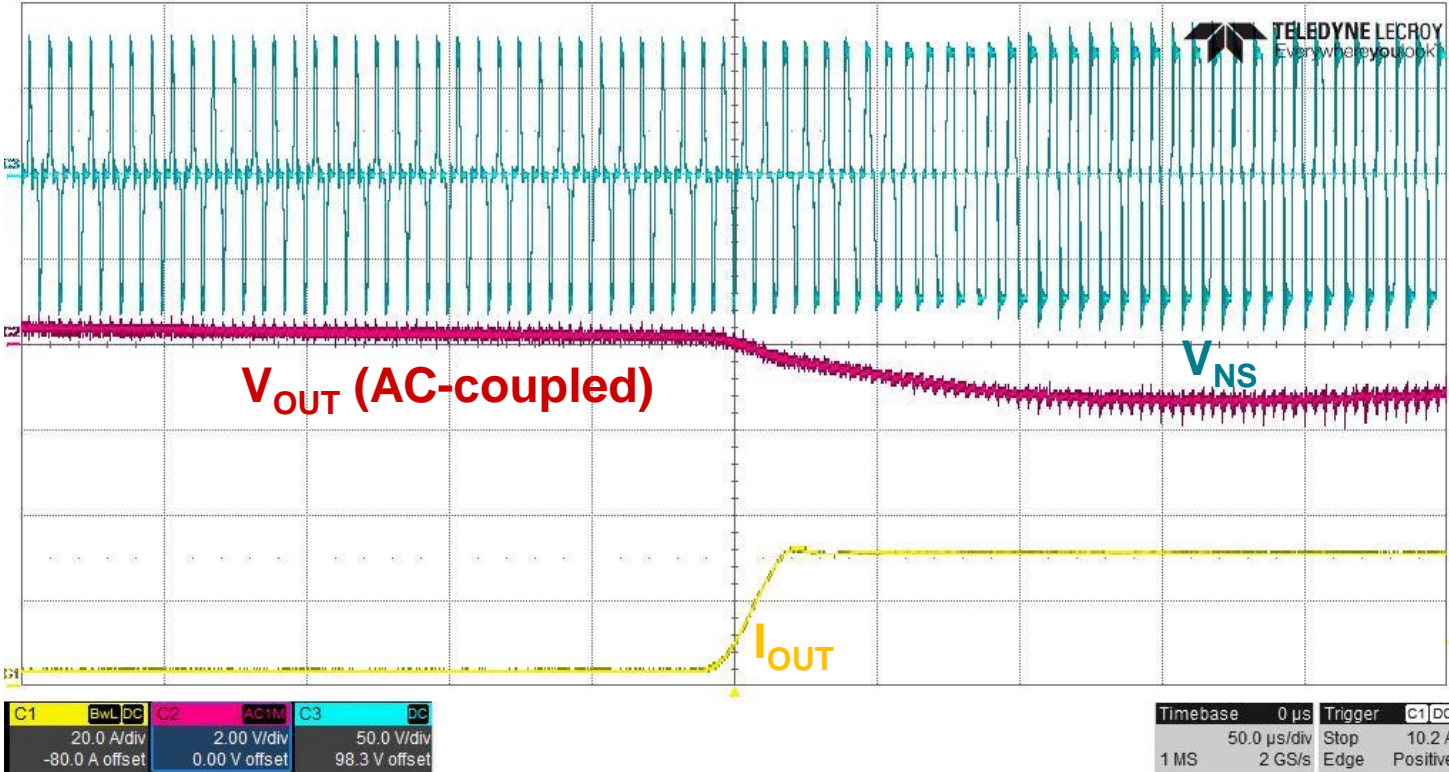


M-CRPS load transient, 50% load step (3 A to 31 A)

Load current



M-CRPS load transient, 50% load step (3 A to 31 A)



Summary

- A PSFB is a good candidate for applications that require a wide input/output voltage range and fast load transient response
- Characteristics among a PSFB and other isolated topologies
- PSFB operation principles
- Discussed different types of rectifiers, rectifier clamping options, modes of control
- Showcased a PSFB reference design meeting M-CRPS specifications with active clamping
- Calls to action in the following slides

Calls to action: Power trends and specifications

- **Power trends:**

- Yin, Richard. “Power Tips #109: Five major trends in power supply design for servers.” EDN Power Tips series, August 2022. <https://www.edn.com/five-major-trends-in-power-supply-design-for-servers/>

- **80 Plus standard:**

- <https://www.clearexult.com/80plus/program-details#program-details-table>

- **OCP M-CRPS specification:**

- <https://www.opencompute.org/wiki/Server/Working>

- **OCP Open Rack v3 specification:**

- https://www.opencompute.org/wiki/Open_Rack/SpecsAndDesigns

Calls to action: Topology comparisons

- **Resonant converter vs. DAB:**

- Yu, Sheng-Yang, et al. “Designing a high-power bidirectional AC/DC power supply using SiC FETs.” Texas Instruments Power Supply Design Seminar SEM2400, literature No. SLUP399, 2020. <https://www.ti.com/seclit/ml/slup399/slup399.pdf>

- **Resonant converter vs. PSFB:**

- Gillmor, Colin. “Comparison of PSFB and FB-LLC for high power DC/DC conversion” in Texas Instruments video library.
 - Part 1: <https://www.ti.com/video/5979520091001>
 - Part 2: <https://www.ti.com/video/5980232599001>
 - Part 3: <https://www.ti.com/video/5980257698001>
 - Part 4: <https://www.ti.com/video/5980260615001>
 - Part 5: <https://www.ti.com/video/5980344049001>
 - Part 6: <https://www.ti.com/video/5980375024001>

Calls to action: PSFB operation and rectifiers

- **PSFB operation and how to achieve PSFB soft switching:**
 - Sabate, J.A., et al. “Design considerations for high-voltage high-power full-bridge zero-voltage-switched PWM converter.” In Proc. APEC, 1990, pp. 275-284.
- **PSFB output rectifiers:**
 - Balogh, Laszlo. “The current-doubler rectifier: an alternative rectification technique for push-pull and bridge converters.” Texas Instruments application note, literature No. SLUA121. <https://www.ti.com/lit/an/sl原因121/sl原因121.pdf>

Calls to action: PSFB clamping options

- **Passive clamp:**

- Lin, Song-Yi, et al. “Analysis and design for RCD clamped snubber used in output rectifier of phase-shift full-bridge ZVS converters.” In IEEE Transactions on Industrial Electronics 45, no. 2 (April 1998), pp. 358-359.

- **Primary clamp:**

- Redl, Richard. “Optimum ZVS Full-Bridge DC/DC Converter with PWM Phase-Shift Control: Analysis, Design Considerations, and Experimental Results.” In Proc. APEC, 1994, pp. 159-165, vol. 1.

- **Active clamp:**

- Yu, Sheng-Yang, et al. “Achieving high converter efficiency with an active clamp in a PSFB converter.” Texas Instruments Analog Design Journal, literature No. SLYT835, Q1 2023. <https://www.ti.com/lit/an/slyt835/slyt835.pdf>

Calls to action: PSFB control and design examples

- **PSFB control:**

- Wang, Shi-song, et al. “Small-Signal Modeling of Phase-Shift Full-Bridge Converter with Peak Current Mode Control.” 2020 IEEE ASEM, Tianjin, China, 2020, pp. 1-2.
- Ahmed, M.R., et al. “Enhanced Models for Current-Mode Controllers of the Phase-Shifted Full Bridge Converter with Current Doubler Rectifier.” In IEEE ECCE Asia 2019, pp. 3271-3278.
- Vlatkovic V., et al. “Small-signal analysis of the phase-shifted PWM converter.” Published in IEEE Transactions on Power Electronics 7, issue 1 (January 1992): pp. 128-135.
- Basso, Christophe. “Transfer Functions of Switching Converters.” Faraday Press, 2021.

- **Design examples:**

- 3-kW (400 V to 12 V) Phase-Shifted Full Bridge with Active Clamp Reference Design:
<https://www.ti.com/tool/PMP23126>
- 3-kW (400 V to 54 V) Phase-Shifted Full Bridge with Active Clamp Reference Design:
<https://www.ti.com/tool/PMP22951>



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