

## ISOx5 隔离式 3.3V 半双工和全双工 RS-485 收发器

### 1 特性

- 符合或超出 TIA/EIA RS-485 的要求
- 1/8 单位负载，一条总线上多达 256 个节点
- 信号传输速率高达 1Mbps
- 热关断保护
- 低总线电容 - 16pF (典型值)
- 50kV/ $\mu$ s 典型瞬态抗扰度
- 针对总线开路、短路及空闲状态的失效防护接收器
- 可耐受 5V 电压的 3.3V 输入
- 安全及管理批准
  - 符合 DIN EN IEC 60747-17 (VDE 0884-17) 标准的 4000V<sub>PK</sub> V<sub>IOTM</sub>, 560V<sub>PK</sub> V<sub>IORM</sub>
  - 符合 UL 1577 标准的 2500V<sub>RMS</sub> 隔离额定值
  - 符合 CSA 62368-1 标准的 2500V<sub>RMS</sub> 隔离额定值。

### 2 应用

- 安全系统
- 化学品生产
- 工厂自动化
- 电机和运动控制
- HVAC 及楼宇自动化网络
- 联网安检站

### 3 说明

ISO15 是一款隔离式半双工差分线路收发器，而 ISO35 是一款适用于 TIA/EIA 485/422 应用的隔离式全双工差分线路驱动器和接收器。ISO15M 和 ISO35M 具有更宽的环境温度范围，为 -55°C 至 125°C；而 ISO15 和 ISO35 在 -40°C 至 85°C 的额定工作温度范围工作。

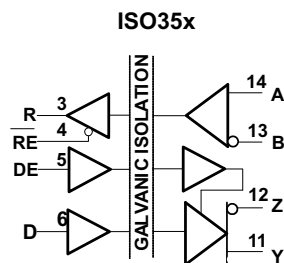
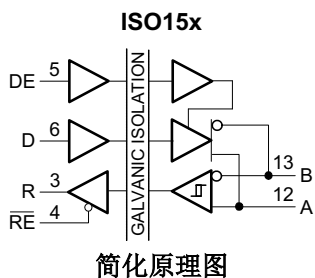
由于接地环路断开，这类器件能够在多得多的共模电压范围内运行，非常适合于远距离传输线路。经测试，该器件的对称隔离栅可在总线收发器和逻辑电平接口之间提供符合 VDE 标准的 4000V<sub>PK</sub> 隔离以及符合 UL 和 CSA 标准的 2500V<sub>RMS</sub> 隔离。

所有带线缆的 I/O 都容易遭受来自各种信号源的电瞬态噪声影响。这些瞬态噪声如果具有足够的幅度和持续时间，就有可能导致收发器和/或邻近的敏感电路受到损坏。此类隔离器件能够显著地提高保护水平并降低昂贵控制电路受损的风险。

#### 器件信息(1)

器件型号	封装	封装尺寸 (标称值)
ISO15	SOIC (16)	10.30mm x 7.50mm
ISO35		

(1) 如需了解所有可用封装，请参阅数据表末尾的可订购产品附录。



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## 4 Revision History

注：以前版本的页码可能与当前版本的页码不同

Changes from Revision G (March 2015) to Revision H (August 2023)	Page
• 将 CSA 标准更新为 CSA 62368-1，将 VDE 标准更新为 DIN EN IEC 60747-17 (VDE 0884-17).....	1
• 更新了整个文档中的表格、图和交叉参考的编号格式.....	1
• Updated Thermal Characteristics, Safety Limiting Values, and Thermal Derating Curves to provide more accurate system-level thermal calculations.....	7

Changes from Revision F (January 2012) to Revision G (October 2014)	Page
• 添加了引脚配置和功能部分、ESD 等级表、特性说明部分、器件功能模式、应用和实施部分、电源相关建议部分、布局部分、器件和文档支持部分以及机械、封装和可订购信息部分.....	1
• 将 VDE 标准更改为 DIN V VDE V 0884-10 (VDE V 0884-10):2006-12.....	1

Changes from Revision E (April 2010) to Revision F (January 2012)	Page
• 将特性从 $4000V_{\text{peak}} 560V_{\text{peak}} V_{\text{IORM}}$ (IEC...修订版 2) 更改为 $4000V_{\text{PK}} V_{\text{IOTM}}, 560V_{\text{PK}} V_{\text{IORM}}$ , IEC 60747-5-2 (VDE 0884, 修订版 2).....	1
• 将说明从“经测试, ...隔离。”更改为“经测试, 该器件的对称隔离栅可在...接口之间提供符合 VDE 标准的 $4000V_{\text{PK}}$ 隔离以及符合 UL 和 CSA 标准的 $2500V_{\text{RMS}}$ 隔离。”.....	1
• Updated electrical and switching characteristics to match device performance.....	9

Changes from Revision D (March 2009) to Revision E (April 2010)	Page
• 向数据表添加了器件 ISO15M 和 ISO35M.....	1
• 将说明从“ISO15 和 ISO35 适用的温度范围为 $-40^{\circ}\text{C}$ 至 $85^{\circ}\text{C}$ 。”更改为“ISO15M 和 ISO35M 具有更宽的环境温度范围, 为 $-55^{\circ}\text{C}$ 至 $125^{\circ}\text{C}$ ; 而 ISO15 和 ISO35 在 $-40^{\circ}\text{C}$ 至 $85^{\circ}\text{C}$ 的额定工作温度范围工作。”.....	1
• Added the Driver output pins Note for 图 7-1 through 图 7-4.....	14
• Changed the Driver output pins Note for 图 7-5 through 图 7-6.....	14

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**Changes from Revision B (July 2008) to Revision C (December 2008) Page**

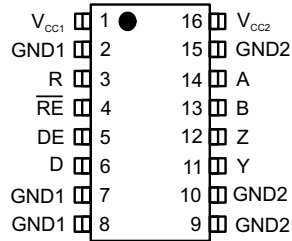
- 添加了 IEC...已批准..... **1**

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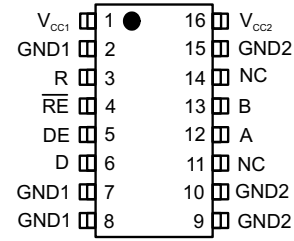
**Changes from Revision A (June 2008) to Revision B (July 2008) Page**

- 将 4000Vpeak 隔离更改为 4000Vpeak 隔离，560Vpeak VIORM UL 1577，IEC 60747-5-2 ( VDE 0884，修订版 2 ) ..... **1**
- Changed [图 7-13](#), Full-Duplex Common-Mode Transient Immunity Test Circuit..... **14**

## 5 Pin Configuration and Functions



**图 5-1. ISO35x DW Package  
16-Pin SOIC  
Top View**



**图 5-2. ISO15x DW Package  
16-Pin SOIC  
Top View**

**表 5-1. Pin Functions**

NAME	PIN		I/O	DESCRIPTION
	ISO15x NO.	ISO35x NO.		
A	12	14	I/O	ISO15x: Noninverting bus input or output
			I	ISO35x: Noninverting bus input
B	13	13	I/O	ISO15x: Inverting bus input or output
			I	ISO35x: Inverting bus input
D	6	6	I	Driver input
DE	5	5	I	Driver logic-high enable input
GND1	2,7,8	2,7,8	—	Logic side ground; internally connected
GND2	9,10,15	9,10,15	—	Bus side ground; internally connected
NC	11,14	—	—	Not connected internally; may be left floating
R	3	3	O	Receiver output
RE	4	4	I	Receiver logic-low enable
V <sub>CC1</sub>	1	1	—	Logic side power supply
V <sub>CC2</sub>	16	16	—	Bus side power supply
Y	—	11	O	Noninverting bus output
Z	—	12	O	Inverting bus output

## 6 Specifications

### 6.1 Absolute Maximum Ratings

Over operating free-air temperature range (unless otherwise noted)<sup>(1)</sup>

		MIN	MAX	UNIT
V <sub>CC1</sub>	Supply voltage, side 1 <sup>(2)</sup>	-0.3	6	V
V <sub>CC2</sub>	Supply voltage, side 2 <sup>(2)</sup>	-0.3	6	V
V <sub>O</sub>	Voltage at any bus I/O terminal	-9	14	V
V <sub>it</sub>	Voltage input, transient pulse, A, B, Y, and Z (through 100Ω, see Figure 13)	-50	50	V
V <sub>I</sub>	Voltage input at any D, DE or RE terminal	-0.5	6	V
I <sub>O</sub>	Receiver output current	-10	10	mA
T <sub>J</sub>	Junction temperature		150	°C
T <sub>STG</sub>	Storage temperature	-65	150	°C

- (1) Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under *Recommended Operating Conditions*. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
- (2) All voltage values except differential I/O bus voltages are with respect to network ground terminal and are peak voltage values.

### 6.2 ESD Ratings

			VALUE	UNIT
V <sub>(ESD)</sub>	Human body model (HBM), per ANSI/ESDA/JEDEC JS-001 <sup>(1)</sup>	Bus pins and GND1	±6000	V
V <sub>(ESD)</sub>	Human body model (HBM), per ANSI/ESDA/JEDEC JS-001 <sup>(1)</sup>	Bus pins and GND2	±16000	V
V <sub>(ESD)</sub>	Human body model (HBM), per ANSI/ESDA/JEDEC JS-001 <sup>(1)</sup>	All pins	±4000	V
V <sub>(ESD)</sub>	Charged device model (CDM), per JEDEC specification JESD22-C101, all pins <sup>(2)</sup>		±1000	V
V <sub>(ESD)</sub>	Machine model ANSI/ESDS5.2-1996		±200	V

- (1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.
- (2) JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.

### 6.3 Recommended Operating Conditions

		MIN	TYP	MAX	UNIT
V <sub>CC1</sub>	Supply Voltage, Side 1	3.15	3.3	3.6	V
V <sub>CC2</sub>	Supply Voltage, Side 2	3.15	3.3	3.6	V
V <sub>OC</sub>	Common Mode voltage at any bus terminal: A or B	-7		12	V
V <sub>IH</sub>	High-level input voltage (D, DE, $\overline{RE}$ inputs)	2		V <sub>CC1</sub>	V
V <sub>IL</sub>	Low-level input voltage (D, DE, RE inputs)	0		0.8	V
V <sub>ID</sub>	Differential input voltage, A with respect to B	-12		12	V
R <sub>L</sub>	Differential load resistance	54	60		Ω
I <sub>O</sub>	Output current, Driver	-60		60	mA
1/t <sub>UI</sub>	Signaling rate ISO15x and ISO35x			1	Mbps
T <sub>A</sub>	Operating ambient temperature (ISO15 and ISO35)	-40	25	85	°C
	Operating ambient temperature (ISO15M and ISO35M)	-55	25	125	°C

		<b>MIN</b>	<b>TYP</b>	<b>MAX</b>	<b>UNIT</b>
T <sub>J</sub>	Operating junction temperature (ISO15 and ISO35)	-40		150	°C
	Operating junction temperature (ISO15M and ISO35M)	-55		150	°C

## 6.4 Thermal Information

THERMAL METRIC <sup>(1)</sup>		ISO15, ISO35	UNIT
		DW (SOIC)	
		16 PINS	
$R_{\theta JA}$	Junction-to-ambient thermal resistance	79.6	°C/W
$R_{\theta JC(top)}$	Junction-to-case (top) thermal resistance	39.7	°C/W
$R_{\theta JB}$	Junction-to-board thermal resistance	44.6	°C/W
$\psi_{JT}$	Junction-to-top characterization parameter	11.8	°C/W
$\psi_{JB}$	Junction-to-board characterization parameter	44	°C/W
$R_{\theta JC(bot)}$	Junction-to-case (bottom) thermal resistance	—	°C/W

(1) For more information about traditional and new thermal metrics, see the [Semiconductor and IC Package Thermal Metrics application report](#).

## 6.5 Power Ratings

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
$P_D$	Maximum power dissipation (both sides) $V_{CC1} = V_{CC2} = 3.6\text{ V}$ , $T_J = 150^\circ\text{C}$ , $C_L = 15\text{ pF}$ , Input a 0.5 MHz 50% duty cycle square wave			220	mW

## 6.6 Insulation Specifications

PARAMETER		TEST CONDITIONS	VALUE	UNIT
			DW-16	
CLR	External clearance <sup>(1)</sup>	Shortest terminal-to-terminal distance through air	8	mm
CPG	External creepage <sup>(1)</sup>	Shortest terminal-to-terminal distance across the package surface	8	mm
DTI	Distance through the insulation	Minimum internal gap (internal clearance)	8	um
CTI	Comparative tracking index	DIN EN 60112 (VDE 0303-11); IEC 60112	>400	V
	Material group	According to IEC 60664-1	II	
	Overvoltage category per IEC 60664-1	Rated mains voltage $\leq 150\text{ V}_{RMS}$	I-IV	
		Rated mains voltage $\leq 300\text{ V}_{RMS}$	I-III	
<b>DIN EN IEC 60747-17 (VDE 0884-17) <sup>(2)</sup></b>				
$V_{IORM}$	Maximum repetitive peak isolation voltage	AC voltage (bipolar)	560	$V_{PK}$
$V_{IOTM}$	Maximum transient isolation voltage	$V_{TEST} = V_{IOTM}$ , $t = 60\text{ s}$ (qualification); $V_{TEST} = 1.2 \times V_{IOTM}$ , $t = 1\text{ s}$ (100% production)	4000	$V_{PK}$
$q_{pd}$	Apparent charge <sup>(3)</sup>	Method b; At routine test (100% production) $V_{ini} = 1.2 \times V_{IOTM}$ , $t_{ini} = 1\text{ s}$ ; $V_{pd(m)} = 1.5 \times V_{IORM}$ , $t_m = 1\text{ s}$	$\leq 5$	pC
$C_{IO}$	Barrier capacitance, input to output <sup>(4)</sup>	$V_{IO} = 0.4 \times \sin(2\pi ft)$ , $f = 1\text{ MHz}$	2	pF
$C_i$	Input capacitance to ground	$V_i = V_{CC}/2 + 0.4 \times \sin(2\pi ft)$ , $f = 1\text{ MHz}$ , $V_{CC} = 3.3\text{ V}$	2	pF
$R_{IO}$	Isolation resistance <sup>(4)</sup>	$V_{IO} = 500\text{ V}$ , $T_A = 25^\circ\text{C}$	$>10^{12}$	$\Omega$
		$V_{IO} = 500\text{ V}$ , $T_S = 150^\circ\text{C}$	$>10^9$	
	Pollution degree		2	
	Climatic Category		40/125/21	
<b>UL 1577</b>				

PARAMETER		TEST CONDITIONS	VALUE	UNIT
			DW-16	
V <sub>ISO</sub>	Maximum withstanding isolation voltage	V <sub>TEST</sub> = V <sub>ISO</sub> , t = 60 s (qualification), V <sub>TEST</sub> = 1.2 x V <sub>ISO</sub> , t = 1 s (100% production)	2500	V <sub>RMS</sub>

- Creepage and clearance requirements should be applied according to the specific equipment isolation standards of an application. Care should be taken to maintain the creepage and clearance distance of a board design to ensure that the mounting pads of the isolator on the printed-circuit board do not reduce this distance. Creepage and clearance on a printed-circuit board become equal in certain cases. Techniques such as inserting grooves and/or ribs on a printed-circuit board are used to help increase these specifications.
- This coupler is suitable for *basic electrical insulation* only within the safety ratings. Compliance with the safety ratings shall be ensured by means of suitable protective circuits.
- Apparent charge is electrical discharge caused by a partial discharge (pd).
- All pins on each side of the barrier tied together creating a two-terminal device.

### 6.7 Safety-Related Certifications

VDE	CSA	UL
Certified according to DIN EN IEC 60747-17 (VDE 0884-17)	Certified according to IEC 60950-1 and IEC 62368-1	Certified according to UL 1577 Component Recognition Program
Basic insulation, 4000 V <sub>PK</sub> Maximum transient isolation voltage, 560 V <sub>PK</sub> Maximum repetitive peak isolation voltage	2500 V <sub>RMS</sub> Isolation rating, Reinforced insulation per CSA 60950-1 and IEC 60950-1 148V <sub>RMS</sub> working voltage; Basic insulation per CSA 62368-1 and IEC 62368-1 300V <sub>RMS</sub> working voltage	Single protection, 2500 V <sub>RMS</sub>
Certificate number: 40047657	Master contract number: 220991	File number: E181974

### 6.8 Safety Limiting Values

Safety limiting<sup>(1)</sup> intends to minimize potential damage to the isolation barrier upon failure of input or output circuitry.

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
<b>DW-16 PACKAGE</b>						
I <sub>S</sub>	Safety input, output, or supply current	R <sub>θJA</sub> = 79.6°C/W, V <sub>I</sub> = 3.6 V, T <sub>J</sub> = 150°C, T <sub>A</sub> = 25°C.			436	mA
T <sub>S</sub>	Maximum safety temperature				150	°C

- The maximum safety temperature, T<sub>S</sub>, has the same value as the maximum junction temperature, T<sub>J</sub>, specified for the device. The I<sub>S</sub> and P<sub>S</sub> parameters represent the safety current and safety power respectively. The maximum limits of I<sub>S</sub> and P<sub>S</sub> should not be exceeded. These limits vary with the ambient temperature, T<sub>A</sub>.  
The junction-to-air thermal resistance, R<sub>θJA</sub>, in the table is that of a device installed on a high-K test board for leaded surface-mount packages. Use these equations to calculate the value for each parameter:  
T<sub>J</sub> = T<sub>A</sub> + R<sub>θJA</sub> × P, where P is the power dissipated in the device.  
T<sub>J(max)</sub> = T<sub>S</sub> = T<sub>A</sub> + R<sub>θJA</sub> × P<sub>S</sub>, where T<sub>J(max)</sub> is the maximum allowed junction temperature.  
P<sub>S</sub> = I<sub>S</sub> × V<sub>I</sub>, where V<sub>I</sub> is the maximum input voltage.



## 6.9 Electrical Characteristics: Driver

All typical specs are at  $V_{CC1}=3.3V$ ,  $V_{CC2}=5V$ ,  $T_A=27^\circ C$ , (Min/Max specs are over recommended operating conditions unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
V <sub>OD</sub>	Driver differential-output voltage magnitude	I <sub>O</sub> = 0 mA, no load	2.5		V <sub>CC2</sub>	V
		R <sub>L</sub> = 54 Ω, See Figure 3	1.5	2		V
		R <sub>L</sub> = 100 Ω (RS-422), See Figure 3	2	2.3		V
		V <sub>test</sub> from - 7 V to +12 V, See Figure 4	1.5			V
Δ V <sub>OD</sub>	Change in differential output voltage between two states	See Figure 3 and Figure 4	- 200		200	mV
V <sub>OC</sub>	Common-mode output voltage	See Figure 5	1	2.6	3	V
ΔV <sub>OC(SS)</sub>	change in steady-state common-mode output voltage between two states	See Figure 5	- 100		100	mV
V <sub>OC(PP)</sub>	Peak-to-peak common-mode output voltage	See Figure 5		0.5		V
I <sub>I</sub>	Input current	D, DE, V <sub>I</sub> at 0 V or V <sub>CC1</sub>	- 10		10	μA
I <sub>OZ</sub>	High-impedance state output current	ISO15 See receiver input current				
		ISO35 V <sub>Y</sub> or V <sub>Z</sub> = 12 V			90	μA
		ISO35 V <sub>Y</sub> or V <sub>Z</sub> = 12 V, V <sub>CC</sub> = 0			90	μA
		ISO35 V <sub>Y</sub> or V <sub>Z</sub> = - 7 V		-10		μA
		ISO35 V <sub>Y</sub> or V <sub>Z</sub> = - 7 V, V <sub>CC</sub> = 0		-10		μA
I <sub>OS</sub>	Short-circuit output current	V <sub>A</sub> or V <sub>B</sub> at - 7 V	- 250		250	mA
		V <sub>A</sub> or V <sub>B</sub> at 12 V	- 250		250	mA
C <sub>OD</sub>	Differential output capacitance	V <sub>I</sub> = 0.4 sin (4E6 π t) + 0.5 V, DE at 0 V		16		pF
CMTI	Common-mode transient immunity	V <sub>I</sub> = V <sub>CC</sub> or 0 V, See Figure 14 and Figure 15	25	50		kV/μs

## 6.10 Electrical Characteristics: Receiver

All typical specs are at  $V_{CC1}=3.3V$ ,  $V_{CC2}=5V$ ,  $T_A=27^\circ C$ , (Min/Max specs are over recommended operating conditions unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
V <sub>IT+</sub>	Positive-going input threshold voltage	I <sub>O</sub> = - 8 mA			- 20	mV
V <sub>IT-</sub>	Negative-going input threshold voltage	I <sub>O</sub> = 8 mA	- 200			mV
V <sub>hys</sub>	Input hysteresis (V <sub>IT+</sub> - V <sub>IT-</sub> )			50		mV
V <sub>O</sub>	Output Voltage	V <sub>ID</sub> = 200mV, I <sub>O</sub> = -8mA	2.4			V
		V <sub>ID</sub> = - 200mV, I <sub>O</sub> = 8mA			0.4	V
I <sub>OZ</sub>	Output high-impedance current on the R pin	V <sub>I</sub> = - 7 to 12 V, Other input = 0 V	- 1		1	μA
I <sub>A</sub> or I <sub>B</sub>	Bus input current	-55°C ≤ T <sub>A</sub> ≤ 85°C V <sub>A</sub> or V <sub>B</sub> = 12 V		50	100	μA
		-55°C ≤ T <sub>A</sub> ≤ 85°C V <sub>A</sub> or V <sub>B</sub> = 12 V, V <sub>CC</sub> = 0		50	100	μA
		85°C ≤ T <sub>A</sub> ≤ 125°C V <sub>A</sub> or V <sub>B</sub> = 12 V			200	μA
		85°C ≤ T <sub>A</sub> ≤ 125°C V <sub>A</sub> or V <sub>B</sub> = 12 V, V <sub>CC</sub> = 0			200	μA
		-55°C ≤ T <sub>A</sub> ≤ 125°C V <sub>A</sub> or V <sub>B</sub> = - 7 V	-100	-40		μA
		-55°C ≤ T <sub>A</sub> ≤ 125°C V <sub>A</sub> or V <sub>B</sub> = - 7 V, V <sub>CC</sub> = 0	-100	-30		μA
I <sub>IH</sub>	High-level input current, RE	V <sub>IH</sub> = 2 V	-10			μA

All typical specs are at  $V_{CC1}=3.3V$ ,  $V_{CC2}=5V$ ,  $T_A=27^{\circ}C$ , (Min/Max specs are over recommended operating conditions unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
$I_{IL}$	Low-level input current, RE	$V_{IL} = 0.8 V$	-10			$\mu A$
$R_{ID}$	Differential input resistance	A, B	48			kohm
$C_{ID}$	Differential input capacitance	$V_I = 0.4 \sin(4E6 \pi t) + 0.5V$ , DE at 0 V		16		pF

## 6.11 Supply Current

Bus loaded or unloaded (over recommended operating conditions unless otherwise noted)

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
<b>DRIVER ENABLED, RECEIVER DISABLED</b>					
I <sub>CC1</sub>	ISO35x and ISO15x: $\overline{RE}$ at 0 V or V <sub>CC</sub> , DE at 0 V, No load (driver disabled)			8	mA
I <sub>CC1</sub>	ISO35x and ISO15x: $\overline{RE}$ at 0 V or V <sub>CC</sub> , DE at V <sub>CC</sub> , No load (driver enabled)			8	mA
I <sub>CC2</sub>	ISO35x and ISO15x: $\overline{RE}$ at 0 V or V <sub>CC</sub> , DE at 0 V, No load (driver disabled)			15	mA
I <sub>CC2</sub>	ISO35x and ISO15x: $\overline{RE}$ at 0 V or V <sub>CC</sub> , DE at V <sub>CC</sub> , No Load (driver enabled)			19	mA

## 6.12 Switching Characteristics: Driver

All typical specs are at  $V_{CC1}=3.3V$ ,  $V_{CC2}= 3.3V$ ,  $T_A=27^{\circ}C$ , (Min/Max specs are over recommended operating conditions unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
<b>500-kbps DEVICES</b>						
$t_r, t_f$	Differential output rise time and fall time	ISO15M and ISO35M	120	180	350	ns
$t_{PHZ}$	Propagation delay, high-level-to-high-impedance output	See Figure 7			205	ns
$t_{PLZ}$	Propagation delay, low-level to high-impedance output	See Figure 8			330	ns
$t_{PZL}$	Propagation delay, standby-to-low-level output	See Figure 8			530	ns
$t_{PHL}, t_{PLH}$	Propagation delay	See Figure 6			340	ns
tsk(p)	Pulse skew ( $ t_{PHL} - t_{PLH} $ )	See Figure 6		6		ns
$t_r, t_f$	Differential output rise time and fall time	ISO15 and ISO35	120	180	300	ns
$t_{PZH}$	Propagation delay, high-impedance-to-high-level output	See Figure 7			530	ns

## 6.13 Switching Characteristics: Receiver

All typical specs are at  $V_{CC1}=3.3V$ ,  $V_{CC2}= 3.3V$ ,  $T_A=27^{\circ}C$ , (Min/Max specs are over recommended operating conditions unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
<b>500-kbps DEVICES</b>						
$t_{PHL}, t_{PLH}$	Propagation delay	See Figure 10			100	ns
tsk(p)	Pulse skew ( $ t_{PHL} - t_{PLH} $ )	ISO15 and ISO35			13	ns
tsk(p)	Pulse skew ( $ t_{PHL} - t_{PLH} $ )	ISO15M and ISO35M			18	ns
$t_r, t_f$	Differential output rise time and fall time	ISO15 and ISO35		2	4	ns
$t_r, t_f$	Differential output rise time and fall time	ISO15M and ISO35M		2	6	ns
$t_{PHZ}, t_{PLZ}$	Propagation delay, high-impedance-to-high-level output, Propagation delay, high-impedance-to-low-level output	DE at 0 V, See Figure 11 and Figure 12		13	25	ns
$t_{PZH}, t_{PZL}$	Propagation delay, high-level-to-high-impedance output, Propagation delay, low-level to high-impedance output	DE at 0 V, See Figure 11 and Figure 12		13	25	ns

### 6.14 Insulation Characteristics Curves

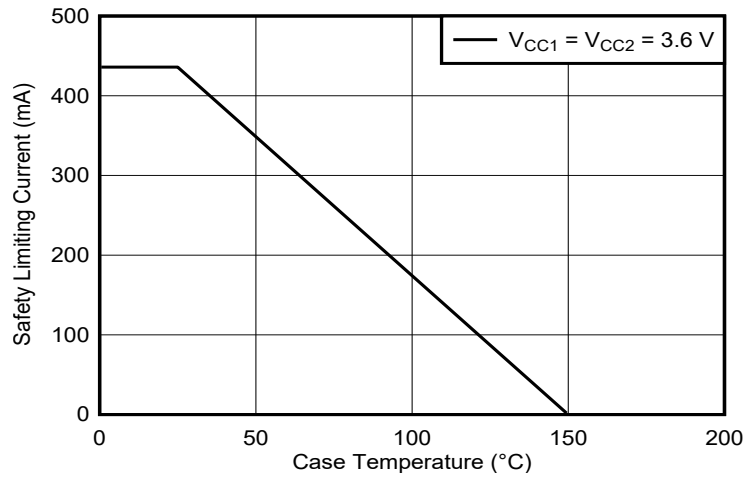


图 6-1. Thermal Derating Curve for Safety Limiting Power for DW-16 Package

### 6.15 Typical Characteristics

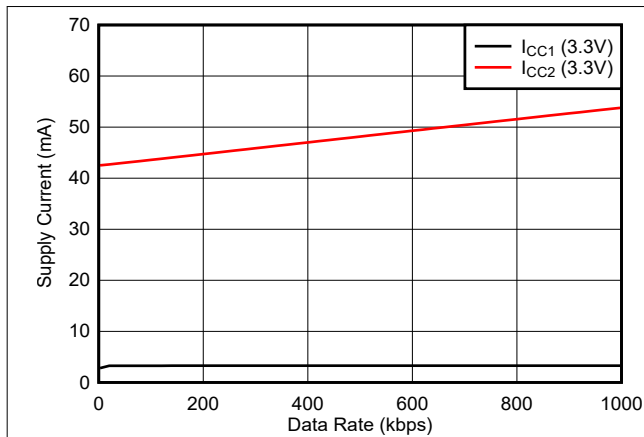


图 6-2. ISOx5 Supply Current vs Data Rate With Load

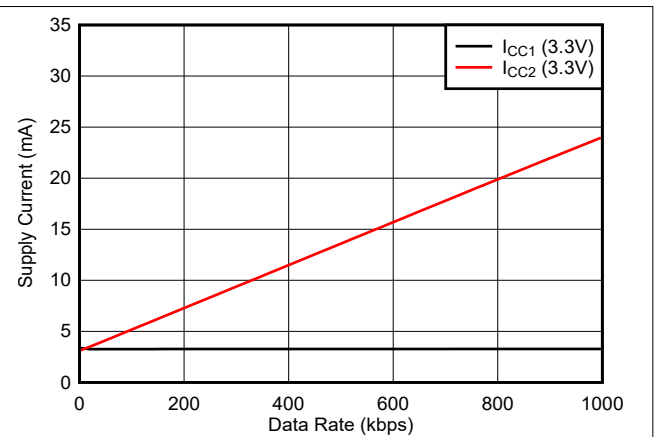


图 6-3. ISOx5 Supply Current vs Data Rate With No Load

## 7 Parameter Measurement Information

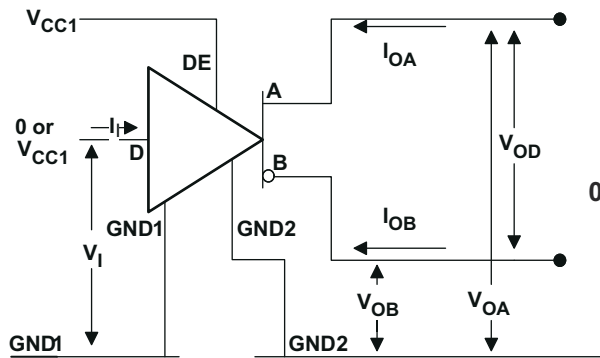


图 7-1. Driver  $V_{OD}$  Test and Current Definitions

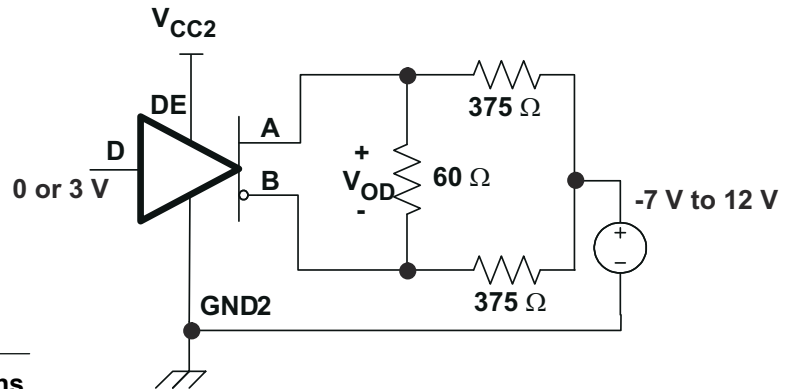


图 7-2. Driver  $V_{OD}$  With Common-Mode Loading Test Circuit

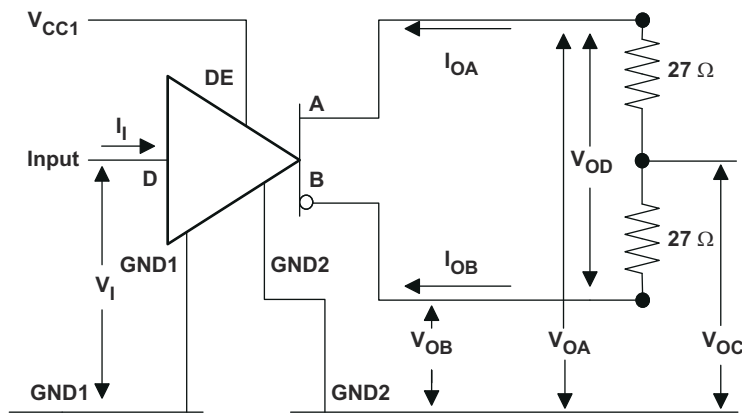
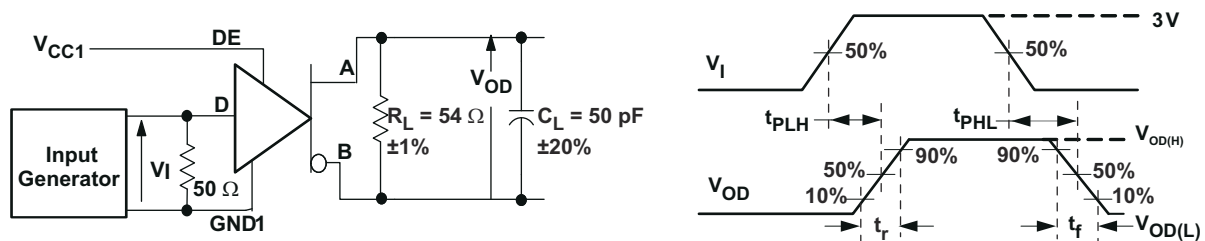


图 7-3. Test Circuit and Waveform Definitions for the Driver Common-Mode Output Voltage



Generator: PRR = 500 kHz, 50% duty cycle,  $t_r < 6\text{ns}$ ,  $t_f < 6\text{ns}$ ,  $Z_0 = 50\ \Omega$   $C_L$  includes fixture and Instrumentation Capacitance

图 7-4. Driver Switching Test Circuit and Voltage Waveforms

### 备注

Driver output pins are A and B for the ISO15 (see 图 7-1 through 图 7-4). These correspond to ISO35 pins Y and Z

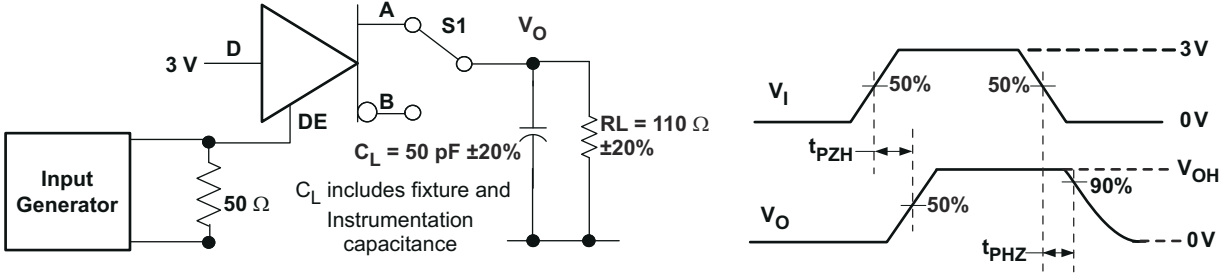


图 7-5. Driver High-Level Output Enable and Disable Time Test Circuit and Voltage Waveforms

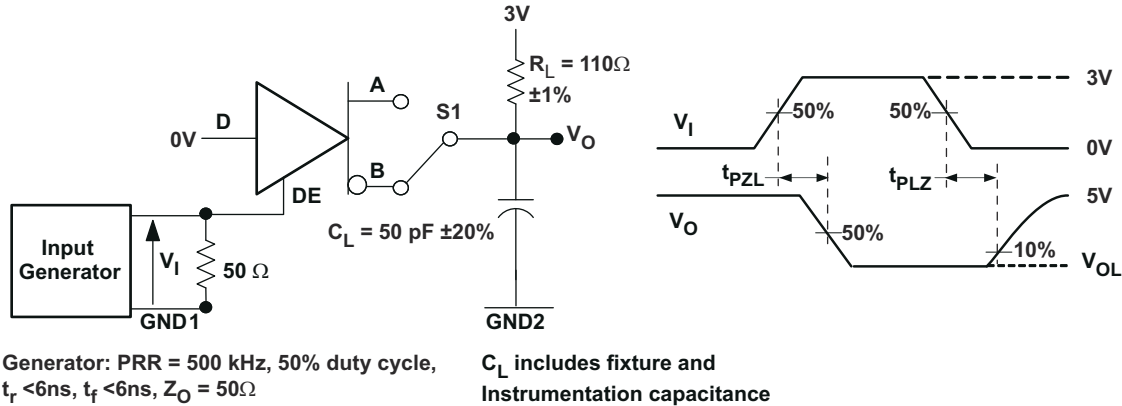


图 7-6. Driver Low-Level Output Enable and Disable Time Test Circuit and Voltage Waveform

备注

Driver output pins are A and B for the ISO15 (see 图 7-5 through 图 7-6). These correspond to ISO35 pins Y and Z

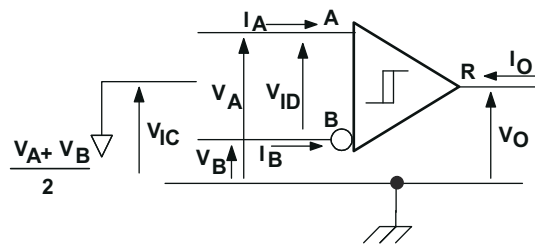


图 7-7. Receiver Voltage and Current Definitions

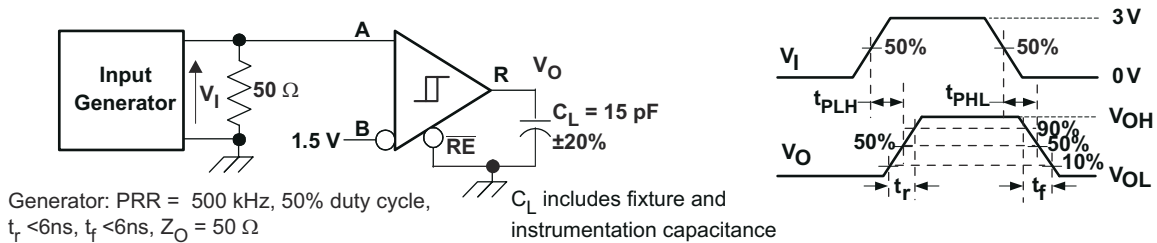


图 7-8. Receiver Switching Test Circuit and Waveforms

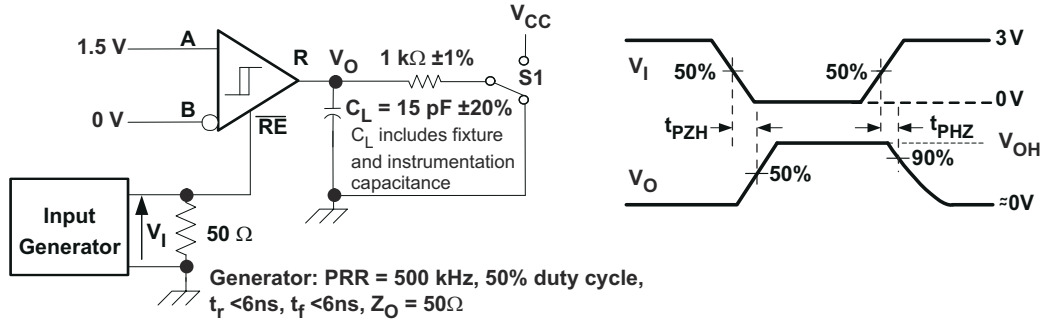


图 7-9. Receiver Enable Test Circuit and Waveforms, Data Output High

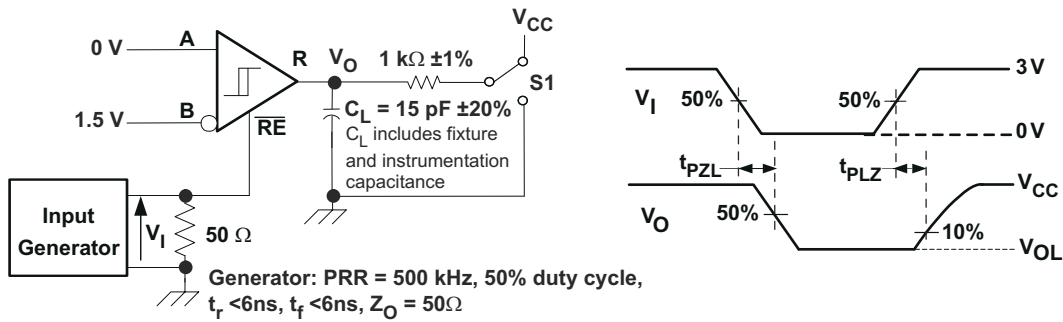
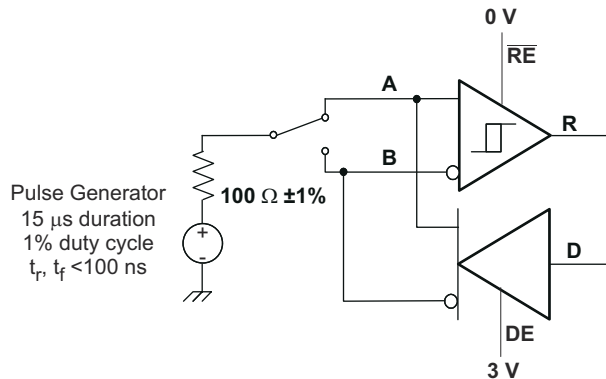


图 7-10. Receiver Enable Test Circuit and Waveforms, Data Output Low



Note: This test is conducted to test survivability only. Data stability at the R output is not specified.

图 7-11. Transient Overvoltage Test Circuit



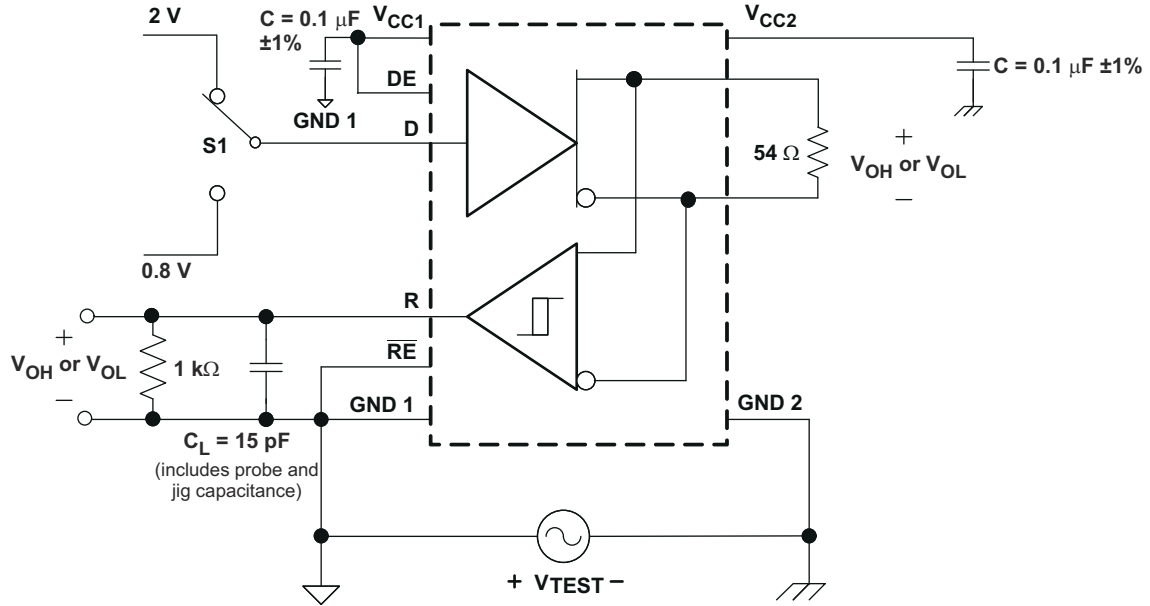


图 7-12. Half-Duplex Common-Mode Transient Immunity Test Circuit

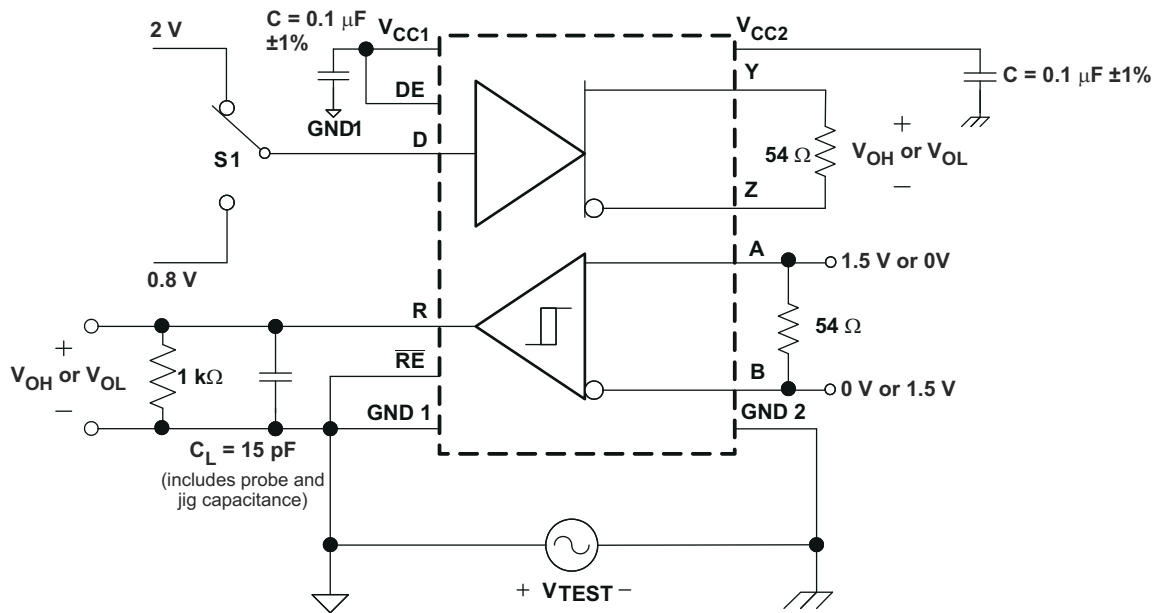


图 7-13. Full-Duplex Common-Mode Transient Immunity Test Circuit

## 8 Detailed Description

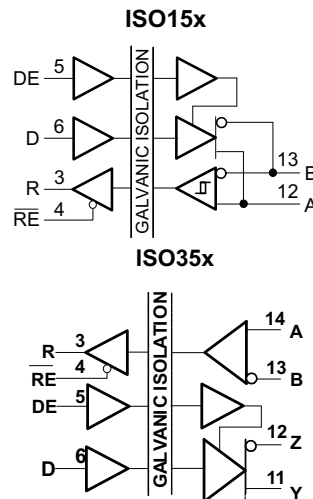
### 8.1 Overview

The ISO15 and ISO15M are isolated half-duplex differential line drivers and receivers while the ISO35 and ISO35M are isolated full-duplex differential line transceivers for TIA/EIA 485/422 applications. They are rated to provide galvanic isolation of up to 2500  $V_{rms}$  for 60 sec as per the standard. They have active-high driver enables and active-low receiver enables to control the data flow.

When the driver enable pin, DE, is logic high, the differential outputs Y and Z follow the logic states at data input D. A logic high at D causes Y to turn high and Z to turn low. In this case the differential output voltage defined as  $V_{OD} = V_{(Y)} - V_{(Z)}$  is positive. When D is low, the output states reverse, Z turns high, Y becomes low, and  $V_{OD}$  is negative. When DE is low, both outputs turn high-impedance. In this condition the logic state at D is irrelevant. The DE pin has an internal pulldown resistor to ground, thus when left open the driver is disabled (high-impedance) by default. The D pin has an internal pullup resistor to  $V_{CC}$ , thus, when left open while the driver is enabled, output Y turns high and Z turns low.

When the receiver enable pin, RE, is logic low, the receiver is enabled. When the differential input voltage defined as  $V_{ID} = V_{(A)} - V_{(B)}$  is positive and higher than the positive input threshold,  $V_{IT+}$ , the receiver output, R, turns high. When  $V_{ID}$  is negative and less than the negative and lower than the negative input threshold,  $V_{IT-}$ , the receiver output, R, turns low. If  $V_{ID}$  is between  $V_{IT+}$  and  $V_{IT-}$  the output is indeterminate. When RE is logic high or left open, the receiver output is high-impedance and the magnitude and polarity of  $V_{ID}$  are irrelevant. Internal biasing of the receiver inputs causes the output to go failsafe-high when the transceiver is disconnected from the bus (open-circuit), the bus lines are shorted (short-circuit), or the bus is not actively driven (idle bus).

### 8.2 Functional Block Diagrams



### 8.3 Device Functional Modes

表 8-1. Driver Function Table<sup>(2)</sup>

$V_{CC1}$	$V_{CC2}$	INPUT (D)	ENABLE INPUT (DE)	OUTPUTS <sup>(1)</sup>	
				Y / A	Z / B
PU	PU	H	H	H	L
PU	PU	L	H	L	H
PU	PU	X	L	Hi-Z	Hi-Z
PU	PU	X	OPEN	Hi-Z	Hi-Z
PU	PU	OPEN	H	H	L
PD	PU	X	X	Hi-Z	Hi-Z
PU	PD	X	X	Hi-Z	Hi-Z

**表 8-1. Driver Function Table<sup>(2)</sup> (continued)**

V <sub>CC1</sub>	V <sub>CC2</sub>	INPUT (D)	ENABLE INPUT (DE)	OUTPUTS <sup>(1)</sup>	
				Y / A	Z / B
PD	PD	X	X	Hi-Z	Hi-Z

(1) Driver output pins are Y and Z for full-duplex devices and A & B for half-duplex devices.

(2) PU = Powered Up; PD = Powered Down; H = Logic High; L = Logic Low; X = Irrelevant, Hi-Z = High Impedance (off)

表 8-2. Receiver Function Table<sup>(1)</sup>

V <sub>CC1</sub>	V <sub>CC2</sub>	DIFFERENTIAL INPUT V <sub>ID</sub> = (V <sub>A</sub> - V <sub>B</sub> )	ENABLE (RE)	OUTPUT (R)
PU	PU	$-0.01\text{ V} \leq V_{ID}$	L	H
PU	PU	$-0.2\text{ V} < V_{ID} < -0.01\text{ V}$	L	?
PU	PU	$V_{ID} \leq -0.2\text{ V}$	L	L
PU	PU	X	H	Hi-Z
PU	PU	X	OPEN	Hi-Z
PU	PU	Open circuit	L	H
PU	PU	Short Circuit	L	H
PU	PU	Idle (terminated) bus	L	H
PD	PU	X	X	Hi-Z
PU	PD	X	L	H

(1) PU = Powered Up; PD = Powered Down; H = Logic High; L = Logic Low; X = Irrelevant, Hi-Z = High Impedance (off), ? = Indeterminate

### 8.3.1 Device I/O Schematics

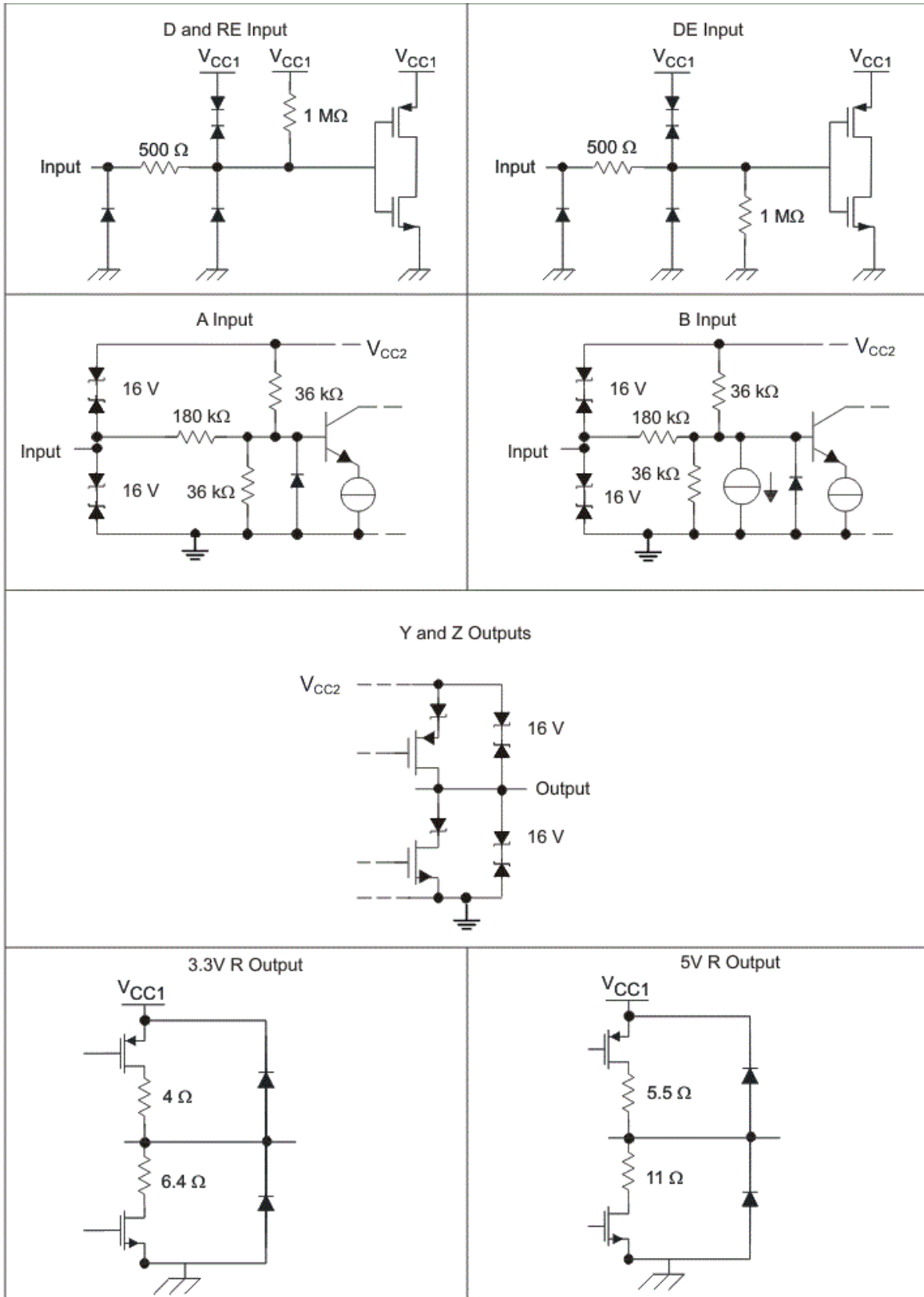


图 8-1. Device I/O Schematics

## 9 Application and Implementation

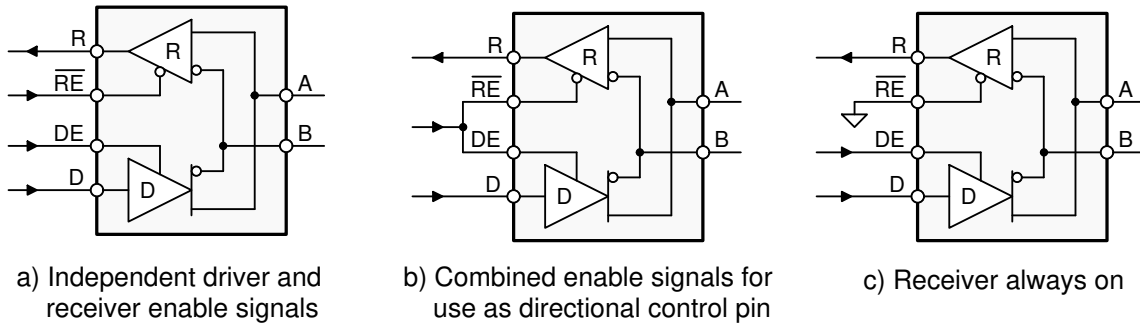
### 备注

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### 9.1 Application Information

The ISO15x and ISO35x family consists of RS-485 transceivers commonly used for asynchronous data transmissions. Full-duplex implementation requires two signal pairs (four wires), and allows each node to transmit data on one pair while simultaneously receiving data on the other pair. For half-duplex transmission there is only one pair which shared for both transmission and reception of data. To eliminate line reflections, each cable end is terminated with a termination resistor,  $R(T)$ , whose value matches the characteristic impedance,  $Z_0$ , of the cable. This method, known as parallel termination, allows for higher data rates over longer cable length.

### 9.2 Typical Application



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图 9-1. Half-Duplex Transceiver Configurations

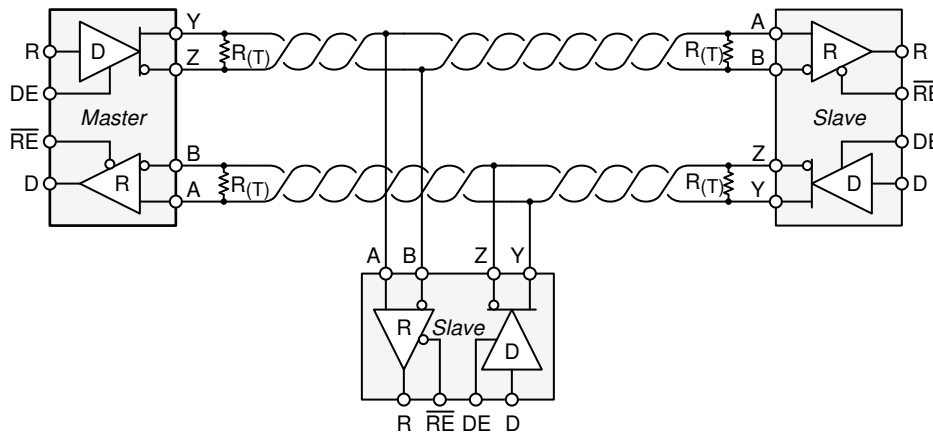


图 9-2. Typical RS-485 Network With Full-Duplex Transceivers

### 9.2.1 Design Requirements

RS-485 is a robust electrical standard suitable for long-distance networking that may be used in a wide range of applications with varying requirements, such as distance, data rate, and number of nodes.

表 9-1. Design Parameters

PARAMETER	VALUE
Pullup and Pulldown Resistors	1 k $\Omega$ to 10 k $\Omega$
Decoupling Capacitors	100 nF

### 9.2.2 Detailed Design Procedure

There is an inverse relationship between data rate and cable length, which means the higher the data rate, the shorter the cable length; and conversely, the lower the data rate, the longer the cable length. When connecting a node to the bus, the distance between the transceiver inputs and the cable trunk, known as the stub, should be as short as possible. Stubs present a nonterminated piece of bus line which can introduce reflections as the length of the stub increases. As a general guideline, the electrical length, or round-trip delay, of a stub should be less than one-tenth of the rise time of the driver. The RS-485 standard specifies that a compliant driver must be able to driver 32 unit loads (UL), where 1 unit load represents a load impedance of approximately 12 k $\Omega$ . Because these devices consists of 1/8 UL transceivers, connecting up to 256 receivers to the bus is possible.

### 9.2.3 Application Curve

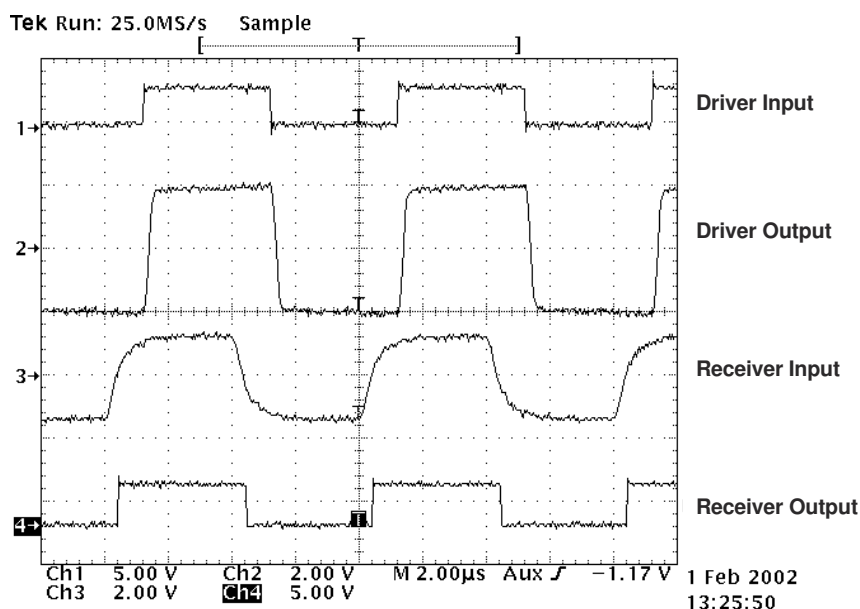


图 9-3. Typical Input and Output Waveforms

## Power Supply Recommendations

To ensure reliable operation at all data rates and supply voltages, a 0.1- $\mu$ F bypass capacitor is recommended at input and output supply pins ( $V_{CC1}$  and  $V_{CC2}$ ). The capacitors should be placed as close to the supply pins as possible. If only a single primary-side power supply is available in an application, isolated power can be generated for the secondary-side with the help of a transformer driver such as Texas Instruments' SN6501. For such applications, detailed power supply design and transformer selection recommendations are available in SN6501 data sheet ([SLLSEA0](#)).

## 10 Layout

### 10.1 Layout Guidelines

On-chip IEC-ESD protection is good for laboratory and portable equipment but never sufficient for EFT and surge transients occurring in industrial environments. Therefore robust and reliable bus node design requires the use of external transient protection devices. Because ESD and EFT transients have a wide frequency bandwidth from approximately 3-MHz to 3-GHz, high-frequency layout techniques must be applied during PCB design. A minimum of four layers is required to accomplish a low EMI PCB design (see [Figure 10-1](#)).

- Layer stacking should be in the following order (top-to-bottom): high-speed signal layer, ground plane, power plane and low-frequency signal layer.
- Place the protection circuitry close to the bus connector to prevent noise transients from penetrating the board.
- Use VCC and ground planes to provide low-inductance. High-frequency currents might follow the path of least inductance and not necessarily the path of least resistance.
- Design the protection components into the direction of the signal path. Do not force the transient currents to divert from the signal path to reach the protection device.
- Apply 100-nF to 220-nF bypass capacitors as close as possible to the VCC-pins of transceiver, UART, controller ICs on the board (see [Figure 10-1](#)).
- Use at least two vias for VCC and ground connections of bypass capacitors and protection devices to minimize effective via-inductance (see [Figure 10-1](#)).
- Use 1-k $\Omega$  to 10-k $\Omega$  pullup and pulldown resistors for enable lines to limit noise currents in these lines during transient events (see [Figure 10-1](#)).
- Insert pulse-proof resistors into the A and B bus lines if the TVS clamping voltage is higher than the specified maximum voltage of the transceiver bus pins. These resistors limit the residual clamping current into the transceiver and prevent it from latching up (see [Figure 10-1](#)).
- While pure TVS protection is sufficient for surge transients up to 1 kV, higher transients require metal-oxide varistors (MOVs) which reduce the transients to a few hundred volts of clamping voltage, and transient blocking units (TBUs) that limit transient current to less than 1 mA.
- Routing the high-speed traces on the top layer avoids the use of vias (and the introduction of their inductances) and allows for clean interconnects between the isolator and the transmitter and receiver circuits of the data link.
- Placing a solid ground plane next to the high-speed signal layer establishes controlled impedance for transmission line interconnects and provides an excellent low-inductance path for the return current flow.
- Placing the power plane next to the ground plane creates additional high-frequency bypass capacitance of approximately 100 pF/in<sup>2</sup>.
- Routing the slower speed control signals on the bottom layer allows for greater flexibility as these signal links usually have margin to tolerate discontinuities such as vias.

If an additional supply voltage plane or signal layer is needed, add a second power and ground plane system to the stack to keep it symmetrical. This makes the stack mechanically stable and prevents it from warping. Also the power and ground plane of each power system can be placed closer together, thus increasing the high-frequency bypass capacitance significantly.



备注

For detailed layout recommendations, see Application Note [SLLA284](#), *Digital Isolator Design Guide*.

### 10.2 Layout Example

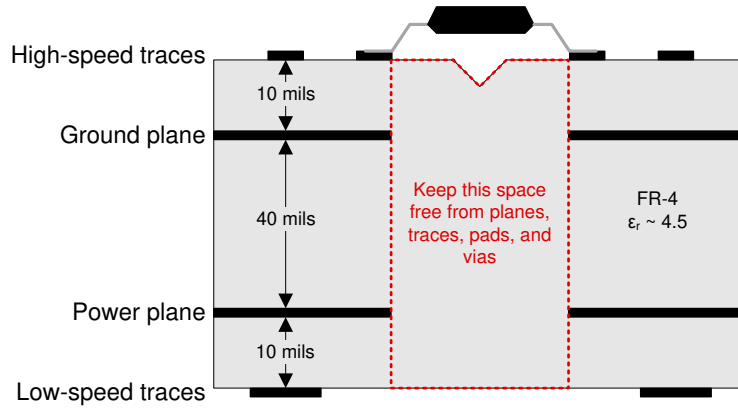


图 10-1. Recommended Layer Stack

## 11 Device and Documentation Support

### 11.1 Documentation Support

#### 11.1.1 Related Documentation

For related documentation, see the following:

- Texas Instruments, [Digital Isolator Design Guide data sheet](#)
- Texas Instruments, [Transformer Driver for Isolated Power Supplies application report](#)
- Texas Instruments, [Isolation Glossary application report](#)

#### 11.2 接收文档更新通知

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#### 11.3 支持资源

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ESD 的损坏小至导致微小的性能降级，大至整个器件故障。精密的集成电路可能更容易受到损坏，这是因为非常细微的参数更改都可能会导致器件与其发布的规格不相符。

#### 11.6 术语表

[TI 术语表](#) 本术语表列出并解释了术语、首字母缩略词和定义。

## 13 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

**PACKAGING INFORMATION**

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead finish/ Ball material (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
ISO15DWR	ACTIVE	SOIC	DW	16	2000	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 85	ISO15	<a href="#">Samples</a>
ISO35DWR	ACTIVE	SOIC	DW	16	2000	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 85	ISO35	<a href="#">Samples</a>

(1) The marketing status values are defined as follows:

**ACTIVE:** Product device recommended for new designs.

**LIFEBUY:** TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

**NRND:** Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

**PREVIEW:** Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

(2) **RoHS:** TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

**RoHS Exempt:** TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

**Green:** TI defines "Green" to mean the content of Chlorine (Cl) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

(6) Lead finish/Ball material - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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**TAPE AND REEL INFORMATION**

**QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE**


\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
ISO15DWR	SOIC	DW	16	2000	330.0	16.4	10.75	10.7	2.7	12.0	16.0	Q1
ISO15MDWR	SOIC	DW	16	2000	330.0	16.4	10.75	10.7	2.7	12.0	16.0	Q1
ISO35DWR	SOIC	DW	16	2000	330.0	16.4	10.75	10.7	2.7	12.0	16.0	Q1
ISO35MDWR	SOIC	DW	16	2000	330.0	16.4	10.75	10.7	2.7	12.0	16.0	Q1

## TAPE AND REEL BOX DIMENSIONS



\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
ISO15DWR	SOIC	DW	16	2000	350.0	350.0	43.0
ISO15MDWR	SOIC	DW	16	2000	350.0	350.0	43.0
ISO35DWR	SOIC	DW	16	2000	350.0	350.0	43.0
ISO35MDWR	SOIC	DW	16	2000	350.0	350.0	43.0

**TUBE**


\*All dimensions are nominal

Device	Package Name	Package Type	Pins	SPQ	L (mm)	W (mm)	T (μm)	B (mm)
ISO15DW	DW	SOIC	16	40	506.98	12.7	4826	6.6
ISO15DWG4	DW	SOIC	16	40	506.98	12.7	4826	6.6
ISO15MDW	DW	SOIC	16	40	506.98	12.7	4826	6.6
ISO35DW	DW	SOIC	16	40	506.98	12.7	4826	6.6
ISO35DWG4	DW	SOIC	16	40	506.98	12.7	4826	6.6
ISO35MDW	DW	SOIC	16	40	506.98	12.7	4826	6.6

## GENERIC PACKAGE VIEW

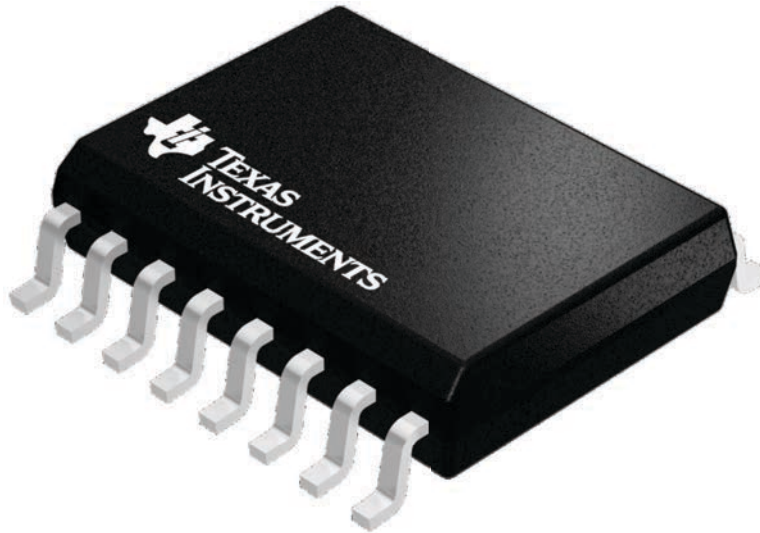
**DW 16**

**SOIC - 2.65 mm max height**

7.5 x 10.3, 1.27 mm pitch

SMALL OUTLINE INTEGRATED CIRCUIT

This image is a representation of the package family, actual package may vary.  
Refer to the product data sheet for package details.



4224780/A





# DW0016B

# PACKAGE OUTLINE

## SOIC - 2.65 mm max height

SOIC



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### NOTES:

1. All linear dimensions are in millimeters. Dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.15 mm, per side.
4. This dimension does not include interlead flash. Interlead flash shall not exceed 0.25 mm, per side.
5. Reference JEDEC registration MS-013.

# EXAMPLE BOARD LAYOUT

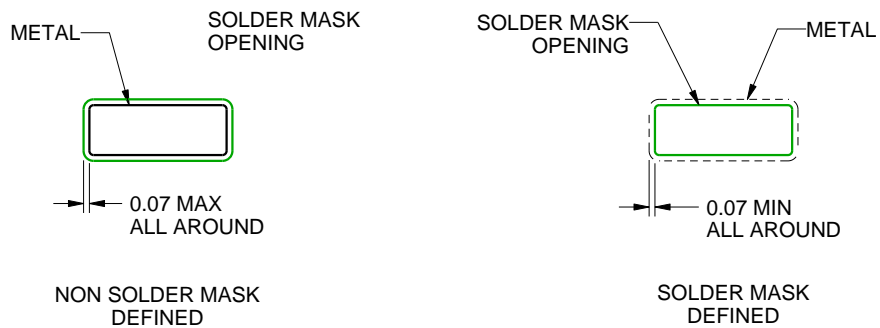
DW0016B

SOIC - 2.65 mm max height

SOIC



LAND PATTERN EXAMPLE  
SCALE:4X



SOLDER MASK DETAILS

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NOTES: (continued)

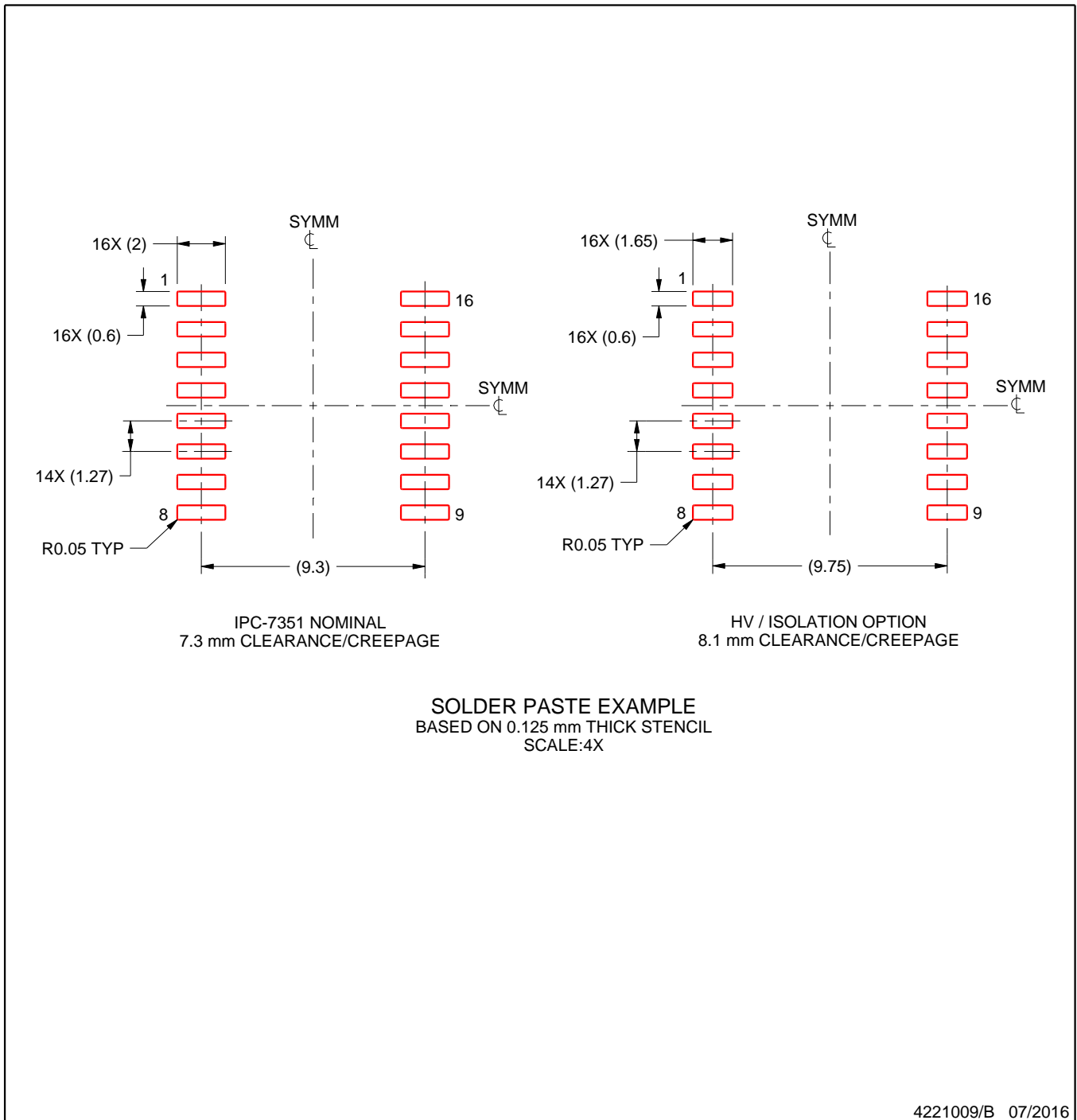
- 6. Publication IPC-7351 may have alternate designs.
- 7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.

# EXAMPLE STENCIL DESIGN

DW0016B

SOIC - 2.65 mm max height

SOIC



NOTES: (continued)

8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
9. Board assembly site may have different recommendations for stencil design.

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