

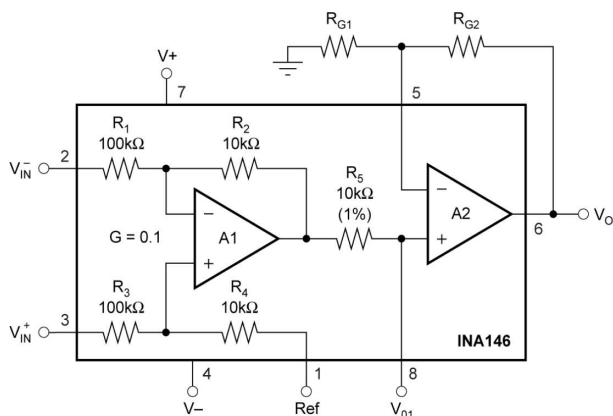
INA146 高电压可编程增益差分放大器

1 特性

- 高共模电压：
 - $V_S = 5V$ 时为 $40V$
 - $V_S = \pm 15V$ 时为 $\pm 100V$
- 差分增益 = $0.1V/V$ 至 $100V/V$ ：
 - 可通过外部电阻器设定
- 低静态电流： $570\mu A$
- 宽电源电压范围：
 - 单电源： $4.5V$ 至 $36V$
 - 双电源： $\pm 2.25V$ 至 $\pm 18V$
- 低增益误差： 0.025%
- 高共模抑制： $80dB$

2 应用

- 电芯化成和测试设备
- 交流驱动器控制模块
- HVAC 控制器
- 专业音频放大器 (机架式)
- 可编程直流电源
- 数据采集 (DAQ)



INA146 简化方框图

3 说明

INA146 是一款精密差分放大器，用于精确衰减高差分电压并抑制高共模电压，以便与常见信号处理电压电平兼容。高电压功能还可提供固有的输入保护。输入共模范围扩展到超出两个电源轨，从而使 INA146 成为单电源和双电源应用的理想选择。

片上精密电阻器经过激光修整，可实现精确增益和高共模抑制。这些电阻器的出色 TCR 跟踪功能可在温度范围内提供持续的高精度。

当输出放大器用作单位增益缓冲器时，10:1 差分放大器可提供 $0.1V/V$ 增益。在此配置中，可以测量高达 $\pm 100V$ 的输入电压。可使用一个外部电阻器对来设定大于 $0.1V/V$ 的增益，而不会影响共模输入范围。

INA146 采用 SO-8 表面贴装封装，可在 $-40^\circ C$ 至 $85^\circ C$ 的额定工业工作温度范围内运行。

封装信息

器件型号	封装 ⁽¹⁾	额定温度范围	封装尺寸 ⁽²⁾
INA146	SOIC (8)	$-40^\circ C$ 至 $85^\circ C$	$4.90mm \times 6.00mm$

- 如需了解所有可用封装，请参阅数据表末尾的可订购产品附录。
- 封装尺寸 (长 × 宽) 为标称值，并包括引脚 (如适用)。



Table of Contents

1 特性	1	6 Application and Implementation	11
2 应用	1	6.1 Application Information.....	11
3 说明	1	7 Device and Documentation Support	17
4 Pin Configuration and Functions	3	7.1 第三方产品免责声明.....	17
5 Specifications	4	7.2 Documentation Support.....	17
5.1 Absolute Maximum Ratings.....	4	7.3 接收文档更新通知.....	17
5.2 ESD Ratings	4	7.4 支持资源.....	17
5.3 Recommended Operating Conditions.....	4	7.5 Trademarks.....	17
5.4 Thermal Information.....	4	7.6 静电放电警告.....	17
5.5 Electrical Characteristics $V_S = \pm 2.25\text{ V to } \pm 18\text{ V}$	5	7.7 术语表.....	17
5.6 Electrical Characteristics $V_S = 5\text{ V Single Supply}$	6	8 Revision History	18
5.7 Amplifier A1, A2 Performance.....	7	9 Mechanical, Packaging, and Orderable Information ..	19
5.8 Typical Performance Curves.....	8		

4 Pin Configuration and Functions

Top View

SO-8

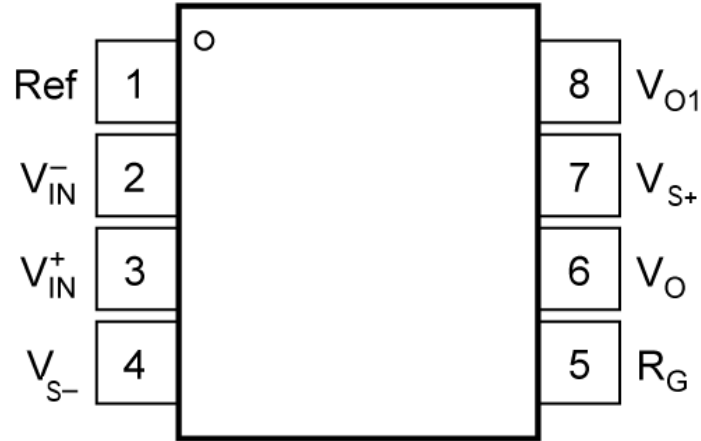


图 4-1. INA146 D Package, 8-Pin SOIC (Top View)

表 4-1. Pin Functions

PIN		TYPE ⁽¹⁾	DESCRIPTION
NAME	NO.		
Ref	1	I	Reference input. This pin must be driven by a low impedance source.
V_{IN-}	2	I	Negative (inverting) input
V_{IN+}	3	I	Positive (non-inverting) input
V_{S-}	4	-	Negative supply
R_G	5	I	Gain setting input. Place a resistor network between pin 1 and pin 5.
V_O	6	O	Output of amplifier A2
V_{S+}	7	-	Positive supply
V_{O1}	8	O	Output of amplifier A1

(1) Signal Types: I = Input, O = Output, I/O = Input or Output.

5 Specifications

5.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted)⁽¹⁾

			MIN	MAX	UNIT
V _S	Supply voltage	Dual supply, V _S = (V _{S+}) - (V _{S-})		±18	V
		Single supply, V _S = (V _{S+}) - 0 V		36	
V _{IN+} , V _{IN-}	Signal input voltage			±100	V
	Signal input current			±1	mA
	Output short-circuit ⁽²⁾		Continuous		
T _A	Operating temperature		- 55	125	°C
T _{stg}	Storage temperature		- 55	125	°C
T _J	Junction temperature			150	°C
	Lead temperature (soldering, 10 s)			240	°C

(1) Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under *Recommended Operating Conditions*. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

(2) Short-circuit to V_S / 2.

5.2 ESD Ratings

			VALUE	UNIT
V _(ESD)	Electrostatic discharge	Human-body model (HBM), per ANSI/ESDA/JEDEC JS-001 ⁽¹⁾	±1000	V
		Charged-device model (CDM), per JEDEC specification JESD22-C101 ⁽²⁾	±1000	

(1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.

(2) JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.

5.3 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)

			MIN	TYP	MAX	UNIT
V _S	Supply voltage	Single-supply	4.5	30	36	V
		Dual-supply	±2.25	±15	±18	
T _A	Specified temperature		- 40		85	°C

5.4 Thermal Information

THERMAL METRIC ⁽¹⁾		INA146	UNIT
		SO-8	
		8 PINS	
R _{θJA}	Junction-to-ambient thermal resistance	110.3	°C/W
R _{θJC(top)}	Junction-to-case (top) thermal resistance	43.6	°C/W
R _{θJB}	Junction-to-board thermal resistance	55.5	°C/W
ψ _{JT}	Junction-to-top characterization parameter	6.2	°C/W
ψ _{JB}	Junction-to-board characterization parameter	54.3	°C/W

(1) For more information about traditional and new thermal metrics, see the [Semiconductor and IC Package Thermal Metrics](#) application report.

5.5 Electrical Characteristics $V_S = \pm 2.25\text{ V to } \pm 18\text{ V}$

 at $T_A = 25^\circ\text{C}$, $V_S = \pm 2.25\text{ V to } \pm 18\text{ V}$, $R_L = 10\text{ k}\Omega$, $V_{REF} = 0\text{ V}$, $V_{CM} = V_S / 2$, and $G = 0.1$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS	TEST CONDITIONS	MIN	TYP	MAX	UNIT
INPUT							
V_{OS}	Offset voltage, V_O	RTI, $V_S = \pm 15\text{ V}$, $V_{CM} = 0\text{ V}$			± 1	± 5	mV
	Offset voltage, V_{O1}	RTI			± 1		mV
	Offset voltage drift	RTI, $T_A = -40^\circ\text{C to } 85^\circ\text{C}$			± 10		$\mu\text{V}/^\circ\text{C}$
PSRR	Power-supply rejection ratio	RTI, $V_S = \pm 1.35\text{ V to } \pm 18\text{ V}$			± 100	± 600	$\mu\text{V}/\text{V}$
V_{CM}	Common-mode voltage ⁽¹⁾	$V_S = \pm 15\text{ V}$, $V_{IN} = 0\text{ V}$		- 100		100	V
CMRR	Common-mode voltage rejection	RTI, $11(-V_S) < V_{CM} < 11(V_S - 1)$, $R_S = 0\ \Omega$		70	80		dB
		$T_A = -40^\circ\text{C to } 85^\circ\text{C}$		64	74		
	Differential input impedance	Non-inverting input			110		k Ω
		Inverting input			91.7		
	Common-mode input impedance				55		k Ω
BIAS CURRENT							
I_B	Bias Current	$V_{CM} = V_S / 2$			± 50		nA
I_{OS}	Offset Current				± 5		nA
NOISE							
e_N	Voltage noise	RTI, $f_B = 0.1\text{ Hz to } 10\text{ Hz}$			12		μV_{PP}
		RTI, $f = 1\text{ kHz}$			550		$\text{nV}/\sqrt{\text{Hz}}$
GAIN							
	Gain			0.1		100	V/V
GE	Gain error	$V_O = (V_-) + 0.15\text{ V to } (V_+) - 1\text{ V}$, $R_L = 100\text{ k}\Omega$, $G = 1$			± 0.025	± 0.1	%
		$V_O = (V_-) + 0.3\text{ V to } (V_+) - 1.25\text{ V}$, $R_L = 10\text{ k}\Omega$, $G = 1$			± 0.025	± 0.1	
	Gain error drift ⁽²⁾	$T_A = -40^\circ\text{C to } 85^\circ\text{C}$	$V_O = (V_-) + 0.25\text{ V to } (V_+) - 1\text{ V}$, $R_L = 100\text{ k}\Omega$, $G = 1$		± 1	± 10	ppm/ $^\circ\text{C}$
			$V_O = (V_-) + 0.5\text{ V to } (V_+) - 1.25\text{ V}$, $R_L = 10\text{ k}\Omega$, $G = 1$		± 1	± 10	
	Gain nonlinearity	$V_O = (V_-) + 0.3\text{ V to } (V_+) - 1.25\text{ V}$, $G = 1$			± 0.001	± 0.01	% of FSR
OUTPUT							
	Output voltage	$R_L = 100\text{ k}\Omega$, $G = 1$		$(V_-) + 0.15$		$(V_+) - 1$	V
			$T_A = -40^\circ\text{C to } 85^\circ\text{C}$	$(V_-) + 0.25$		$(V_+) - 1$	
		$R_L = 10\text{ k}\Omega$, $G = 1$		$(V_-) + 0.3$		$(V_+) - 1.25$	
			$T_A = -40^\circ\text{C to } 85^\circ\text{C}$	$(V_-) + 0.5$		$(V_+) - 1.25$	
C_L	Load capacitance	Stable operation			1		nF
I_{SC}	Short-circuit current	Continuous to $V_S / 2$			± 15		mA
FREQUENCY RESPONSE							
BW	Bandwidth, - 3 dB	$G = 0.1$			550		kHz
		$G = 1$			50		
SR	Slew rate				0.3		V/ μs
t_s	Settling time	To 0.1%,	$V_O = 10\text{ V-step}$		40		μs
		To 0.01%	$V_O = 10\text{ V-step}$		80		
	Overload recovery	50% input overload			40		μs
POWER SUPPLY							

at $T_A = 25^\circ\text{C}$, $V_S = \pm 2.25\text{ V}$ to $\pm 18\text{ V}$, $R_L = 10\text{ k}\Omega$, $V_{\text{REF}} = 0\text{ V}$, $V_{\text{CM}} = V_S / 2$, and $G = 0.1$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS	TEST CONDITIONS	MIN	TYP	MAX	UNIT
I_Q	Quiescent current	$V_{\text{IN}} = 0\text{ V}$		± 570	± 700	μA
		$T_A = -40^\circ\text{C}$ to 85°C			± 750	

- (1) Input common-mode voltage varies with output voltage; see *Typical Characteristics*.
(2) Specified by wafer test.

5.6 Electrical Characteristics $V_S = 5\text{ V}$ Single Supply

at $T_A = 25^\circ\text{C}$, $V_S = 5\text{ V}$, $R_L = 10\text{ k}\Omega$, $V_{\text{REF}} = 0\text{ V}$, $V_{\text{CM}} = V_S / 2$, and $G = 0.1$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS	TEST CONDITIONS	MIN	TYP	MAX	UNIT
INPUT						
V_{OS}	Offset voltage, V_O	RTI, $V_{\text{CM}} = 0\text{ V}$		± 3	± 10	mV
	Offset voltage, V_{O1}	RTI, $V_{\text{CM}} = 0\text{ V}$		± 1		mV
	Offset voltage drift	RTI, $T_A = -40^\circ\text{C}$ to 85°C		± 10		$\mu\text{V}/^\circ\text{C}$
PSRR	Power-supply rejection ratio	RTI, $V_S = \pm 1.35\text{ V}$ to $\pm 18\text{ V}$		± 100	± 600	$\mu\text{V}/\text{V}$
V_{CM}	Common-mode voltage ⁽¹⁾	$V_{\text{IN}} = 0\text{ V}$	- 25		19	V
CMRR	Common-mode voltage rejection	$V_{\text{CM}} = -25\text{ V}$ to 19 V , $R_S = 0\ \Omega$	70	80		dB
		$T_A = -40^\circ\text{C}$ to 85°C	64	74		
	Differential input impedance	Non-inverting input		110		$\text{k}\Omega$
		Inverting input		91.7		
	Common-mode input impedance			55		$\text{k}\Omega$
BIAS CURRENT						
I_B	Bias Current	$V_{\text{CM}} = V_S / 2$		± 50		nA
I_{OS}	Offset Current	$V_{\text{CM}} = V_S / 2$		± 5		nA
NOISE						
e_N	Voltage noise	RTO, $f_B = 0.1\text{ Hz}$ to 10 Hz		12		μV_{PP}
		RTO, $f = 1\text{ kHz}$		550		$\text{nV}/\sqrt{\text{Hz}}$
GAIN						
	Gain		0.1		100	V/V
GE	Gain error	$V_O = 0.15\text{ V}$ to 4 V , $R_L = 100\text{ k}\Omega$		± 0.025	± 0.1	%
		$V_O = 0.3\text{ V}$ to 3.75 V , $R_L = 10\text{ k}\Omega$		± 0.025	± 0.1	
	Gain error drift ⁽²⁾	$T_A = -40^\circ\text{C}$ to 85°C	$V_O = 0.25\text{ V}$ to 4 V , $R_L = 100\text{ k}\Omega$	± 1	± 10	ppm/ $^\circ\text{C}$
			$V_O = 0.5\text{ V}$ to 3.75 V , $R_L = 10\text{ k}\Omega$	± 1	± 10	
	Gain nonlinearity	$V_O = 0.3\text{ V}$ to 3.75 V		± 0.001	± 0.01	% of FSR
OUTPUT						
	Output voltage	$R_L = 100\text{ k}\Omega$		0.15	4	V
			$T_A = -40^\circ\text{C}$ to 85°C	0.25	4	
		$R_L = 10\text{ k}\Omega$		0.3	3.75	
			$T_A = -40^\circ\text{C}$ to 85°C	0.5	3.75	
C_L	Load capacitance	Stable operation		1		nF
I_{SC}	Short-circuit current	Continuous to $V_S / 2$		± 15		mA
FREQUENCY RESPONSE						
BW	Bandwidth, - 3 dB	$G = 0.1$		550		kHz
		$G = 1$		50		
SR	Slew rate			0.3		V/ μs
t_s	Settling time	To 0.1%,	$V_O = 10\text{ V-step}$	40		μs
		To 0.01%	$V_O = 10\text{ V-step}$	80		

at $T_A = 25^\circ\text{C}$, $V_S = 5\text{ V}$, $R_L = 10\text{ k}\Omega$, $V_{\text{REF}} = 0\text{ V}$, $V_{\text{CM}} = V_S / 2$, and $G = 0.1$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Overload recovery	50% input overload			40		μs
POWER SUPPLY						
I_Q	Quiescent current	$V_{\text{IN}} = 0\text{ V}$		± 570	± 700	μA
		$T_A = -40^\circ\text{C}$ to 85°C			± 750	

- (1) Input common-mode voltage varies with output voltage; see *Typical Characteristics*.
(2) Specified by wafer test.

5.7 Amplifier A1, A2 Performance

at $T_A = 25^\circ\text{C}$, $R_L = 10\text{ k}\Omega$, $V_{\text{REF}} = 0\text{ V}$, $V_{\text{CM}} = V_S / 2$, and $G = 0.1$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS	TEST CONDITIONS	MIN	TYP	MAX	UNIT
INPUT						
V_{OS}	Offset voltage, V_O	RTI, $V_S = \pm 15\text{ V}$, $V_{\text{CM}} = V_O = 0\text{ V}$		± 0.5		mV
	Offset voltage drift	RTI, $T_A = -40^\circ\text{C}$ to 85°C		± 1		$\mu\text{V}/^\circ\text{C}$
V_{CM}	Common-mode voltage (1)	$V_{\text{IN}} = V_O = 0\text{ V}$		V_{S-} to $(V_{S+}) - 1$		V
CMRR	Common-mode voltage rejection	$V_{\text{CM}} = V_{S-}$ to $(V_{S+}) - 1$		90		dB
GAIN						
A_{OL}	Open Loop Gain			110		dB
BIAS CURRENT						
I_B	Bias Current			± 50		nA
I_{OS}	Offset Current			± 5		nA
OUTPUT						
	Resistor at V_{O1}	Initial		10		$\text{k}\Omega$
	Error at V_{O1}			± 1		%
	Error drift at V_{O1}			± 100		$\text{ppm}/^\circ\text{C}$

- (1) Input common-mode voltage varies with output voltage; see *Typical Characteristics*.

5.8 Typical Performance Curves

At $T_A = 25^\circ\text{C}$, $V_S = \pm 15\text{ V}$, $G = 0.1$, $R_L = 10\text{ k}\Omega$ connected to ground and Ref pin connected to ground, unless otherwise noted.

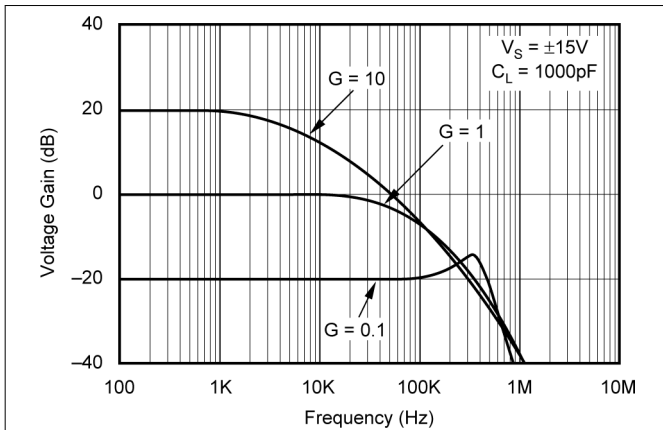


图 5-1. Gain vs Frequency

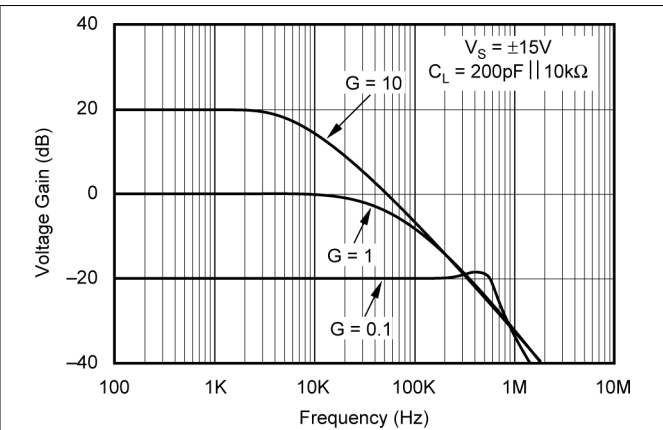


图 5-2. Gain vs Frequency

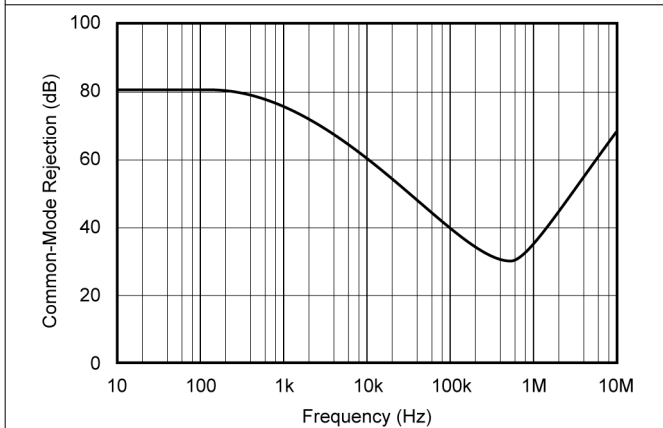


图 5-3. Common-mode Rejection vs Frequency

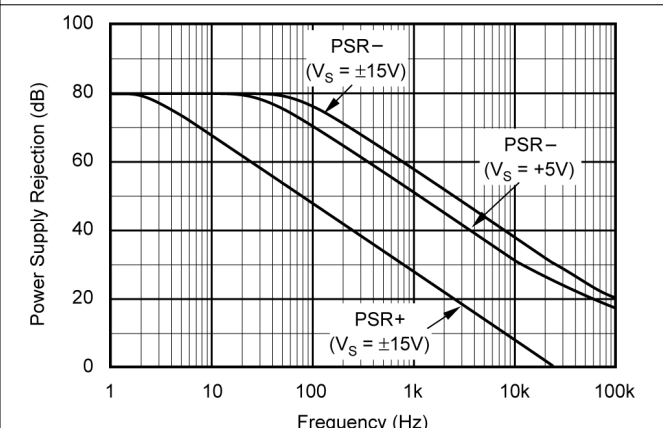


图 5-4. Power Supply Rejection vs Frequency

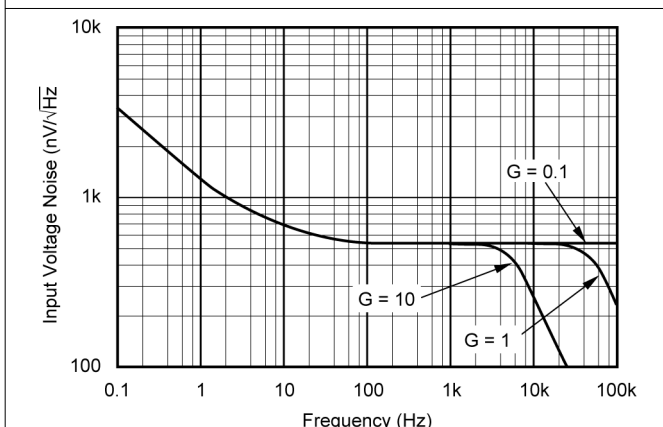


图 5-5. Input Voltage Noise Density

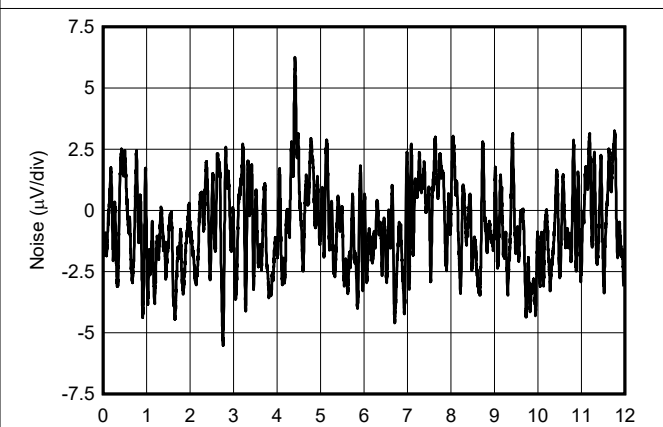


图 5-6. 0.1-Hz to 10-Hz Voltage Noise (R_{ti})

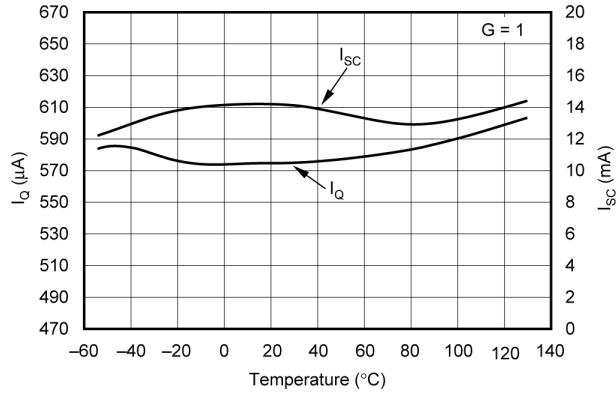


图 5-7. Quiescent Current and Short-circuit Current vs Temperature

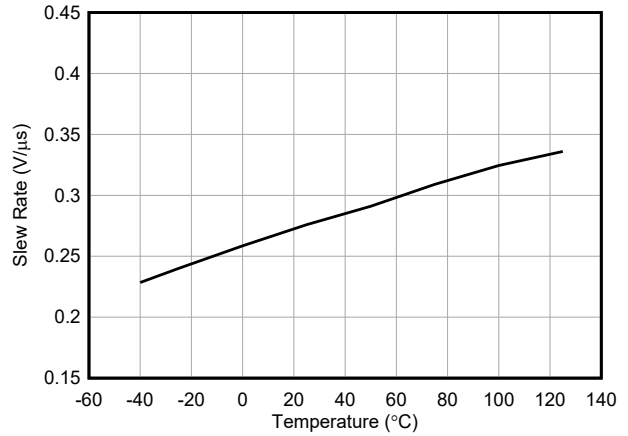


图 5-8. Slew Rate vs Temperature

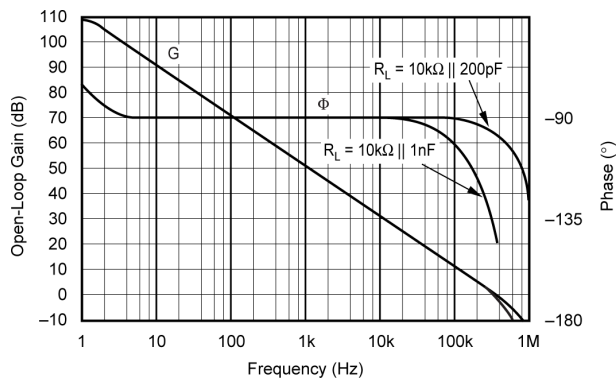


图 5-9. Gain and Phase vs Frequency Op Amp A1 and A2

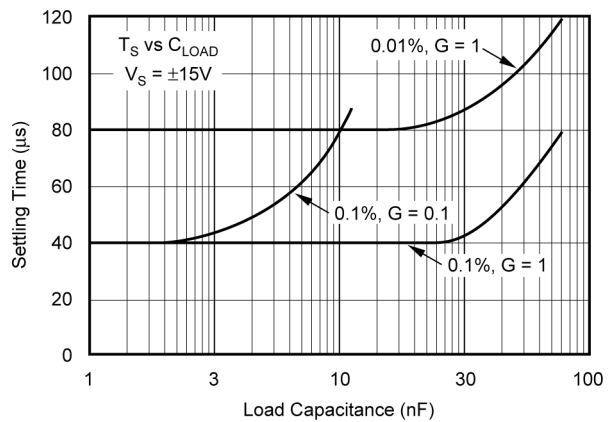


图 5-10. Settling Time vs Load Capacitance

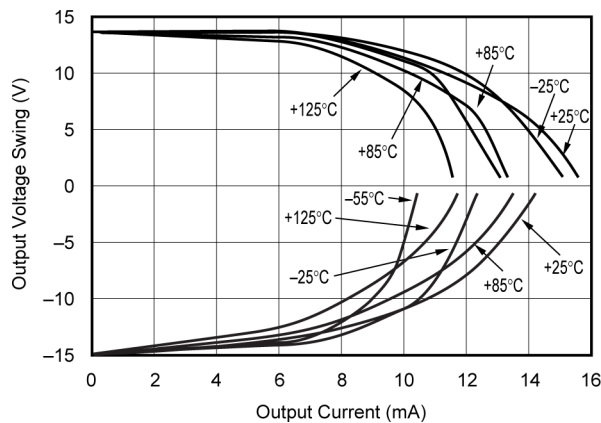


图 5-11. Maximum Output Voltage Swing vs Output Current

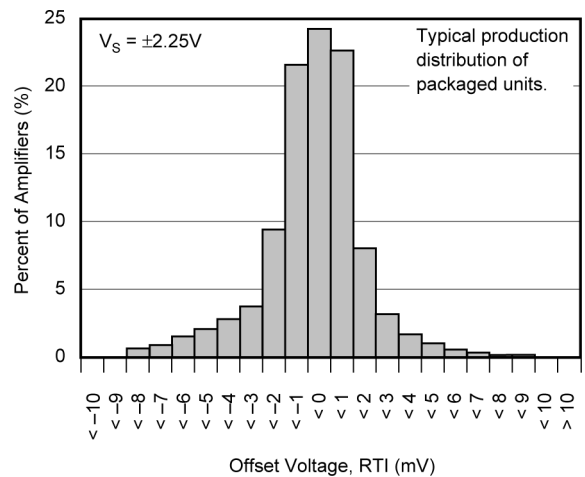


图 5-12. Offset Voltage Production Distribution

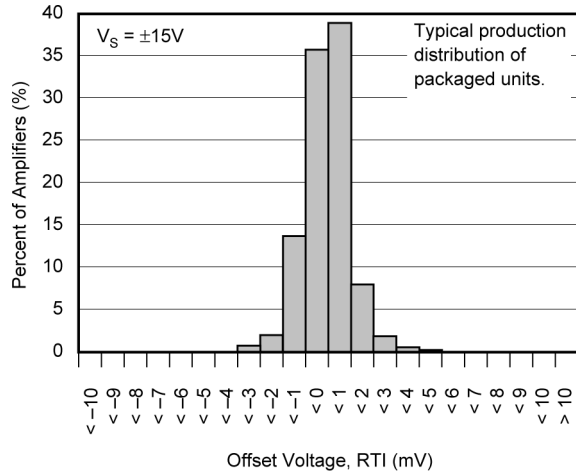


图 5-13. Offset Voltage Production Distribution

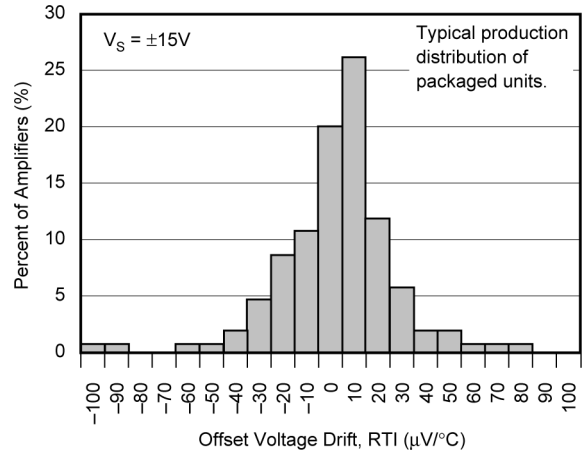


图 5-14. Offset Voltage Drift Production Distribution

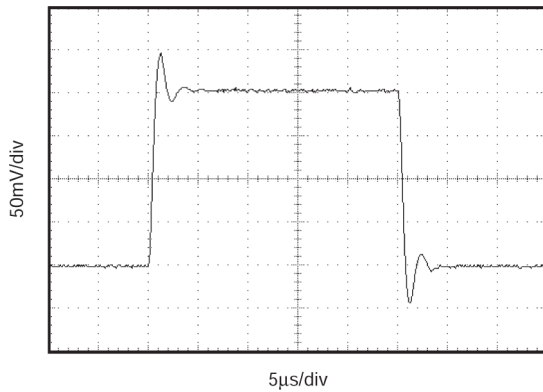


图 5-15. Small-signal Step Response ($G = 0.1$, $R_L = 10\text{ k}\Omega$, $C_L = 200\text{ pF}$)

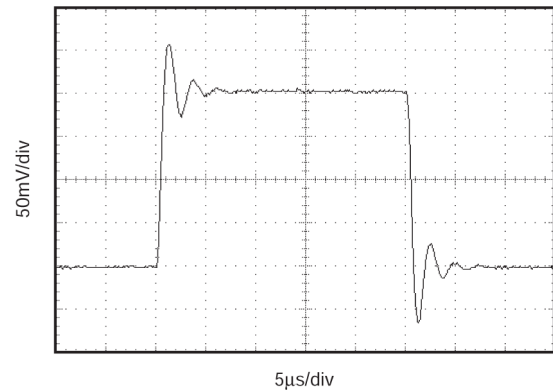


图 5-16. Small-signal Step Response ($G = 0.1$, $C_L = 1000\text{ pF}$)

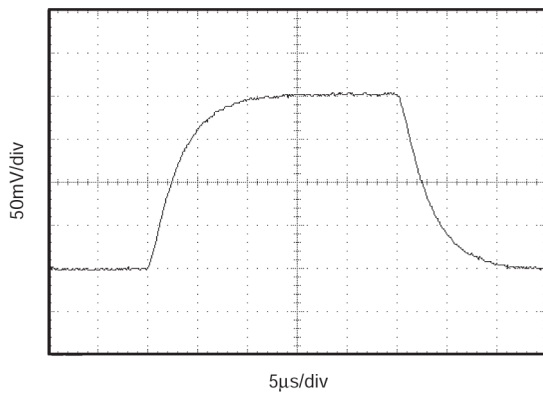


图 5-17. Small-signal Step Response ($G = 1$, $C_L = 1000\text{ pF}$)

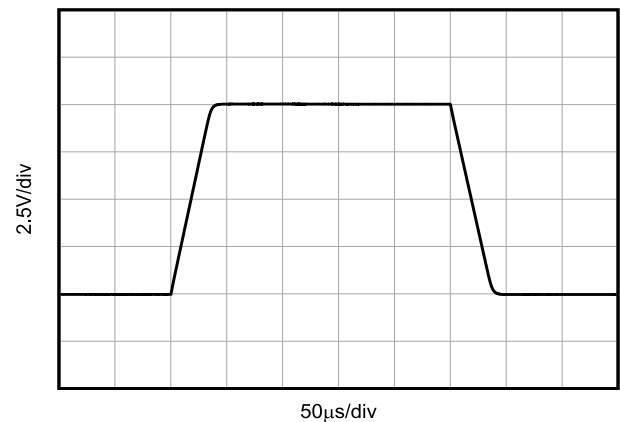


图 5-18. Large-signal Step Response ($G = 1$, $R_L = 10\text{ k}\Omega$, $C_L = 200\text{ pF}$)

6 Application and Implementation

备注

以下应用部分中的信息不属于 TI 器件规格的范围，TI 不担保其准确性和完整性。TI 的客户应负责确定器件是否适用于其应用。客户应验证并测试其设计，以确保系统功能。

6.1 Application Information

The INA146 is a programmable gain difference amplifier consisting of a gain of 0.1 difference amplifier and a programmable-gain output buffer stage. Basic circuit connections are shown in [图 6-1](#). Power supply bypass capacitors must be connected close to pins 4 and 7, as shown. The amplifier is programmable in the range of $G = 0.1$ to $G = 50$ with two external resistors.

The output of A1 is connected to the noninverting input of A2 through a 10-k Ω resistor which is trimmed to $\pm 1\%$ absolute accuracy. The A2 input is available for applications such as a filter or a precision current source. See application figures for examples.

6.1.1 Operating Voltage

The INA146 is fully specified for supply voltages from ± 2.25 V to ± 18 V with key parameters specified over the temperature range -40°C to 85°C . The INA146 can be operated with single or dual supplies with excellent performance. Parameters that vary significantly with operating voltage, load conditions or temperature are shown in the typical performance curves.

6.1.2 Setting the Gain

The gain of the INA146 is set by using two external resistors, R_{G1} and R_{G2} , according to the equation:

$$G = 0.1 \times (1 + R_{G2}/R_{G1})$$

For a total gain of 0.1, A2 is connected as a buffer amplifier with no R_{G1} . A feedback resistor, $R_{G2} = 10$ k Ω , must be used in the buffer connection. This provides bias current cancellation (in combination with internal R_5) to assure specified offset voltage performance. Commonly used values are shown in the table of [图 6-1](#). Resistor values for other gains must be chosen to provide a 10-k Ω parallel resistance.

6.1.3 Common-mode Range

The 10:1 input resistor ratio of the INA146 provides an input common-mode range that can extend well beyond the power supply rails. Exact range depends on the power supply voltage and the voltage applied to the Ref terminal (pin 1). For proper operation, the voltage at the non-inverting input of A1 (an internal node) must be within the linear operating range. The voltage is determined by the simple 10:1 voltage divider between pin 3 and pin 1. This voltage must be between V^- and $(V^+) - 1$ V.

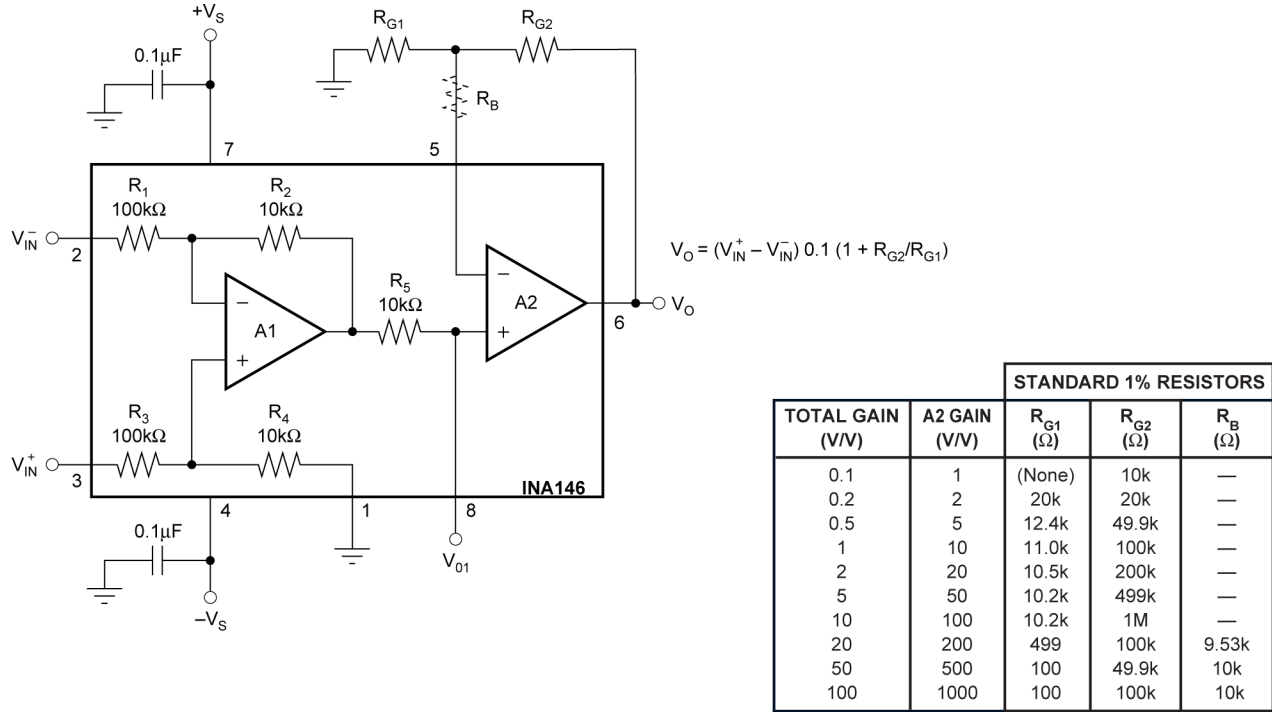


图 6-1. Basic Circuit Connections

6.1.4 Offset Trim

The INA146 is laser-trimmed for low offset voltage and drift. Most applications require no external offset adjustment. 图 6-2 shows an optional circuit for trimming the offset voltage. A voltage applied to the Ref terminal is summed with the output signal. This feature can be used to null offset voltage. To maintain good common-mode rejection, the source impedance of a signal applied to the Ref terminal must be less than 10 Ω and a resistor added to the positive input terminal must be 10 times that, or 100 Ω. Alternatively, the trim voltage can be buffered with an operational amplifier such as the OPA277.

6.1.5 Input Impedance

The input impedance of the INA146 is determined by the input resistor network and is approximately 100 kΩ. The source impedance at the two input terminals must be nearly equal to maintain good common-mode rejection. A 12-Ω mismatch in impedance between the two inputs causes the typical common-mode rejection to be degraded to approximately 72 dB. 图 6-7 shows a common application measuring power supply current through a shunt resistor. The source impedance of the shunt resistor, RS, is balanced by an equal compensation resistor, RC.

Source impedances greater than 800 Ω are not recommended, even if the source impedances are perfectly matched. Internal resistors are laser trimmed for accurate ratios, not to absolute values. Adding equal resistors greater than 800 Ω can cause a mismatch in the total resistor ratios, degrading CMR.

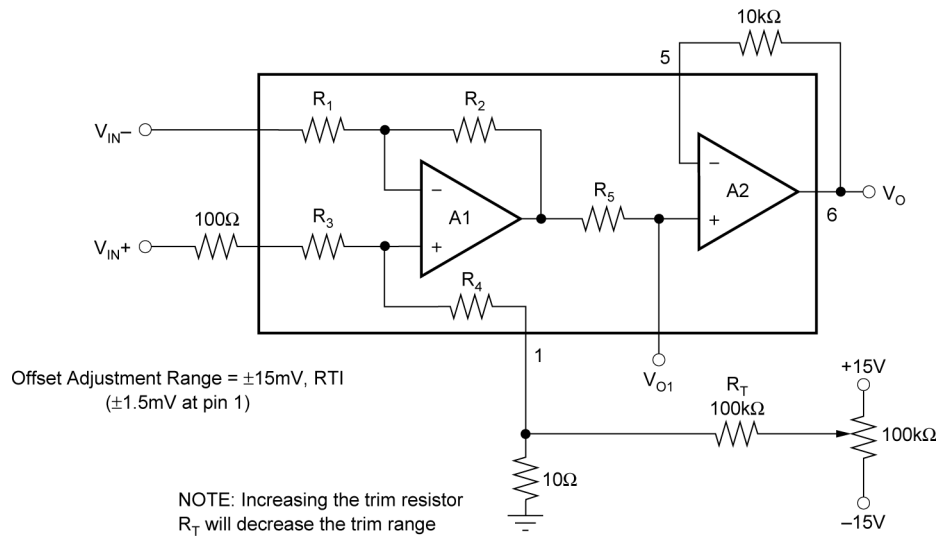


图 6-2. Optional Offset Trim Circuit

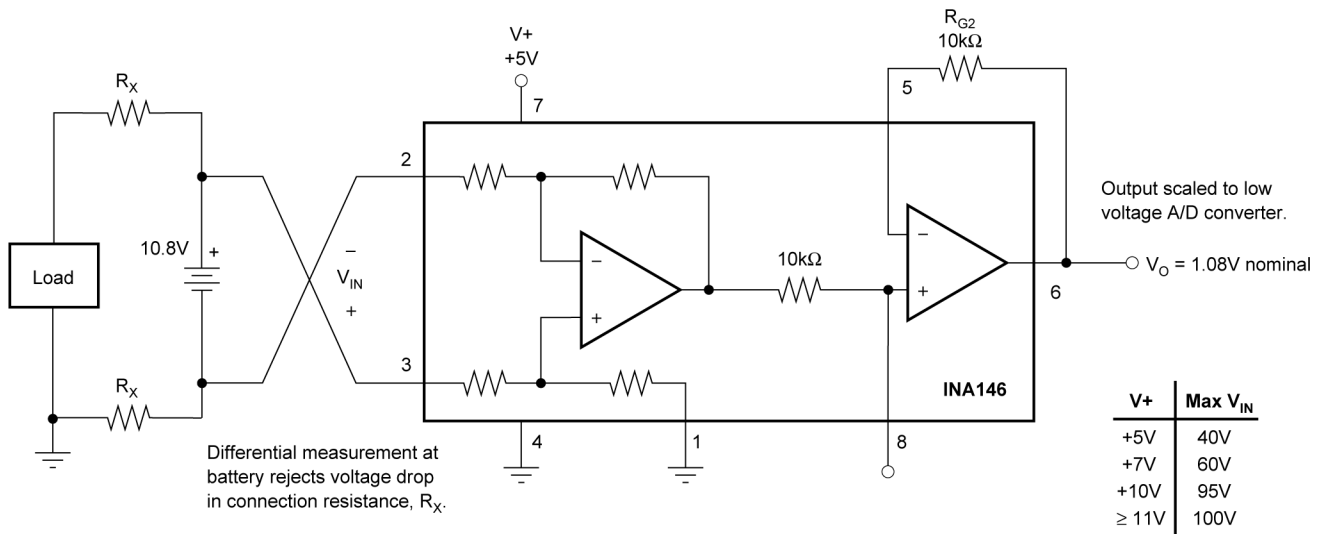


图 6-3. Measuring Voltages Greater Than Supply Voltage

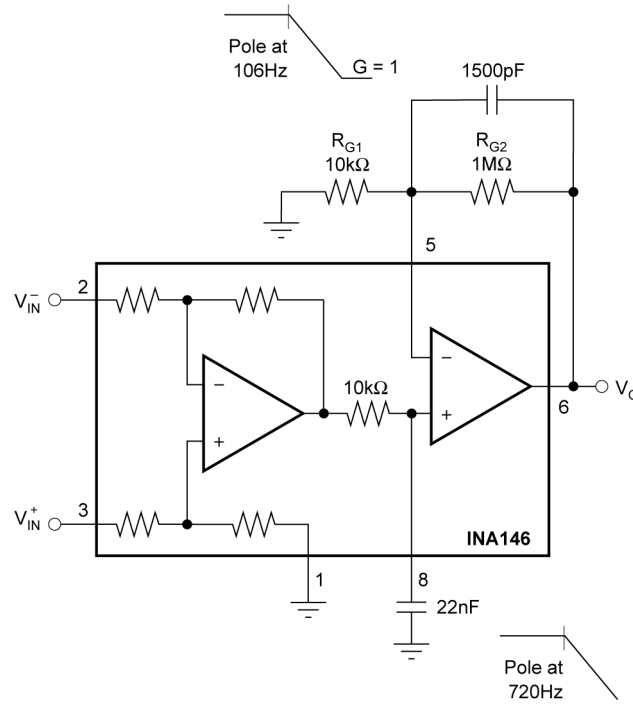


图 6-4. Noise Filtering

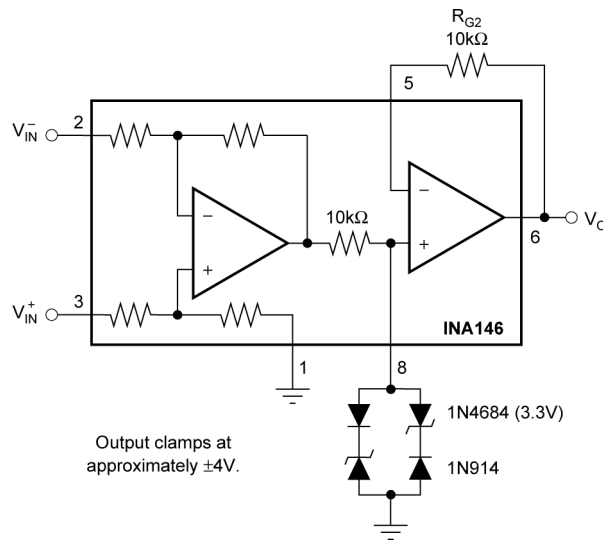
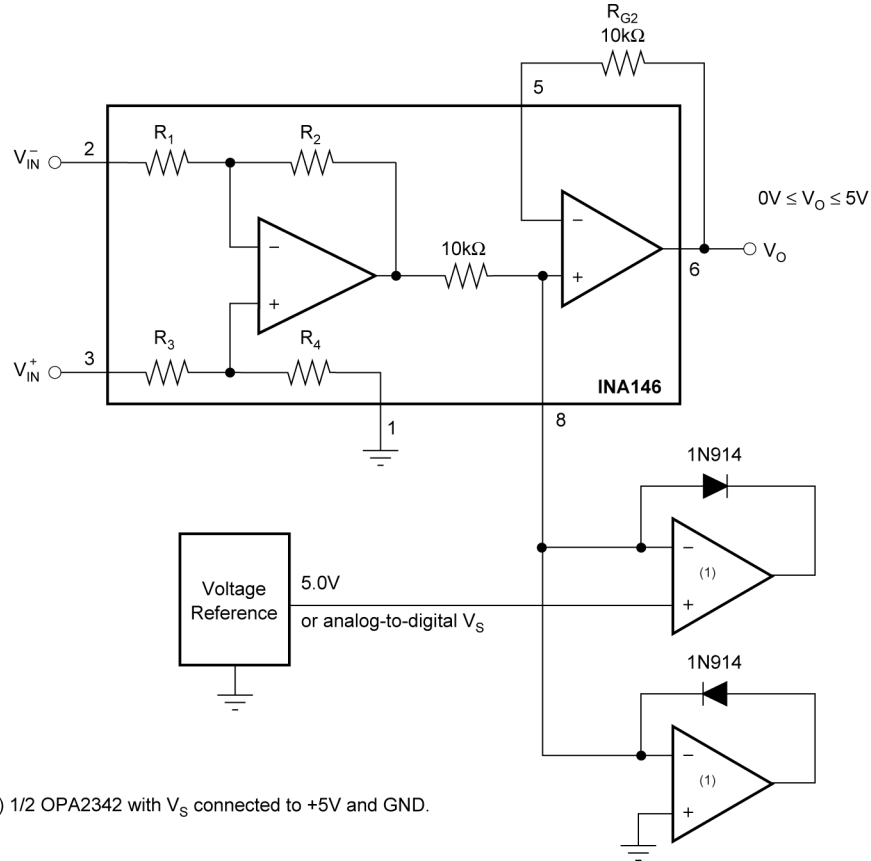
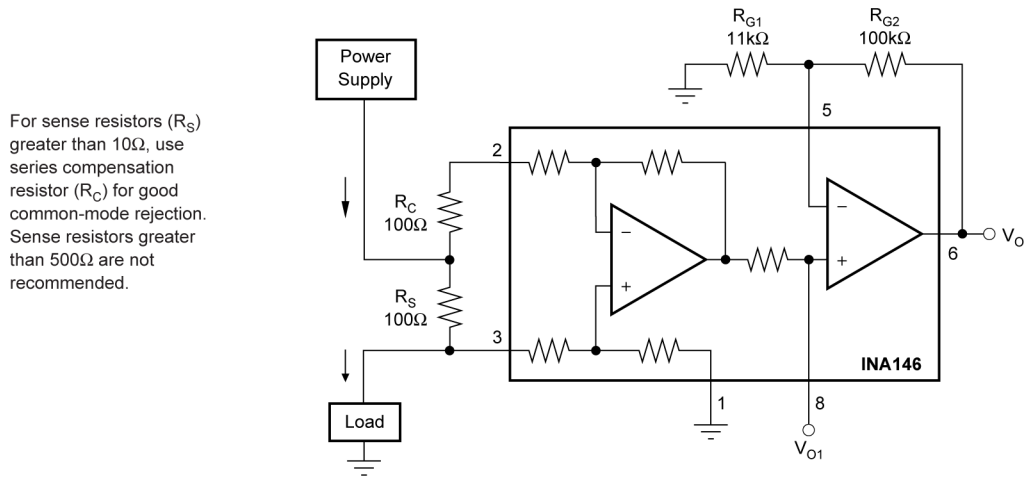


图 6-5. Output Clamp



NOTE: (1) 1/2 OPA2342 with V_S connected to +5V and GND.

图 6-6. Precision Clamp



For sense resistors (R_S) greater than 10Ω , use series compensation resistor (R_C) for good common-mode rejection. Sense resistors greater than 500Ω are not recommended.

图 6-7. Current Monitor, $G = 1$

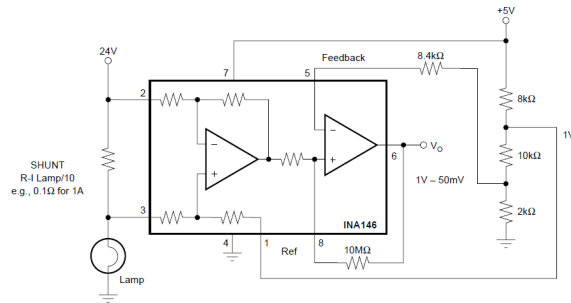


图 6-8. Comparator Output With Optional Hysteresis Application to Sense Lamp Burn-Out

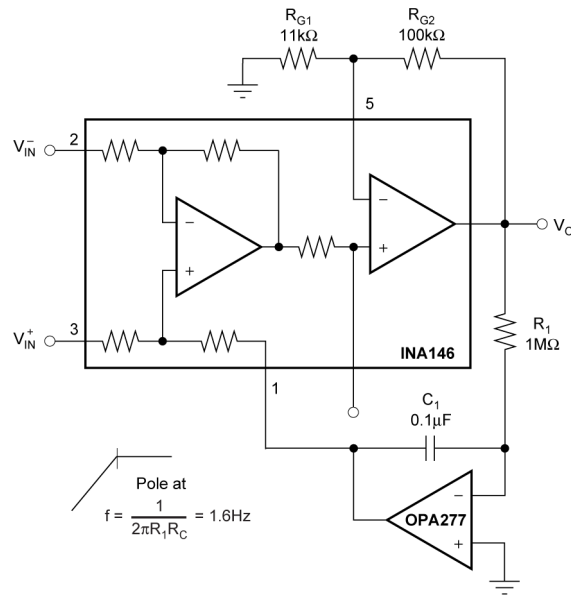


图 6-9. AC Coupling (DC Restoration)

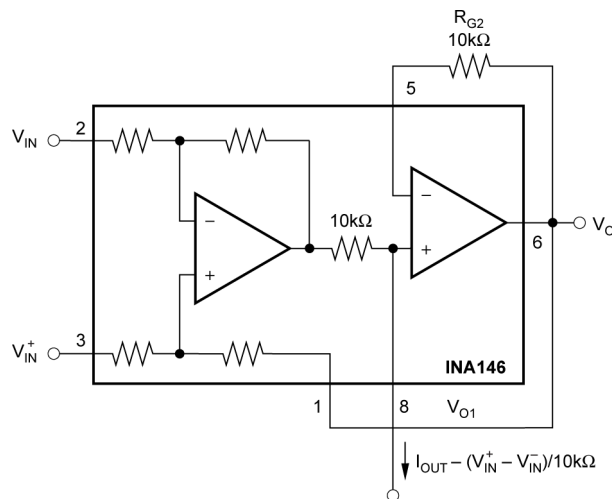


图 6-10. Precision Current Source

7 Device and Documentation Support

TI offers an extensive line of development tools. Tools and software to evaluate the performance of the device, generate code, and develop solutions are listed below.

7.1 第三方产品免责声明

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7.2 Documentation Support

7.2.1 Related Documentation

For related documentation see the following:

- Texas Instruments, [Comprehensive Error Calculation for Instrumentation Amplifiers](#), application note
- Texas Instruments, [High-Voltage Signal Conditioning for Low Voltage ADCs](#), application note
- Texas Instruments, [Analog Engineer's Calculator](#), application

7.3 接收文档更新通知

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7.4 支持资源

[TI E2E™ 中文支持论坛](#) 是工程师的重要参考资料，可直接从专家处获得快速、经过验证的解答和设计帮助。搜索现有解答或提出自己的问题，获得所需的快速设计帮助。

链接的内容由各个贡献者“按原样”提供。这些内容并不构成 TI 技术规范，并且不一定反映 TI 的观点；请参阅 TI 的 [使用条款](#)。

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ESD 的损坏小至导致微小的性能降级，大至整个器件故障。精密的集成电路可能更容易受到损坏，这是因为非常细微的参数更改都可能会导致器件与其发布的规格不相符。

7.7 术语表

[TI 术语表](#) 本术语表列出并解释了术语、首字母缩略词和定义。

8 Revision History

注：以前版本的页码可能与当前版本的页码不同

Changes from Revision * (September 1999) to Revision A (October 2023)	Page
• 更新了整个文档中的表格、图和交叉参考的编号格式。.....	1
• Changed pin 7 from V+ to V_{S+} and pin 4 from V- to V_{S-}	3
• Added $V_{REF} = 0\text{ V}$, $V_{CM} = V_S / 2$ "unless otherwise noted" conditions in <i>Electrical Characteristics</i> and <i>Typical Characteristics</i> for clarity.....	5
• Changed from Offset Voltage vs Power Supply to Power-supply rejection ratio for more clarity.....	5
• Change test condition V_{CM} formula for more clarity.....	5
• Added test condition "TA = -40°C to 85°C" for "Gain error vs temperature" in <i>Electrical Characteristics</i> and renamed to "Gain drift" for clarity.....	5
• Added test condition "TA = -40°C to 85°C" for Output over Temperature in <i>Electrical Characteristics</i>	5
• Added test condition "Continuous to $V_S / 2$ " short-circuit current specification in <i>Electrical Characteristics</i> for clarity.....	5
• Deleted redundant voltage range, operating temperature range, and specification temperature range specifications from <i>Electrical Characteristics</i>	5
• Added $V_{REF} = 0\text{ V}$, $V_{CM} = V_S / 2$, to "unless otherwise noted" conditions in <i>Electrical Characteristics</i> and <i>Typical Characteristics</i> for clarity.....	6
• Changed from Offset Voltage vs Power Supply to Power-supply rejection ratio for more clarity.....	6
• Added test condition "TA = -40°C to 85°C" for "Gain error vs temperature" in <i>Electrical Characteristics</i> and renamed to "Gain drift" for clarity.....	6
• Added test condition "TA = -40°C to 85°C" to "Output voltage" in <i>Electrical Characteristics</i>	6
• Added test condition "Continuous to $V_S / 2$ " short-circuit current specification in <i>Electrical Characteristics</i> for clarity.....	6
• Deleted redundant voltage range, operating temperature range, and specification temperature range specifications from <i>Electrical Characteristics</i>	6
• Added $V_{REF} = 0\text{ V}$, $V_{CM} = V_S / 2$ to "unless otherwise noted" conditions in <i>Electrical Characteristics</i> and <i>Typical Characteristics</i> for clarity.....	7

9 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead finish/ Ball material (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
INA146UA	OBSOLETE	SOIC	D	8		TBD	Call TI	Call TI	-40 to 85	INA 146UA	
INA146UA/2K5	ACTIVE	SOIC	D	8	2500	RoHS & Green	NIPDAU	Level-3-260C-168 HR	-40 to 85	INA 146UA	Samples
INA146UAE4	NRND	SOIC	D	8	75	TBD	Call TI	Call TI	-40 to 85		

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

(2) **RoHS:** TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

Green: TI defines "Green" to mean the content of Chlorine (Cl) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "-" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

(6) Lead finish/Ball material - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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TAPE AND REEL INFORMATION

QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE


*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
INA146UA/2K5	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
INA146UA/2K5	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1

TAPE AND REEL BOX DIMENSIONS


*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
INA146UA/2K5	SOIC	D	8	2500	353.0	353.0	32.0
INA146UA/2K5	SOIC	D	8	2500	367.0	367.0	35.0

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