Optimize RF Sampling Receiver Performance using Frequency Planning

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1

Frequency Planning

- 1. ADC SFDR
	- Low order spurs H2-H5
	- Interleaving spurs
	- High order harmonics
	- SFDR vs amplitude/frequency?
	- Nyquist zones and aliasing
- 2. Concept of frequency planning
- 3. Evolution from Traditional IF sampling to Modern RF sampling

Introduction

- High-speed ADCs (> 100Msps) are used in wide range of different applications.
- In frequency domain applications (such as communications or radar), the ADC spur free dynamic range (SFDR) is a key receiver performance differentiator.
- ADC manufacturers spend a lot of time and power to improve spurious performance.

SFDR – Where do the spurs come from?

Pipeline design is one of most commonly used architectures for high-speed ADCs

- Input buffer, amplifiers and other "non-ideal" analog blocks have non-linear behavior
- Large input signal amplitudes make degradation worse
- Later stages and digital noise coupling are responsible for higher order harmonics

Additional SFDR Impairments

SFDR gets worse with higher input frequencies

- Layout and matching gets harder as frequencies increase
	- Fin = 100MHz $\qquad \Rightarrow$ HD2 = 200MHz
	- Fin = 1 GHz \qquad => HD2 = 2 GHz

Low order harmonics (HD2...HD5) tend to improve with backoff (ie lower amplitude)

– Amplifier drives smaller swing at higher frequencies

Additional SFDR Impairments (2)

- Fast Gsps ADCs are often times interleaved ADCs
- Interleaved ADCs have additional spurs originating from gain, offset, bandwidth and timing mismatch
- A 4x interleaved ADC (e.g. ADC12J4000, ADC32RF45) will have spurs at:

ADC Frequency Spectrum

TEXAS INSTRUMENTS

Typical HS-ADC FFT Plot

Sampling Rate FS = 500Msps, Input Signal at 62MHz

- Nyquist zone = $0...250$ MHz (FS/2)
- Larger amplitude low order harmonic spurs (HD2… HD5)

• Lower amplitude high order harmonic spurs (HD6+)

Nyquist Sampling vs. Undersampling vs. RF Sampling

- Nyquist theorem dictates that $fs > 2$ * fin, max to avoid aliasing
- Beyond Nyquist, if fs $> 2 * \Delta f$ (signal bandwidth) and with careful frequency planning, spectral overlap can be avoided as signal folds back to 1st Nyquist zone
- With RF sampling, the undersampling concept gets expanded to higher Nyquist zones and offers more benefits and flexibility

Adopted from: TIPL ADCs – Bandwidth vs. Frequency (Sub-sampling concepts), <https://training.ti.com/ti-precision-labs-adcs-bandwidth-vs-frequency>

Practical Aliasing Example - Undersampling

Practical Aliasing Example - Undersampling

Are all ADCs suitable for undersampling / RF Sampling?

- Analog Input Bandwidth is key parameter determining undersampling capability
- Recent RF sampling ADCs have pushed usable input bandwidth up as high as 10GHz
- Trend continues to provide >> 10GHz direct sampling capability for microwave and millimeterwave systems

ADC12DJ3200 6.4-GSPS Single C 12-bit, RF-Sampling Analo

- **Features**
- ADC Core:
	- 12-bit Resolution
	- Up to 6.4 GSPS in single channel mode -
	- Up to 3.2 GSPS in dual channel mode
- Buffered Analog Inputs with V_{CM} of 0 V

- Usable input frequency range: >10 GHz
- Full-scale input voltage (V_{FS} , default): 0.8 V_{PP}

Concept of Frequency Planning

Frequency Planning Concept

- Until ~ 2010 the primary path to achieving highest SFDR was to spend more power on the ADC.
- As sampling rate requirements increased quickly that direction got more and more impractical

Let's look at an example:

Input frequency band Fin

=> ADC generates primary harmonic bands HD2 to HD5 (remember HD2 = twice as wide as Fin etc)

Frequency Planning Concept (2)

Now let's add the Nyquist zone boundary of the ADC

=> Bands of harmonic frequencies wrap around Nyquist zone edges

Frequency Planning Concept (3)

In receiver application there are 2 different types of interferer signals:

- In-band: within the pass-band of the external filter experience no attenuation Harmonic distortions in ADC from this interferer will be larger
- Out-of-band: get attenuated by external filter

Harmonic distortions in ADC will be much smaller as ADC SFDR improves with backoff

=> Frequency planning concept focuses on in-band interferer

Frequency Planning Concept (4)

Wanted band Fin has partial overlap with HD2/HD3 and full overlap with HD4/HD5 => The harmonic distortions of In-band Interferer fall on top of wanted signal

Frequency Planning Concept (5)

By increasing Nyquist zone (faster ADC sampling rate), the input band can be placed such that all low order harmonics from In-band Interferer fall out of band

Rule of thumb: ADC Sampling Rate needs > 10x of Signal Bandwidth

Digital Decimation Filter is often used to reduce output data rate

 \Rightarrow Oversampling + decimation (similar to ΔΣ ADC)

Frequency Planning Concept (6)

Summary

- Concept of frequency planning has been successfully used in last several years to achieve extraordinary SFDR performance with the ADC
- Nothing specific to RF sampling
- System Designer picks the IF center frequency and sampling rate such that the ADC dominant harmonics fall outside band of interest
	- Change input frequency range to plan around hd2, hd3 and dominant spurs.
	- **Increase** sampling frequency to work around fixed input frequency bands
	- Optionally use decimation filter to reduce data rate again

Evolution to RF sampling

Traditional IF Sampling

- Desired signal at some RF frequency anywhere from 700M to 6GHz
- Gets filtered and mixed to intermediate (IF) frequency (0-500MHz)
- Additional amplification and filtering is needed

Traditional IF Sampling (2)

- In-band interferer at RF can't be filtered out
- Mixer down converts Image and ADC Alias, which fall on top of wanted signal after sampling
- => Both can contain out-of-band interferer which needs to be filtered out primarily at RF

Evolution to Direct RF Sampling

Remove an IF sampling stage

- Less components (higher integration)
- Less power consumption
- Higher overall channel count possible
- Relax RF Filtering (no close Image to filter)
- RF ADC can replace multiple IF chains
- Can use frequency planning

Super wide BW signal (500+MHz BW ultra wideband)

- Like traditional IF sampling, just at RF.
- Limited possibility of frequency planning

Direct RF Sampling

- In-band interferers still can't be filtered out. Need frequency planning to avoid its harmonics
- Out-of-band interferers get attenuated by filter.
	- => Filtering gets little more relaxed as no close by Image to filter out

Tools for RF Sampling Receiver Designs

1. Frequency Planning Tool

- Quick analysis for in-band interferer
- 2. Analog Filter Design Tool
	- External RF Filter design
- 3. Decimation Filter Spur Calculator
	- Tool to trace back spur origin when using decimation filter

Frequency Planning Tool

Basic Idea:

Develop a tool that shows location of unwanted spurs in frequency domain Main parameters to adjust:

- ADC Sampling Rate
- Frequency band of interest (center frequency and bandwidth)

TEXAS INSTRUMENTS

Frequency Planning Tool

- 1. Enter ADC sampling rate
- 2. Enter ADC interleaving factor (e.g. ADC32RF45, ADC12J4000 = 4x)
- 3. Enter Signal Center Frequency and Bandwidth

Tool calculates:

- ADC Nyquist Boundary
- Locations of dominant spurs in frequency domain
- Displays if there is overlap with input signal

Frequency Planning Tool (2)

What is HD2/3 Image?

- In interleaved ADCs, the low order harmonics (HD2-HD5) are mixing with the internal, interleaved clock frequencies.
- This creates images of low order harmonics, just like interleaving spurs with amplitudes significantly better than the harmonics but possibly worse than spur noise floor.
- For example: A 4x interleaved ADC would have HD2 image at Fs/2-HD2 and FS+/- HD2

Good vs bad frequency plan – fixed RF frequency

Fixed RF input frequency: 100MHz RX band centered at 2050MHz

Frequency Plan Frequency Plan User Entry User Entry ADC Sampling Rate 2900 Msps **ADC Sampling Rate** 3000 Msps Interleaving Factor (ADC) $2x$ Overlap? α Interleaving Factor (ADC) $2x$ Overlap? HD₂ Signal Center 2050 MHz **Signal Center** 2050 MHz Signal BW 100 MHz 100 MHz **Signal BW Frequency Plan Chart Frequency Plan Chart** 6 -6 -5 5 RX band RX band $\overline{\mathbf{A}}$ \overline{a} -HD₂ $-mn2$ Harmonic No. g ----- $-HD3$ ----- $-$ HD3 Harmonic
w -HD4 $-$ HD4 **HD5** -HD5 - Nyquist — Nyquist $\overline{2}$ $\overline{2}$ $-$ HD2 Image $-$ HD2 Image $\mathbf{1}$ $\mathbf{1}$ $-$ HD3 Image $-$ HD3 Image $\bf{0}$ $\mathbf{0}$ 200 Ω 200 400 600 800 1000 1200 1400 1600 - 0 400 600 800 1000 1200 1400 1600 **Frequency (MHz) Frequency (MHz)**

Good Frequency Plan Bad Frequency Plan

Good vs bad frequency plan – fixed ADC clock rate

Fixed ADC clock rate of 3Gsps and 200MHz RX band

Bad Frequency Plan Good Frequency Plan

Multiband Option

ADC32RF45 – Good Frequency Plan

- $Fs = 3Gsps$, $Fin = 1940MHz$ with 100MHz BW
- Overlap with HD3 image and HD4

ADC32RF45 – Bad Frequency Plan

- $Fs = 3 Gsps$, $Fin = 2000 MHz$ with 100MHz BW
- Overlap with HD2, HD5

RF Filter Design Tool

TI Confidential – NDA Restrictions

Background

- Out-of-band interferers can generate large harmonic spurs that fall inside signal band
- Need to attenuate out-of-band signals sufficiently
- Also need to consider ADC alias bands

RF Filter Design Tool (1)

Let's look at the following example:

=> Remember HD2 of interferer is twice as wide as interferer itself

RF Filter Design Tool (2)

There are 2 other Nyquist zones with signal aliases we need to consider for filter design.

1st Nyquist zone:

 $\text{Alias} = 1150 - 1250 \text{ MHz}$

=> Filter for out-of-band interferer at 575-625 MHz

3rd Nyquist zone:

 $\text{Alias} = 4150 - 4250 \text{ MHz}$

=> Filter for HD2 of out-of-band interferer at 2075-2125 MHz

RF Filter Design Tool (3)

User Entry:

- ADC setup (sampling rate, interleaving factor, signal bandwidth and center)
- SFDR limit for out-of-band interferer
- Information about ADC SFDR Performance

The RF Filter Tool performs a frequency spectrum sweep from 0 to 4 GHz and returns:

- Table with required filter attenuation needed for each dominant spur and filter offset from input signal.
- A filter mask indicating required out-of-band filter attenuation vs frequency

RF Filter Design Tool (4)

A few other points to note:

- High order spurs (non HD23) require broad band filtering
- Band alias in other Nyquist zones requires full out-of-band attenuation as interferer directly falls on top of wanted signal
- This is a simple tool for first hand check. Signal backoff due to filtering is not taken into account. As interferer amplitude is reduced (due to filtering), HD2-HD5 will further improve.

DDC Calculator Tool

TI Confidential – NDA Restrictions

Background

- With decimation it can be difficult to trace back spurs in FFT plots.
- FFT plot shows Fs = 3Gsps with 8x decimation
- => Few dominant spurs in FFT plot are not labelled
	- $F = 150$ MHz
	- $-$ F = 50 MHz
	- $-$ F = 15 MHz
	- $-$ F = -50 MHz
	- $F = -150 MHz$

DDC Calculator

- Developed a tool that shows where dominant spurs are before and after complex decimation
- Requires ADC setup information (Sampling Rate, interleaving factor, input frequency, complex decimation factor and NCO frequency)

Spectrum after Sampling

- This basically shows the full ADC Nyquist zone without any decimation applied.
- The plot shows input aliased to 1st Nyquist zone (and it's image in the negative frequency)
- Overlaid on top are the NCO frequency along with decimation filter response

Spectrum after NCO Shift

- Next, the location of input signal and all relevant spurs is calculated for both the alias and negative image. This is prior to decimation.
- Frequencies located within the passband of the DDC are highlighted in red. In this example we can see one interleaving spur (FS/4+Fin), HD4 and one HD3 image pass through the decimation filter.

Spectrum after Decimation

• Finally the frequency locations after decimation filtering are calculated. This is a little bit more complicated as in complex decimation frequencies wrap around instead of

Comparison of FFT plots vs Calculation

Calculation correctly predicted spurs at

- $F = 150$ MHz
- \cdot F = 50 MHz
- $F = -150$ MHz

Spurs at $F = +15$ and -50MHz were not determined with the tool

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