# Optimize RF Sampling Receiver Performance using Frequency Planning

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## **Frequency Planning**

- 1. ADC SFDR
  - Low order spurs H2-H5
  - Interleaving spurs
  - High order harmonics
  - SFDR vs amplitude/frequency?
  - Nyquist zones and aliasing
- 2. Concept of frequency planning
- 3. Evolution from Traditional IF sampling to Modern RF sampling



### Introduction

- High-speed ADCs (> 100Msps) are used in wide range of different applications.
- In frequency domain applications (such as communications or radar), the ADC spur free dynamic range (SFDR) is a key receiver performance differentiator.
- ADC manufacturers spend a lot of time and power to improve spurious performance.



## **SFDR – Where do the spurs come from?**

Pipeline design is one of most commonly used architectures for high-speed ADCs

- Input buffer, amplifiers and other "non-ideal" analog blocks have non-linear behavior
- Large input signal amplitudes make degradation worse
- Later stages and digital noise coupling are responsible for higher order harmonics





## **Additional SFDR Impairments**

#### SFDR gets worse with higher input frequencies

- Layout and matching gets harder as frequencies increase
  - Fin = 100MHz => HD2 = 200MHz
  - Fin = 1GHz => HD2 = 2GHz

Low order harmonics (HD2...HD5) tend to <u>improve with backoff</u> (ie lower amplitude)

- Amplifier drives smaller swing at higher frequencies





## **Additional SFDR Impairments (2)**

- Fast Gsps ADCs are often times interleaved ADCs
- Interleaved ADCs have additional spurs originating from gain, offset, bandwidth and timing mismatch
- A 4x interleaved ADC (e.g. ADC12J4000, ADC32RF45) will have spurs at:





## **ADC Frequency Spectrum**



Texas Instruments

## **Typical HS-ADC FFT Plot**

Sampling Rate FS = 500Msps, Input Signal at 62MHz

- Nyquist zone = 0...250 MHz (FS/2)
- Larger amplitude low order harmonic spurs (HD2... HD5) •



Lower amplitude high order harmonic spurs (HD6+) ٠



## Nyquist Sampling vs. Undersampling vs. RF Sampling

- Nyquist theorem dictates that fs > 2 \* fin,max to avoid aliasing
- Beyond Nyquist, if fs > 2 \* ∆f (signal bandwidth) and with careful frequency planning, spectral overlap can be avoided as signal folds back to 1<sup>st</sup> Nyquist zone
- With RF sampling, the undersampling concept gets expanded to higher Nyquist zones and offers more benefits and flexibility



Adopted from: TIPL ADCs – Bandwidth vs. Frequency (Sub-sampling concepts), https://training.ti.com/ti-precision-labs-adcs-bandwidth-vs-frequency



### **Practical Aliasing Example - Undersampling**





### **Practical Aliasing Example - Undersampling**



![](_page_10_Picture_2.jpeg)

### Are all ADCs suitable for undersampling / RF Sampling?

- Analog Input Bandwidth is key parameter determining undersampling capability
- Recent RF sampling ADCs have pushed usable input bandwidth up as high as 10GHz
- Trend continues to provide >> 10GHz direct sampling capability for microwave and millimeterwave systems

![](_page_11_Picture_4.jpeg)

#### ADC12DJ3200 6.4-GSPS Single C 12-bit, RF-Sampling Analo

- 1 Features
- ADC Core:
  - 12-bit Resolution
  - Up to 6.4 GSPS in single channel mode
  - Up to 3.2 GSPS in dual channel mode
- Buffered Analog Inputs with V<sub>CMI</sub> of 0 V

Analog input bandwidth (-3 dB): 8.0 GHz
Usable input frequency range: >10 GHz

Full-scale input voltage (V<sub>ES</sub>, default): 0.8 V<sub>PP</sub>

![](_page_11_Figure_14.jpeg)

![](_page_11_Picture_15.jpeg)

# **Concept of Frequency Planning**

![](_page_12_Picture_1.jpeg)

## **Frequency Planning Concept**

- Until ~ 2010 the primary path to achieving highest SFDR was to spend more power on the ADC.
- As sampling rate requirements increased quickly that direction got more and more impractical

Let's look at an example:

Input frequency band Fin

=> ADC generates primary harmonic bands HD2 to HD5 (remember HD2 = twice as wide as Fin etc)

![](_page_13_Figure_6.jpeg)

![](_page_13_Picture_7.jpeg)

## **Frequency Planning Concept (2)**

Now let's add the Nyquist zone boundary of the ADC

=> Bands of harmonic frequencies wrap around Nyquist zone edges

![](_page_14_Figure_3.jpeg)

## **Frequency Planning Concept (3)**

In receiver application there are 2 different types of interferer signals:

- In-band: within the pass-band of the external filter experience no attenuation
   Harmonic distortions in ADC from this interferer will be larger
- Out-of-band: get attenuated by external filter
  - Harmonic distortions in ADC will be much smaller as ADC SFDR improves with backoff

=> Frequency planning concept focuses on in-band interferer

![](_page_15_Figure_6.jpeg)

![](_page_15_Picture_7.jpeg)

## **Frequency Planning Concept (4)**

Wanted band Fin has partial overlap with HD2/HD3 and full overlap with HD4/HD5

=> The harmonic distortions of In-band Interferer fall on top of wanted signal

![](_page_16_Figure_3.jpeg)

![](_page_16_Picture_4.jpeg)

## **Frequency Planning Concept (5)**

By increasing Nyquist zone (faster ADC sampling rate), the input band can be placed such that all low order harmonics from In-band Interferer fall out of band

Rule of thumb: ADC Sampling Rate needs > 10x of Signal Bandwidth

Digital Decimation Filter is often used to reduce output data rate

=> Oversampling + decimation (similar to  $\Delta\Sigma$  ADC)

![](_page_17_Figure_5.jpeg)

## **Frequency Planning Concept (6)**

### Summary

- Concept of frequency planning has been successfully used in last several years to achieve extraordinary SFDR performance with the ADC
- Nothing specific to RF sampling
- System Designer picks the IF center frequency and sampling rate such that the ADC dominant harmonics fall outside band of interest
  - Change input frequency range to plan around hd2, hd3 and dominant spurs.
  - **Increase** sampling frequency to work around fixed input frequency bands
  - Optionally use decimation filter to reduce data rate again

![](_page_18_Picture_8.jpeg)

# **Evolution to RF sampling**

![](_page_19_Picture_1.jpeg)

## **Traditional IF Sampling**

- Desired signal at some RF frequency anywhere from 700M to 6GHz
- Gets filtered and mixed to intermediate (IF) frequency (0-500MHz)
- Additional amplification and filtering is needed

![](_page_20_Figure_4.jpeg)

![](_page_20_Picture_5.jpeg)

## **Traditional IF Sampling (2)**

- In-band interferer at RF can't be filtered out
- Mixer down converts Image and ADC Alias, which fall on top of wanted signal after sampling
- => Both can contain out-of-band interferer which needs to be filtered out primarily at RF

![](_page_21_Figure_4.jpeg)

![](_page_21_Picture_5.jpeg)

## **Evolution to Direct RF Sampling**

#### Remove an IF sampling stage

- Less components (higher integration)
- Less power consumption
- Higher overall channel count possible
- Relax RF Filtering (no close Image to filter)
- RF ADC can replace multiple IF chains
- Can use frequency planning

Super wide BW signal (500+MHz BW ultra wideband)

- Like traditional IF sampling, just at RF.
- Limited possibility of frequency planning

![](_page_22_Figure_11.jpeg)

![](_page_22_Picture_12.jpeg)

## **Direct RF Sampling**

- In-band interferers still can't be filtered out. Need frequency planning to avoid its harmonics
- Out-of-band interferers get attenuated by filter.
  - => Filtering gets little more relaxed as no close by Image to filter out

![](_page_23_Figure_4.jpeg)

![](_page_23_Picture_5.jpeg)

# **Tools for RF Sampling Receiver Designs**

### 1. Frequency Planning Tool

- Quick analysis for in-band interferer
- 2. Analog Filter Design Tool
  - External RF Filter design
- 3. Decimation Filter Spur Calculator
  - Tool to trace back spur origin when using decimation filter

![](_page_24_Picture_7.jpeg)

## **Frequency Planning Tool**

Basic Idea:

Develop a tool that shows location of unwanted spurs in frequency domain Main parameters to adjust:

- ADC Sampling Rate
- Frequency band of interest (center frequency and bandwidth)

![](_page_25_Figure_5.jpeg)

![](_page_25_Picture_6.jpeg)

## **Frequency Planning Tool**

- 1. Enter ADC sampling rate
- 2. Enter ADC interleaving factor (e.g. ADC32RF45, ADC12J4000 = 4x)
- 3. Enter Signal Center Frequency and Bandwidth

![](_page_26_Figure_4.jpeg)

Tool calculates:

- ADC Nyquist Boundary
- Locations of dominant spurs in frequency domain
- Displays if there is overlap with input signal

![](_page_26_Picture_9.jpeg)

## **Frequency Planning Tool (2)**

![](_page_27_Figure_1.jpeg)

![](_page_27_Picture_2.jpeg)

## What is HD2/3 Image?

- In interleaved ADCs, the low order harmonics (HD2-HD5) are mixing with the internal, interleaved clock frequencies.
- This creates images of low order harmonics, just like interleaving spurs with amplitudes significantly better than the harmonics but possibly worse than spur noise floor.
- For example: A 4x interleaved ADC would have HD2 image at Fs/2-HD2 and FS+/- HD2

![](_page_28_Figure_4.jpeg)

![](_page_28_Picture_5.jpeg)

## Good vs bad frequency plan – fixed RF frequency

Fixed RF input frequency: 100MHz RX band centered at 2050MHz

Good Frequency Plan Frequency Plan Frequency Plan User Entry **User Entry** ADC Sampling Rate 2900 Msps ADC Sampling Rate 3000 Msps Interleaving Factor (ADC) 2 x Overlap? OK Interleaving Factor (ADC) 2 x Overlap? 2050 MHz Signal Center Signal Center 2050 MHz 100 MHz 100 MHz Signal BW Signal BW Frequency Plan Chart Frequency Plan Chart 6 6 5 5 RX band RX band 4 4 HD2 HD2 Harmonic No. 8 Ň, \_ \_ \_ \_ \_ HD3 \_\_\_\_ HD3 Harmonic HD4 HD4 HD5 HD5 — Nyquist — Nyquist 2 2 HD2 Image HD2 Image 1 1 HD3 Image HD3 Image 0 0 200 0 200 400 600 800 1000 1200 1400 1600 0 400 600 800 1000 1200 1400 1600 Frequency (MHz) Frequency (MHz)

#### **Bad Frequency Plan**

![](_page_29_Picture_4.jpeg)

## Good vs bad frequency plan – fixed ADC clock rate

Fixed ADC clock rate of 3Gsps and 200MHz RX band

Bad Frequency Plan

![](_page_30_Figure_3.jpeg)

Good Frequency Plan

![](_page_30_Picture_5.jpeg)

## **Multiband Option**

![](_page_31_Figure_1.jpeg)

![](_page_31_Picture_2.jpeg)

## ADC32RF45 – Good Frequency Plan

- Fs = 3Gsps, Fin = 1940MHz with 100MHz BW
- Overlap with HD3 image and HD4

![](_page_32_Figure_3.jpeg)

![](_page_32_Picture_4.jpeg)

## ADC32RF45 – Bad Frequency Plan

- Fs = 3 Gsps, Fin = 2000 MHz with 100MHz BW
- Overlap with HD2, HD5

![](_page_33_Figure_3.jpeg)

![](_page_33_Picture_4.jpeg)

# **RF Filter Design Tool**

TI Confidential – NDA Restrictions

![](_page_34_Picture_2.jpeg)

## Background

- Out-of-band interferers can generate large harmonic spurs that fall inside signal band
- Need to attenuate out-of-band signals sufficiently
- Also need to consider ADC alias bands

![](_page_35_Figure_4.jpeg)

![](_page_35_Picture_5.jpeg)

## **RF Filter Design Tool (1)**

Let's look at the following example:

ADC Sampling Rate	3 Gsps
RX Band:	100MHz centered at 1800MHz (1750 to 1850MHz)
In-band SFDR Requirement:	100 dB
HD2 of the RF ADC is:	65 dB
=> Filter attenuation for <u>HD2</u> is:	35 dB (100 dB – 65 dB)
At primary frequency location:	875 to 925 MHz (1750/2 to 1850/2)

=> Remember HD2 of interferer is twice as wide as interferer itself

![](_page_36_Figure_4.jpeg)

![](_page_36_Picture_5.jpeg)

## **RF Filter Design Tool (2)**

There are 2 other Nyquist zones with signal aliases we need to consider for filter design.

1<sup>st</sup> Nyquist zone:

3<sup>rd</sup> Nyquist zone:

Alias = 1150-1250 MHz

=> Filter for out-of-band interferer at 575-625 MHz

Alias = 4150-4250 MHz

=> Filter for HD2 of out-of-band interferer at 2075-2125 MHz

![](_page_37_Figure_8.jpeg)

![](_page_37_Picture_9.jpeg)

## **RF Filter Design Tool (3)**

User Entry:

- ADC setup (sampling rate, interleaving factor, signal bandwidth and center)
- SFDR limit for out-of-band interferer
- Information about ADC SFDR Performance

The RF Filter Tool performs a frequency spectrum sweep from 0 to 4 GHz and returns:

- Table with required filter attenuation needed 
  for each dominant spur and filter offset from
  input signal.
- A filter mask indicating required out-of-band filter attenuation vs frequency

![](_page_38_Figure_8.jpeg)

![](_page_38_Picture_9.jpeg)

## **RF Filter Design Tool (4)**

A few other points to note:

- High order spurs (non HD23) require broad band filtering
- Band alias in other Nyquist zones requires full out-of-band attenuation as interferer directly falls on top of wanted signal
- This is a simple tool for first hand check. Signal backoff due to filtering is not taken into account. As interferer amplitude is reduced (due to filtering), HD2-HD5 will further improve.

![](_page_39_Figure_5.jpeg)

![](_page_39_Picture_6.jpeg)

## **DDC Calculator Tool**

TI Confidential – NDA Restrictions

![](_page_40_Picture_2.jpeg)

## Background

- With decimation it can be difficult to trace back spurs in FFT plots.
- FFT plot shows Fs = 3Gsps with 8x decimation
- => Few dominant spurs in FFT plot are not labelled
  - F = 150 MHz
  - F = 50 MHz
  - F = 15 MHz
  - F = -50 MHz
  - F = -150 MHz

![](_page_41_Figure_9.jpeg)

![](_page_41_Picture_10.jpeg)

## **DDC Calculator**

- Developed a tool that shows where dominant spurs are before and after complex decimation
- Requires ADC setup information (Sampling Rate, interleaving factor, input frequency, complex decimation factor and NCO frequency)

![](_page_42_Figure_3.jpeg)

![](_page_42_Picture_4.jpeg)

## **Spectrum after Sampling**

- This basically shows the full ADC Nyquist zone without any decimation applied.
- The plot shows input aliased to 1<sup>st</sup> Nyquist zone (and it's image in the negative frequency)
- Overlaid on top are the NCO frequency along with decimation filter response

![](_page_43_Figure_4.jpeg)

![](_page_43_Picture_5.jpeg)

## **Spectrum after NCO Shift**

- Next, the location of input signal and all relevant spurs is calculated for both the alias and negative image. This is prior to decimation.
- Frequencies located within the passband of the DDC are highlighted in red. In this example we can see one interleaving spur (FS/4+Fin), HD4 and one HD3 image pass through the decimation filter.

![](_page_44_Figure_3.jpeg)

![](_page_44_Picture_4.jpeg)

## **Spectrum after Decimation**

• Finally the frequency locations after decimation filtering are calculated. This is a little bit more complicated as in complex decimation frequencies wrap around instead of

![](_page_45_Figure_2.jpeg)

![](_page_45_Picture_3.jpeg)

## **Comparison of FFT plots vs Calculation**

Calculation correctly predicted spurs at

- F = 150 MHz
- F = 50 MHz
- F = -150 MHz

Spurs at F = +15 and -50MHz were not determined with the tool

DDC									
	User Entry	Spectrum after:	Sampling		shift b	shift by NCO		Decimation	
ADC Sampling Rate	3000	Msps		Alias	Image	Alias	Image	Alias	Image
nterleaving Factor (ADC)	4	x	Input Frequency	1150	1850	-600	100	150	100
nput Frequency	1850	MHz	IL1 (FS/2 - FIN)	350	2650	-1400	900	100	150
Decimation by	8	x	IL2 (FS/4 - FIN)	400	2600	-1350	850	150	100
NCO	1750	MHz	IL3 (FS/4 + FIN)	1100	1900	-650		100	150
			HD2	700	2300	-1050	550	75	175
			HD3	450	2550	-1300	800	-175	50
			HD2 Image	800	2200	-950	450	175	75
			HD2 Image	50	2950	-1700	1200	175	75
			HD2 Image	1450	1550	-300	-200	75	175
			FS/4	750	2250	-1000	500	125	125
			HD4	1400	1600	-350		25	-150
			HD5	250	2750	-1500	1000	0	-125
			HD3 Image	1050	1950	-700	200	50	-175
			HD3 Image	1200	1800	-550		-175	50
			HD3 Image	300	2700	-1450	950	50	-175
			FIN + NCO	600	2400	-1150	650	-25	-100

![](_page_46_Figure_7.jpeg)

![](_page_46_Picture_8.jpeg)

![](_page_47_Picture_0.jpeg)

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