

# TLV700xx-Q1

## 200-mA, Low- $I_Q$ , Low-Dropout Regulator (LDO) for Portable Devices

### 1 Features

- Qualified for Automotive Applications
- AEC-Q100 Qualified With the Following Results:
  - Device Temperature Grade 1:  $-40^{\circ}\text{C}$  to  $125^{\circ}\text{C}$  Ambient Operating Temperature Range
  - Device HBM ESD Classification Level H2
  - Device CDM ESD Classification Level C4B
- 2% Accuracy
- Low  $I_Q$ : 31  $\mu\text{A}$
- Fixed Output Voltage Combination Possible From 1.9 V to 4.8 V
- High PSRR: 68 dB at 1 kHz
- Stable With Effective Capacitance of 0.1  $\mu\text{F}$
- Thermal Shutdown and Overcurrent Protection
- Latch-Up Performance Meets 100 mA Per AEC-Q100, Level I
- Available in SOT-23-5 and SC70 Packages

### 2 Applications

- Automotive Camera Modules
- Image Sensor Power
- Microprocessor Rails
- Automotive Infotainment Head Units
- Automotive Body Electronics

### 3 Description

The TLV700xx-Q1 family of low-dropout (LDO) linear regulators are low-quiescent-current devices with excellent line and load transient performance. These LDOs are designed for power-sensitive applications. A precision band-gap and error amplifier provides overall 2% accuracy. Low output noise, very high power-supply rejection ratio (PSRR), and low dropout voltage make this series of devices ideal for most battery-operated handheld equipment. All device versions have thermal shutdown and current limit for safety.

Furthermore, these devices are stable with an effective output capacitance of only 0.1  $\mu\text{F}$ . This feature enables the use of cost-effective capacitors that have higher bias voltages and temperature derating. The devices regulate to specified accuracy with no output load.

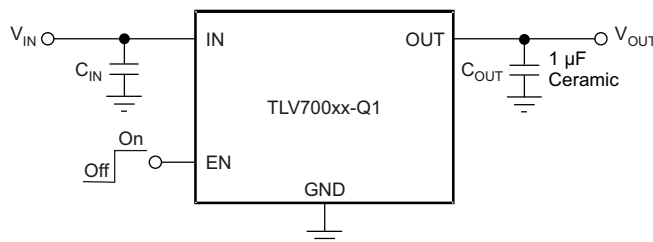
The TLV700xx-Q1 LDOs are available in SOT-23-5 and SC70 packages.

#### Device Information<sup>(1)</sup>

PART NUMBER	PACKAGE	BODY SIZE (NOM)
TLV700xx-Q1	SC70 (5)	2.00 mm x 1.25 mm
	SOT (5)	2.90 mm x 1.60 mm

(1) For all available packages, see the orderable addendum at the end of the data sheet.

#### Typical Application Circuit (Fixed-Voltage Versions)



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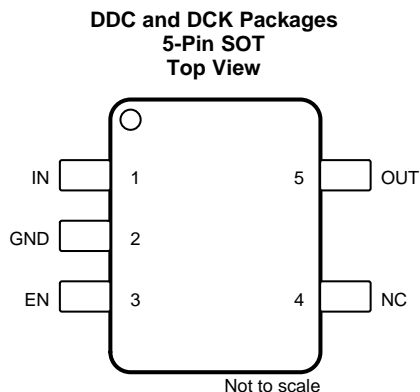
## 4 Revision History

Changes from Revision B (October 2016) to Revision C	Page
• Added DCK (SC70) package to document; note that TLV70025-Q1 and TLV70033-Q1 were previously listed in SLVSA61 .....	1
• Changed <i>Fixed Output Voltages</i> bullet to <i>Fixed Output Voltage Combination</i> in <i>Features</i> section.....	1
• Changed last paragraph of Description section to include the SC70 package .....	1
• Added SC70 row to <i>Device Information</i> table .....	1
• Added DCK package to <i>Pin Configuration and Functions</i> section .....	3
• Added T <sub>J</sub> parameter to <i>Absolute Maximum Ratings</i> table.....	3
• Changed T <sub>J</sub> parameter to T <sub>A</sub> in <i>Recommended Operating Conditions</i> table and changed <i>junction</i> to <i>ambient</i> in parameter name .....	4
• Added <i>TLV70033-Q1 PSRR Ratio</i> figure .....	8

Changes from Revision A (September 2016) to Revision B	Page
• Changed maximum specification of V <sub>EN</sub> parameter in <i>Absolute Maximum Ratings</i> table .....	3
• Changed I <sub>OUT</sub> parameter name in <i>Recommended Operating Conditions</i> table .....	4

Changes from Original (July 2016) to Revision A	Page
• Released to production; note that TLV70028QDDCRQ1 and TLV70032QDDCRQ1 were previously listed in SLVSA61 ....	1

## 5 Pin Configuration and Functions



### Pin Functions

PIN			I/O	DESCRIPTION
NAME	SC70	SOT		
EN	3	3	I	Enable pin. Driving EN over 0.9 V turns on the regulator. Driving EN below 0.4 V puts the regulator into shutdown mode and reduces operating current to 1 $\mu$ A, nominal.
GND	2	2	—	Ground pin
IN	1	1	I	Input pin. A small 1- $\mu$ F ceramic capacitor is recommended from this pin to ground to assure stability and good transient performance. See <a href="#">Input and Output Capacitor Requirements</a> in the <a href="#">Application and Implementation</a> section for more details.
NC	4	4	—	No connection. This pin can be tied to ground to improve thermal dissipation.
OUT	5	5	O	Regulated output voltage pin. A small 1- $\mu$ F ceramic capacitor is needed from this pin to ground to assure stability. See <a href="#">Input and Output Capacitor Requirements</a> in the <a href="#">Application and Implementation</a> section for more details.

## 6 Specifications

### 6.1 Absolute Maximum Ratings

at  $T_A = -40^\circ\text{C}$  to  $+125^\circ\text{C}$  (unless otherwise noted); all voltages are with respect to GND<sup>(1)</sup>

		MIN	MAX	UNIT
$V_{IN}$	Input voltage	-0.3	6	V
$V_{EN}$	Enable voltage	-0.3	$V_{IN} + 0.3$	V
$V_{OUT}$	Output voltage	-0.3	6	V
$I_{OUT}$	Maximum output current	Internally limited		
	Output short-circuit duration	Indefinite		
$T_A$	Operating ambient temperature	-40	150	$^\circ\text{C}$
$T_J$	Operating junction temperature	-40	150	$^\circ\text{C}$
$T_{stg}$	Storage temperature	-55	150	$^\circ\text{C}$

- (1) Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under *Recommended Operating Conditions*. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

### 6.2 ESD Ratings

		VALUE	UNIT
$V_{(ESD)}$	Electrostatic discharge	Human-body model (HBM), per AEC Q100-002 <sup>(1)</sup>	$\pm 2000$
		Charged-device model (CDM), per AEC Q100-011	$\pm 750$

- (1) AEC Q100-002 indicates that HBM stressing shall be in accordance with the ANSI/ESDA/JEDEC JS-001 specification.

### 6.3 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)

		MIN	NOM	MAX	UNIT
V <sub>IN</sub>	Input voltage	2		5.5	V
V <sub>EN</sub>	Enable voltage	0		5.5	V
I <sub>OUT</sub>	Output current		200		mA
C <sub>IN</sub>	Input capacitor	0	1		μF
C <sub>OUT</sub>	Output capacitor	0.22	1		μF
T <sub>A</sub>	Operating ambient temperature	–40		125	°C

### 6.4 Thermal Information

THERMAL METRIC <sup>(1)</sup>	TLV700xx-Q1		UNIT
	DDC (SOT)		
	5 PINS		
R <sub>θJA</sub>	Junction-to-ambient thermal resistance	262.8	°C/W
R <sub>θJC(top)</sub>	Junction-to-case (top) thermal resistance	68.2	°C/W
R <sub>θJB</sub>	Junction-to-board thermal resistance	81.6	°C/W
ψ <sub>JT</sub>	Junction-to-top characterization parameter	1.1	°C/W
ψ <sub>JB</sub>	Junction-to-board characterization parameter	80.9	°C/W
R <sub>θJC(bot)</sub>	Junction-to-case (bottom) thermal resistance	n/a	°C/W

- (1) For more information about traditional and new thermal metrics, see the [Semiconductor and IC Package Thermal Metrics](#) application report.

## 6.5 Electrical Characteristics

at  $T_A = -40^\circ\text{C}$  to  $+125^\circ\text{C}$ ,  $V_{IN} = V_{OUT(TYP)} + 0.3\text{ V}$  or  $2\text{ V}$  (whichever is greater),  $I_{OUT} = 10\text{ mA}$ ,  $V_{EN} = V_{IN}$ , and  $C_{OUT} = 1\text{ }\mu\text{F}$  (unless otherwise noted); typical values are at  $T_A = 25^\circ\text{C}$

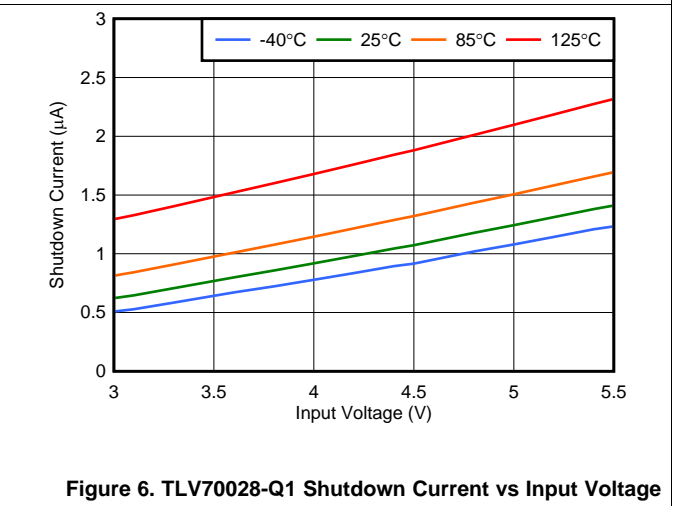
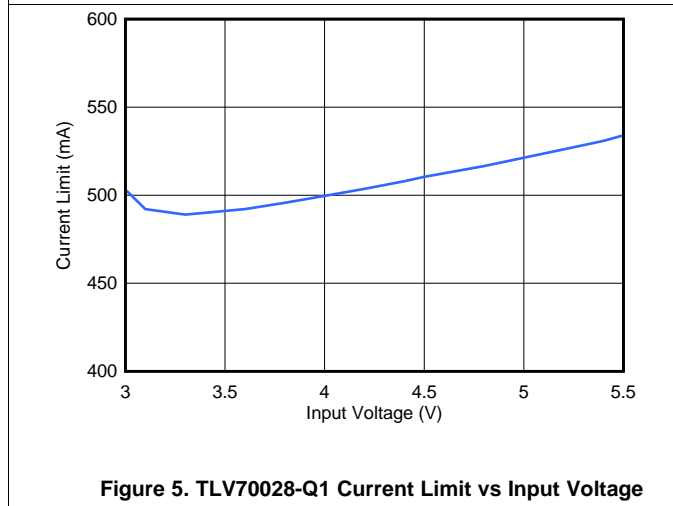
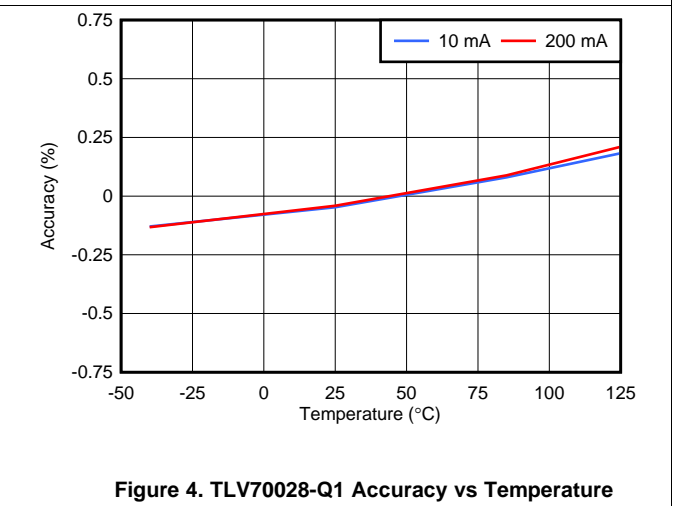
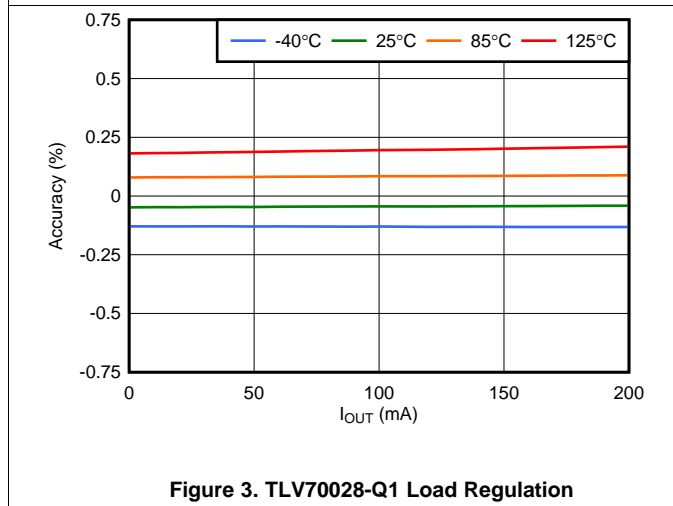
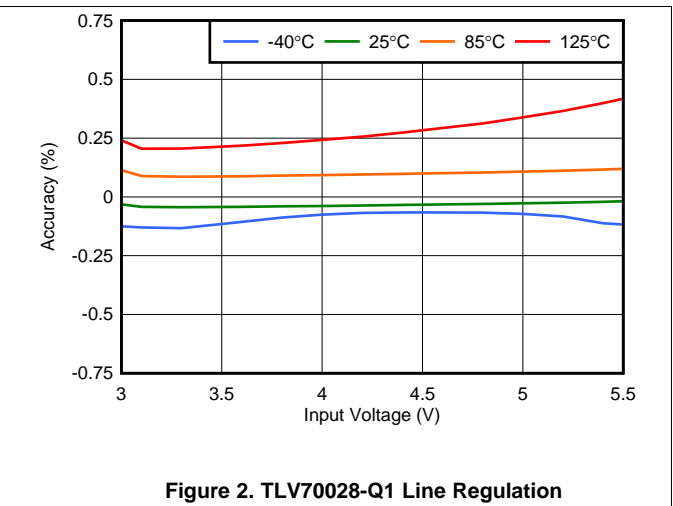
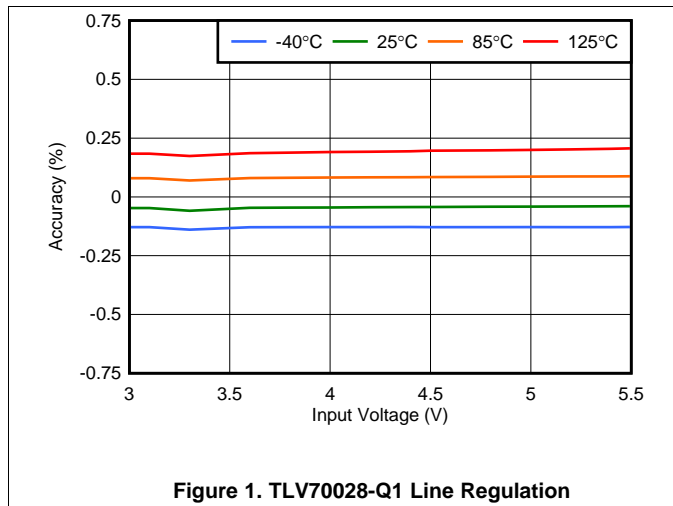
PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
$V_{IN}$	Input voltage range		2		5.5	V
$V_{OUT}$	DC output accuracy	$-40^\circ\text{C} \leq T_A \leq +125^\circ\text{C}$ , $V_{OUT} \geq 1\text{ V}$	-2%		2%	
$\Delta V_O/\Delta V_{IN}$	Line regulation	$V_{OUT(NOM)} + 0.5\text{ V} \leq V_{IN} \leq 5.5\text{ V}$ , $I_{OUT} = 10\text{ mA}$		1	5	mV
$\Delta V_O/\Delta I_{OUT}$	Load regulation	$0\text{ mA} \leq I_{OUT} \leq 200\text{ mA}$			15	mV
$V_{DO}$	Dropout voltage <sup>(1)</sup>	$V_{IN} = 0.98 \times V_{OUT(NOM)}$ , $I_{OUT} = 200\text{ mA}$		175	250	mV
$I_{CL}$	Output current limit	$V_{OUT} = 0.9 \times V_{OUT(NOM)}$	220	350	860	mA
$I_{GND}$	Ground pin current	$I_{OUT} = 0\text{ mA}$		31	55	$\mu\text{A}$
		$I_{OUT} = 200\text{ mA}$ , $V_{IN} = V_{OUT} + 0.5\text{ V}$		270		
$I_{SHDN}$	Ground pin current (shutdown)	$V_{EN} \leq 0.4\text{ V}$ , $2.0\text{ V} \leq V_{IN} \leq 4.5\text{ V}$		1	2.5	$\mu\text{A}$
PSRR	Power-supply rejection ratio	$V_{IN} = 2.3\text{ V}$ , $V_{OUT} = 1.8\text{ V}$ , $I_{OUT} = 10\text{ mA}$ , $f = 1\text{ kHz}$		68		dB
$V_N$	Output noise voltage	$BW = 100\text{ Hz to }100\text{ kHz}$ , $V_{IN} = 2.3\text{ V}$ , $V_{OUT} = 1.8\text{ V}$ , $I_{OUT} = 10\text{ mA}$		48		$\mu\text{V}_{RMS}$
$t_{STR}$	Startup time <sup>(2)</sup>	$C_{OUT} = 1\text{ }\mu\text{F}$ , $I_{OUT} = 200\text{ mA}$		100		$\mu\text{s}$
$V_{EN(HI)}$	Enable pin high (enabled)		0.9		$V_{IN}$	V
$V_{EN(LO)}$	Enable pin low (disabled)		0		0.4	V
$I_{EN}$	Enable pin current	$V_{EN} = 5.5\text{ V}$ , $I_{OUT} = 10\text{ }\mu\text{A}$		0.04	0.5	$\mu\text{A}$
UVLO	Undervoltage lockout	$V_{IN}$ rising		1.9		V
$T_{SD}$	Thermal shutdown temperature	Shutdown, temperature increasing		160		$^\circ\text{C}$
		Reset, temperature decreasing		140		
$T_A$	Operating ambient temperature		-40		125	$^\circ\text{C}$

(1)  $V_{DO}$  is measured for devices with  $V_{OUT(NOM)} \geq 2.35\text{ V}$ .

(2) Startup time = time from EN assertion to  $0.98 \times V_{OUT(NOM)}$ .

### 6.6 Typical Characteristics

at  $T_J = -40^\circ\text{C}$  to  $+125^\circ\text{C}$ ,  $V_{IN} = V_{OUT(TYP)} + 0.5\text{ V}$  or  $2\text{ V}$  (whichever is greater),  $I_{OUT} = 10\text{ mA}$ ,  $V_{EN} = V_{IN}$ ,  $C_{IN} = 1\ \mu\text{F}$ , and  $C_{OUT} = 1\ \mu\text{F}$  (unless otherwise noted); typical values are at  $T_J = 25^\circ\text{C}$



### Typical Characteristics (continued)

at  $T_J = -40^\circ\text{C}$  to  $+125^\circ\text{C}$ ,  $V_{IN} = V_{OUT(TYP)} + 0.5\text{ V}$  or  $2\text{ V}$  (whichever is greater),  $I_{OUT} = 10\text{ mA}$ ,  $V_{EN} = V_{IN}$ ,  $C_{IN} = 1\text{ }\mu\text{F}$ , and  $C_{OUT} = 1\text{ }\mu\text{F}$  (unless otherwise noted); typical values are at  $T_J = 25^\circ\text{C}$

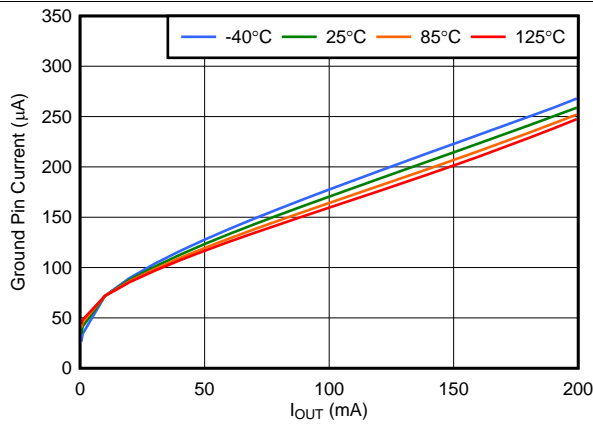


Figure 7. TLV70028-Q1 Ground Pin Current vs Output Current

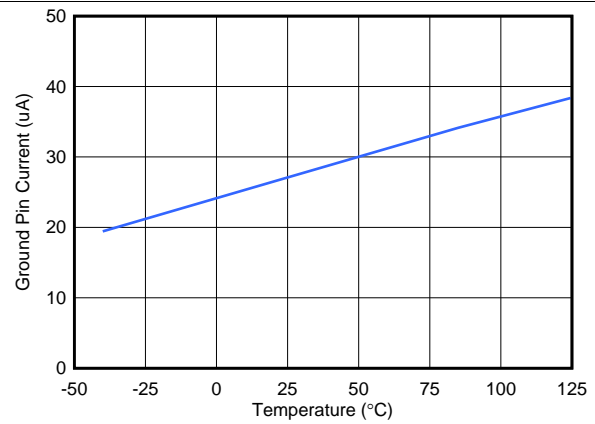


Figure 8. TLV70028-Q1 Ground Pin Current vs Temperature

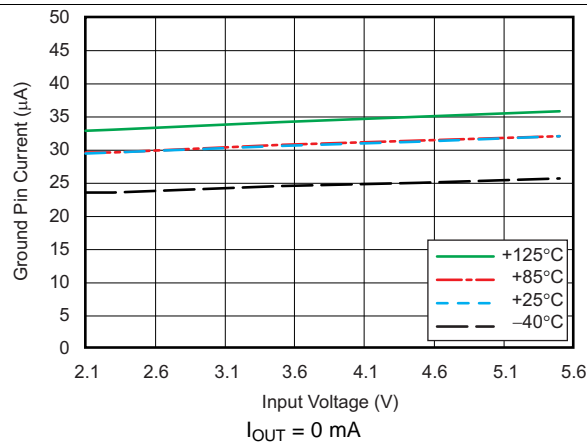


Figure 9. TLV70048-Q1 Ground Pin Current vs Input Voltage

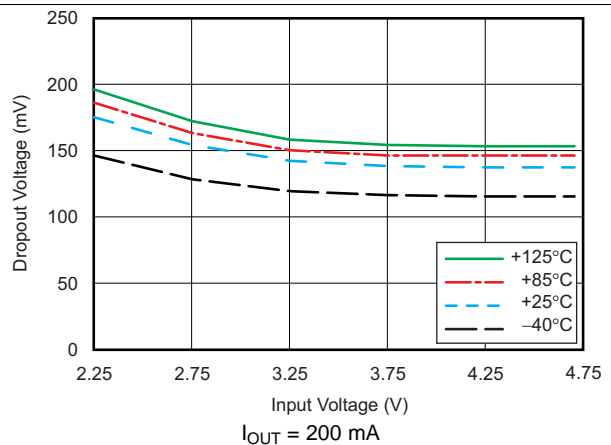


Figure 10. TLV70048-Q1 Dropout Voltage vs Input Voltage

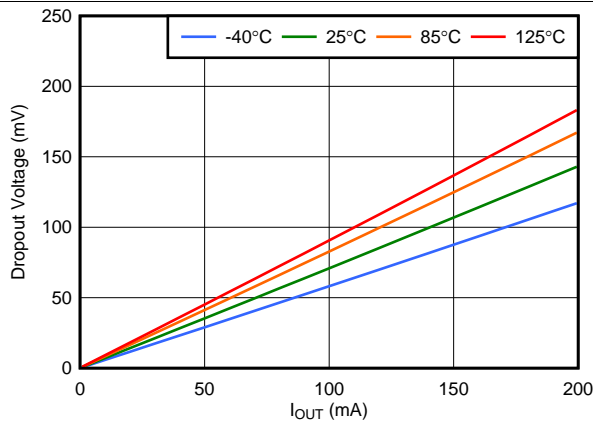


Figure 11. TLV70028-Q1 Dropout Voltage vs Output Current

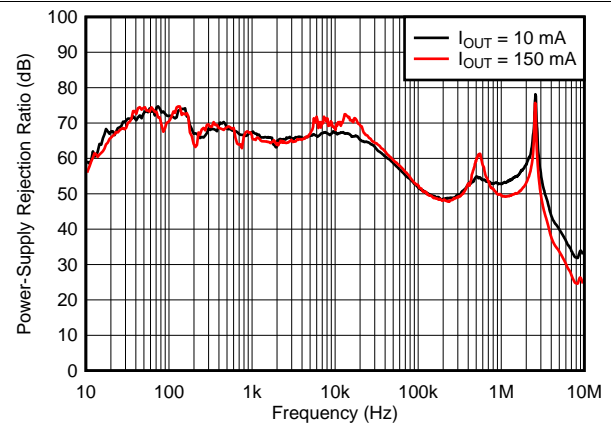


Figure 12. TLV70028-Q1 Power-Supply Rejection Ratio vs Frequency

Typical Characteristics (continued)

at  $T_J = -40^{\circ}\text{C}$  to  $+125^{\circ}\text{C}$ ,  $V_{IN} = V_{OUT(TYP)} + 0.5\text{ V}$  or  $2\text{ V}$  (whichever is greater),  $I_{OUT} = 10\text{ mA}$ ,  $V_{EN} = V_{IN}$ ,  $C_{IN} = 1\ \mu\text{F}$ , and  $C_{OUT} = 1\ \mu\text{F}$  (unless otherwise noted); typical values are at  $T_J = 25^{\circ}\text{C}$

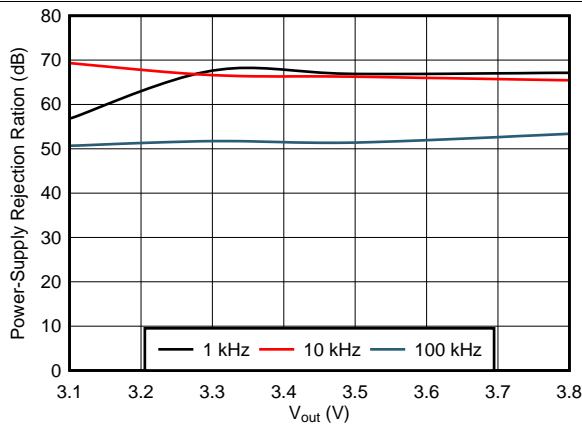


Figure 13. TLV70028-Q1 Power-Supply Rejection Ratio vs Output Voltage

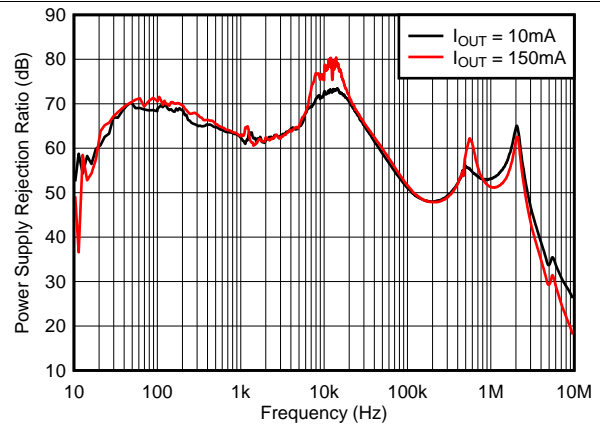


Figure 14. TLV70033-Q1 PSRR Ratio

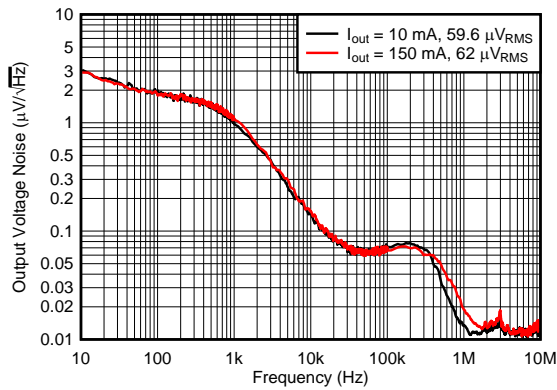
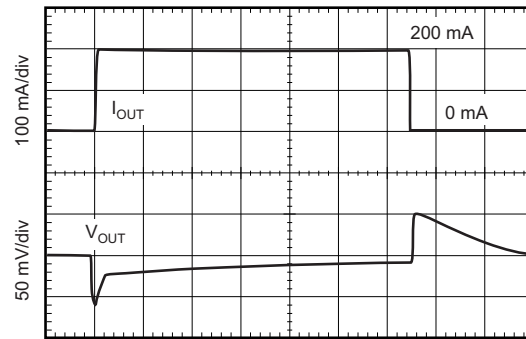
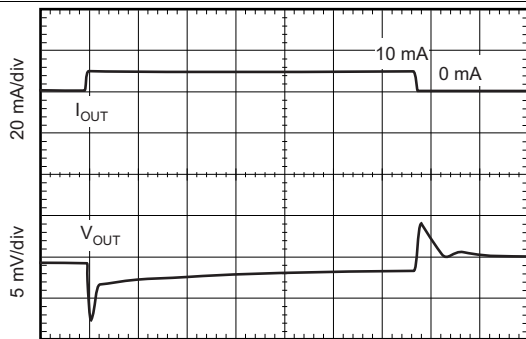


Figure 15. TLV70028-Q1 Output Spectral Noise Density vs Frequency



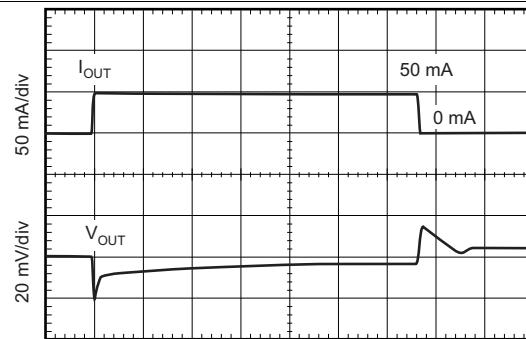
$t_R = t_F = 1\ \mu\text{s}$ ,  $V_{IN} = 2.1\text{ V}$

Figure 16. Load Transient Response



$t_R = t_F = 1\ \mu\text{s}$ ,  $V_{IN} = 2.3\text{ V}$

Figure 17. Load Transient Response



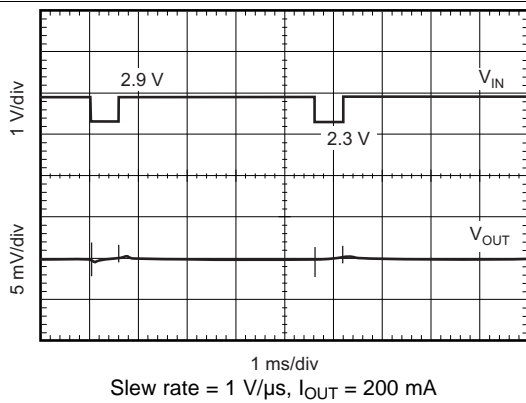
$t_R = t_F = 1\ \mu\text{s}$ ,  $V_{IN} = 2.3\text{ V}$

Figure 18. Load Transient Response

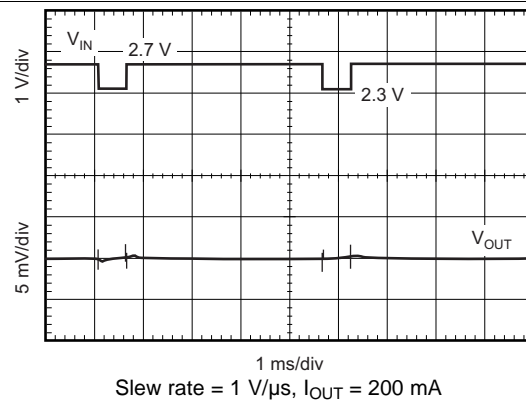


**Typical Characteristics (continued)**

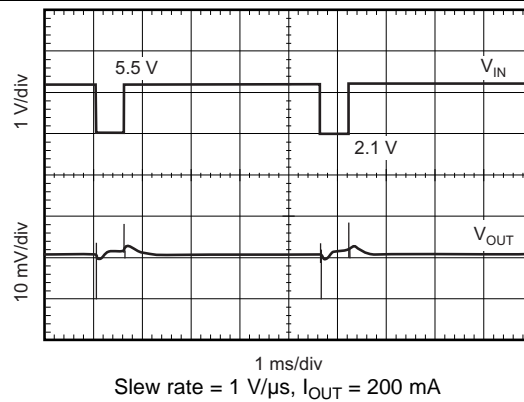
at  $T_J = -40^{\circ}\text{C}$  to  $+125^{\circ}\text{C}$ ,  $V_{IN} = V_{OUT(TYP)} + 0.5\text{ V}$  or  $2\text{ V}$  (whichever is greater),  $I_{OUT} = 10\text{ mA}$ ,  $V_{EN} = V_{IN}$ ,  $C_{IN} = 1\text{ }\mu\text{F}$ , and  $C_{OUT} = 1\text{ }\mu\text{F}$  (unless otherwise noted); typical values are at  $T_J = 25^{\circ}\text{C}$



**Figure 19. Line Transient Response**



**Figure 20. Line Transient Response**



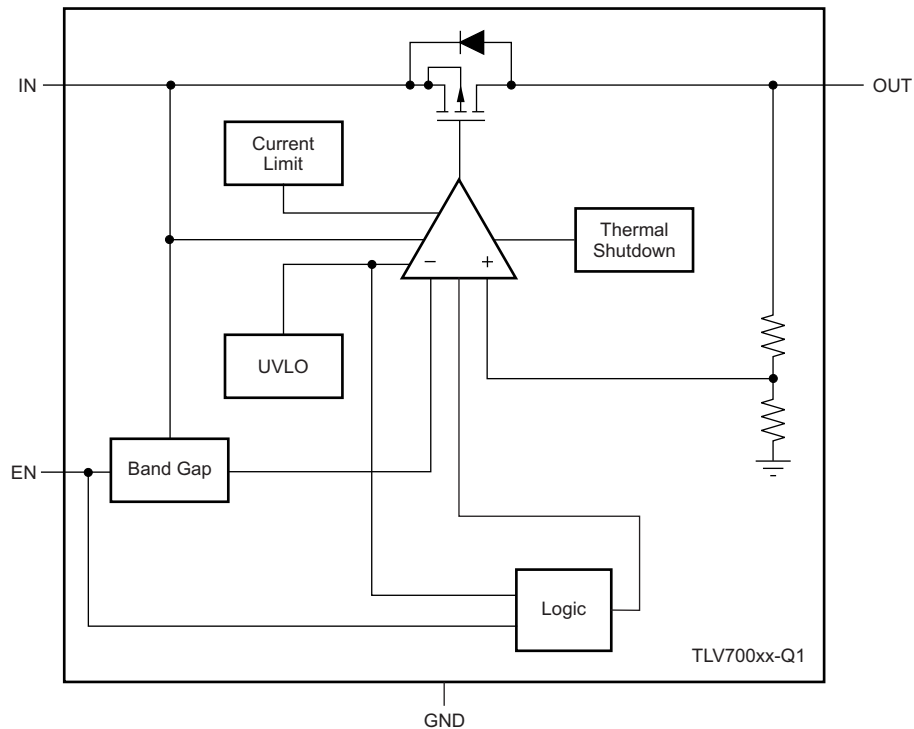
**Figure 21. Line Transient Response**

## 7 Detailed Description

### 7.1 Overview

The TLV700xx-Q1 low-dropout (LDO) linear regulators are low-quiescent-current devices with excellent line and load transient performance. These LDOs are designed for power-sensitive applications. A precision band-gap and error amplifier provides overall 2% accuracy together with low output noise, very high power-supply rejection ratio (PSRR), and low dropout voltage.

### 7.2 Functional Block Diagram



### 7.3 Feature Description

#### 7.3.1 Internal Current Limit

The TLV700xx-Q1 internal current limit helps protect the regulator during fault conditions. During current limit, the output sources a fixed amount of current that is largely independent of the output voltage. In such a case, the output voltage is not regulated, and is  $V_{OUT} = I_{LIMIT} \times R_{LOAD}$ . The PMOS pass transistor dissipates  $(V_{IN} - V_{OUT}) \times I_{LIMIT}$  until thermal shutdown is triggered and the device turns off. When the TLV700xx-Q1 cools down, the device is turned on by the internal thermal-shutdown circuit. If the fault condition continues, the device cycles between current limit and thermal shutdown; see the [Thermal Protection](#) section for more details.

The PMOS pass element in the TLV700xx-Q1 has a built-in body diode that conducts current when the voltage at OUT exceeds the voltage at IN. This current is not limited, so if extended reverse voltage operation is anticipated, external limiting to 5% of the rated output current is recommended.

#### 7.3.2 Shutdown

The enable pin (EN) is active high and is compatible with standard and low-voltage transistor-transistor logic, complementary metal oxide semiconductor (TTL-CMOS) levels. When shutdown capability is not required, EN can be connected to the IN pin.

## Feature Description (continued)

### 7.3.3 Dropout Voltage

The TLV700xx-Q1 uses a PMOS pass transistor to achieve low dropout. When  $(V_{IN} - V_{OUT})$  is less than the dropout voltage ( $V_{DO}$ ), the PMOS pass device is in the linear region of operation and the input-to-output resistance is the  $r_{DS(on)}$  of the PMOS pass element.  $V_{DO}$  scales approximately with output current because the PMOS device behaves as a resistor in dropout.

As with any linear regulator, PSRR and transient response are degraded when  $(V_{IN} - V_{OUT})$  approaches dropout. This effect is illustrated in [Figure 13](#) in the [Typical Characteristics](#) section.

### 7.3.4 Undervoltage Lockout (UVLO)

The TLV700xx-Q1 uses an undervoltage lockout circuit to keep the output shut off until the internal circuitry is operating properly.

## 7.4 Device Functional Modes

### 7.4.1 Operation with $V_{IN}$ Less Than 2 V

The TLV700xx-Q1 family of devices operates with input voltages above 2 V. The typical UVLO voltage is 1.9 V and the device operates at an input voltage above 2 V. When the input voltage falls below the UVLO voltage, the device is shutdown.

### 7.4.2 Operation with $V_{IN}$ Greater Than 2 V

When  $V_{IN}$  is greater than 2 V, if the input voltage is higher than the desired output voltage plus dropout voltage, the output voltage is equal to the desired value. Otherwise, the output voltage is  $V_{IN}$  minus the dropout voltage.

## 8 Application and Implementation

### NOTE

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

### 8.1 Application Information

The TLV700xx-Q1 devices belong to a family of next-generation-value LDO regulators. The devices consume low quiescent current and deliver excellent line and load transient performance. These characteristics, combined with low noise and very good PSRR with little ( $V_{IN} - V_{OUT}$ ) headroom, make this device family ideal for RF portable applications. This family of regulators offers sub-band-gap output voltages down to 0.7 V, current limit, and thermal protection, and is specified from  $-40^{\circ}\text{C}$  to  $+125^{\circ}\text{C}$ .

#### 8.1.1 Input and Output Capacitor Requirements

Ceramic, 1.0- $\mu\text{F}$ , X5R- and X7R-type capacitors are recommended because these capacitors have minimal variation in value and equivalent series resistance (ESR) over temperature.

However, the TLV700xx-Q1 devices are designed to be stable with an effective capacitance of 0.1  $\mu\text{F}$  or larger at the output. Thus, these devices are stable with capacitors of other dielectric types as well, as long as the effective capacitance under operating bias voltage and temperature is greater than 0.1  $\mu\text{F}$ . This effective capacitance refers to the capacitance under the operating bias voltage and temperature conditions; that is, the capacitance after taking both bias voltage and temperature derating into consideration. In addition to allowing the use of cheaper dielectrics, this capability of being stable with 0.1- $\mu\text{F}$  effective capacitances also enables the use of smaller-footprint capacitors that have higher derating in size- and space-constrained applications.

Note that using a 0.1- $\mu\text{F}$  rated capacitor at the output of the LDO does not ensure stability because the effective capacitance under the specified operating conditions is less than 0.1  $\mu\text{F}$ . Maximum ESR must be less than 200 m $\Omega$ .

Although an input capacitor is not required for stability, good analog design practice is to connect a 0.1- $\mu\text{F}$  to 1- $\mu\text{F}$ , low-ESR capacitor across the IN pin and the GND pin of the regulator. This capacitor counteracts reactive input sources and improves transient response, noise rejection, and ripple rejection. A higher-value capacitor may be necessary if large, fast, rise-time load transients are anticipated, or if the device is not located close to the power source. If source impedance is more than 2  $\Omega$ , a 0.1- $\mu\text{F}$  input capacitor may be necessary to ensure stability.

#### 8.1.2 Transient Response

As with any regulator, increasing the size of the output capacitor reduces over- and undershoot magnitude but increases the duration of the transient response.

#### 8.1.3 Thermal Protection

Thermal protection disables the output when the junction temperature rises to approximately  $160^{\circ}\text{C}$ , allowing the device to cool. When the junction temperature cools to approximately  $140^{\circ}\text{C}$ , the output circuitry is again enabled. Depending on power dissipation, thermal resistance, and ambient temperature, the thermal protection circuit can cycle on and off. This cycling limits the dissipation of the regulator, protecting the regulator from damage as a result of overheating.

Any tendency to activate the thermal protection circuit indicates excessive power dissipation or an inadequate heatsink. For reliable operation, limit junction temperature to  $125^{\circ}\text{C}$  (maximum). To estimate the margin of safety in a complete design (including heatsink), increase the ambient temperature until the thermal protection is triggered; use worst-case loads and signal conditions. For good reliability, thermal protection must trigger at least  $35^{\circ}\text{C}$  above the maximum expected ambient condition of the particular application. This configuration produces a worst-case junction temperature of  $125^{\circ}\text{C}$  at the highest-expected ambient temperature and worst-case load.

## Application Information (continued)

The internal protection circuitry of the TLV700xx-Q1 is designed to protect against overload conditions. This circuitry is not intended to replace proper heatsinking. Continuously running the TLV700xx-Q1 into thermal shutdown degrades device reliability.

### 8.2 Typical Application

The TLV700xx-Q1 devices are 200-mA, low quiescent current, low-noise, high-PSRR, fast start-up LDO linear regulators with excellent line and load transient response. The [TLV700xxEVM-503 user's guide](#) (SLUU391) evaluation module (EVM) helps designers evaluate the operation and performance of the TLV700xx-Q1 family.

Figure 22 shows a typical application for the TLV700xx-Q1 device.

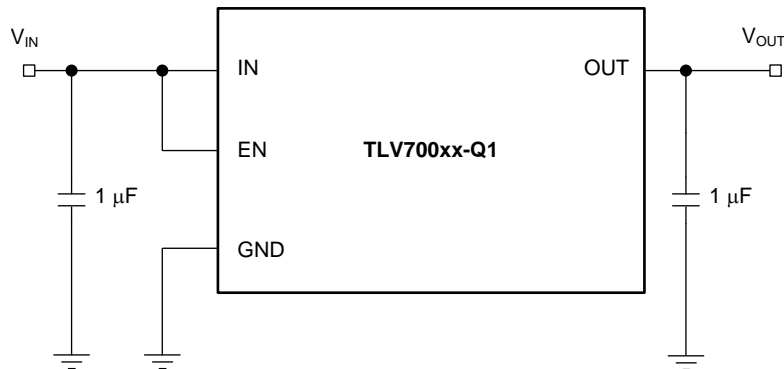


Figure 22. TLV700xx-Q1 Typical Application

#### 8.2.1 Design Requirements

Table 1 shows example design parameters and values for this typical application.

Table 1. Design Parameters

PARAMETER	VALUE
Input voltage range	2 V to 5.5 V
Output voltage	2.2 V, 2.8 V, 3.2 V
Output current rating	200 mA
Effective output capacitor range	> 0.1 µF
Maximum output capacitor ESR range	< 200 mΩ

#### 8.2.2 Detailed Design Procedure

##### 8.2.2.1 Input Capacitance

Although not required for stability, connecting a 0.1-µF to 1-µF low-ESR capacitor across the IN pin and GND pin the regulator is good analog design practice.

##### 8.2.2.2 Output Capacitance

Effect capacitance of 0.1 µF or larger is required to ensure stable operation. The maximum ESR must be less than 200 mΩ.

### 8.2.2.3 Thermal Calculation

Equation 1 shows the thermal calculation.

$$P_D = I_{OUT} \times (V_{IN} - V_{OUT}) + I_Q \times V_{IN}$$

where

- $P_D$  = continuous power dissipation
- $I_{OUT}$  = output current
- $V_{IN}$  = input voltage
- $V_{OUT}$  = output voltage
- Because  $I_Q \ll I_{OUT}$ , the term  $I_Q \times V_{IN}$  is always ignored (1)

For a device under operation at a given ambient air temperature ( $T_A$ ), use Equation 2 to calculate the junction temperature ( $T_J$ ).

$$T_J = T_A + (R_{\theta JA} \times P_D)$$

where

- $Z_{\theta JA}$  = junction-to-ambient air thermal impedance (2)

Use Equation 3 to calculate the rise in junction temperature because of power dissipation.

$$\Delta T = T_J - T_A = (R_{\theta JA} \times P_D)$$
(3)

For a given maximum junction temperature ( $T_{Jmax}$ ), use Equation 4 to calculate the maximum ambient air temperature ( $T_{Amax}$ ) at which the device can operate.

$$T_{Amax} = T_{Jmax} - (R_{\theta JA} \times P_D)$$
(4)

### 8.2.3 Application Curve

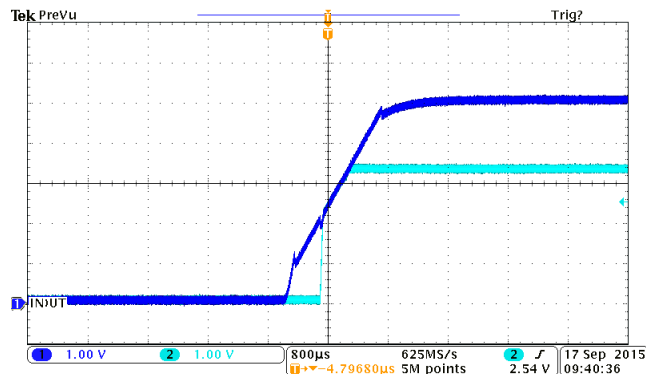


Figure 23. Power-Up

## 9 Power Supply Recommendations

The device is designed to operate from an input-voltage supply range between 2 V and 5.5 V. This input supply must be well regulated. If the input supply is located more than a few inches from the TPS7B69xx-Q1 device, a capacitor with a value of 0.1  $\mu$ F and a ceramic bypass capacitor are recommended to be added at the input.

## 10 Layout

### 10.1 Layout Guidelines

When laying out the board for the TLV700xx-Q1, the board is recommended to be designed with separate ground planes for  $V_{IN}$  and  $V_{OUT}$  that are only connected at the GND pin of the device, as shown in Figure 24. Also, the ground connection for the bypass capacitor must be connected directly to the GND pin of the device. Improve the PSRR performance of the TLV700xx-Q1 by following these layout guidelines.

### 10.2 Board Layout Recommendations to Improve PSRR and Noise Performance

Place input and output capacitors as close to the device pins as possible. To improve ac performance (such as PSRR, output noise, and transient response), the board is recommended to be designed with separate ground planes for  $V_{IN}$  and  $V_{OUT}$ , with the ground plane connected only at the GND pin of the device. In addition, connect the ground connection for the output capacitor directly to the GND pin of the device. High-ESR capacitors can degrade PSRR performance.

### 10.3 Layout Example

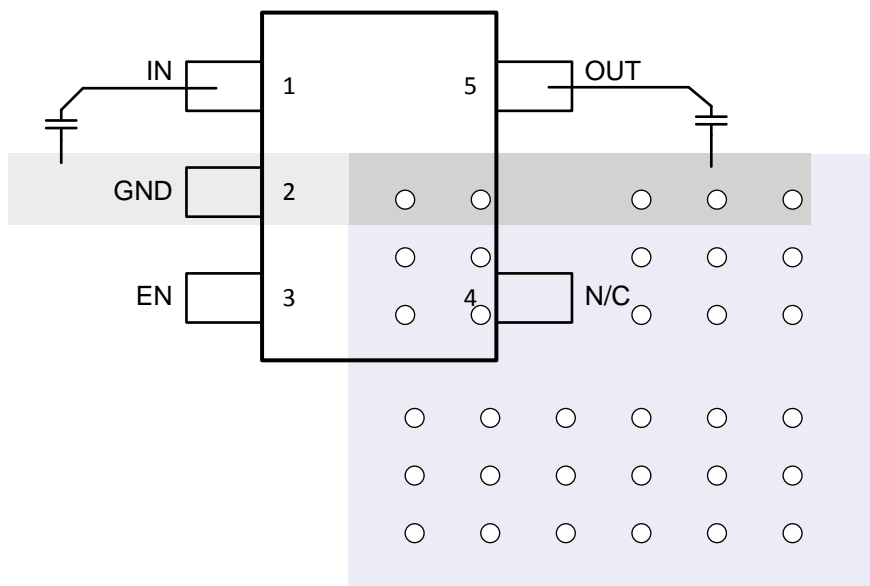


Figure 24. TLV700xx-Q1 Layout Example

## 11 Device and Documentation Support

### 11.1 Documentation Support

#### 11.1.1 Related Documentation

For related documentation see the following:

[Using the TLV700xxEVM-503](#)

#### 11.2 Receiving Notification of Documentation Updates

To receive notification of documentation updates — go to the product folder for your device on ti.com. In the upper right-hand corner, click the *Alert me* button to register and receive a weekly digest of product information that has changed (if any). For change details, check the revision history of any revised document.

#### 11.3 Community Resources

The following links connect to TI community resources. Linked contents are provided "AS IS" by the respective contributors. They do not constitute TI specifications and do not necessarily reflect TI's views; see TI's [Terms of Use](#).

**TI E2E™ Online Community** *TI's Engineer-to-Engineer (E2E) Community*. Created to foster collaboration among engineers. At e2e.ti.com, you can ask questions, share knowledge, explore ideas and help solve problems with fellow engineers.

**Design Support** *TI's Design Support* Quickly find helpful E2E forums along with design support tools and contact information for technical support.

#### 11.4 Trademarks

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All other trademarks are the property of their respective owners.

#### 11.5 Electrostatic Discharge Caution



This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

#### 11.6 Glossary

[SLYZ022](#) — *TI Glossary*.

This glossary lists and explains terms, acronyms, and definitions.

## 12 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.



**PACKAGING INFORMATION**

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead finish/ Ball material (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
TLV70025QDDCRQ1	ACTIVE	SOT-23-THIN	DDC	5	3000	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	QVC	<a href="#">Samples</a>
TLV70028QDDCRQ1	ACTIVE	SOT-23-THIN	DDC	5	3000	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	SJU	<a href="#">Samples</a>
TLV70032QDDCRQ1	ACTIVE	SOT-23-THIN	DDC	5	3000	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	SKA	<a href="#">Samples</a>
TLV70033QDDCRQ1	ACTIVE	SOT-23-THIN	DDC	5	3000	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	OFL	<a href="#">Samples</a>

(1) The marketing status values are defined as follows:

**ACTIVE:** Product device recommended for new designs.

**LIFEBUY:** TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

**NRND:** Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

**PREVIEW:** Device has been announced but is not in production. Samples may or may not be available.

**OBSELETE:** TI has discontinued the production of the device.

(2) **RoHS:** TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

**RoHS Exempt:** TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

**Green:** TI defines "Green" to mean the content of Chlorine (Cl) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

(3) **MSL, Peak Temp.** - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

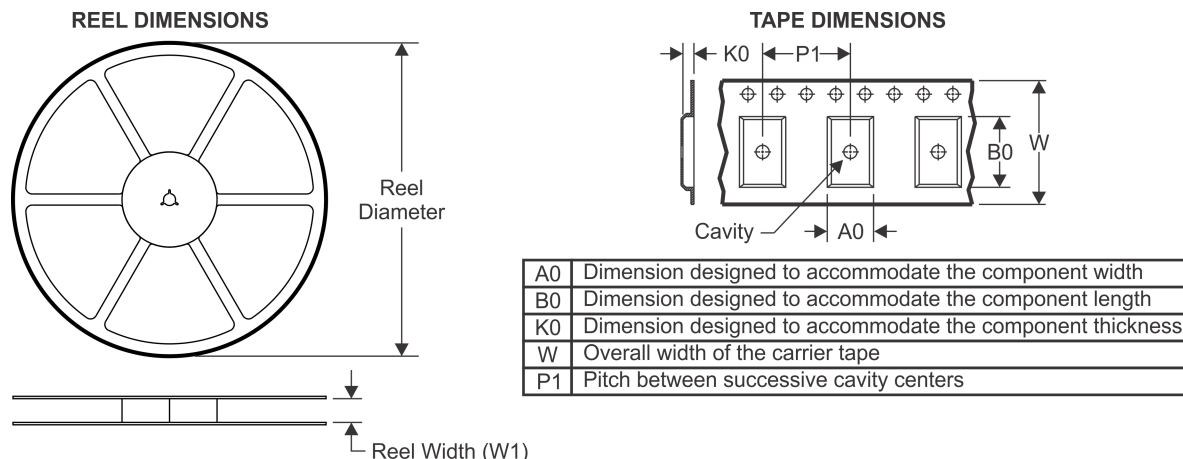
(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "-" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

(6) **Lead finish/Ball material** - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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**TAPE AND REEL INFORMATION**

**QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE**


\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TLV70025QDDCRQ1	SOT-23-THIN	DDC	5	3000	179.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3
TLV70028QDDCRQ1	SOT-23-THIN	DDC	5	3000	180.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3
TLV70032QDDCRQ1	SOT-23-THIN	DDC	5	3000	179.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3
TLV70033QDDCRQ1	SOT-23-THIN	DDC	5	3000	180.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3

**TAPE AND REEL BOX DIMENSIONS**


\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TLV70025QDDCRQ1	SOT-23-THIN	DDC	5	3000	213.0	191.0	35.0
TLV70028QDDCRQ1	SOT-23-THIN	DDC	5	3000	213.0	191.0	35.0
TLV70032QDDCRQ1	SOT-23-THIN	DDC	5	3000	213.0	191.0	35.0
TLV70033QDDCRQ1	SOT-23-THIN	DDC	5	3000	213.0	191.0	35.0

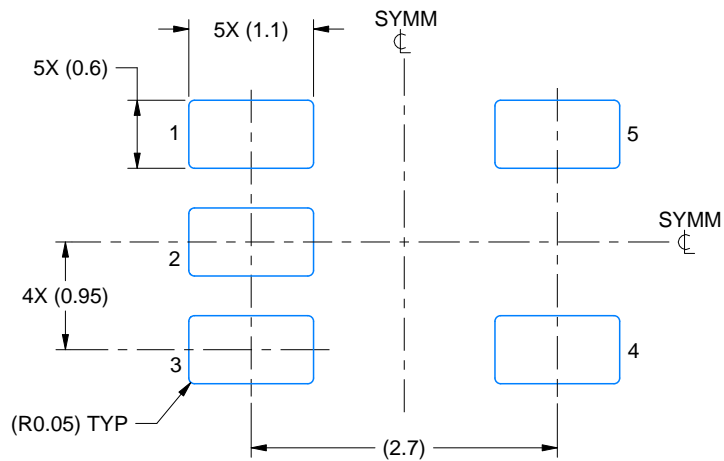


# EXAMPLE BOARD LAYOUT

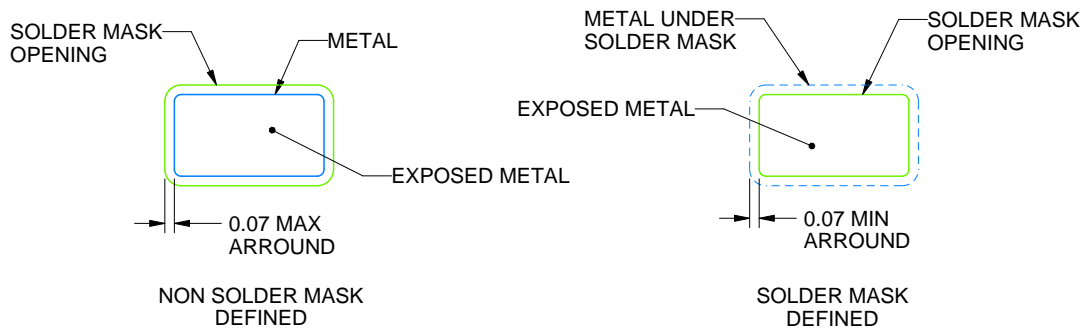
DDC0005A

SOT-23 - 1.1 max height

SMALL OUTLINE TRANSISTOR



LAND PATTERN EXAMPLE  
EXPLODED METAL SHOWN  
SCALE:15X



SOLDERMASK DETAILS

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NOTES: (continued)

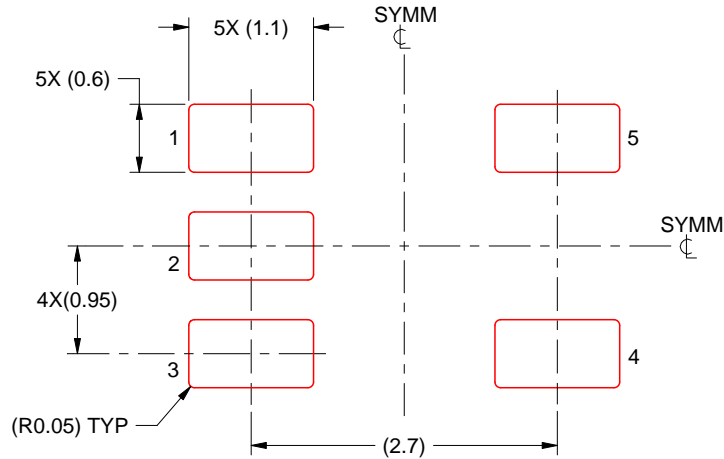
- 4. Publication IPC-7351 may have alternate designs.
- 5. Solder mask tolerances between and around signal pads can vary based on board fabrication site.

# EXAMPLE STENCIL DESIGN

DDC0005A

SOT-23 - 1.1 max height

SMALL OUTLINE TRANSISTOR



SOLDER PASTE EXAMPLE  
BASED ON 0.125 THICK STENCIL  
SCALE:15X

4220752/C 08/2024

NOTES: (continued)

6. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
7. Board assembly site may have different recommendations for stencil design.

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