# Low-Noise and Long-Range PIR Sensor Conditioner Circuit with MSP430™ Smart Analog Combo



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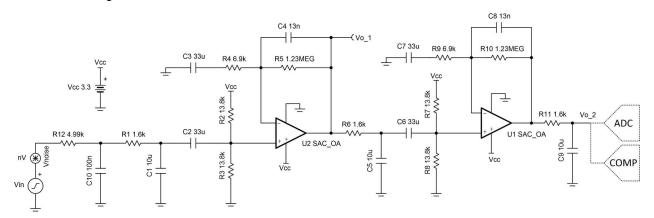
### **Design Goals**

AC Gain	Filter Cut-Off Frequency		Sup	pply
90dB	f <sub>L</sub>	f <sub>H</sub>	V <sub>cc</sub>	V <sub>ee</sub>
	0.7Hz	10Hz	3.3V	0V

#### **Design Description**

Some MSP430™ microcontrollers (MCUs) contain configurable integrated signal chain elements such as opamps, DACs, and programmable gain stages. These elements make up a peripheral called the Smart Analog Combo (SAC). For information on the different types of SACs and how to leverage configurable analog signal chain capabilities, visit MSP430 MCUs Smart Analog Combo Training. To get started with your design, download the Low-Noise Long-Range PIR Sensor Conditioner Circuit Design Files.

This design leverages two of the four integrated op-amp blocks (SACs) in the MSP430FR2355 MCU. Two SAC\_L3 peripherals are configured as cascaded op-amps in general-purpose mode to amplify and filter the signal from a passive infrared (PIR) sensor. The circuit includes multiple low-pass and high-pass filters to reduce noise at the output of the circuit to be able to detect motion at long distances and reduce false triggers. The output of the second-stage op-amp in this circuit can be internally or externally connected to other integrated peripherals in the MSP430FR2355 MCU. For example, the analog-to-digital converter (ADC) window comparator can sample this output periodically (with no CPU intervention) and trigger an interrupt when the signal crosses a threshold, indicating motion or an alert.



# **Design Notes**

- The common-mode voltage and output-bias voltage are set using the resistor dividers between R<sub>2</sub> and R<sub>3</sub> (and R<sub>7</sub> and R<sub>8</sub>).
- Two or more amplifier stages must be used to allow for sufficient loop gain.
- · Additional low-pass and high-pass filters can be added to further reduce noise.
- Capacitors C<sub>4</sub> and C<sub>8</sub> filter noise by decreasing the bandwidth of the circuit and help stabilize the amplifiers.



- RC filters on the output of the amplifiers (for example, R<sub>6</sub> and C<sub>5</sub>) are required to reduce the total integrated noise of the amplifier.
- The maximum gain of the circuit can be affected by the cut-off frequencies of the filters. Adjust the cut-off frequencies to achieve the desired gain.
- For this design, two SAC\_L3 peripherals in the MSP430FR2355 MCU are configured as cascaded op-amps in general-purpose mode.
- This design can also be implemented by using the transimpedance amplifier (TIA) and SAC\_L1 peripheral in the MSP430FR2311 MCU for the cascaded op-amps, but since the maximum input voltage of the TIA is limited to VCC/2, limit the common-mode voltage and gain accordingly.
- The Low-Noise Long-Range PIR Sensor Conditioner Circuit Design Files include a code example demonstrating how to properly configure the SAC\_L3 and ADC window comparator peripherals in the MSP430FR2355 MCU.

#### **Design Steps**

1. Choose large-valued capacitors C<sub>1</sub>, C<sub>5</sub>, and C<sub>9</sub> for the low-pass filters. Select these capacitors first because large-valued capacitors have limited standard values to select from compared to standard resistor values.

$$C_1 = C_5 = C_9 = 10 \mu F$$

2. Calculate resistor values for  $R_1$ ,  $R_6$ , and  $R_{11}$  to form the low-pass filters.

$$\begin{split} &R_1=R_6=R_{11}=\frac{1}{2\pi\times f_H\times C_1}=\frac{1}{2\pi\times 10Hz\times 10\mu F}=1.592k\Omega \\ &\text{Choose} \quad R_1=R_6=R_{11}=1.6k\Omega \quad \Big(\text{Standard value}\Big) \end{split}$$

3. Select capacitor values for C2, C3, C6, and C7 for the high-pass filters.

$$C_2 = C_3 = C_6 = C_7 = 33\mu F$$

4. Calculate the resistor values for  $R_4$  and  $R_9$  for the high-pass filters.

$$\begin{split} R_4 &= R_9 = \frac{1}{2\pi \times f_L \times C_2} = \frac{1}{2\pi \times 0.7 \text{Hz} \times 33 \mu F} = 6.89 \text{k}\Omega \\ \text{Choose} \quad R_4 &= R_9 = 6.9 \text{k}\Omega \quad \Big( \text{Standard value} \Big) \end{split}$$

5. Set the common-mode voltage of the amplifier to mid-supply using a voltage divider. The equivalent resistance of the voltage divider should be equal to R<sub>4</sub> to properly set the corner frequency of the high-pass filter.

$$R_2 = R_3 = R_7 = R_8 = 2 \times R_4 = 2 \times 6.9 k\Omega = 13.8 k\Omega$$
  
Choose  $R_2 = R_3 = R_7 = R_8 = 13.8 k\Omega$  (Standard value)

6. Calculate the gain required by each gain stage to achieve the total gain requirement. Distribute the total gain target of the circuit evenly between both gain stages.

Gain = 
$$\frac{90dB}{2}$$
 = 45dB = 177.828 $\frac{V}{V}$ 

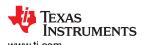
7. Calculate R<sub>5</sub> to set the gain of the first stage.

$$\begin{split} R_5 &= (Gain-1) \times R_4 = \left(177.828\frac{V}{V} - 1\right) \times 6.9 k\Omega = 1.22 M\Omega \\ Choose \ R_5 &= 1.23 M\Omega \ \left(Standard\ value\right) \end{split}$$

8. Calculate C<sub>4</sub> to set the low-pass filter cut-off frequency.

$$\begin{split} &C_4 = \frac{1}{2\pi \times f_H \times R_5} = \frac{1}{2\pi \times 10 Hz \times 1.23 M\Omega} = 12.939 nF \\ &Choose \quad C_4 = 13 nF \quad \Big( Standard \quad value \Big) \end{split}$$

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9. Since the gain and cut-off frequency of the first gain stage is equal to the second gain stage, set all component values of both stages equal to each other.

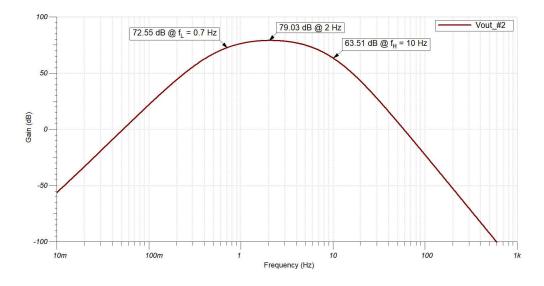
$$\begin{split} R_1 &= R_6 = 1.6 k\Omega \\ R_7 &= R_8 = 13.8 k\Omega \\ R_9 &= R_4 = 6.9 k\Omega \\ R_{10} &= R_5 = 1.23 M\Omega \\ C_8 &= C_4 = 13 nF \end{split}$$

10. Calculate R<sub>11</sub> to set the cut-off frequency of the low-pass filter at the output of the circuit.

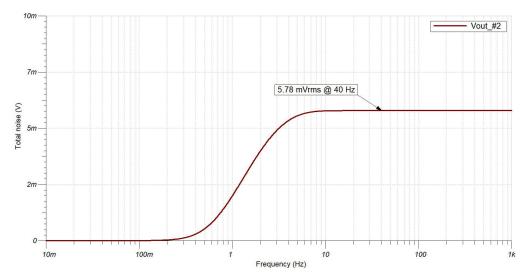
$$\begin{split} R_{11} &= \frac{1}{2\pi \times f_H \times C_9} = \frac{1}{2\pi \times 10 Hz \times 10 \mu F} = 1.592 k\Omega \\ \text{Choose} \quad R_{11} &= 1.6 k\Omega \quad \Big( \text{Standard value} \Big) \end{split}$$

# **Design Simulations**

## **AC Simulation Results**



#### **Noise Simulation Results**



# **Target Applications**

- Motion detector
- Occupancy detection (people tracking and people counting)
- Building automation
- IP network camera
- · Lighting sensor
- Thermostat
- Video doorbell

#### References

- 1. Texas Instruments, Low-Noise Long-Range PIR Sensor Conditioner Circuit, design files
- 2. Texas Instruments, 16MHz integrated analog microcontroller with 3.75-KB FRAM, Op Amp, TIA, comparator with DAC, 10-bit ADC, product page
- 3. Texas Instruments, How to Use the Smart Analog Combo in MSP430<sup>TM</sup> MCUs, application report
- 4. Texas Instruments, MSP430 MCUs Smart Analog Combo, training video

**Design Featured Op Amp** 

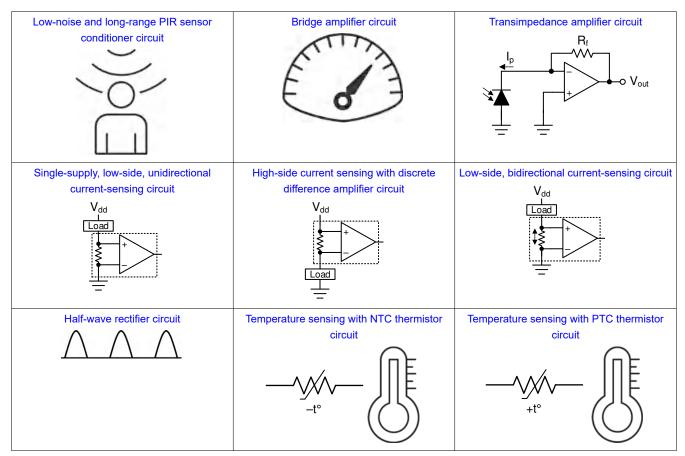
MSP430FRxx Smart Analog Combo					
	MSP430FR2311 SAC_L1	MSP430FR2355 SAC_L3			
V <sub>cc</sub>	2.0V to 3.6V				
V <sub>CM</sub>	-0.1V to V <sub>CC</sub> + 0.1V				
V <sub>out</sub>	Rail-to-rail				
V <sub>os</sub>	±5mV				
A <sub>OL</sub>	100dB				
ı	350μA (high-speed mode)				
Iq	120µA (low-power mode)				
l <sub>b</sub>	50pA				
UGBW	4MHz (high-speed mode)	2.8MHz (high-speed mode)			
OGBW	1.4MHz (low-power mode)	1MHz (low-power mode)			
SR	3V/μs (high-speed mode)				
JK .	1V/μs (low-power mode)				
Number of channels	1	4			
	MSP430FR2311	MSP430FR2355			

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### **Design Alternate Op Amp**

MSP430FR2311 Transimpedance Amplifier				
V <sub>cc</sub>	2.0V to 3.6V			
V <sub>CM</sub>	-0.1V to V <sub>CC</sub> /2V			
V <sub>out</sub>	Rail-to-rail			
V <sub>os</sub>	±5mV			
A <sub>OL</sub>	100dB			
ı	350μA (high-speed mode)			
Iq	120μA (low-power mode)			
L	5pA (TSSOP-16 with OA-dedicated pin input)			
l <sub>b</sub>	50pA (TSSOP-20 and VQFN-16)			
UGBW	5MHz (high-speed mode)			
OGDW	1.8MHz (low-power mode)			
SR	4V/µs (high-speed mode)			
SK .	1V/μs (low-power mode)			
Number of channels	1			
MSP430FR2311				

### **Related MSP430 Circuits**



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Revision History www.ti.com

# **Revision History**

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.	
Changes from Revision A (February 2020) to Revision B (October 2024)	Page
Updated the format for tables, figures, and cross-references throughout the document	1
Changes from Revision * (December 2019) to Revision A (February 2020)	Paga
	Page
Added Related MSP430 Circuits section	1

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