

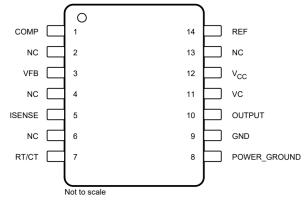
TLx84x Current-Mode PWM Controllers

1 Features

- Optimized for off-line and DC-to-DC converters
- Low start-up current (< 1mA)
- Automatic feed-forward compensation
- Pulse-by-pulse current limiting
- Enhanced load-response, characteristics
- Undervoltage lockout with hysteresis
- Double-pulse suppression
- High-current totem-pole output
- Internally trimmed bandgap reference
- 500kHz operation
- Error amplifier with low output resistance
- Designed to be interchangeable with UC2842 and UC3842 series

2 Applications

- Switching regulators of any polarity
- Transformer-coupled DC/DC convertors



D Package 14-Pin SOIC Top View

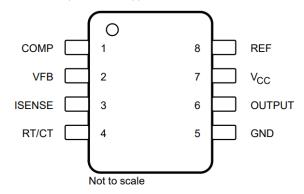
3 Description

The TL284x and TL384x series of control integrated circuits provide the features that are necessary to implement off-line or DC-to-DC fixed-frequency current-mode control schemes, with a minimum number of external components. Some of the internally implemented circuits are an undervoltage lockout (UVLO), featuring a start-up current of less than 1mA, and a precision reference trimmed for accuracy at the error amplifier input. Other internal circuits include logic to ensure latched operation, a pulse-width modulation (PWM) comparator (that also provides current-limit control), and a totem-pole output stage designed to source or sink high-peak current. The output stage, suitable for driving N-channel MOSFETs, is low when it is in the off state.

Package Information

PART NUMBER	PACKAGE ⁽¹⁾	PACKAGE SIZE ⁽²⁾
	D (SOIC, 8)	4.90mm × 6.00mm
TLx84x	D (SOIC, 14)	8.65mm × 6.00mm
	P (PDIP, 8)	9.81mm × 9.43mm

- For all available packages, see the orderable addendum at the end of the data sheet.
- The package size (length × width) is a nominal value and includes pins, where applicable.



NC — No internal connection

D or P Package 8-Pin SOIC or PDIP Top View



Table of Contents

1 Features	6.4 Device Functional Modes
2 Applications	7 Application and Implementation10
3 Description	7.1 Typical Application10
4 Pin Configuration and Functions	7.2 Power Supply Recommendations12
5 Specifications	7.3 Layout
5.1 Absolute Maximum Ratings	8 Device and Documentation Support14
5.2 ESD Ratings	
5.3 Recommended Operating Conditions4	8.2 Support Resources14
5.4 Thermal Information	8.3 Trademarks14
5.5 Electrical Characteristics	8.4 Electrostatic Discharge Caution14
5.6 Typical Characteristics6	8.5 Glossary14
6 Detailed Description	
6.1 Overview	10 Mechanical, Packaging, and Orderable
6.2 Functional Block Diagram	
6.3 Feature Description	}



4 Pin Configuration and Functions

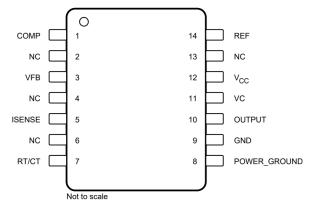
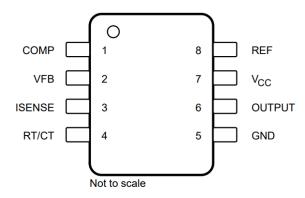


Figure 4-1. D Package 14-Pin SOIC Top View



NC — No internal connection

Figure 4-2. D or P Package 8-Pin SOIC or PDIP Top View

Table 4-1. Pin Functions

	PIN		TYPE ⁽¹⁾	DESCRIPTION						
NAME	D	D or P	IIFE(/	DESCRIPTION						
COMP	1	1	I/O	Error amplifier compensation pin						
GND	9	5	_	Device power supply ground terminal						
ISENSE	5	3	I	Current sense comparator input						
NC	2, 4, 6, 13	_	_	Do not connect						
OUTPUT	10	6	0	PWM Output						
POWER GROUND	8	_	_	Output PWM ground terminal						
REF	14	8	0	Oscillator voltage reference						
RT/CT	7	4	I/O	Oscillator RC input						
VC	11	_	_	Output PWM positive voltage supply						
V _{CC}	12	7	_	Device positive voltage supply						
VFB	3	2	I	Error amplifier input						

⁽¹⁾ I = Input; O = Output; I/O = Input or Output

5 Specifications

5.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted)(1)

		MIN	MAX	UNIT
V _{CC}	Supply Voltage ⁽²⁾	Self lir	niting	_
VI	Analog input voltage range, VFB and ISENSE	-0.3	6.3	V
Vo	Output Voltage		35	V
VI	Input Voltage, VC and D Package only		35	V
I _{CC}	Supply current		30	mA
Io	Output current		±1	Α
	error amplifier output sink current		10	mA
TJ	Virtual junction temperature		150	°C
	Output energy (capacitive load)		5	μJ
T _{stg}	Storage temperature	-65	150	°C

⁽¹⁾ Operation outside the Absolute Maximum Ratings may cause permanent device damage. Absolute Maximum Ratings do not imply functional operation of the device at these or any other conditions beyond those listed under Recommended Operating Conditions. If used outside the Recommended Operating Conditions but within the Absolute Maximum Ratings, the device may not be fully functional, and this may affect device reliability, functionality, performance, and shorten the device lifetime.

5.2 ESD Ratings

			VALUE	UNIT
V	Electrostatic discharge	Human body model (HBM), per ANSI/ESDA/JEDEC JS-001, all pins ⁽¹⁾	±3000	V
V(ESD)	Electrostatic discharge	Charged device model (CDM), per JEDEC specification JESD22-C101, all pins ⁽²⁾	±2000	

⁽¹⁾ JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.

5.3 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)

			MIN	TYP	MAX	UNIT
V _{CC} and VC ⁽¹⁾	Supply Voltage				30	V
V _I , RT/CT	Input Voltage		0		5.5	V
V _I , VFB and ISENSE	Input Voltage		0		5.5	V
V _O , OUTPUT	Output voltage		0		30	V
V _O , POWER GROUND ⁽¹⁾	Output voltage	-0.1		1	V	
I _{CC}	Supply current, externally limited				25	mA
Io	Average output current				200	mA
I _{O(ref)}	Reference output current				-20	mA
fosc	Oscillator frequency		100	500	kHz	
T _A	Operating free-air temperature	TL284x	-40		85	°C
'A	Operating nee-an temperature	TL384x	0		70	U

⁽¹⁾ These recommended voltages for VC and POWER GROUND apply only to the D package.

5.4 Thermal Information

			TLx84x		
	THERMAL METRIC(1)	D (SOIC)	D (SOIC)	P (PDIP)	UNIT
		8 PINS	14 PINS	8 PINS	
$R_{\theta JA}$	Junction-to-ambient thermal resistance	117.4	87.9	74.1	°C/W

For more information about traditional and new thermal metrics, see the Semiconductor and IC Package Thermal Metrics application note.

⁽²⁾ All voltages are with respect to the device GND pin.

⁽²⁾ JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.



5.5 Electrical Characteristics

over operating free-air temperature range, V_{CC} = 15 $V^{(1)}$, R_T = 10 $k\Omega$, C_T = 3.3 nF (unless otherwise noted)

DADAMETED		TEST CONDITIONS ⁽²⁾		TL284x			TL384x			
PARAMETER			MIN	TYP ⁽²⁾	MAX	MIN	TYP ⁽²⁾	MAX	UNIT	
Reference Section										
Output voltage		I _O = 1 mA, T _A = 25°C	4.95	5	5.05	4.9	5	5.1	V	
Line regulation		V _{CC} = 12 V to 25 V		6	20		6	20	mV	
Load regulation		I _O = 1 mA to 20 mA		6	25		6	25	mV	
Temperature coefficient of ou	tput voltage			0.2	0.4		0.2	0.4	mV/°C	
Output voltage with worst-ca	se variation	V _{CC} = 12 V to 25 V, I _O = 1 mA to 20 mA	4.9		5.1	4.82		5.18	V	
Output noise voltage		f = 10 Hz to 10 kHz, T _A = 25°C	-	50			50		μV	
Output-voltage long-term drif	t	After 1000 h at T _A = 25°C		5	25		5	25	mV	
Short-circuit output current			-30	-100	-180	-30	-100	-180	mA	
Oscillator Section										
Oscillator frequency ⁽³⁾		T _A = 25°C	47	52	57	47	52	57	kHz	
Frequency change with supp	ly voltage	V _{CC} = 12 V to 25 V		2	10		2	10	Hz/kHz	
Frequency change with temp	erature			50			50		Hz/kHz	
peak-to-peak amplitude at R	Г/СТ			1.7			1.7		V	
Error-Amplifier Section										
Feedback input voltage		COMP at 2.5 V	2.45	2.50	2.55	2.42	2.50	2.58	V	
Input bias current				-0.3	-1		-0.3	-2	μA	
Open-loop voltage amplificat	ion	V _O = 2 V to 4 V	65	90		65	90		dB	
Gain-bandwidth product			0.7	1		0.7	1		MHz	
Supply-voltage rejection ratio)	V _{CC} = 12 V to 25 V	60	70		60	70		dB	
Output sink current		VFB, at 2.7 V, COMP at 1.1 V	2	6		2	6		mA	
Output source current		VFB, at 2.3 V, COMP at 5 V	-0.5	-0.8		-0.5	-0.8		mA	
Hihg-level output voltage		VFB, at 2.3 V, R_L = 15 k Ω to GND	5	6		5	6		٧	
Low-level output voltage		VFB, at 2.7 V, R_L = 15 k Ω to GND		0.7	1.1		0.7	1.1	٧	
Current-sense Section										
Voltage amplification		See ^{(4) (5)}	2.85	3	3.13	2.85	3	3.15	V/V	
Current-sense comparator th	reshold	COMP at 5 V, see ⁽⁴⁾	0.9	1	1.1	0.9	1	1.1	V	
Supply-voltage rejection ratio)	V _{CC} = 12 V to 25 V, see ⁽⁴⁾		70			70		dB	
Input bias current				-2	-10		-2	-10	μA	
Delay time to output				150	300		150	300	ns	
Output Section										
High-level output voltage		I _{OH} = -20 mA	13	13.5		13	13.5		V	
nigri-level output voltage		I _{OH} = -200 mA	12	13.5		13	13.5		V	
Low-level output voltage		I _{OH} = 20 mA		0.1	0.4		0.1	0.4	V	
Low-level output voltage		I _{OH} = 200 mA		1.5	2.2		1.5	2.2	V	
Rise time		C _L = 1 nF, T _A = 25°C		25	150		25	150	ns	
fall time		C _L = 1 nF, T _A = 25°C		25	150		25	150	ns	
Undervoltage-Lockout Section										
Start threshold voltage	TLx842, TLx844		15	16	17	14.5	16	17.5	V	
	TLx843, TLx845		7.8	8.4	9	7.8	8.4	9	v	
Minimum operating voltage	TLx842, TLx844		9	10	11	8.5	10	11.5	V	
after startup	TLx843, TLx845	¬	7	7.6	8.2	7	7.6	8.02	٧	



5.5 Electrical Characteristics (continued)

over operating free-air temperature range, V_{CC} = 15 $V^{(1)}$, R_T = 10 $k\Omega$, C_T = 3.3 nF (unless otherwise noted)

PARAMETER		TEST CONDITIONS ⁽²⁾	TL284x				UNIT		
		TEST CONDITIONS	MIN	TYP ⁽²⁾	MAX	MIN	TYP ⁽²⁾	MAX	UNII
Maximum duty cycle	TLx842, TLx843		92%	97%	100%	92%	97%	100%	
Maximum duty cycle	TLx844, TLx845		46%	48%	50%	46%	48%	50%	
Minimum duty cycle					0%			0%	
Supply Voltage									
Start-up current			0.5	1		0.5	1	mA	
Operating supply current	VFB and ISENSE at 0 V		11	17		11	17	mA	
Limiting voltage		I _{CC} = 25 mA		39			39		V

- (1) Adjust V_{CC} above the start threshold before setting it to 15 V.
- (2) All typical values are at TA = 25°C.
- (3) Output frequency equals oscillator frequency for the TLx842 and TLx843. Output frequency is one-half the oscillator frequency for the TLx844 and TLx845.
- (4) These parameters are measured at the trip point of the latch, with VFB at 0 V.
- (5) Voltage amplification is measured between ISENSE and COMP, with the input changing from 0 V to 0.8 V.

5.6 Typical Characteristics

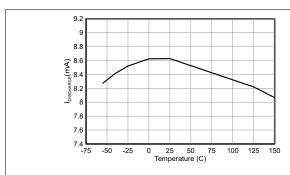


Figure 5-1. Oscillator Discharge Current vs Temperature for V_{IN} = 15 V and V_{OSC} = 2V

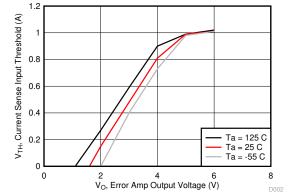


Figure 5-2. Current Sense Input Threshold vs Error Amplifier Output Voltage for V_{IN} = 15 V

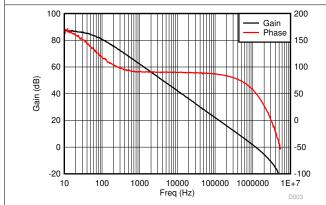


Figure 5-3. Error Amplifier Open-Loop Gain and Phase vs Frequency V_{CC} = 15 V, R_L = 100 k Ω , and T_A = 25 °C

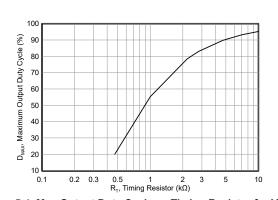
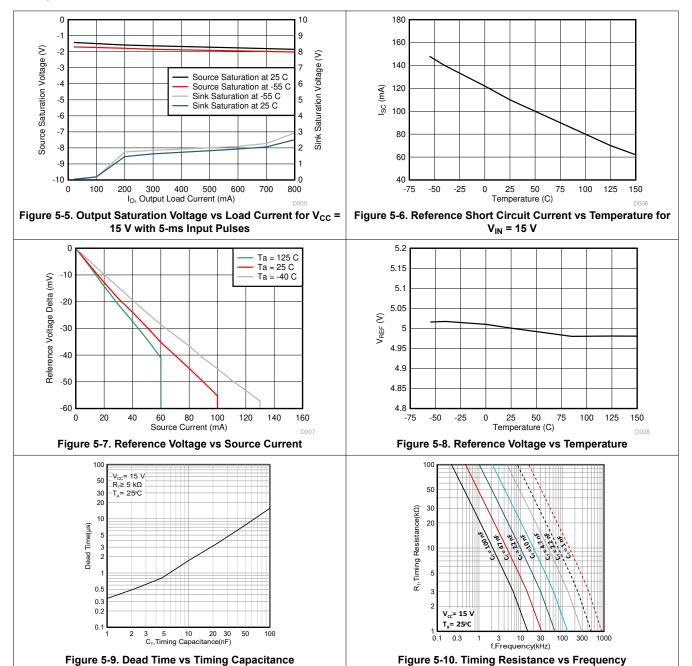


Figure 5-4. Max Output Duty Cycle vs Timing Resistor for V_{CC} = 15, C_T = 3.3 nF, T_A = 25 °C



5.6 Typical Characteristics (continued)





6 Detailed Description

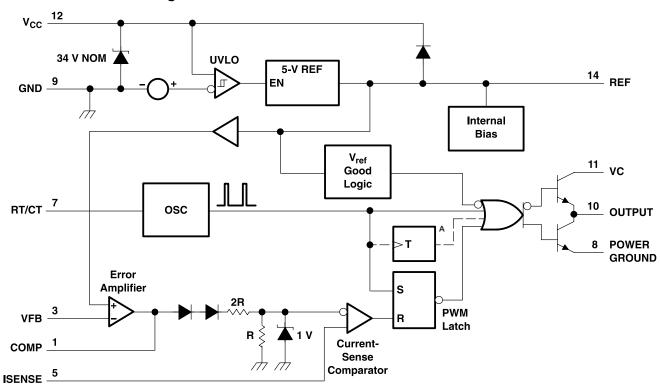
6.1 Overview

The TL284x and TL384x series of control integrated circuits provide the features that are necessary to implement off-line or DC-to-DC fixed-frequency current-mode control schemes, with a minimum number of external components. Some of the internally implemented circuits are an undervoltage lockout (UVLO), featuring a start-up current of less than 1 mA, and a precision reference trimmed for accuracy at the error amplifier input. Other internal circuits include logic to ensure latched operation, a pulse-width modulation (PWM) comparator (that also provides current-limit control), and a totem-pole output stage designed to source or sink high-peak current. The output stage, suitable for driving N-channel MOSFETs, is low when it is in the off state.

Major differences between members of these series are the UVLO thresholds and maximum duty-cycle ranges. Typical UVLO thresholds of 16 V (on) and 10 V (off) on the TLx842 and TLx844 devices make them ideally suited to off-line applications. The corresponding typical thresholds for the TLx843 and TLx845 devices are 8.4 V (on) and 7.6 V (off). The TLx842 and TLx843 devices can operate to duty cycles approaching 100%. A duty-cycle range of 0 to 50% is obtained by the TLx844 and TLx845 by the addition of an internal toggle flip-flop, which blanks the output off every other clock cycle.

The TL284x-series devices are characterized for operation from −40°C to +85°C. The TL384x devices are characterized for operation from 0°C to 70°C.

6.2 Functional Block Diagram



A. The toggle flip-flop is present only in TL2844, TL2845, TL3844, and TL3845. Pin numbers shown are for the D (14-pin) package.

6.3 Feature Description

6.3.1 Pulse-by-Pulse Current Limiting

Pulse-by-pulse limiting is inherent in the control scheme. An upper limit on the peak current can be established by simply clamping the error voltage. Accurate current limiting allows optimization of magnetic and power semiconductor elements while ensuring reliable supply operation



6.3.2 Error Amplifier With Low Output Resistance

With a low output resistance, various impedance networks may be used on the compensation pin input for error amplifier feedback.

6.3.3 High-Current Totem-Pole Output

The output of the TLx84x devices can sink or source up to 1 A of current.

6.4 Device Functional Modes

6.4.1 Shutdown Technique

The PWM controller (see Figure 6-1) can be shut down by two methods: either raise the voltage at ISENSE above 1 V or pull the COMP terminal below a voltage two diode drops above ground. Either method causes the output of the PWM comparator to be high (see *Functional Block Diagram*). The PWM latch is reset dominant so that the output remains low until the next clock cycle after the shutdown condition at the COMP or ISENSE terminal is removed. In one example, an externally latched shutdown can be accomplished by adding an SCR that resets by cycling VCC below the lower UVLO threshold. At this point, the reference turns off, allowing the SCR to reset.

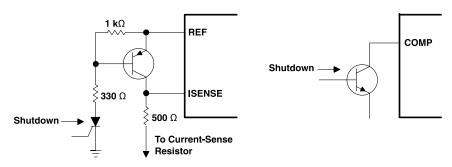


Figure 6-1. Shutdown Techniques

6.4.2 Slope Compensation

A fraction of the oscillator ramp can be summed resistively with the current-sense signal to provide slope compensation for converters requiring duty cycles over 50% (see Figure 6-2).

Note

Capacitor C forms a filter with R2 to suppress the leading-edge switch spikes.

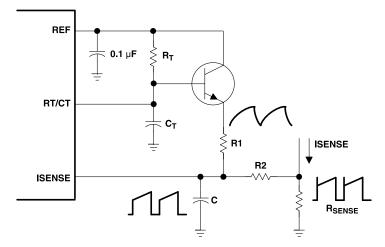


Figure 6-2. Slope Compensation



7 Application and Implementation

Note

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes, as well as validating and testing their design implementation to confirm system functionality.

7.1 Typical Application

The following application is an open-loop laboratory test fixture. This circuit demonstrates the setup and use of the TL284x and TL384x devices and their internal circuitry.

In the open-loop laboratory test fixture (see Figure 7-1), high peak currents associated with loads necessitate careful grounding techniques. Timing and bypass capacitors should be connected close to the GND terminal in a single-point ground. The transistor and $5-k\Omega$ potentiometer sample the oscillator waveform and apply an adjustable ramp to the ISENSE terminal.

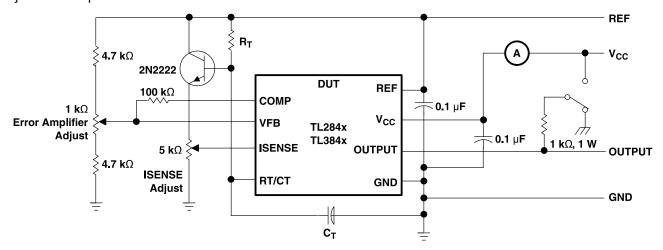


Figure 7-1. Open-Loop Laboratory Test Fixture

7.1.1 Design Requirements

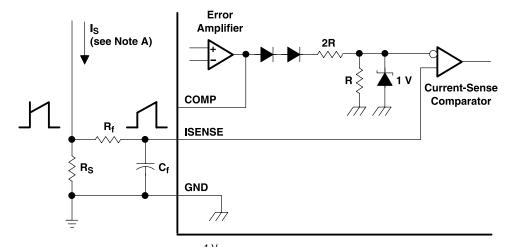
The design techniques in the following sections may be used for power supply PWM applications which fall within the following requirements.

- 500-kHz or lower operation
- · 30-V or less output voltage
- 200-mA or less output current



7.1.2 Detailed Design Procedure

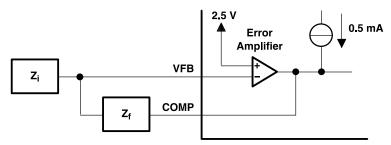
7.1.2.1 Current-Sense Circuit



A. Peak current (IS) is determined by the formula: $I_{S(max)} = \frac{\dot{r}}{R_s}$ A small RC filter formed by resistor Rf and capacitor Cf may be required to suppress switch transients.

Figure 7-2. Current-Sense Circuit Schematic

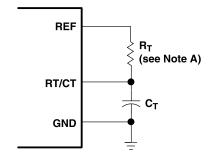
7.1.2.2 Error-Amplifier Configuration



A. Error amplifier can source or sink up to 0.5 mA.

Figure 7-3. Error-Amplifier Configuration Schematic

7.1.2.3 Oscillator Section



 $A. \quad \text{ For } R_T > 5 \text{ k}\Omega \text{: } f \approx \frac{1.72}{R_{\scriptscriptstyle T} C_{\scriptscriptstyle T}}$

Figure 7-4. Oscillator Section Schematic



7.1.3 Application Curve

The application curve shows oscillator characteristics for chosen capacitor and resistor values.

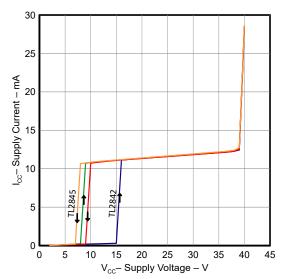


Figure 7-5. Supply Current vs Supply Voltage

7.2 Power Supply Recommendations

See *Recommended Operating Conditions* for the recommended power supply voltages for the TL284x and TL384x devices. TI also recommends to have a decoupling capacitor on the output of the device's power supply to limit noise on the device input.

7.3 Layout

7.3.1 Layout Guidelines

Always try to use a low EMI inductor with a ferrite type closed core. Some examples would be toroid and encased E core inductors. Open core can be used if they have low EMI characteristics and are located a bit more away from the low power traces and components. Make the poles perpendicular to the PCB as well if using an open core. Stick cores usually emit the most unwanted noise.

7.3.1.1 Feedback Traces

Try to run the feedback trace as far from the inductor and noisy power traces as possible. Also, keep the feedback trace to be as direct as possible and somewhat thick. These two sometimes involve a trade-off, but keeping it away from inductor EMI and other noise sources is the more critical of the two. Run the feedback trace on the side of the PCB opposite of the inductor with a ground plane separating the two.

7.3.1.2 Input/Output Capacitors

When using a low value ceramic input filter capacitor, it should be located as close to the V_{CC} pin of the IC as possible. This will eliminate as much trace inductance effects as possible and give the internal IC rail a cleaner voltage supply. Some designs require the use of a feed-forward capacitor connected from the output to the feedback pin as well, usually for stability reasons. In this case it should also be positioned as close to the IC as possible. Using surface mount capacitors also reduces lead length and lessens the chance of noise coupling into the effective antenna created by through-hole components.

7.3.1.3 Compensation Components

External compensation components for stability should also be placed close to the IC. Surface mount components are recommended here as well for the same reasons discussed for the filter capacitors. These should not be located very close to the inductor either.



7.3.1.4 Traces and Ground Planes

Make all of the power (high current) traces as short, direct, and thick as possible. It is good practice on a standard PCB board to make the traces an absolute minimum of 15 mils (0.381 mm) per Ampere. The inductor, output capacitors, and output diode should be as close to each other possible. This helps reduce the EMI radiated by the power traces due to the high switching currents through them. This will also reduce lead inductance and resistance as well, which in turn reduces noise spikes, ringing, and resistive losses that produce voltage errors.

The grounds of the IC, input capacitors, output capacitors, and output diode (if applicable) should be connected close together directly to a ground plane. It would also be a good idea to have a ground plane on both sides of the PCB. This will reduce noise as well by reducing ground loop errors as well as by absorbing more of the EMI radiated by the inductor. For multi-layer boards with more than two layers, a ground plane can be used to separate the power plane (where the power traces and components are) and the signal plane (where the feedback and compensation and components are) for improved performance. On multi-layer boards the use of vias will be required to connect traces and different planes. It is good practice to use one standard via per 200 mA of current if the trace will need to conduct a significant amount of current from one plane to the other.

Arrange the components so that the switching current loops curl in the same direction. Due to the way switching regulators operate, there are two power states. One state when the switch is on and one when the switch is off. During each state there will be a current loop made by the power components that are currently conducting. Place the power components so that during each of the two states the current loop is conducting in the same direction. This prevents magnetic field reversal caused by the traces between the two half-cycles and reduces radiated EMI.

7.3.2 Layout Example

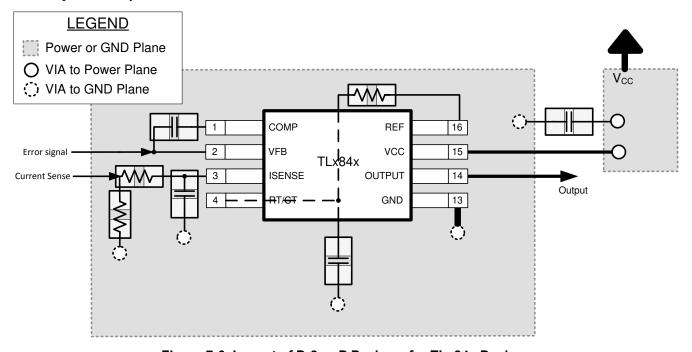


Figure 7-6. Layout of D-8 or P Package for TLx84x Devices



8 Device and Documentation Support

8.1 Receiving Notification of Documentation Updates

To receive notification of documentation updates, navigate to the device product folder on ti.com. Click on *Notifications* to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

8.2 Support Resources

TI E2E[™] support forums are an engineer's go-to source for fast, verified answers and design help — straight from the experts. Search existing answers or ask your own question to get the quick design help you need.

Linked content is provided "AS IS" by the respective contributors. They do not constitute TI specifications and do not necessarily reflect TI's views; see TI's Terms of Use.

8.3 Trademarks

TI E2E[™] is a trademark of Texas Instruments.

All trademarks are the property of their respective owners.

8.4 Electrostatic Discharge Caution



This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

8.5 Glossary

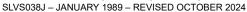
TI Glossary

This glossary lists and explains terms, acronyms, and definitions.

9 Revision History

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

C	hanges from Revision I (July 2016) to Revision J (October 2024)	Page
•	Updated the numbering format for tables, figures, and cross-references throughout the document	4 5 cal
•	Changed the TOTAL STANDBY CURRENT, VCC Zener voltage, typical value from 34V to 39V in the Electrical Characteristics section	Ct, Rt
С	hanges from Revision H (January 2015) to Revision I (July 2016)	Page
•	Updated pinout imagesChanged TL984x to TL384x in <i>Recommended Operating Conditions</i> Changed TLx842, TLx844 to TLx842, TLx843 and TLx843, TLx845 to TLx844, TLx845 in <i>Pulse-Width</i> -	





Changes from Revision G	(February 2008) to Revision H	(January 2015)

Page

Added Applications, Device Information table, Pin Functions table, ESD Ratings table, Thermal Information table, Typical Characteristics, Feature Description section, Device Functional Modes, Application and Implementation section, Power Supply Recommendations section, Layout section, Device and

10 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.



www.ti.com 27-Sep-2024

PACKAGING INFORMATION

Orderable Device	Status	Package Type	Package Drawing	Pins	Package Qty	Eco Plan	Lead finish/ Ball material	MSL Peak Temp	Op Temp (°C)	Device Marking (4/5)	Samples
TL2842D	OBSOLETE	SOIC	D	14		TBD	Call TI	Call TI	-40 to 85	TL2842	
TL2842D-8	OBSOLETE	SOIC	D	8		TBD	Call TI	Call TI	-40 to 85	TL2842	
TL2842DR	ACTIVE	SOIC	D	14	2500	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 85	TL2842	Samples
TL2842DR-8	ACTIVE	SOIC	D	8	2500	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 85	TL2842	Samples
TL2842P	ACTIVE	PDIP	Р	8	50	RoHS & Green	NIPDAU	N / A for Pkg Type	-40 to 85	TL2842P	Samples
TL2843D-8	OBSOLETE	SOIC	D	8		TBD	Call TI	Call TI	-40 to 85	TL2843	
TL2843DR	ACTIVE	SOIC	D	14	2500	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 85	TL2843	Samples
TL2843DR-8	ACTIVE	SOIC	D	8	2500	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 85	TL2843	Samples
TL2843DRE4	ACTIVE	SOIC	D	14	2500	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 85	TL2843	Samples
TL2843DRG4-8	ACTIVE	SOIC	D	8	2500	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 85	TL2843	Samples
TL2843P	ACTIVE	PDIP	Р	8	50	RoHS & Green	NIPDAU	N / A for Pkg Type	-40 to 85	TL2843P	Samples
TL2844D	OBSOLETE	SOIC	D	14		TBD	Call TI	Call TI	-40 to 85	TL2844	
TL2844D-8	OBSOLETE	SOIC	D	8		TBD	Call TI	Call TI	-40 to 85	TL2844	
TL2844DR	ACTIVE	SOIC	D	14	2500	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 85	TL2844	Samples
TL2844DR-8	ACTIVE	SOIC	D	8	2500	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 85	TL2844	Samples
TL2844DRG4	ACTIVE	SOIC	D	14	2500	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 85	TL2844	Samples
TL2844P	ACTIVE	PDIP	Р	8	50	RoHS & Green	NIPDAU	N / A for Pkg Type	-40 to 85	TL2844P	Samples
TL2844PE4	ACTIVE	PDIP	Р	8	50	RoHS & Green	NIPDAU	N / A for Pkg Type	-40 to 85	TL2844P	Samples
TL2845D	OBSOLETE	SOIC	D	14		TBD	Call TI	Call TI	-40 to 85	TL2845	
TL2845D-8	OBSOLETE	SOIC	D	8		TBD	Call TI	Call TI	-40 to 85	TL2845	
TL2845DR	ACTIVE	SOIC	D	14	2500	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 85	TL2845	Samples
TL2845DR-8	ACTIVE	SOIC	D	8	2500	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 85	TL2845	Samples





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27-Sep-2024

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan	Lead finish/ Ball material	MSL Peak Temp	Op Temp (°C)	Device Marking (4/5)	Samples
TL2845DRG4	ACTIVE	SOIC	D	14	2500	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 85	TL2845	Samples
TL2845P	ACTIVE	PDIP	Р	8	50	RoHS & Green	NIPDAU	N / A for Pkg Type	-40 to 85	TL2845P	Samples
TL3842D-8	OBSOLETE	SOIC	D	8		TBD	Call TI	Call TI	0 to 70	TL3842	
TL3842DR	ACTIVE	SOIC	D	14	2500	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	0 to 70	TL3842	Samples
TL3842DR-8	ACTIVE	SOIC	D	8	2500	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	0 to 70	TL3842	Samples
TL3842DRE4-8	ACTIVE	SOIC	D	8	2500	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	0 to 70	TL3842	Samples
TL3842P	ACTIVE	PDIP	Р	8	50	RoHS & Green	NIPDAU	N / A for Pkg Type	0 to 70	TL3842P	Samples
TL3842PE4	ACTIVE	PDIP	Р	8	50	RoHS & Green	NIPDAU	N / A for Pkg Type	0 to 70	TL3842P	Samples
TL3843D	OBSOLETE	SOIC	D	14		TBD	Call TI	Call TI	0 to 70	TL3843	
TL3843D-8	OBSOLETE	SOIC	D	8		TBD	Call TI	Call TI	0 to 70	TL3843	
TL3843DR	ACTIVE	SOIC	D	14	2500	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	0 to 70	TL3843	Samples
TL3843DR-8	ACTIVE	SOIC	D	8	2500	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	0 to 70	TL3843	Samples
TL3843P	ACTIVE	PDIP	Р	8	50	RoHS & Green	NIPDAU	N / A for Pkg Type	0 to 70	TL3843P	Samples
TL3844D	OBSOLETE	SOIC	D	14		TBD	Call TI	Call TI	0 to 70	TL3844	
TL3844D-8	OBSOLETE	SOIC	D	8		TBD	Call TI	Call TI	0 to 70	TL3844	
TL3844DR	ACTIVE	SOIC	D	14	2500	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	0 to 70	TL3844	Samples
TL3844DR-8	ACTIVE	SOIC	D	8	2500	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	0 to 70	TL3844	Samples
TL3844P	ACTIVE	PDIP	Р	8	50	RoHS & Green	NIPDAU	N / A for Pkg Type	0 to 70	TL3844P	Samples
TL3844PE4	ACTIVE	PDIP	Р	8	50	RoHS & Green	NIPDAU	N / A for Pkg Type	0 to 70	TL3844P	Samples
TL3845D	ACTIVE	SOIC	D	14	50	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	0 to 70	TL3845	Samples
TL3845D-8	OBSOLETE	SOIC	D	8		TBD	Call TI	Call TI	0 to 70	TL3845	
TL3845DR	ACTIVE	SOIC	D	14	2500	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	0 to 70	TL3845	Samples
TL3845DR-8	ACTIVE	SOIC	D	8	2500	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	0 to 70	TL3845	Samples

www.ti.com 27-Sep-2024

Orderable Device	Status	Package Type	Package Drawing	Pins	Package Qty	Eco Plan	Lead finish/ Ball material	MSL Peak Temp	Op Temp (°C)	Device Marking (4/5)	Samples
TL3845P	ACTIVE	PDIP	Р	8	50	RoHS & Green	NIPDAU	N / A for Pkg Type	0 to 70	TL3845P	Samples
TL3845PE4	ACTIVE	PDIP	Р	8	50	RoHS & Green	NIPDAU	N / A for Pkg Type	0 to 70	TL3845P	Samples

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

(2) RoHS: TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

Green: TI defines "Green" to mean the content of Chlorine (CI) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

- (3) MSL, Peak Temp. The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.
- (4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.
- (5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.
- (6) Lead finish/Ball material Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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www.ti.com 25-Sep-2024

TAPE AND REEL INFORMATION



A0	Dimension designed to accommodate the component width
В0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TL2842DR	SOIC	D	14	2500	330.0	16.4	6.5	9.0	2.1	8.0	16.0	Q1
TL2842DR-8	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
TL2843DR	SOIC	D	14	2500	330.0	16.4	6.5	9.0	2.1	8.0	16.0	Q1
TL2843DR-8	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
TL2844DR	SOIC	D	14	2500	330.0	16.4	6.5	9.0	2.1	8.0	16.0	Q1
TL2844DR-8	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
TL2845DR	SOIC	D	14	2500	330.0	16.4	6.5	9.0	2.1	8.0	16.0	Q1
TL2845DR-8	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
TL3842DR	SOIC	D	14	2500	330.0	16.4	6.5	9.0	2.1	8.0	16.0	Q1
TL3842DR-8	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
TL3843DR	SOIC	D	14	2500	330.0	16.4	6.5	9.0	2.1	8.0	16.0	Q1
TL3843DR-8	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
TL3844DR	SOIC	D	14	2500	330.0	16.4	6.5	9.0	2.1	8.0	16.0	Q1
TL3844DR-8	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
TL3845DR	SOIC	D	14	2500	330.0	16.4	6.5	9.0	2.1	8.0	16.0	Q1
TL3845DR-8	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1



www.ti.com 25-Sep-2024



*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TL2842DR	SOIC	D	14	2500	356.0	356.0	35.0
TL2842DR-8	SOIC	D	8	2500	353.0	353.0	32.0
TL2843DR	SOIC	D	14	2500	356.0	356.0	35.0
TL2843DR-8	SOIC	D	8	2500	353.0	353.0	32.0
TL2844DR	SOIC	D	14	2500	353.0	353.0	32.0
TL2844DR-8	SOIC	D	8	2500	353.0	353.0	32.0
TL2845DR	SOIC	D	14	2500	356.0	356.0	35.0
TL2845DR-8	SOIC	D	8	2500	353.0	353.0	32.0
TL3842DR	SOIC	D	14	2500	356.0	356.0	35.0
TL3842DR-8	SOIC	D	8	2500	353.0	353.0	32.0
TL3843DR	SOIC	D	14	2500	356.0	356.0	35.0
TL3843DR-8	SOIC	D	8	2500	353.0	353.0	32.0
TL3844DR	SOIC	D	14	2500	353.0	353.0	32.0
TL3844DR-8	SOIC	D	8	2500	353.0	353.0	32.0
TL3845DR	SOIC	D	14	2500	356.0	356.0	35.0
TL3845DR-8	SOIC	D	8	2500	353.0	353.0	32.0



www.ti.com 25-Sep-2024

TUBE



*All dimensions are nominal

Device	Package Name	Package Type	Pins	SPQ	L (mm)	W (mm)	T (µm)	B (mm)
TL2842P	Р	PDIP	8	50	506	13.97	11230	4.32
TL2843P	Р	PDIP	8	50	506	13.97	11230	4.32
TL2844P	Р	PDIP	8	50	506	13.97	11230	4.32
TL2844PE4	Р	PDIP	8	50	506	13.97	11230	4.32
TL2845P	Р	PDIP	8	50	506	13.97	11230	4.32
TL3842P	Р	PDIP	8	50	506	13.97	11230	4.32
TL3842P	Р	PDIP	8	50	506	13.97	11230	4.32
TL3842PE4	Р	PDIP	8	50	506	13.97	11230	4.32
TL3842PE4	Р	PDIP	8	50	506	13.97	11230	4.32
TL3843P	Р	PDIP	8	50	506	13.97	11230	4.32
TL3843P	Р	PDIP	8	50	506	13.97	11230	4.32
TL3844P	Р	PDIP	8	50	506	13.97	11230	4.32
TL3844P	Р	PDIP	8	50	506	13.97	11230	4.32
TL3844PE4	Р	PDIP	8	50	506	13.97	11230	4.32
TL3844PE4	Р	PDIP	8	50	506	13.97	11230	4.32
TL3845D	D	SOIC	14	50	506.6	8	3940	4.32
TL3845P	Р	PDIP	8	50	506	13.97	11230	4.32
TL3845P	Р	PDIP	8	50	506	13.97	11230	4.32
TL3845PE4	Р	PDIP	8	50	506	13.97	11230	4.32
TL3845PE4	Р	PDIP	8	50	506	13.97	11230	4.32





NOTES:

- 1. All linear dimensions are in millimeters. Dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.

 2. This drawing is subject to change without notice.

 3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not
- exceed 0.15 mm, per side.
- 4. This dimension does not include interlead flash. Interlead flash shall not exceed 0.43 mm, per side.
- 5. Reference JEDEC registration MS-012, variation AB.





NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.

7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.





NOTES: (continued)

- 8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
- 9. Board assembly site may have different recommendations for stencil design.







NOTES:

- 1. Linear dimensions are in inches [millimeters]. Dimensions in parenthesis are for reference only. Controlling dimensions are in inches. Dimensioning and tolerancing per ASME Y14.5M.
- 2. This drawing is subject to change without notice.
- 3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed .006 [0.15] per side.
- 4. This dimension does not include interlead flash.
- 5. Reference JEDEC registration MS-012, variation AA.





NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.

7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.





NOTES: (continued)

- 8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
- 9. Board assembly site may have different recommendations for stencil design.



P (R-PDIP-T8)

PLASTIC DUAL-IN-LINE PACKAGE



NOTES:

- A. All linear dimensions are in inches (millimeters).
- B. This drawing is subject to change without notice.
- C. Falls within JEDEC MS-001 variation BA.



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