

AMC1351 高精度、5V 入力、強化絶縁型アンプ

1 特長

- リニア入力電圧範囲: $-0.25\text{V} \sim 5\text{V}$
- 高い入力インピーダンス: $1.25\text{M}\Omega$ (標準値)
- 固定ゲイン: 0.4V/V
- 小さい DC 誤差:
 - オフセット誤差: $\pm 1.5\text{mV}$ (最大値)
 - オフセット・ドリフト: $\pm 15\mu\text{V}/^\circ\text{C}$ (最大値)
 - ゲイン誤差: $\pm 0.2\%$ (最大値)
 - ゲイン・ドリフト: $\pm 35\text{ppm}/^\circ\text{C}$ (最大値)
 - 非直線性: $\pm 0.02\%$ (最大値)
- ハイサイド、ローサイドでの動作: 3.3V または 5V
- 高い CMTI: $100\text{kv}/\mu\text{s}$ (最小値)
- フェイルセーフ出力
- 安全関連の認証:
 - DIN VDE V 0884-11 に準拠した強化絶縁耐圧: $7070\text{V}_{\text{PK}}2017-01$
 - UL 1577 に準拠した絶縁耐圧: $5000\text{V}_{\text{RMS}}$ (1 分間)
- 拡張産業用温度範囲全体にわたって仕様を完全に規定: $-40^\circ\text{C} \sim +125^\circ\text{C}$

2 アプリケーション

- 次の用途での絶縁型 DC 電圧センシング:
 - モーター・ドライブ
 - 周波数インバータ
 - 太陽光インバータ
 - 電源

3 概要

AMC1351 は高精度の絶縁型アンプで、磁気干渉に対して高い耐性のある絶縁バリアにより、入力側と出力側の回路が分離されています。この絶縁バリアは、VDE V 0884-11 および UL1577 に従って最大 5kV_{RMS} の強化ガルバニック絶縁を達成していることが認証されており、最大 $1.5\text{kV}_{\text{RMS}}$ の使用電圧に対応しています。

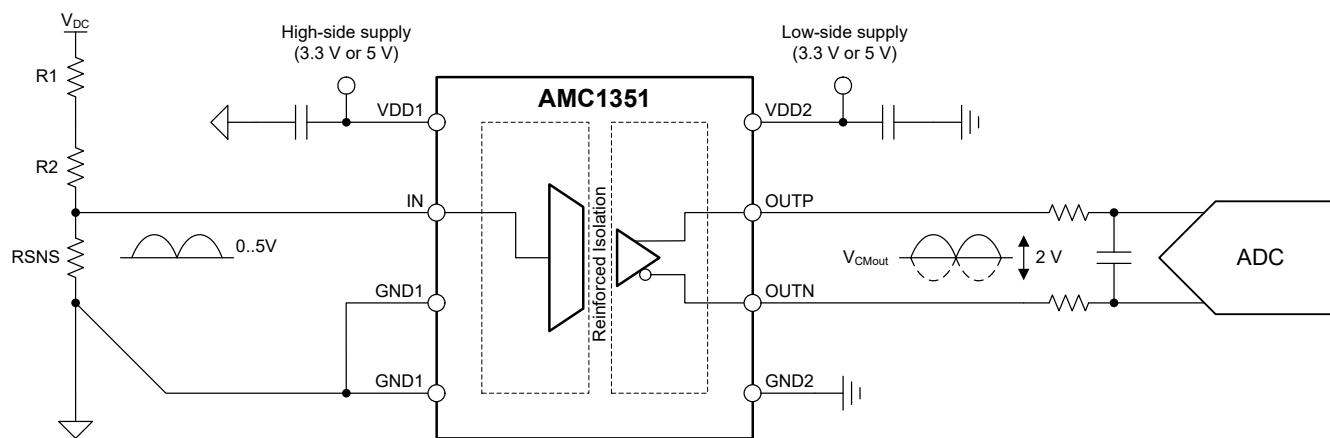
この絶縁バリアは、異なる同相電圧レベルで動作するシステム領域を分離し、潜在的に危険な電圧と損傷から低電圧側を保護します。

AMC1351 の高インピーダンス入力は、高インピーダンスの抵抗分圧器や出力抵抗の高い他の電圧信号源と接続するよう最適化されています。優れた精度と低い温度ドリフトにより、拡張産業用温度範囲の $-40^\circ\text{C} \sim +125^\circ\text{C}$ にわたって、DC/DC コンバータ、周波数インバータ、モーター・ドライブ、その他のアプリケーションで高精度の DC 電圧センシングをサポートします。

製品情報⁽¹⁾

部品番号	パッケージ	本体サイズ (公称)
AMC1351	SOIC (8)	5.85mm × 7.50mm

- (1) 利用可能なすべてのパッケージについては、このデータシートの末尾にある注文情報を参照してください。



代表的なアプリケーション



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4 Revision History

資料番号末尾の英字は改訂を表しています。その改訂履歴は英語版に準じています。

DATE	REVISION	NOTES
December 2021	*	Initial Release

5 Pin Configuration and Functions

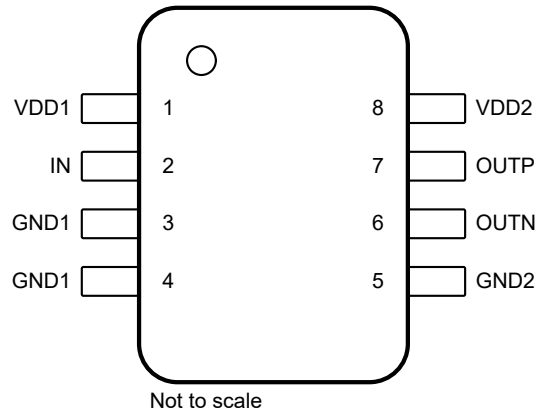


图 5-1. DWV Package, 8-Pin SOIC (Top View)

表 5-1. Pin Functions

PIN		TYPE	DESCRIPTION
NO.	NAME		
1	VDD1	High-side power	High-side power supply ⁽¹⁾
2	IN	Analog input	Analog input
3	GND1	High-side ground	High-side analog ground reference for input amplifier. Connect to pin 4. Do not leave unconnected.
4	GND1	High-side ground	High-side analog ground
5	GND2	Low-side ground	Low-side analog ground
6	OUTN	Analog output	Inverting analog output
7	OUTP	Analog output	Noninverting analog output
8	VDD2	Low-side power	Low-side power supply ⁽¹⁾

(1) See the [Power Supply Recommendations](#) section for power-supply decoupling recommendations.

6 Specifications

6.1 Absolute Maximum Ratings

see⁽¹⁾

		MIN	MAX	UNIT
Power-supply voltage	High-side VDD1 to GND1	-0.3	6.5	V
	Low-side VDD2 to GND2	-0.3	6.5	
Analog input voltage	IN	-1	15	V
Analog output voltage	OUTP, OUTN	GND2 - 0.5	VDD2 + 0.5	V
Input current	Continuous, any pin except power-supply pins	-10	10	mA
Temperature	Junction, T _J		150	°C
	Storage, T _{stg}	-65	150	

- (1) Operation outside the *Absolute Maximum Ratings* may cause permanent device damage. *Absolute Maximum Ratings* do not imply functional operation of the device at these or any other conditions beyond those listed under *Recommended Operating Conditions*. If used outside the Recommended Operating Conditions but within the Absolute Maximum Ratings, the device may not be fully functional, and this may affect device reliability, functionality, performance, and shorten the device lifetime

6.2 ESD Ratings

			VALUE	UNIT
V _(ESD)	Electrostatic discharge	Human-body model (HBM), per ANSI/ESDA/JEDEC JS-001 ⁽¹⁾	±2000	V
		Charged-device model (CDM), per per ANSI/ESDA/JEDEC JS-002 ⁽²⁾	±1000	

- (1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.
 (2) JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.

6.3 Recommended Operating Conditions

over operating ambient temperature range (unless otherwise noted)

				MIN	NOM	MAX	UNIT
POWER SUPPLY							
VDD1	High-side power-supply	VDD1 to GND1		3	5	5.5	V
VDD2	Low-side power-supply	VDD2 to GND2		3	3.3	5.5	V
ANALOG INPUT							
V _{Clipping}	Input voltage before clipping output				6.25		V
V _{FSR}	Specified linear full-scale voltage			-0.25		5	V
ANALOG OUTPUT							
C _{LOAD}	Capacitive load	On OUTP or OUTN to GND2				500	pF
		OUTP to OUTN				250	
R _{LOAD}	Resistive load	On OUTP or OUTN to GND2			10	1	kΩ
TEMPERATURE RANGE							
T _A	Operating ambient temperature			-55		125	°C
	Specified ambient temperature			-40		125	

6.4 Thermal Information

THERMAL METRIC ⁽¹⁾		AMC1351	UNIT
		DWV (SOIC)	
		8 PINS	
$R_{\theta JA}$	Junction-to-ambient thermal resistance	84.6	°C/W
$R_{\theta JC(top)}$	Junction-to-case (top) thermal resistance	28.3	°C/W
$R_{\theta JB}$	Junction-to-board thermal resistance	41.1	°C/W
Ψ_{JT}	Junction-to-top characterization parameter	4.9	°C/W
Ψ_{JB}	Junction-to-board characterization parameter	39.1	°C/W
$R_{\theta JC(bot)}$	Junction-to-case (bottom) thermal resistance	n/a	°C/W

(1) For more information about traditional and new thermal metrics, see the [Semiconductor and IC Package Thermal Metrics](#) application report.

6.5 Power Ratings

PARAMETER		TEST CONDITIONS	VALUE	UNIT
P_D	Maximum power dissipation (both sides)	VDD1 = VDD2 = 5.5 V	96	mW
P_{D1}	Maximum power dissipation (high-side)	VDD1 = 3.6 V	29	mW
		VDD1 = 5.5 V	51	
P_{D2}	Maximum power dissipation (low-side)	VDD2 = 3.6 V	26	mW
		VDD2 = 5.5 V	45	

6.6 Insulation Specifications

over operating ambient temperature range (unless otherwise noted)

PARAMETER		TEST CONDITIONS	VALUE	UNIT
GENERAL				
CLR	External clearance ⁽¹⁾	Shortest pin-to-pin distance through air	≥ 8.5	mm
CPG	External creepage ⁽¹⁾	Shortest pin-to-pin distance across the package surface	≥ 8.5	mm
DTI	Distance through insulation	Minimum internal gap (internal clearance) of the double insulation	≥ 0.021	mm
CTI	Comparative tracking index	DIN EN 60112 (VDE 0303-11); IEC 60112	≥ 600	V
	Material group	According to IEC 60664-1	I	
	Overvoltage category per IEC 60664-1	Rated mains voltage ≤ 600 V _{RMS}	I-IV	
		Rated mains voltage ≤ 1000 V _{RMS}	I-III	
DIN VDE V 0884-11 (VDE V 0884-11): 2017-01				
V _{IORM}	Maximum repetitive peak isolation voltage	At AC voltage	2120	V _{PK}
V _{IOWM}	Maximum-rated isolation working voltage	At AC voltage (sine wave)	1500	V _{RMS}
		At DC voltage	2120	V _{DC}
V _{IOTM}	Maximum transient isolation voltage	V _{TEST} = V _{IOTM} , t = 60 s (qualification test)	7070	V _{PK}
		V _{TEST} = 1.2 × V _{IOTM} , t = 1 s (100% production test)	8480	
V _{IOSM}	Maximum surge isolation voltage ⁽²⁾	Test method per IEC 60065, 1.2/50-μs waveform, V _{TEST} = 1.6 × V _{IOSM} = 12800 V _{PK} (qualification)	8000	V _{PK}
q _{pd}	Apparent charge ⁽³⁾	Method a, after input/output safety test subgroups 2 and 3, V _{ini} = V _{IOTM} , t _{ini} = 60 s, V _{pd(m)} = 1.2 × V _{IORM} , t _m = 10 s	≤ 5	pC
		Method a, after environmental tests subgroup 1, V _{ini} = V _{IOTM} , t _{ini} = 60 s, V _{pd(m)} = 1.6 × V _{IORM} , t _m = 10 s	≤ 5	
		Method b1, at routine test (100% production) and preconditioning (type test), V _{ini} = V _{IOTM} , t _{ini} = 1 s, V _{pd(m)} = 1.875 × V _{IORM} , t _m = 1 s	≤ 5	
C _{IO}	Barrier capacitance, input to output ⁽⁴⁾	V _{IO} = 0.5 V _{PP} at 1 MHz	~1.5	pF
R _{IO}	Insulation resistance, input to output ⁽⁴⁾	V _{IO} = 500 V at T _A = 25°C	> 10 ¹²	Ω
		V _{IO} = 500 V at 100°C ≤ T _A ≤ 125°C	> 10 ¹¹	
		V _{IO} = 500 V at T _S = 150°C	> 10 ⁹	
	Pollution degree		2	
	Climatic category		55/125/21	
UL1577				
V _{ISO}	Withstand isolation voltage	V _{TEST} = V _{ISO} = 5000 V _{RMS} or 7071 V _{DC} , t = 60 s (qualification), V _{TEST} = 1.2 × V _{ISO} = 6000 V _{RMS} , t = 1 s (100% production test)	5000	V _{RMS}

- (1) Apply creepage and clearance requirements according to the specific equipment isolation standards of an application. Care must be taken to maintain the creepage and clearance distance of a board design to ensure that the mounting pads of the isolator on the printed circuit board (PCB) do not reduce this distance. Creepage and clearance on a PCB become equal in certain cases. Techniques such as inserting grooves, ribs, or both on a PCB are used to help increase these specifications.
- (2) Testing is carried out in air or oil to determine the intrinsic surge immunity of the isolation barrier.
- (3) Apparent charge is electrical discharge caused by a partial discharge (pd).
- (4) All pins on each side of the barrier are tied together, creating a two-pin device.

6.7 Safety-Related Certifications

VDE	UL
Certified according to DIN VDE V 0884-11 (VDE V 0884-11): 2017-01, DIN EN 60950-1 (VDE 0805 Teil 1): 2014-08, and DIN EN 60065 (VDE 0860): 2005-11	Recognized under 1577 component recognition
Reinforced insulation	Single protection
Certificate number: pending	File number: E181974

6.8 Safety Limiting Values

Safety limiting⁽¹⁾ intends to minimize potential damage to the isolation barrier upon failure of input or output circuitry. A failure of the I/O can allow low resistance to ground or the supply and, without current limiting, dissipate sufficient power to over-heat the die and damage the isolation barrier potentially leading to secondary system failures.

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
I _S	Safety input, output, or supply current	R _{θJA} = 84.6°C/W, VDDx = 5.5 V, T _J = 150°C, T _A = 25°C			270	mA
		R _{θJA} = 84.6°C/W, VDDx = 3.6 V, T _J = 150°C, T _A = 25°C			410	
P _S	Safety input, output, or total power	R _{θJA} = 84.6°C/W, T _J = 150°C, T _A = 25°C			1480	mW
T _S	Maximum safety temperature				150	°C

- (1) The maximum safety temperature, T_S, has the same value as the maximum junction temperature, T_J, specified for the device. The I_S and P_S parameters represent the safety current and safety power, respectively. Do not exceed the maximum limits of I_S and P_S. These limits vary with the ambient temperature, T_A.

The junction-to-air thermal resistance, R_{θJA}, in the [Thermal Information](#) table is that of a device installed on a high-K test board for leaded surface-mount packages. Use these equations to calculate the value for each parameter:

T_J = T_A + R_{θJA} × P, where P is the power dissipated in the device.

T_{J(max)} = T_S = T_A + R_{θJA} × P_S, where T_{J(max)} is the maximum junction temperature.

P_S = I_S × VDD_{max}, where VDD_{max} is the maximum supply voltage for high-side and low-side.

6.9 Electrical Characteristics

minimum and maximum specifications apply from $T_A = -40^\circ\text{C}$ to $+125^\circ\text{C}$, $V_{DD1} = 3.0\text{ V}$ to 5.5 V , $V_{DD2} = 3.0\text{ V}$ to 5.5 V , $I_N = -0.25\text{ V}$ to $+5\text{ V}$ (unless otherwise noted); typical specifications are at $T_A = 25^\circ\text{C}$, $V_{DD1} = 5\text{ V}$, and $V_{DD2} = 3.3\text{ V}$

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
ANALOG INPUT						
V_{OS}	Offset voltage ⁽²⁾	$T_A = 25^\circ\text{C}$, $I_N = \text{GND1}$, $4.5\text{ V} \leq V_{DD1} \leq 5.5\text{ V}$ ⁽¹⁾	-1.5	± 0.3	1.5	mV
		$T_A = 25^\circ\text{C}$, $I_N = \text{GND1}$, $3.0\text{ V} \leq V_{DD1} \leq 5.5\text{ V}$ ⁽³⁾	-2.5	-0.8	2.5	
ΔV_{OS}	Offset voltage long-term stability	10 years at $T_A = 55^\circ\text{C}$		0 ⁽⁷⁾		mV
TCV_{OS}	Offset voltage thermal drift ⁽⁵⁾	$I_N = \text{GND1}$	-15	± 3	15	$\mu\text{V}/^\circ\text{C}$
ΔTCV_{OS}	Offset voltage thermal drift long-term stability	10 years at $T_A = 55^\circ\text{C}$, $I_N = \text{GND1}$		0 ⁽⁷⁾		$\text{mV}/^\circ\text{C}$
R_{IN}	Input resistance		1	1.25	1.5	M Ω
ΔR_{IN}	Input resistance long-term stability	10 years at $T_A = 55^\circ\text{C}$		0 ⁽⁷⁾		ppm
TCR_{IN}	Input resistance thermal drift	$-40^\circ\text{C} \leq T_A \leq 85^\circ\text{C}$		5		$\text{ppm}/^\circ\text{C}$
C_{IN}	Input capacitance	$f_{IN} = 275\text{ kHz}$		4		pF
ANALOG OUTPUT						
	Nominal gain			0.40		V/V
E_G	Gain error ⁽¹⁾	$T_A = 25^\circ\text{C}$	-0.2%	$\pm 0.05\%$	0.2%	
ΔE_G	Gain error long-term stability	10 years at $T_A = 55^\circ\text{C}$		0 ⁽⁷⁾		
TCE_G	Gain error thermal drift ^{(1) (6)}		-35	± 10	35	$\text{ppm}/^\circ\text{C}$
ΔTCE_G	Gain error thermal drift long-term stability	10 years at $T_A = 55^\circ\text{C}$		0 ⁽⁷⁾		$\text{ppm}/^\circ\text{C}$
	Nonlinearity ⁽¹⁾		-0.02%	$\pm 0.003\%$	0.02%	
	Nonlinearity thermal drift			0.2		$\text{ppm}/^\circ\text{C}$
THD	Total harmonic distortion ⁽⁴⁾	$V_{IN} = 5 V_{PP}$, $f_{IN} = 10\text{ kHz}$, $BW = 100\text{ kHz}$		-82		dB
SNR	Signal-to-noise ratio	$V_{IN} = 5 V_{PP}$, $f_{IN} = 1\text{ kHz}$, $BW = 10\text{ kHz}$	75	79		dB
		$V_{IN} = 5 V_{PP}$, $f_{IN} = 10\text{ kHz}$, $BW = 100\text{ kHz}$		69		
	Output noise	$I_N = \text{GND1}$, $BW = 100\text{ kHz}$		250		μV_{rms}
PSRR	Power-supply rejection ratio ⁽²⁾	PSRR vs V_{DD1} , DC		-67		dB
		PSRR vs V_{DD2} , DC		-80		
		PSRR vs V_{DD1} with 10-kHz, 100-mV ripple		-65		
		PSRR vs V_{DD2} with 10-kHz, 100-mV ripple		-64		
V_{CMout}	Output common-mode voltage		1.39	1.44	1.49	V
$V_{CLIPout}$	Clipping differential output voltage	$V_{OUT} = (V_{OUTP} - V_{OUTN})$, $V_{IN} > V_{Clipping}$		2.49		V
$V_{Fail-safe}$	Fail-safe differential output voltage	V_{DD1} undervoltage or V_{DD1} missing		-2.57	-2.5	V
BW	Output bandwidth		275	300		kHz
R_{OUT}	Output resistance	On $OUTP$ or $OUTN$		< 0.2		Ω
	Output short-circuit current	On $OUTP$ or $OUTN$, sourcing or sinking, $I_N = \text{GND1}$, outputs shorted to either GND or V_{DD2}		14		mA
CMTI	Common-mode transient immunity		100	150		$\text{kV}/\mu\text{s}$
POWER SUPPLY						

6.9 Electrical Characteristics (continued)

minimum and maximum specifications apply from $T_A = -40^\circ\text{C}$ to $+125^\circ\text{C}$, $V_{DD1} = 3.0\text{ V}$ to 5.5 V , $V_{DD2} = 3.0\text{ V}$ to 5.5 V , $I_N = -0.25\text{ V}$ to $+5\text{ V}$ (unless otherwise noted); typical specifications are at $T_A = 25^\circ\text{C}$, $V_{DD1} = 5\text{ V}$, and $V_{DD2} = 3.3\text{ V}$

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
VDD1 _{UV}	VDD1 undervoltage detection threshold	VDD1 rising	2.5	2.7	2.9	V
		VDD1 falling	2.4	2.6	2.8	
VDD2 _{UV}	VDD2 undervoltage detection threshold	VDD2 rising	2.2	2.45	2.65	V
		VDD2 falling	1.85	2.0	2.2	
I _{DD1}	High-side supply current	3.0 V < VDD1 < 3.6 V		6.0	8.1	mA
		4.5 V < VDD1 < 5.5 V		7.0	9.3	
I _{DD2}	Low-side supply current	3.0 V < VDD2 < 3.6 V		5.3	7.2	mA
		4.5 V < VDD2 < 5.5 V		5.9	8.1	

- (1) The typical value includes one standard deviation (*sigma*) at nominal operating conditions.
- (2) This parameter is input referred.
- (3) The typical value is at $V_{DD1} = 3.3\text{ V}$.
- (4) THD is the ratio of the rms sum of the amplitudes of first five higher harmonics to the amplitude of the fundamental.
- (5) Offset error temperature drift is calculated using the box method, as described by the following equation:

$$TCV_{OS} = (V_{OS,MAX} - V_{OS,MIN}) / TempRange$$
 where $V_{OS,MAX}$ and $V_{OS,MIN}$ refer to the maximum and minimum V_{OS} values measured within the temperature range (-40 to 125°C).
- (6) Gain error temperature drift is calculated using the box method, as described by the following equation:

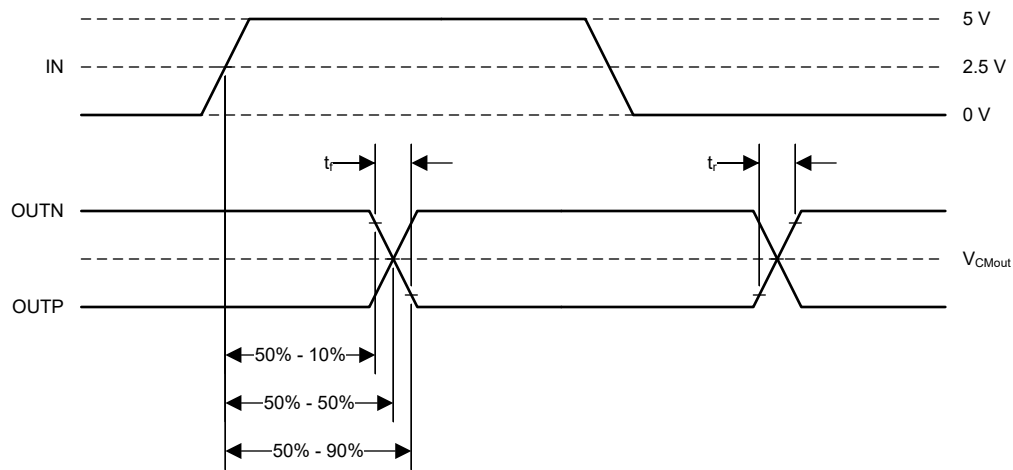
$$TCE_G (ppm) = ((E_{G,MAX} - E_{G,MIN}) / TempRange) \times 10^4$$
 where $E_{G,MAX}$ and $E_{G,MIN}$ refer to the maximum and minimum E_G values (in %) measured within the temperature range (-40 to 125°C).
- (7) Value is below measurement capability.

6.10 Switching Characteristics

over operating ambient temperature range (unless otherwise noted)

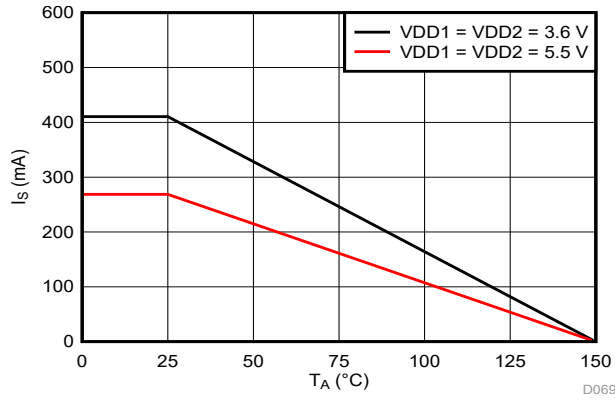
PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
t_r	Output signal rise time			1.3		μs
t_f	Output signal fall time			1.3		μs
	IN to OUTx signal delay (50% – 10%)	Unfiltered output		1	1.5	μs
	IN to OUTx signal delay (50% – 50%)	Unfiltered output		1.6	2.1	μs
	IN to OUTx signal delay (50% – 90%)	Unfiltered output		2.5	3	μs
t_{AS}	Analog settling time	VDD1 step to 3.0 V with VDD2 \geq 3.0 V, to V _{OUTP} and V _{OUTN} valid, 0.1% settling		500	800	μs

6.11 Timing Diagram

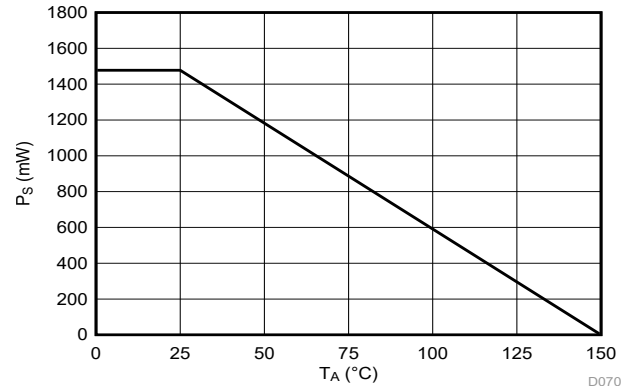


6-1. Rise, Fall, and Delay Time Definition

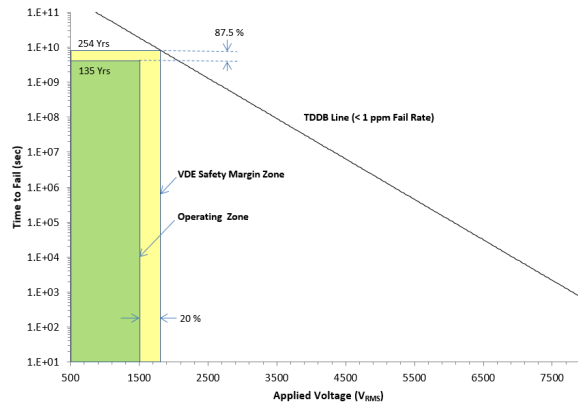
6.12 Insulation Characteristics Curves



6-2. Thermal Derating Curve for Safety-Limiting Current per VDE



6-3. Thermal Derating Curve for Safety-Limiting Power per VDE

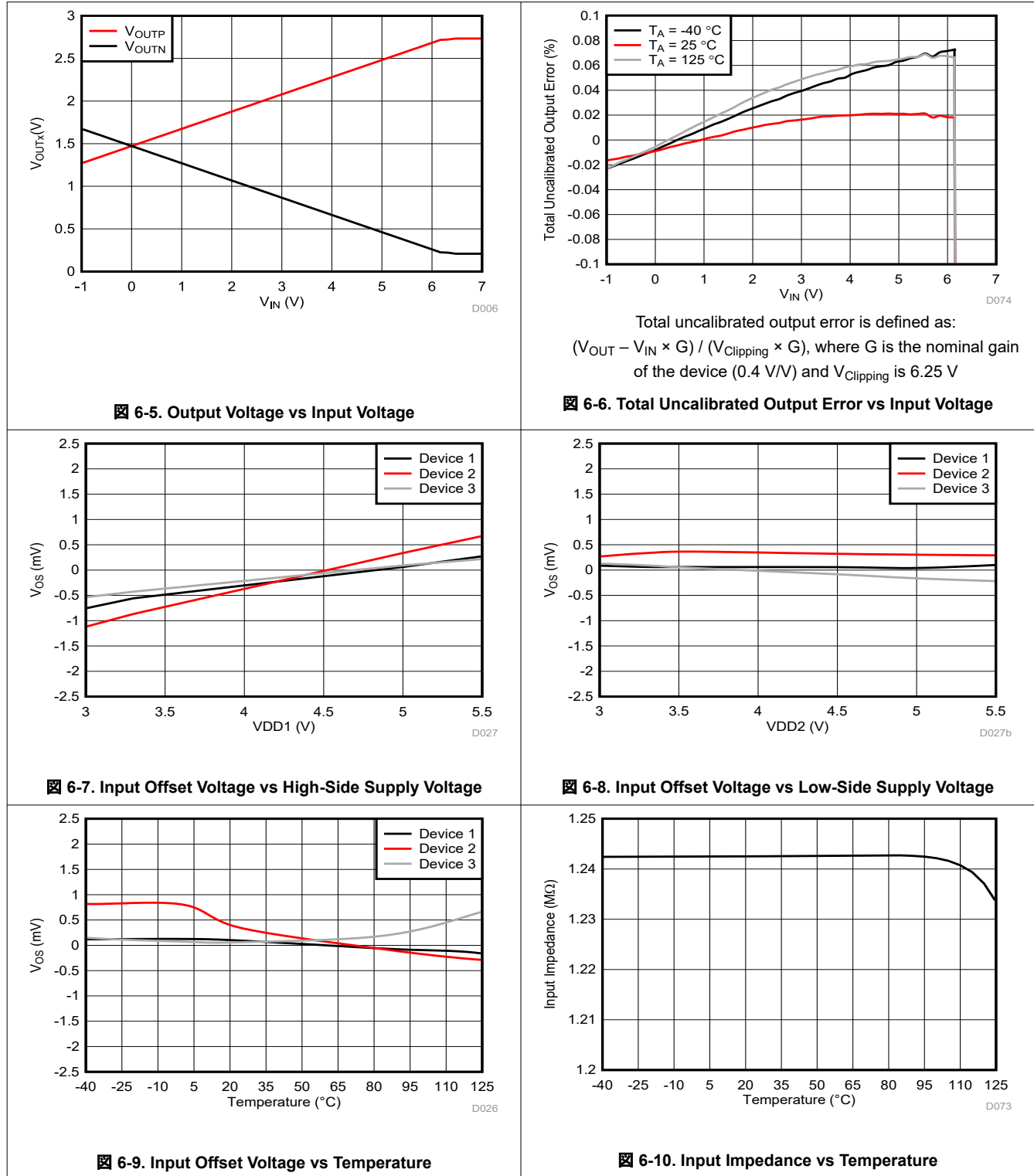


T_A up to 150°C, stress-voltage frequency = 60 Hz, isolation working voltage = 1500 V_{RMS} , operating lifetime = 135 years

6-4. Reinforced Isolation Capacitor Lifetime Projection

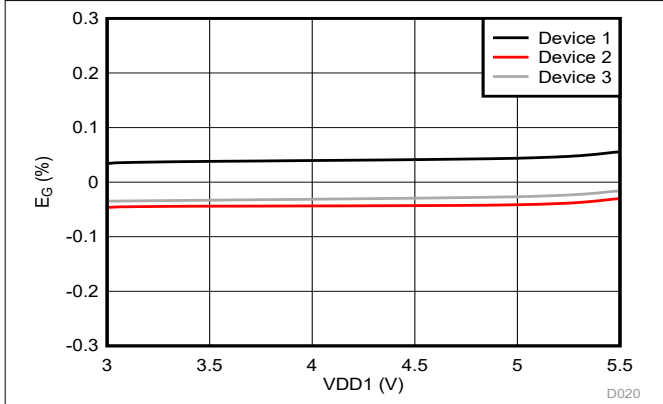
6.13 Typical Characteristics

at VDD1 = 5 V, VDD2 = 3.3 V, IN = 0 V to 5 V, and $f_{IN} = 10$ kHz (unless otherwise noted)

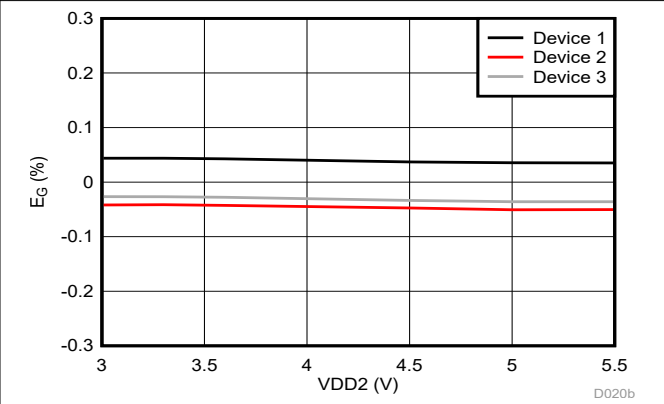


6.13 Typical Characteristics (continued)

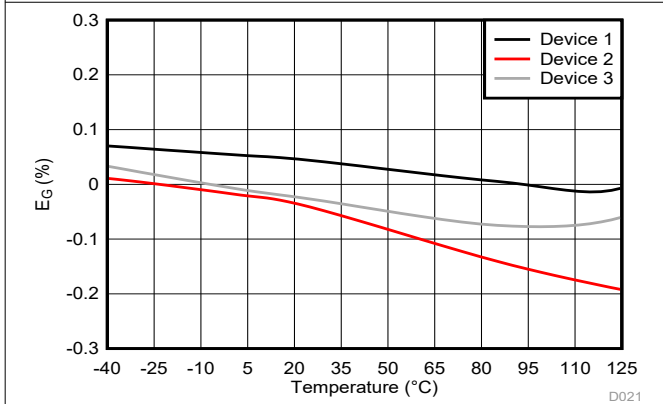
at VDD1 = 5 V, VDD2 = 3.3 V, IN = 0 V to 5 V, and $f_{IN} = 10$ kHz (unless otherwise noted)



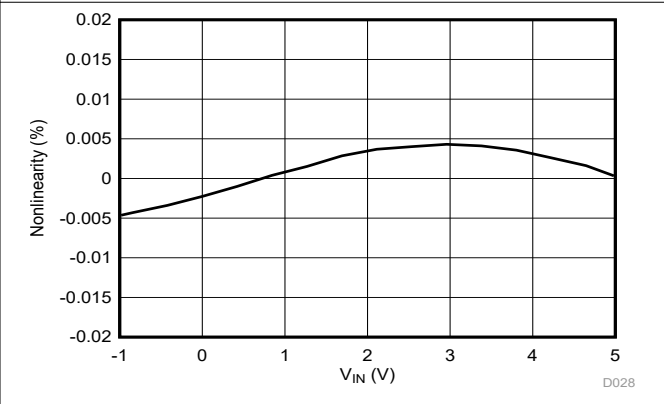
6-11. Gain Error vs High-Side Supply Voltage



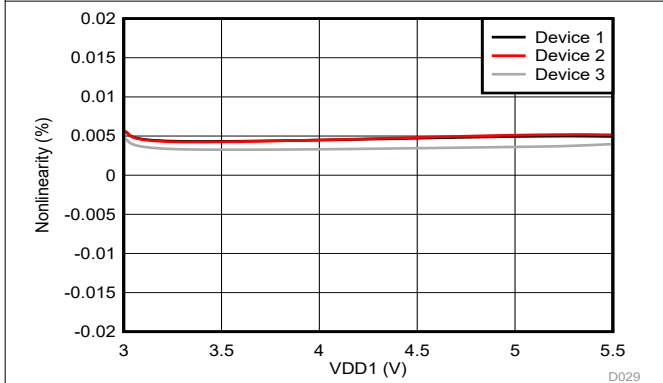
6-12. Gain Error vs Low-Side Supply Voltage



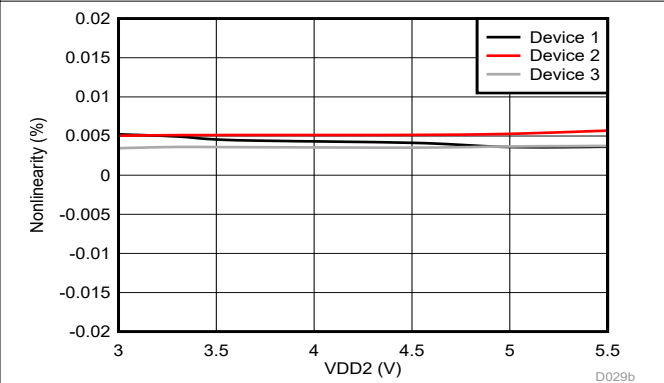
6-13. Gain Error vs Temperature



6-14. Nonlinearity vs Input Voltage



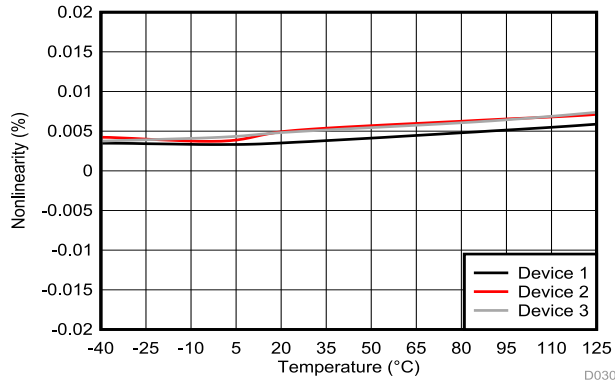
6-15. Nonlinearity vs High-Side Supply Voltage



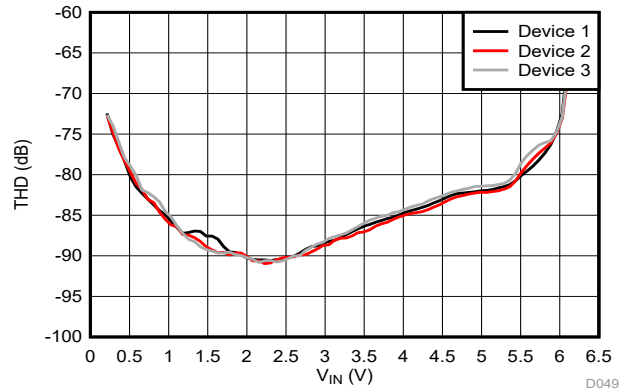
6-16. Nonlinearity vs Low-Side Supply Voltage

6.13 Typical Characteristics (continued)

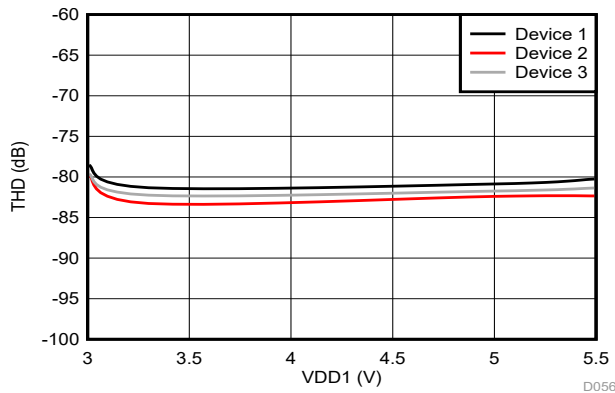
at VDD1 = 5 V, VDD2 = 3.3 V, IN = 0 V to 5 V, and $f_{IN} = 10$ kHz (unless otherwise noted)



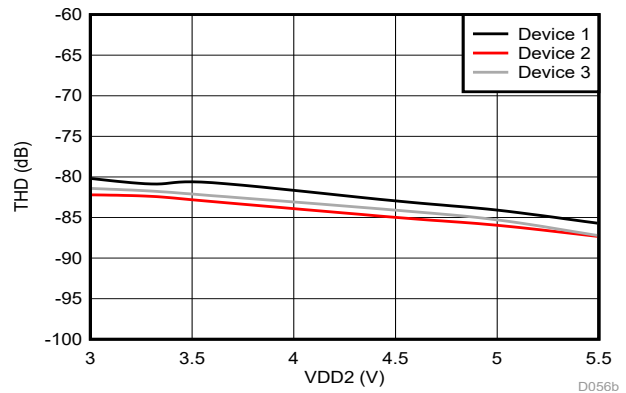
6-17. Nonlinearity vs Temperature



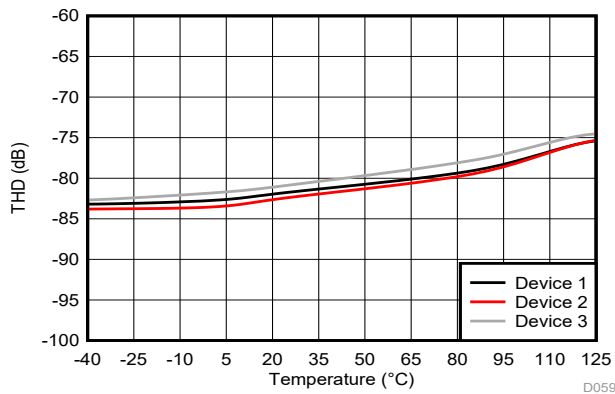
6-18. Total Harmonic Distortion vs Input Voltage



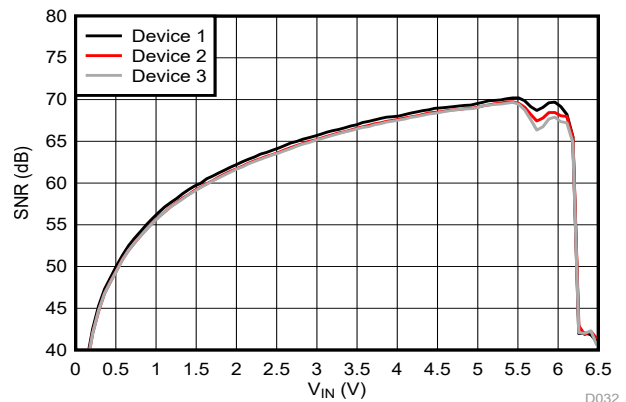
6-19. Total Harmonic Distortion vs High-Side Supply Voltage



6-20. Total Harmonic Distortion vs Low-Side Supply Voltage



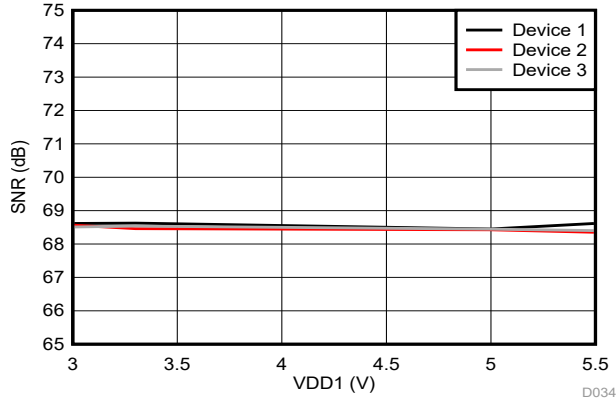
6-21. Total Harmonic Distortion vs Temperature



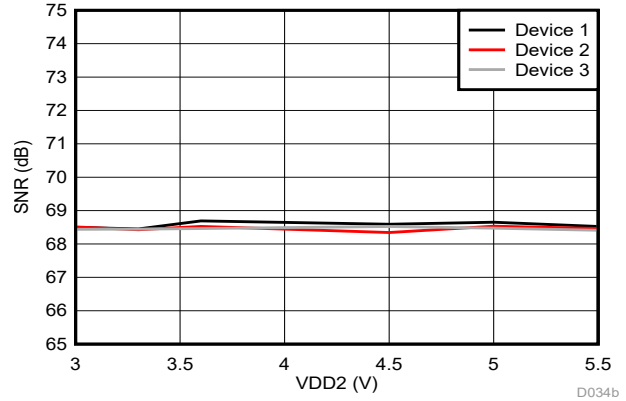
6-22. Signal-to-Noise Ratio vs Input Voltage

6.13 Typical Characteristics (continued)

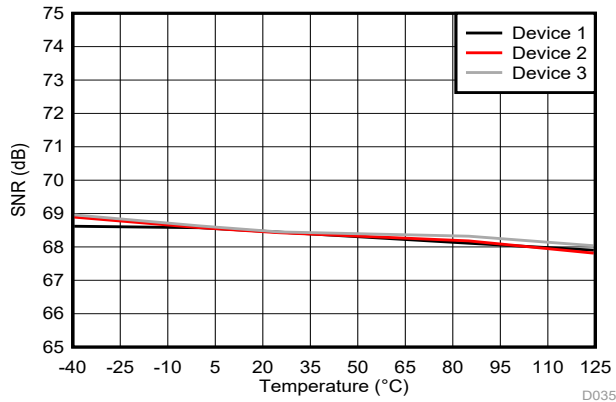
at VDD1 = 5 V, VDD2 = 3.3 V, IN = 0 V to 5 V, and $f_{IN} = 10$ kHz (unless otherwise noted)



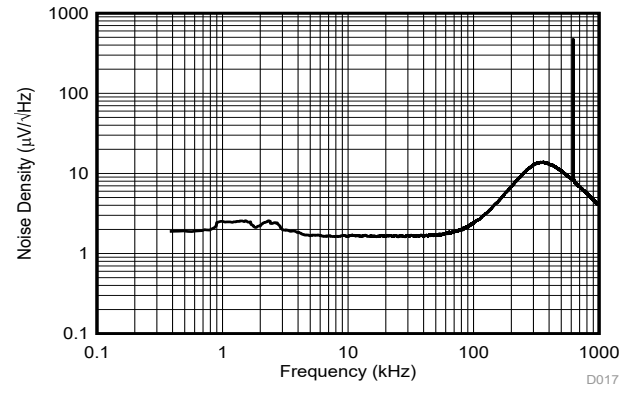
6-23. Signal-to-Noise Ratio vs High-Side Supply Voltage



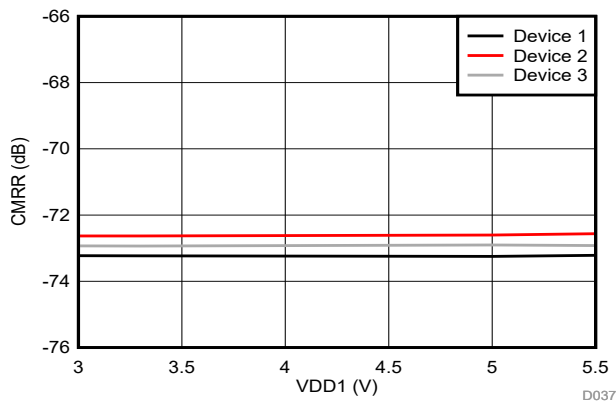
6-24. Signal-to-Noise Ratio vs Low-Side Supply Voltage



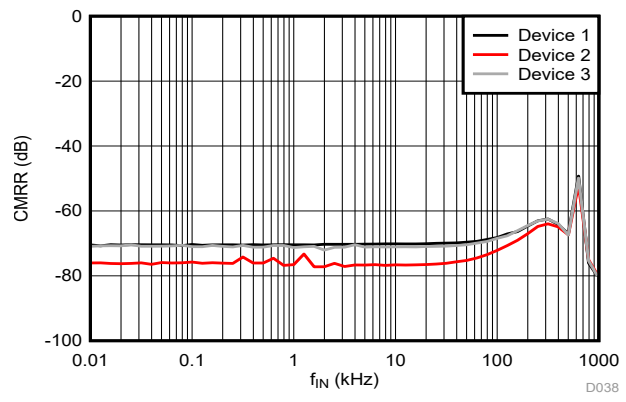
6-25. Signal-to-Noise Ratio vs Temperature



6-26. Input-Referred Noise Density vs Frequency



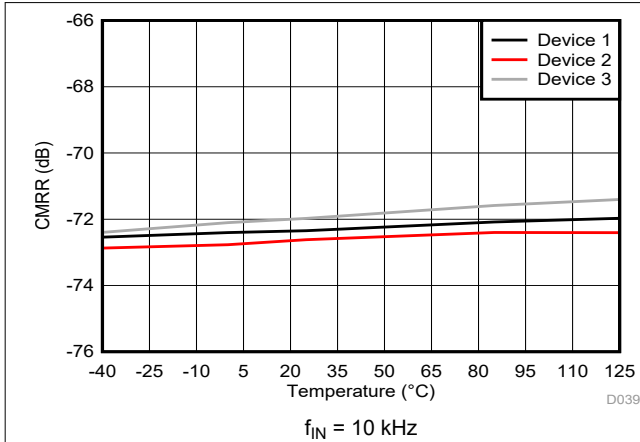
6-27. Common-Mode Rejection Ratio vs Supply Voltage



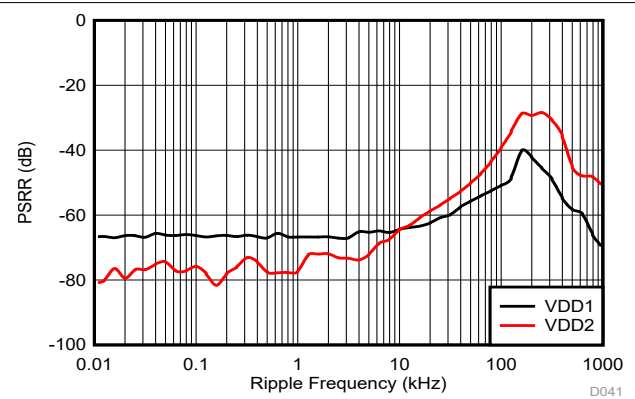
6-28. Common-Mode Rejection Ratio vs Input Frequency

6.13 Typical Characteristics (continued)

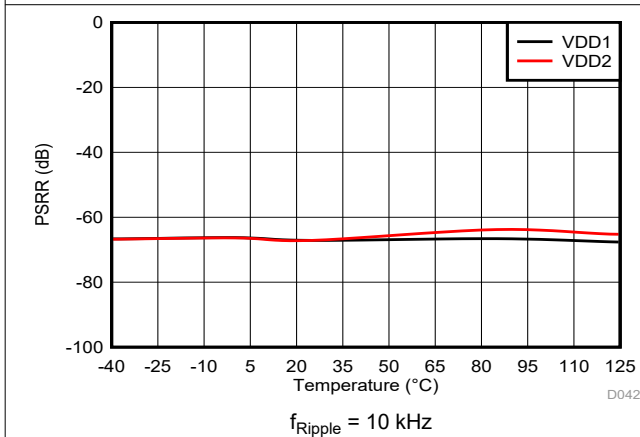
at VDD1 = 5 V, VDD2 = 3.3 V, IN = 0 V to 5 V, and $f_{IN} = 10$ kHz (unless otherwise noted)



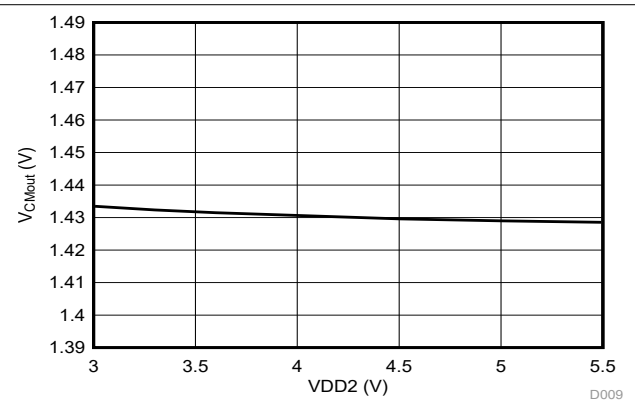
6-29. Common-Mode Rejection Ratio vs Temperature



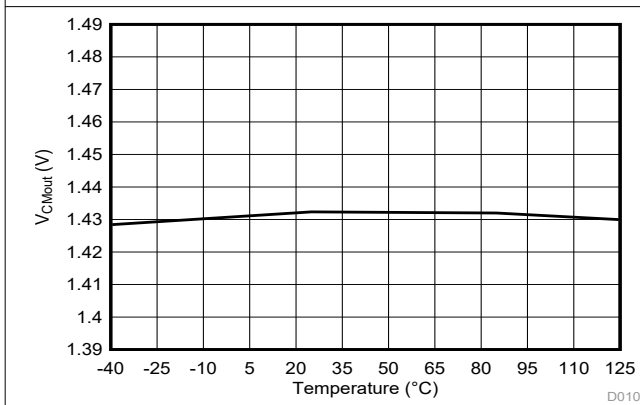
6-30. Power-Supply Rejection Ratio vs Ripple Frequency



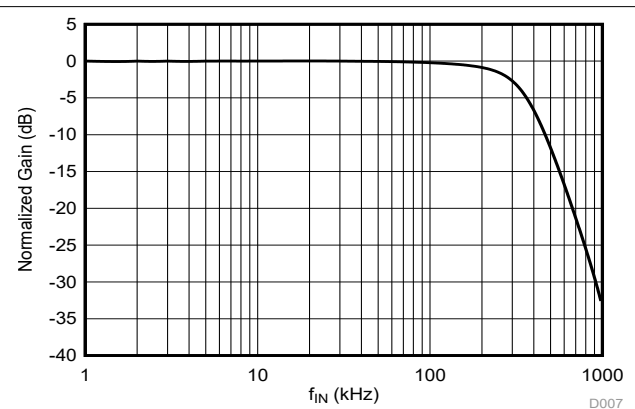
6-31. Power-Supply Rejection Ratio vs Temperature



6-32. Common-Mode Output Voltage vs Supply Voltage



6-33. Common-Mode Output Voltage vs Temperature



6-34. Normalized Gain vs Input Frequency

6.13 Typical Characteristics (continued)

at VDD1 = 5 V, VDD2 = 3.3 V, IN = 0 V to 5 V, and $f_{IN} = 10$ kHz (unless otherwise noted)

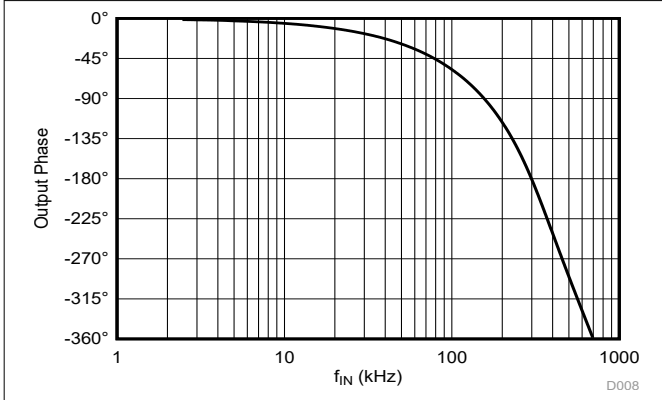


Figure 6-35. Output Phase vs Input Frequency

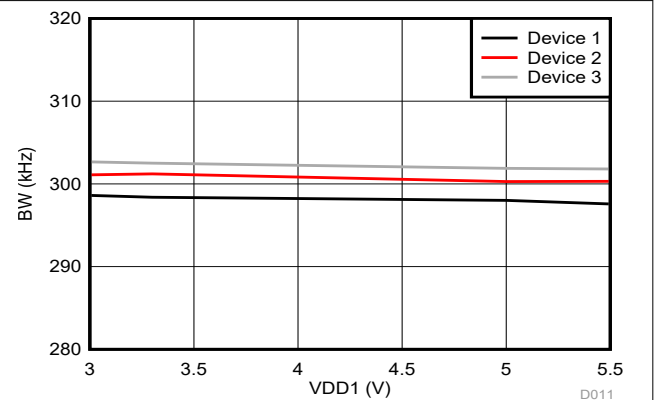


Figure 6-36. Bandwidth vs Supply Voltage

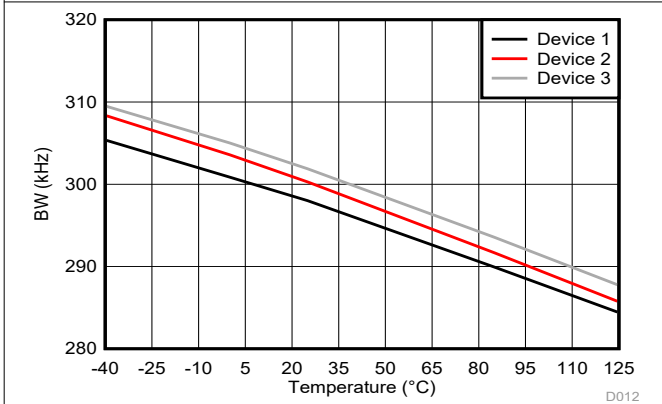


Figure 6-37. Bandwidth vs Temperature

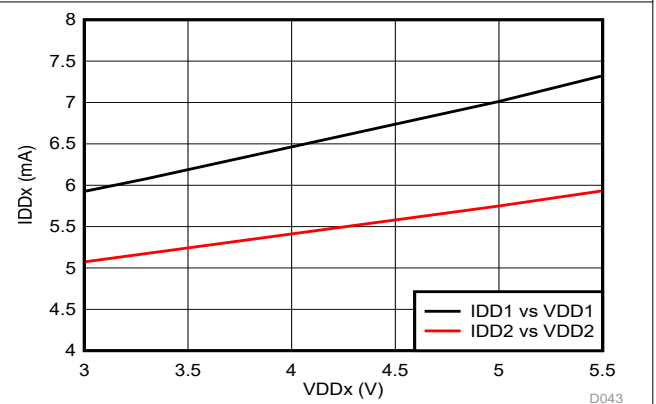


Figure 6-38. Supply Current vs Supply Voltage

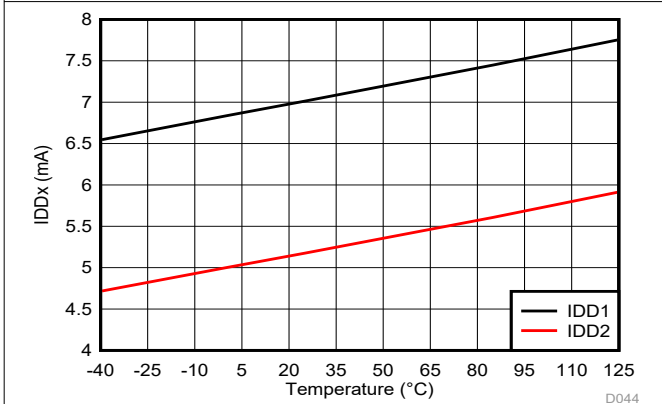


Figure 6-39. Supply Current vs Temperature

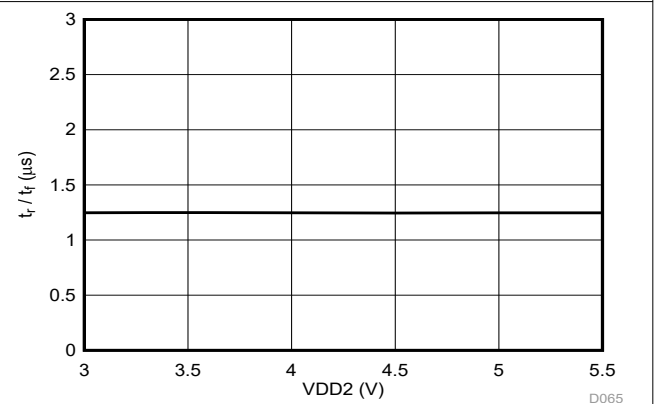
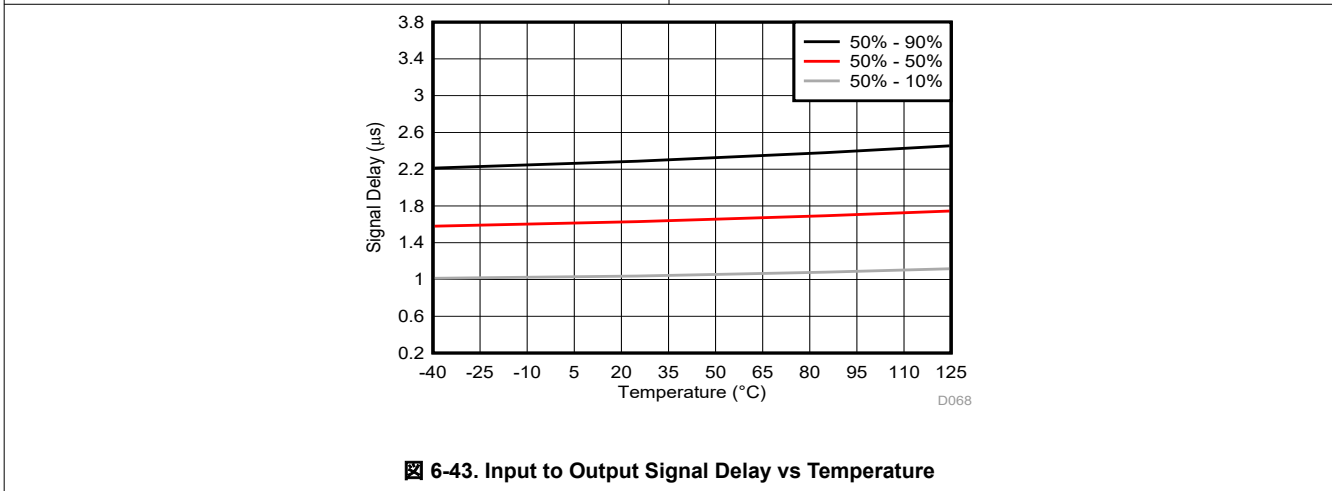
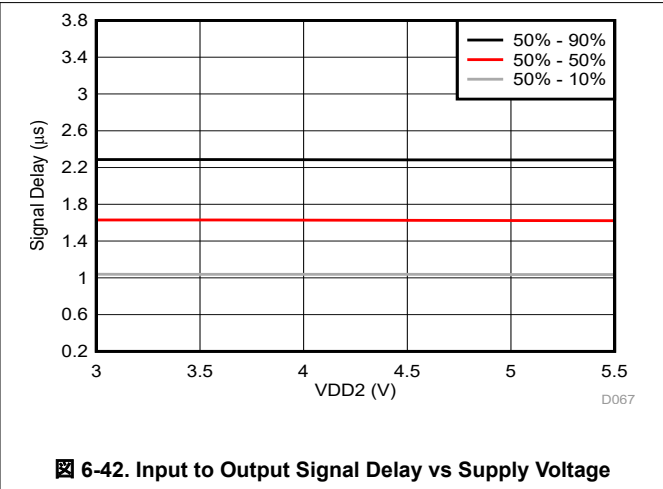
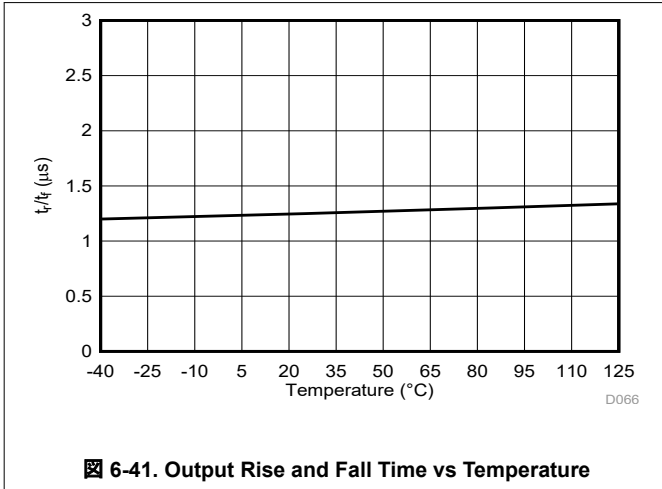


Figure 6-40. Output Rise and Fall Time vs Supply Voltage

6.13 Typical Characteristics (continued)

at VDD1 = 5 V, VDD2 = 3.3 V, IN = 0 V to 5 V, and $f_{IN} = 10$ kHz (unless otherwise noted)



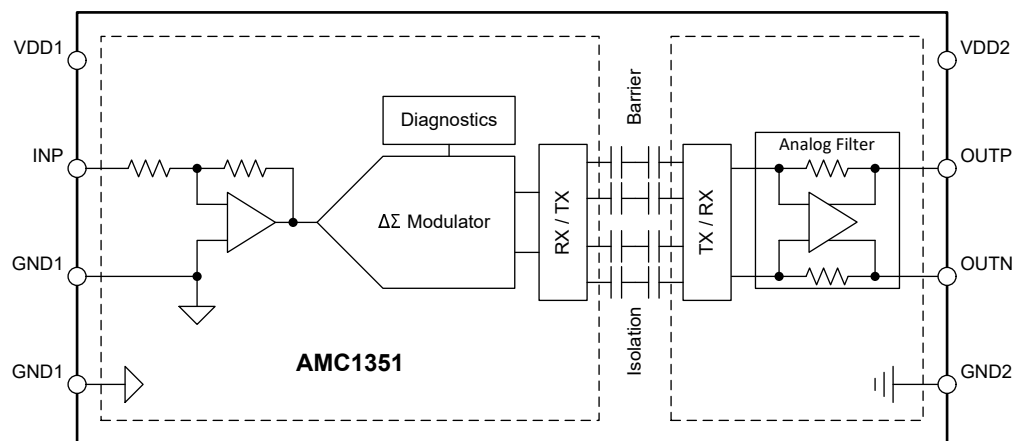
7 Detailed Description

7.1 Overview

The AMC1351 is a single-ended input, precision, isolated amplifier with a high input-impedance and wide input-voltage range. The input stage of the device drives a second-order, delta-sigma ($\Delta\Sigma$) modulator. The modulator converts the analog input signal into a digital bitstream that is transferred across the isolation barrier that separates the high-side from the low-side. On the low-side, the received bitstream is processed by a fourth-order analog filter that outputs a differential signal at the OUPN and OUTN pins proportional to the input signal.

The SiO₂-based, capacitive isolation barrier supports a high level of magnetic field immunity, as described in the [ISO72x Digital Isolator Magnetic-Field Immunity application report](#). The digital modulation used in the AMC1351 to transmit data across the isolation barrier, and the isolation barrier characteristics itself, result in high reliability and common-mode transient immunity.

7.2 Functional Block Diagram



7.3 Feature Description

7.3.1 Analog Input

The single-ended, high-impedance input stage of the AMC1351 feeds a second-order, switched-capacitor, feed-forward $\Delta\Sigma$ modulator. The modulator converts the analog signal into a bitstream that is transferred across the isolation barrier, as described in the [Isolation Channel Signal Transmission](#) section.

There are two restrictions on the analog input signal IN. First, if the input voltage V_{IN} exceeds the range specified in the [Absolute Maximum Ratings](#) table, the input current must be limited to the absolute maximum value because the electrostatic discharge (ESD) protection turns on. In addition, the linearity and parametric performance of the device is ensured only when the analog input voltage remains within the linear full-scale range (V_{FSR}) as specified in the [Recommended Operating Conditions](#) table.

7.3.2 Isolation Channel Signal Transmission

The AMC1351 uses an on-off keying (OOK) modulation scheme, as shown in [Figure 7-1](#), to transmit the modulator output bitstream across the SiO₂-based isolation barrier. The transmit driver (TX) shown in the [Functional Block Diagram](#) transmits an internally-generated, high-frequency carrier across the isolation barrier to represent a digital *one* and does not send a signal to represent a digital *zero*. The nominal frequency of the carrier used inside the AMC1351 is 480 MHz.

The receiver (RX) on the other side of the isolation barrier recovers and demodulates the signal and provides the input to the fourth-order analog filter. The AMC1351 transmission channel is optimized to achieve the highest level of common-mode transient immunity (CMTI) and lowest level of radiated emissions caused by the high-frequency carrier and RX, TX buffer switching.

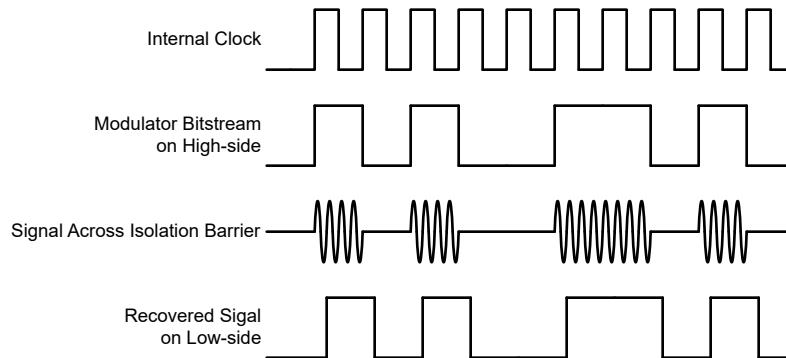


Figure 7-1. OOK-Based Modulation Scheme

7.3.3 Analog Output

The AMC1351 provides a differential analog output on the OUTP and OUTN pins. For input voltages (V_{IN}) in the range from -0.25 V to 5 V , the device provides a linear response with a nominal gain of 0.4 V/V . For example, for an input voltage of 5 V , the differential output voltage ($V_{OUTP} - V_{OUTN}$) is 2 V . At zero input (IN shorted to GND1), both pins output the same common-mode output voltage V_{CMout} , as specified in the [Electrical Characteristics](#) table. For input voltages greater than 5 V but less than approximately 6.25 V , the differential output voltage continues to increase but with reduced linearity performance. The outputs saturate at a differential output voltage of $V_{CLIPout}$, as shown in [Figure 7-2](#), if the input voltage exceeds the $V_{Clipping}$ value.

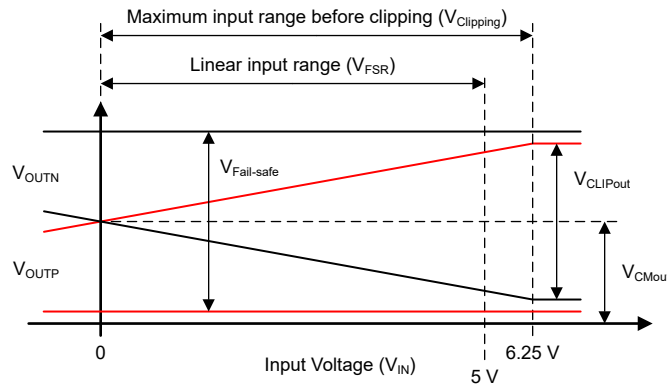


Figure 7-2. Output Behavior of the AMC1351

The AMC1351 output offers a fail-safe feature that simplifies diagnostics on a system level. [Figure 7-2](#) shows the behavior in fail-safe mode, in which the AMC1351 outputs a negative differential output voltage that does not occur under normal operating conditions. The fail-safe output is active:

- When the high-side supply VDD1 of the AMC1351 device is missing
- When the high-side supply VDD1 falls below the undervoltage threshold $VDD1_{UV}$

Use the maximum $V_{Fail-safe}$ voltage specified in the [Electrical Characteristics](#) table as a reference value for fail-safe detection on a system level.

7.4 Device Functional Modes

The AMC1351 is operational when the power supplies VDD1 and VDD2 are applied as specified in the [Recommended Operating Conditions](#) table.

8 Application and Implementation

Note

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes, as well as validating and testing their design implementation to confirm system functionality.

8.1 Application Information

The high input impedance, low input bias current, excellent accuracy, and low temperature drift make the AMC1351 a high-performance solution for industrial applications where voltage sensing in the presence of high common-mode voltage levels is required.

8.2 Typical Application

Isolated amplifiers are widely used for voltage measurements in high-voltage applications that must be isolated from a low-voltage domain. A typical application is the sensing of the DC bus voltage in a frequency inverter.

With its wide, 5-V input voltage range, the AMC1351 is designed for isolated DC voltage-sensing applications where accurate voltage monitoring is required in high-noise environments.

Figure 8-1 shows a simplified schematic of the AMC1351 in a typical motor drive application. The DC bus voltage is divided down to an approximate 5-V level across the bottom resistor (RSNS) of a high-impedance resistor divider that is sensed by the AMC1351. The AMC1351 digitizes the analog input signal on the high-side, transfers the data across the isolation barrier to the low-side, and reconstructs an analog signal that is presented as a differential voltage on the output pins.

The high-impedance input and the high common-mode transient immunity (CMTI) of the AMC1351 ensure reliable and accurate operation even in high-noise environments.

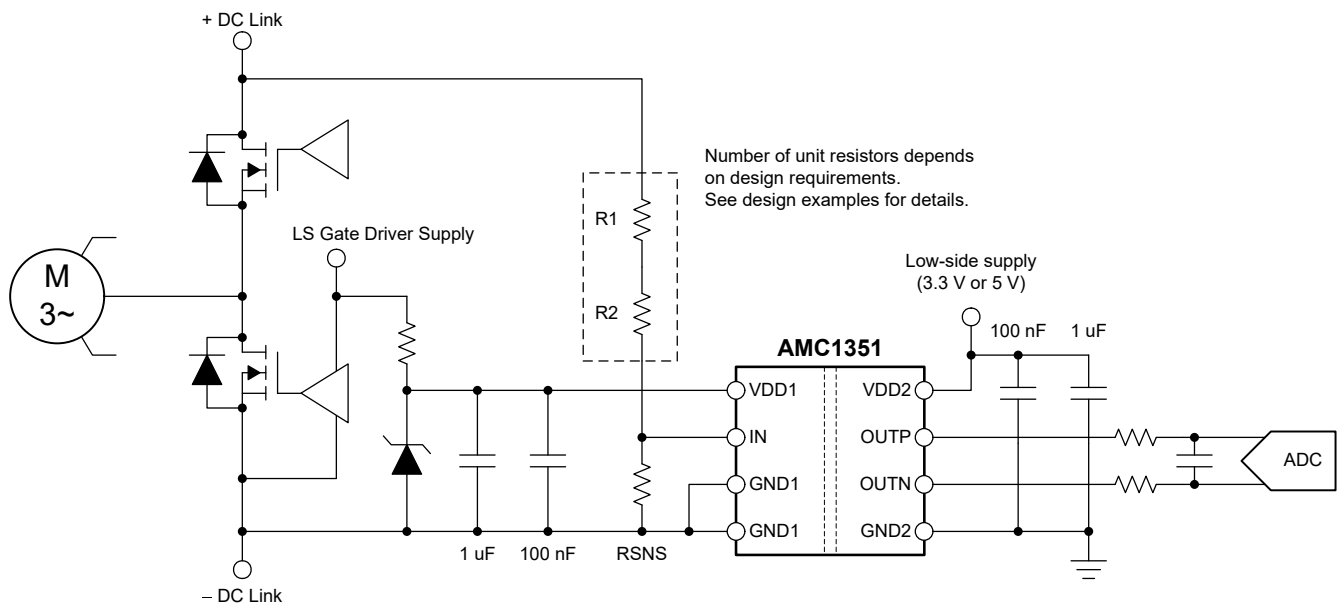


Figure 8-1. Using the AMC1351 for DC Link Voltage Sensing in Frequency Inverters

8.2.1 Design Requirements

表 8-1 lists the parameters for this typical application.

表 8-1. Design Requirements

PARAMETER	190-V _{DC} LINE VOLTAGE	360-V _{DC} LINE VOLTAGE
System input voltage	120 V _{RMS} ±10%, 60 Hz	230 V _{RMS} ±10%, 50 Hz
DC bus voltage (max)	190 V	360 V
High-side supply voltage	3.3 V or 5 V	3.3 V or 5 V
Low-side supply voltage	3.3 V or 5 V	3.3 V or 5 V
Maximum resistor operating voltage	75 V	75 V
Voltage drop across the sense resistor (RSNS) for a linear response	5 V (maximum)	5 V (maximum)
Current through the resistive divider (I _{CROSS})	100 μA	100 μA

8.2.2 Detailed Design Procedure

This discussion covers the 360-V_{DC} example. The procedure for calculating the resistive divider for the 190-V_{DC} use case is identical.

The 100-μA, cross-current requirement at peak input voltage (360 V) determines that the total impedance of the resistive divider is 3.6 MΩ. The impedance of the resistive divider is dominated by the top resistors (shown exemplarily as R1 and R2 in 图 8-1) and the voltage drop across RSNS can be neglected for a short time. The maximum allowed voltage drop per unit resistor is specified as 75 V; therefore, the total minimum number of unit resistors in the top portion of the resistive divider is 360 V / 75 V = 5. The calculated unit value is 3.6 MΩ / 5 = 720 kΩ and the next closest value from the E96 series is 715 kΩ.

The *effective* sense resistor value RSNS_{EFF} is the parallel combination of the external resistor RSNS and the input impedance of the AMC1351, R_{IN}. RSNS_{EFF} is sized such that the voltage drop across the impedance at maximum input voltage (360 V) equals the linear full-scale input voltage (V_{FSR}) of the AMC1351 (that is, 5 V). RSNS_{EFF} is calculated as $RSNS_{EFF} = V_{FSR} / (V_{Peak} - V_{FSR}) \times R_{TOP}$, where R_{TOP} is the total value of top resistor string (5 × 715 kΩ = 3575 kΩ). The resulting value for RSNS_{EFF} is 9.96 kΩ. In a final step, RSNS is calculated as $RSNS = R_{IN} \times RSNS_{EFF} / (R_{IN} - RSNS_{EFF})$. With R_{IN} = 1.25 MΩ (typical), RSNS equals 52.47 kΩ and the next closest value from the E96 series is 52.3 kΩ.

表 8-2 summarizes the design of the resistive divider.

表 8-2. Resistor Value Examples

PARAMETER	190-V _{DC} LINE VOLTAGE	360-V _{DC} LINE VOLTAGE
Unit resistor value (R _{TOP})	634 kΩ	715 kΩ
Number of unit resistors in R _{TOP}	3	5
Sense resistor value (RSNS)	51.1 kΩ	49.9 kΩ
Total resistance value (R _{TOP} + RSNS)	1953.1 kΩ	3624.9 kΩ
Resulting current through resistive divider (I _{CROSS})	97.3 μA	99.3 μA
Resulting full-scale voltage drop across sense resistor RSNS	4.971 V	4.956 V
Peak power dissipated in R _{TOP} unit resistors	6 mW	7.1 mW
Total peak power dissipated in resistive divider	18.5 mW	35.8 mW

8.2.2.1 Input Filter Design

Placing an RC filter in front of the isolated amplifier improves signal-to-noise performance of the signal path. In practice, however, the impedance of the resistor divider is so high that adding a filter capacitor on the IN pin limits the signal bandwidth to an unacceptable low limit, such that the filter capacitor is omitted. When used, design the input filter such that:

- The cutoff frequency of the filter is at least one order of magnitude lower than the sampling frequency (20 MHz) of the internal $\Delta\Sigma$ modulator
- The input bias current does not generate significant voltage drop across the DC impedance of the input filter

Most voltage-sensing applications use high-impedance resistor dividers in front of the isolated amplifier to scale down the input voltage. In that case, no additional resistor is needed and a single capacitor (as shown in [Figure 8-2](#)) is sufficient to filter the input signal.

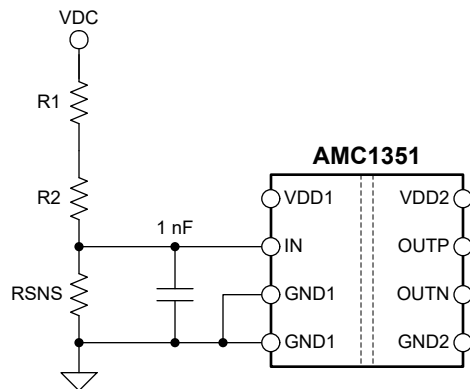


Figure 8-2. Input Filter

8.2.2.2 Differential to Single-Ended Output Conversion

[Figure 8-3](#) shows an example of a TLV6001-based signal conversion and filter circuit for systems using single-ended input ADCs to convert the analog output voltage into digital. With $R1 = R2 = R3 = R4$, the output voltage equals $(V_{OUTP} - V_{OUTN}) + V_{REF}$. Tailor the bandwidth of this filter stage to the bandwidth requirement of the system and use NP0-type capacitors for best performance. For most applications, $R1 = R2 = R3 = R4 = 3.3 \text{ k}\Omega$ and $C1 = C2 = 330 \text{ pF}$ yields good performance.

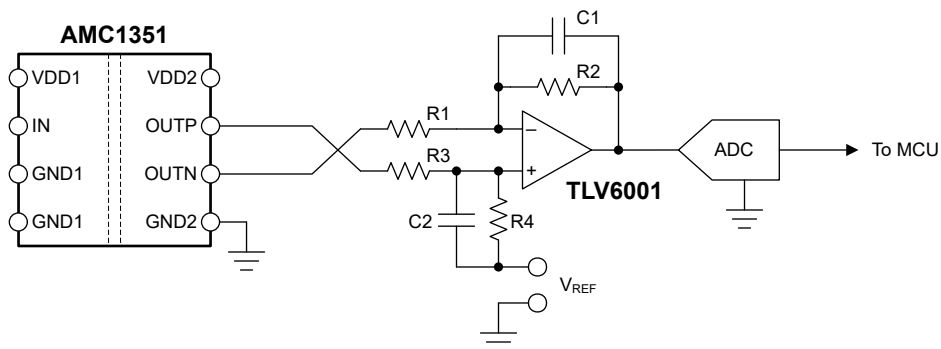

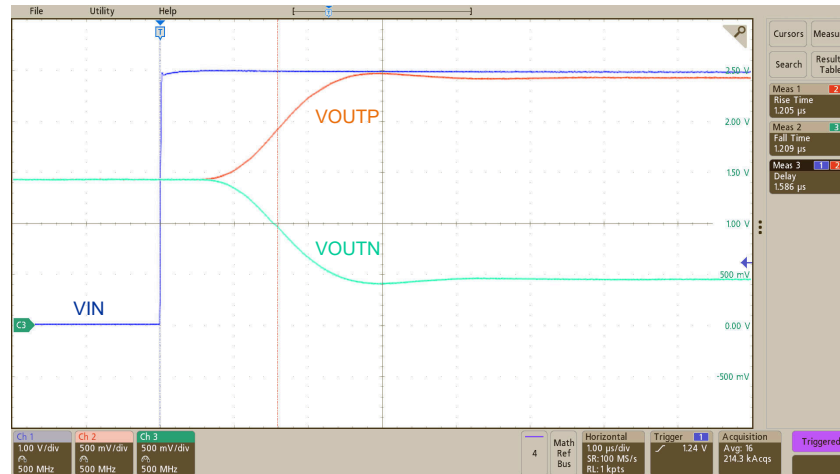


Figure 8-3. Connecting the AMC1351 Output to a Single-Ended Input ADC

For more information on the general procedure to design the filtering and driving stages of SAR ADCs, see the [18-Bit, 1MSPS Data Acquisition Block \(DAQ\) Optimized for Lowest Distortion and Noise](#) and [18-Bit Data Acquisition Block \(DAQ\) Optimized for Lowest Power reference guides](#), available for download at www.ti.com.

8.2.3 Application Curve

One important aspect of system design is the effective detection of an overvoltage condition to protect switching devices and passive components from damage. To power off the system quickly in the event of an overvoltage condition, a low delay caused by the isolated amplifier is required.  8-4 shows the typical full-scale step response of the AMC1351.



 8-4. Step Response of the AMC1351

8.3 What To Do and What Not To Do

Do not leave the analog input (IN) of the AMC1351 unconnected (floating) when the device is powered up on the high-side. If the device input is left floating, the bias current may generate a positive or negative input voltage and the output of the device is undetermined.

Do not connect protection diodes to the input (IN) of the AMC1351. Diode leakage current can introduce significant measurement error especially at high temperatures. The input pin is protected against high voltages by its ESD protection circuit and the high impedance of the external resistive divider

Connect both GND1 pins to the high-side ground potential. Do not leave one of the GND1 pins unconnected.

9 Power Supply Recommendations

In a typical application, the high-side power supply (VDD1) for the AMC1351 is generated either from a gate-driver supply on the high-side (as shown in [Figure 8-1](#)), or from the low-side supply (VDD2) by an isolated DC/DC converter. A low-cost solution is based on the push-pull driver [SN6501](#) and a transformer that supports the desired isolation voltage ratings.

The AMC1351 does not require any specific power-up sequencing. The high-side power supply (VDD1) is decoupled with a low-ESR, 100-nF capacitor (C1) parallel to a low-ESR, 1- μ F capacitor (C2). The low-side power supply (VDD2) is equally decoupled with a low-ESR, 100-nF capacitor (C3) parallel to a low-ESR, 1- μ F capacitor (C4). Place all four capacitors (C1, C2, C3, and C4) as close to the device as possible.

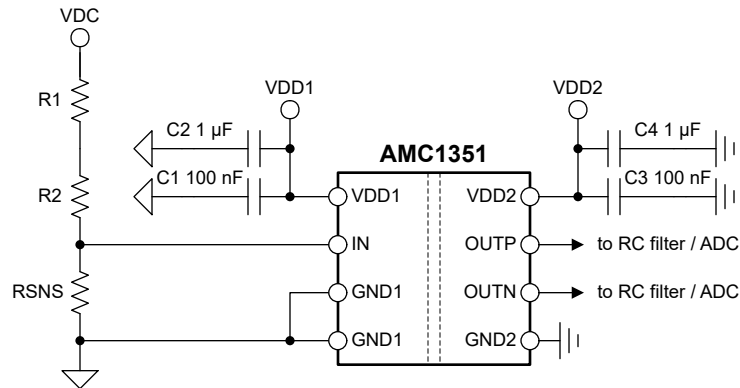



Figure 9-1. Decoupling of the AMC1351

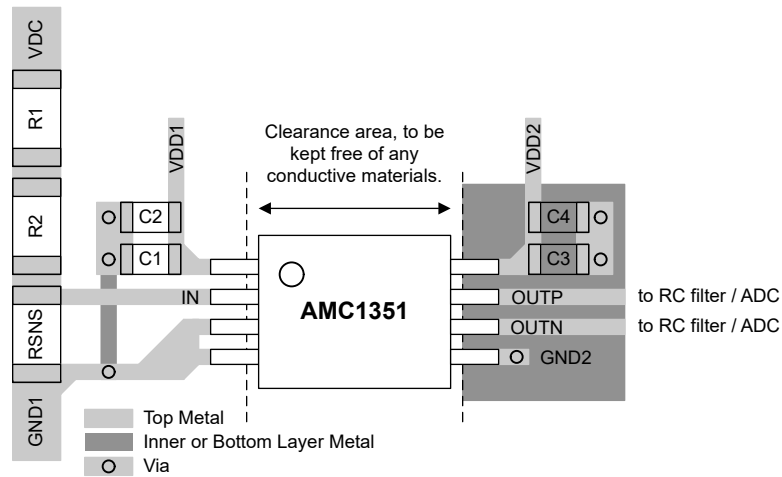
Capacitors must provide adequate effective capacitance under the applicable DC bias conditions they experience in the application. Multilayer ceramic capacitors (MLCC) typically exhibit only a fraction of their nominal capacitance under real-world conditions and this factor must be taken into consideration when selecting these capacitors. This problem is especially acute in low-profile capacitors, in which the dielectric field strength is higher than in taller components. Reputable capacitor manufacturers provide capacitance versus DC bias curves that greatly simplify component selection.

10 Layout

10.1 Layout Guidelines

 10-1 shows a layout recommendation with the critical placement of the decoupling capacitors (as close as possible to the AMC1351 supply pins) and placement of the other components required by the device. For best performance, place the sense resistor close to the device input pin (IN).

10.2 Layout Example



 10-1. Recommended Layout of the AMC1351

11 Device and Documentation Support

11.1 Documentation Support

11.1.1 Related Documentation

For related documentation, see the following:

- Texas Instruments, [Isolation Glossary application report](#)
- Texas Instruments, [Semiconductor and IC Package Thermal Metrics application report](#)
- Texas Instruments, [ISO72x Digital Isolator Magnetic-Field Immunity application report](#)
- Texas Instruments, [TLV600x Low-Power, Rail-to-Rail In/Out, 1-MHz Operational Amplifier for Cost-Sensitive Systems data sheet](#)
- Texas Instruments, [18-Bit, 1-MSPS Data Acquisition Block \(DAQ\) Optimized for Lowest Distortion and Noise reference guide](#)
- Texas Instruments, [18-Bit, 1-MSPS Data Acquisition Block \(DAQ\) Optimized for Lowest Power reference guide](#)
- Texas Instruments, [Isolated Amplifier Voltage Sensing Excel Calculator design tool](#)
- Texas Instruments, [Best in Class Radiated Emissions EMI Performance with the AMC1300B-Q1 Isolated Amplifier technical white paper](#)

11.2 Receiving Notification of Documentation Updates

To receive notification of documentation updates, navigate to the device product folder on [ti.com](#). Click on [Subscribe to updates](#) to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

11.3 サポート・リソース

TI E2E™ サポート・フォーラムは、エンジニアが検証済みの回答と設計に関するヒントをエキスパートから迅速かつ直接得ることができる場所です。既存の回答を検索したり、独自の質問をしたりすることで、設計に必要な支援を迅速に得ることができます。

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11.4 Trademarks

TI E2E™ is a trademark of Texas Instruments.

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11.5 Electrostatic Discharge Caution



This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

11.6 Glossary

[TI Glossary](#) This glossary lists and explains terms, acronyms, and definitions.

12 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead finish/ Ball material (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
AMC1351DWV	ACTIVE	SOIC	DWV	8	64	RoHS & Green	NIPDAU	Level-3-260C-168 HR	-40 to 125	AMC1351	Samples
AMC1351DWVR	ACTIVE	SOIC	DWV	8	1000	RoHS & Green	NIPDAU	Level-3-260C-168 HR	-40 to 125	AMC1351	Samples

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

(2) **RoHS:** TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

Green: TI defines "Green" to mean the content of Chlorine (Cl) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

(6) Lead finish/Ball material - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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In no event shall TI's liability arising out of such information exceed the total purchase price of the TI part(s) at issue in this document sold by TI to Customer on an annual basis.

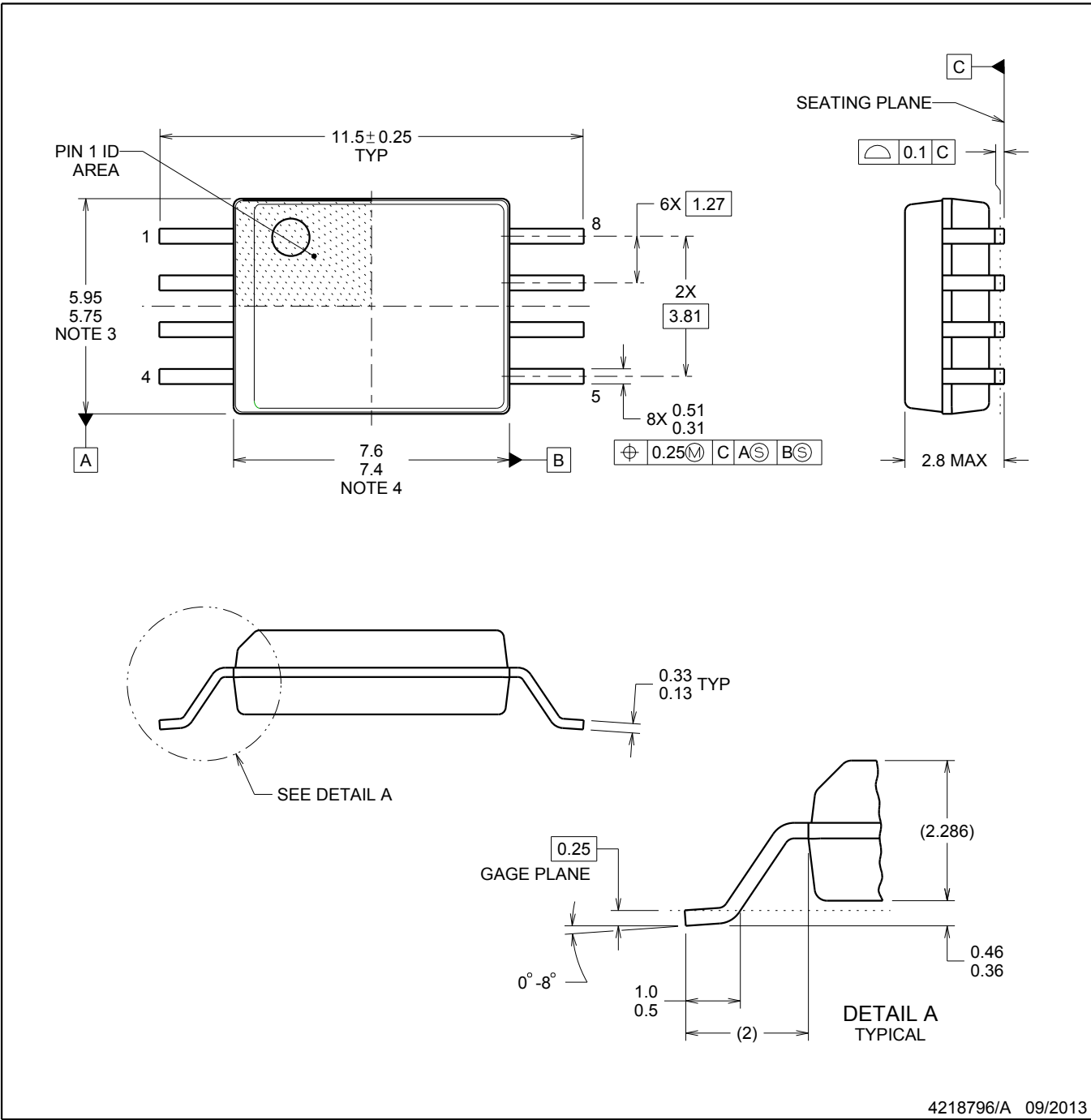
PACKAGE OUTLINE

DWV0008A



SOIC - 2.8 mm max height

SOIC



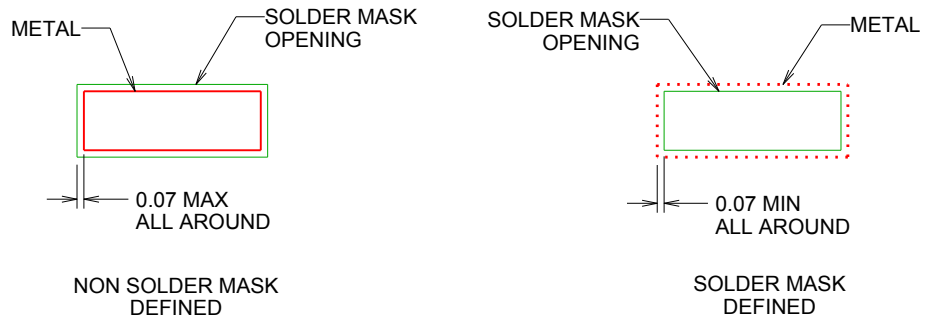
4218796/A 09/2013

NOTES:

1. All linear dimensions are in millimeters. Dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.15 mm, per side.
4. This dimension does not include interlead flash. Interlead flash shall not exceed 0.25 mm, per side.



LAND PATTERN EXAMPLE
9.1 mm NOMINAL CLEARANCE/CREEPAGE
SCALE:6X

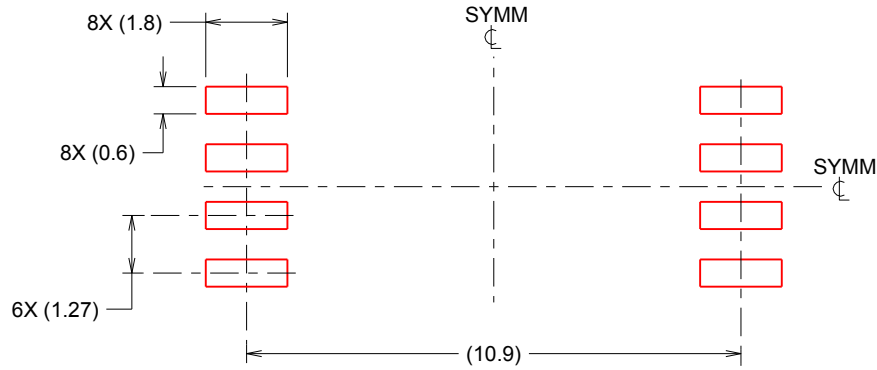


SOLDER MASK DETAILS

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NOTES: (continued)

- 5. Publication IPC-7351 may have alternate designs.
- 6. Solder mask tolerances between and around signal pads can vary based on board fabrication site.



SOLDER PASTE EXAMPLE
 BASED ON 0.125 mm THICK STENCIL
 SCALE:6X

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NOTES: (continued)

- 7. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
- 8. Board assembly site may have different recommendations for stencil design.

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