ISOM811x-Q1 3.75kV$_{RMS}$、車載用シングル チャネル オプト エミュレータ、アナログ トランジスタ出力付き

1 特長

- 業界標準のフォトトランジスタ オプトカプラに対するドロップイン互換性とピン間互換性アップグレード
- 以下の結果で AEC-Q100 認定済み:
  - デバイス温度グレード 1:動作時周囲温度範囲 -40℃～125℃
- 1 チャネルの LED エミュレータ入力
- 電流伝達率 (CTR):I_F = 5mA、V_{CE} = 5V のとき
  - ISOM8110、ISOM8115:100%～155%
  - ISOM8111、ISOM8116:150%～230%
  - ISOM8112、ISOM8117:255%～380%
  - ISOM8113、ISOM8118:375%～560%
- 高いコレクタ - エミッタ電圧:V_{CE} (max) = 80V
- 堅牢な絶縁バリア
  - 絶縁定格:3750V$_{RMS}$
  - 動作電圧:500V$_{RMS}$、707V$_{PK}$
  - サージ耐性:最大 10kV
- 応答時間:V_{CE} = 10V、I_C = 2mA、R_L = 100Ω で 3µs (標準値)
- 機能安全対応
  - 機能安全システムの設計に役立つ資料を利用可能:ISOM811x-Q1
- 安全性関連認証取得予定:
  - UL 1577 認定、3750V$_{RMS}$ の絶縁
  - VDE による DIN EN IEC 60747-17 (VDE 0884-17) 準拠
  - IEC 62368-1 認定、IEC 61010-1 認定
  - CQC GB 4943.1 認定

2 アプリケーション

- スイッチング電源
- 先進運転支援システム (ADAS)
- モータドライブの I/O および位置フィードバック
- ボディ エレクトロニクス & 照明
- インフォテインメント & クラスタ
- HEV/EV のバッテリ管理システム (BMS)

3 概要

ISOM811x-Q1 デバイスは、LED エミュレータ入力とトランジスタ出力を備えたシングル チャネルのオプトカプラ エミュレータです。本デバイスは、従来の多くのオプトカプラとピン互換であり、ドロップイン互換性があるため、PCB の再設計なしで既存システムを拡張できます。

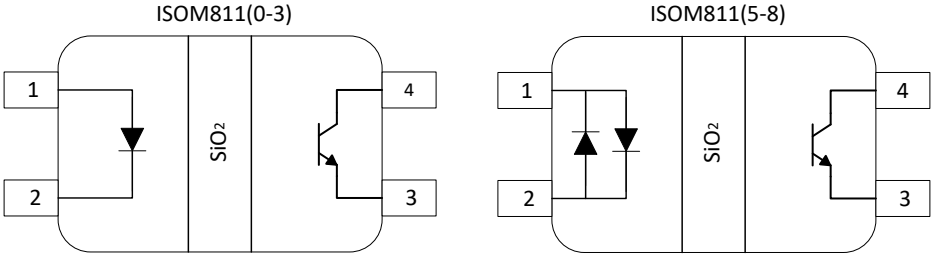
ISOM811x-Q1 オプトカプラ エミュレータは、オプトカプラと比較して信頼性が高く、高帯域幅、短いターンオフ遅延、低消費電力、広い温度範囲、厳格な CTR 制御とプロセス制御を実現しており、部品間スキューが小さい、という性能面での優位性もあります。経年変化や温度変化を補正する必要がないため、エミュレートされた LED 入力段の消費電力はオプトカプラよりも低減されます。

ISOM811x-Q1 デバイスは、2.54mm および 1.27mm ピン ピッチの小型 SOIC-4 パッケージで供給され、3.75kV$_{RMS}$ の絶縁定格で、DC (ISOM811[0-3]) および双方向 DC (ISOM811[5-8]) 入力を選択できます。ISOM811x-Q1 は性能と信頼性が高いため、電源フィードバック設計、モーター ドライブ、産業用コントローラの I/O モジュール、ファクトリ オートメーション アプリケーションなどに使用できます。

パッケージ情報

部品番号	パッケージ (1)	パッケージ サイズ (3)	本体サイズ (公称)
ISOM811x-Q1	SO-4 (DFG)	7.0mm × 3.5mm	4.8mm × 3.5mm
	SO-4 (DFH)(2)	7.0mm × 2.7mm	4.8mm × 2.7mm

(1) 詳細については、セクション 12 を参照してください。
(2) プレビュー版のみ。
(3) パッケージ サイズ (長さ × 幅) は公称値であり、該当する場合はピンも含まれます。



概略回路図

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4 Device Comparison

表 4-1. Device Selection

PART NUMBER(2)	CTR	PACKAGE(1)	PIN PITCH
ISOM8110-Q1, ISOM8115-Q1	100% to 155%	4-pin SOIC (DFG), 4-pin SOIC (DFH)	2.54mm, 1.27mm
ISOM8111-Q1, ISOM8116-Q1	150% to 230%		
ISOM8112-Q1, ISOM8117-Q1	255% to 380%		
ISOM8113-Q1, ISOM8118-Q1	375% to 560%		

(1) DFH package is preview only.
(2) ISOM8111-3 and ISOM8115-8 are preview only.

5 Pin Configuration and Functions

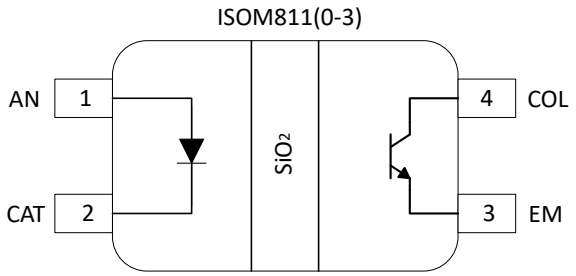


図 5-1. ISOM811(0-3) 4-Pin SOIC (Top View)

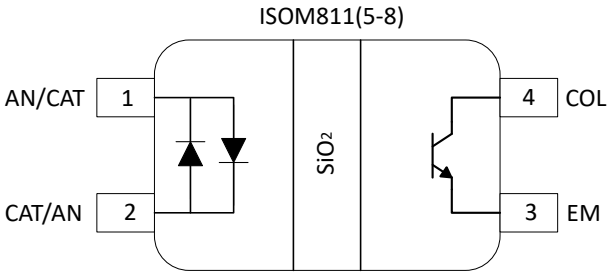


図 5-2. ISOM811(5-8) 4-Pin SOIC (Top View)

表 5-1. Pin Functions

PIN		TYPE(1)	DESCRIPTION
NO.	NAME		
1	AN	I	Anode connection of input LED emulator
2	CAT	I	Cathode connection of input LED emulator
3	EM	O	Emitter for transistor
4	COL	O	Collector for transistor

(1) I = Input, O = Output

6 Specifications

6.1 Absolute Maximum Ratings

See(1) (2)

			MIN	MAX	UNIT
$I_{F(max)}$	Maximum Input forward current	ISOM8110, ISOM8111, ISOM8112, ISOM8113		50	mA
		ISOM8115, ISOM8116, ISOM8117, ISOM8118		±50	mA
V_{CEO}	Collector-emitter voltage	ISOM8110, ISOM8111, ISOM8112, ISOM8113		80	V
		ISOM8115, ISOM8116, ISOM8117, ISOM8118		80	V
V_{ECO}	Emitter-collector voltage	ISOM8110, ISOM8111, ISOM8112, ISOM8113		7	V
		ISOM8115, ISOM8116, ISOM8117, ISOM8118		7	V
I_{FP}	Input pulse forward current (1µs width)	ISOM8110, ISOM8111, ISOM8112, ISOM8113		1	A
I_{FP}	Input pulse forward current (1µs width)	ISOM8115, ISOM8116, ISOM8117, ISOM8118		±1	A
V_R	Input reverse voltage at I_R = 10µA	ISOM8110, ISOM8111, ISOM8112, ISOM8113		7	V
P_I	Input power dissipation	ISOM8110, ISOM8111, ISOM8112, ISOM8113		140	mW
		ISOM8115, ISOM8116, ISOM8117, ISOM8118		140	mW
I_C	Collector current	ISOM8110, ISOM8111, ISOM8112, ISOM8113		50	mA
		ISOM8115, ISOM8116, ISOM8117, ISOM8118		50	mA
P_C	Collector power dissipation	ISOM8110, ISOM8111, ISOM8112, ISOM8113		150	mW
		ISOM8115, ISOM8116, ISOM8117, ISOM8118		150	mW
P_T	Total power dissipation	ISOM8110, ISOM8111, ISOM8112, ISOM8113		290	mW
		ISOM8115, ISOM8116, ISOM8117, ISOM8118		290	mW
T_A	Ambient temperature	ISOM8110, ISOM8111, ISOM8112, ISOM8113	–55	125	°C
		ISOM8115, ISOM8116, ISOM8117, ISOM8118	–55	125	°C
T_J	Operating junction temperature	ISOM8110, ISOM8111, ISOM8112, ISOM8113		150	°C
		ISOM8115, ISOM8116, ISOM8117, ISOM8118		150	°C

(1) Operation outside the Absolute Maximum Ratings may cause permanent device damage. Absolute Maximum Ratings do not imply functional operation of the device at these or any other conditions beyond those listed under the operational sections of this document. If used outside the listed operational conditions but within the Absolute Maximum Ratings, the device may not be fully functional, and this may affect device reliability, functionality, performance, and shorten the device lifetime.

(2) All specifications are at T_A = 25℃ unless otherwise noted

6.2 ESD Ratings

			VALUE	UNIT
$V_{(ESD)}$	Electrostatic discharge	Human-body model (HBM), per AEC Q100-002[1] HBM ESD Classification Level 3A	±2000	V
		Charged-device model (CDM), per AEC Q100-011 CDM ESD Classification Level C6	±1000	

(1) AEC Q100-002 indicates that HBM stressing shall be in accordance with the ANSI/ESDA/JEDEC JS-001 specification.

6.3 Recommended Operating Conditions

		MIN	NOM	MAX	UNIT
$I_{F(ON)}$	Input ON-state forward current	0.7		20	mA
V_{CEO}	Collector-emitter voltage	-5		48	V
T_A	Ambient temperature	-40		125	°C

6.4 Thermal Information

THERMAL METRIC[1]		ISOM811x		UNIT
		DFG (SOIC)	DFH (SOIC)	
		4 PINS	4 PINS	
$R_{\theta JA}$	Junction-to-ambient thermal resistance	283.9	288.8	°C/W
$R_{\theta JC(top)}$	Junction-to-case (top) thermal resistance	173.1	173.6	°C/W
$R_{\theta JB}$	Junction-to-board thermal resistance	201.4	192.9	°C/W
ψ_{JT}	Junction-to-top characterization parameter	125.1	121.5	°C/W
ψ_{JB}	Junction-to-board characterization parameter	198.0	190.0	°C/W

(1) For more information about traditional and new thermal metrics, see Semiconductor and IC Package Thermal Metrics applicaiton note.

6.5 Insulation Specifications

PARAMETER		TEST CONDITIONS	VALUE	UNIT
			4-DFG, 4-DFH	
IEC 60664-1				
CLR	External clearance(1)	Side 1 to side 2 distance through air	> 5	mm
CPG	External creepage(1)	Side 1 to side 2 distance across package surface	> 5	mm
DTI	Distance through the insulation	Minimum internal gap (internal clearance)	>17	µm
CTI	Comparative tracking index	IEC 60112; UL 746A	>400	V
	Material Group	According to IEC 60664-1	II	
	Overvoltage category per IEC 60664-1	Rated mains voltage ≤ 150 V_{RMS}	I-IV	
		Rated mains voltage ≤ 300 V_{RMS}	I-IV	
		Rated mains voltage ≤ 600 V_{RMS}	I-III	
DIN VDE V 0884-11:2017 (6)				
V_{IORM}	Maximum repetitive peak isolation voltage	AC voltage (bipolar)	707	V_{PK}
V_{IOWM}	Maximum isolation working voltage	AC voltage (sine wave); time-dependent dielectric breakdown (TDDB) test	500	V_{RMS}
		DC voltage	707	V_{DC}
V_{IOTM}	Maximum transient isolation voltage	$V_{TEST} = V_{IOTM}$, t = 60s (qualification); $V_{TEST} = 1.2 \times V_{IOTM}$, t = 1s (100% production)	5303	V_{PK}
V_{IMP}	Maximum impulse voltage (2)	Tested in air, 1.2/50µs waveform per IEC 62368-1	7200	V_{PK}
V_{IOSM}	Maximum surge isolation voltage(3)	$V_{ISOM} \geq 1.3 \times V_{IMP}$; tested in oil (qualification test), 1.2/50µs waveform per IEC 62368-1	10000	V_{PK}
q_{pd}	Apparent charge(4)	Method a: After I/O safety test subgroup 2/3, $V_{ini} = V_{IOTM}$, t_{ini} = 60s; $V_{pd(m)} = 1.2 \times V_{IORM}$, t_m = 10s	≤ 5	pC
		Method a: After environmental tests subgroup 1, $V_{ini} = V_{IOTM}$, t_{ini} = 60s; $V_{pd(m)} = 1.6 \times V_{IORM}$, t_m = 10s	≤ 5	
		Method b: At routine test (100% production) and preconditioning (type test), $V_{ini} = 1.2 \times V_{IOTM}$, t_{ini} = 1s; $V_{pd(m)} = 1.875 \times V_{IORM}$, t_m = 1s	≤ 5	
C_{IO}	Barrier capacitance, input to output(5)	V_{IO} = 0.4 × sin (2 πft), f = 1MHz	1	pF
R_{IO}	Insulation resistance, input to output(5)	V_{IO} = 500V, T_A = 25°C	> 10^{12}	Ω
		V_{IO} = 500V, 100°C ≤ T_A ≤ 125°C	> 10^{11}	
		V_{IO} = 500V at T_S = 150°C	> 10^{9}	
	Pollution degree		2	
	Climatic category		40/125/21	
UL 1577				
V_{ISO}	Withstand isolation voltage	$V_{TEST} = V_{ISO}$, t = 60s (qualification); $V_{TEST} = 1.2 \times V_{ISO}$, t = 1s (100% production)	3750	V_{RMS}

(1) Creepage and clearance requirements should be applied according to the specific equipment isolation standards of an application. Care should be taken to maintain the creepage and clearance distance of a board design to ensure that the mounting pads of the isolator on the printed-circuit board do not reduce this distance. Creepage and clearance on a printed-circuit board become equal in certain cases. Techniques such as inserting grooves, ribs, or both on a printed circuit board are used to help increase these specifications.
(2) Testing is carried out in air to determine the surge immunity of the package.
(3) Testing is carried out in oil to determine the intrinsic surge immunity of the isolation barrier.
(4) Apparent charge is electrical discharge caused by a partial discharge (pd).
(5) All pins on each side of the barrier tied together creating a two-pin device.

(6) This coupler is suitable for *safe electrical insulation only* within the safety ratings. Compliance with the safety ratings shall be ensured by means of suitable protective circuits.

6.6 Safety-Related Certifications

VDE	CSA	UL	CQC	TUV
Plan to certify according to DIN EN IEC 60747-17 (VDE 0884-17)	Plan to certify according to IEC 61010-1, IEC 62368-1 and IEC 60601-1	Plan to certify according to UL 1577 Component Recognition Program	Plan to certify according to GB4943.1-2011	Plan to certify according to EN 61010-1:2010/A1:2019 and EN 62368-1:2014
Certificate planned	Certificate planned	Certificate planned	Certificate planned	Certificate planned

6.7 Safety Limiting Values

Safety limiting(1) intends to minimize potential damage to the isolation barrier upon failure of input or output circuitry.

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
SO-4 PACKAGE (DFG)						
I_S	Safety limiting input current	$R_{\theta JA}$= 283.9°C/W, V_F = 1.4V, T_J = 150°C, T_A = 25°C			300	mA
		$R_{\theta JA}$=283.9°C/W, V_{CEO} = 40V, T_J = 150°C, T_A = 25°C			10.5	mA
		$R_{\theta JA}$= 283.9°C/W, V_{CEO} = 24V, T_J =150°C, T_A = 25°C			17.5	mA
		$R_{\theta JA}$= 283.9°C/W, V_{CEO} = 15V, T_J = 150°C, T_A = 25°C			28	mA
P_S	Safety limiting total power	$R_{\theta JA}$= 283.9°C/W, T_J = 150°C, T_A = 25°C			420	mW
T_S	Maximum safety temperature				150	°C

(1) The I_S and P_S parameters represent the safety current and safety power respectively. The maximum limits of I_S and P_S must not be exceeded. These limits vary with the ambient temperature, T_A.
The junction-to-air thermal resistance, $R_{\theta JA}$, in the table is that of a device installed on a high-K test board for leaded surface-mount packages. Use these equations to calculate the value for each parameter:
$T_J = T_A + R_{\theta JA} \times P$, where P is the power dissipated in the device.
$T_{J(max)} = T_S = T_A + R_{\theta JA} \times P_S$, where $T_{J(max)}$ is the maximum allowed junction temperature.
$P_S = I_S \times V_I$, where V_I is the maximum input voltage.

6.8 Electrical Characteristics

All specifications are at T_A = 25℃ unless otherwise noted

PARAMETER		TEST CONDITIONS	GPN	MIN	TYP	MAX	UNIT
INPUT							
V_F	Input forward voltage	I_F = 5mA	ISOM8110, ISOM8111, ISOM8112, ISOM8113		1.2	1.4	V
V_F	Input forward voltage	I_F = 5mA, T_A = -40 to 125°C	ISOM8110, ISOM8111, ISOM8112, ISOM8113		1.2	1.6	V
V_F	Input forward voltage	I_F = ±5mA	ISOM8115, ISOM8116, ISOM8117, ISOM8118		1.2	1.5	V
V_F	Input forward voltage	I_F = ±5mA, T_A = -40 to 125°C	ISOM8115, ISOM8116, ISOM8117, ISOM8118		1.2	1.5	V
I_R	Input reverse current	V_R = 5V, T_A = -40 to 125°C	ISOM8110, ISOM8111, ISOM8112, ISOM8113			10	µA
C_{IN}	Input capacitance	At 1MHz, V_F = 0V	ISOM8110, ISOM8111, ISOM8112, ISOM8113		35		pF
C_{IN}	Input capacitance	At 1MHz, V_F = 0V	ISOM8115, ISOM8116, ISOM8117, ISOM8118		6		pF
OUTPUT							
C_{CE}	Collector-emitter capacitance	1MHz, V_F = 0V	ISOM811x		12		pF
$V_{CE(SAT)}$	Collector-emitter saturation voltage	I_F = 20mA, I_C = 1mA, T_A = -40 to 125°C	ISOM8110, ISOM8111, ISOM8112, ISOM8113			0.3	V
$V_{CE(SAT)}$	Collector-emitter saturation voltage	I_F = ±20mA, I_C = 1mA, T_A = -40 to 125°C	ISOM8115, ISOM8116, ISOM8117, ISOM8118			0.3	V
I_{C_DARK}	Collector dark current	V_{CE} = 20V, I_F = 0mA, T_A = -40 to 125°C	ISOM811x			100	nA
I_{EC}	Reverse current	V_{EC} = 7V, I_F = 0mA, T_A = -40 to 125°C	ISOM811x			100	µA
I_{C_OFF}	OFF_state collector current	V_F = 0.7V, V_{CE} = 48V, T_A = -40 to 125°C	ISOM8110, ISOM8111, ISOM8112, ISOM8113			10	µA
I_{C_OFF}	OFF_state collector current	V_F = ±0.7V, V_{CE} = 48V, T_A = -40 to 125°C	ISOM8115, ISOM8116, ISOM8117, ISOM8118			10	µA
CTR(1)							

All specifications are at T_A = 25℃ unless otherwise noted

PARAMETER		TEST CONDITIONS	GPN	MIN	TYP	MAX	UNIT
CTR	Current Transfer Ratio	I_F = 2mA, V_{CE} = 5V, T_A = -40 to 125°C	ISOM8110	80	130	180	%
			ISOM8115	80	130	180	%
			ISOM8111	120	180	270	%
			ISOM8116	120	180	270	%
			ISOM8112	200	300	450	%
			ISOM8117	200	300	450	%
			ISOM8113	295	440	655	%
			ISOM8118	295	440	655	%
CTR	Current Transfer Ratio	I_F = 5mA, V_{CE} = 5V, T_A = -40 to 125°C	ISOM8110	100	120	155	%
			ISOM8115	100	120	155	%
			ISOM8111	150	180	230	%
			ISOM8116	150	180	230	%
			ISOM8112	255	300	380	%
			ISOM8117	255	300	380	%
			ISOM8113	375	440	560	%
			ISOM8118	375	440	560	%

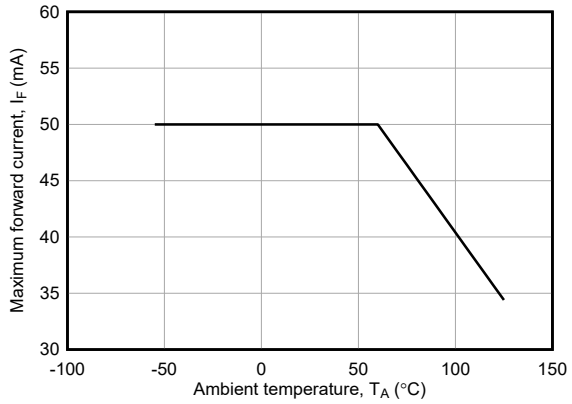
(1) CTR (%) = (I_C / I_F) x 100%

6.9 Switching Characteristics

All specifications are at T_A = 25℃ unless otherwise noted

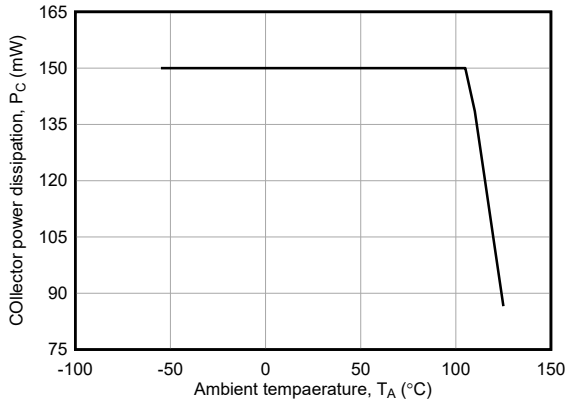
PARAMETER		TEST CONDITIONS	GPN	MIN	TYP	MAX	UNIT
AC							
t_r	Rise time, see 図 7-2 and 図 7-3	V_{CC} = 10V, I_C = 2mA, R_L= 100Ω, C_L = 50pF	ISOM8110		3.2		µs
			ISOM8113		1.1		µs
t_f	Fall time, see 図 7-2 and 図 7-3	V_{CC} = 10V, I_C = 2mA, R_L= 100Ω, C_L = 50pF	ISOM8110		4.0		µs
			ISOM8113		7.5		µs
T_{ON}	Turn on time, see 図 7-2 and 図 7-3	V_{CC} = 10V, I_C = 2mA, R_L= 100Ω, C_L = 50pF	ISOM8110, ISOM8115		5.7		µs
			ISOM8111, ISOM8116		9.5		µs
			ISOM8112, ISOM8117		8.1		µs
			ISOM8113, ISOM8118		20		µs
T_{OFF}	Turn off time, see 図 7-2 and 図 7-3	V_{CC} = 10V, I_C = 2mA, R_L= 100Ω, C_L = 50pF	ISOM8110, ISOM8115		3.6		µs
			ISOM8111, ISOM8116		2.3		µs
			ISOM8112, ISOM8117		1.7		µs
			ISOM8113, ISOM8118		0.68		µs
t_s	Storage time; time required for the output waveform to change from 0% (100%) to 10% (90%) when input is turned on and back off, see 図 7-3	V_{CC} = 5V, I_F = 1.6mA, R_L = 4.7kΩ	ISOM811x			21	µs
BW	Bandwidth, see 図 7-4 and 図 7-5	V_{IN_DC} = 5V, V_{IN_AC} = 1Vpk, R_{IN} = 2kΩ V_{CC}= 5V, R_{LOAD} = 100Ω, C_L = 50pF, measured at V_{CE} –3dB sinewave	ISOM8110, ISOM8115		680		kHz
			ISOM8111, ISOM8116		680		kHz
			ISOM8112, ISOM8117		680		kHz
			ISOM8113, ISOM8118		680		kHz

6.10 Typical Characteristics



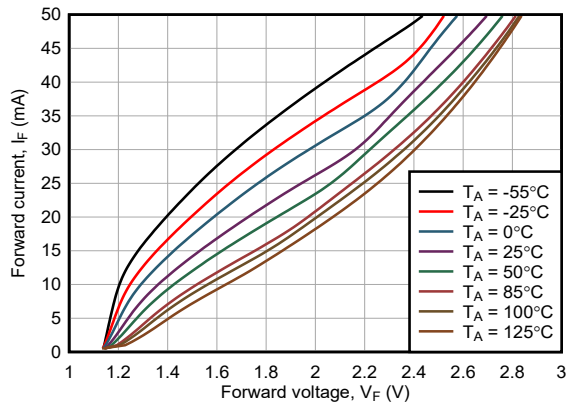
ISOM811x

図 6-1. Maximum Forward Current vs Ambient Temperature



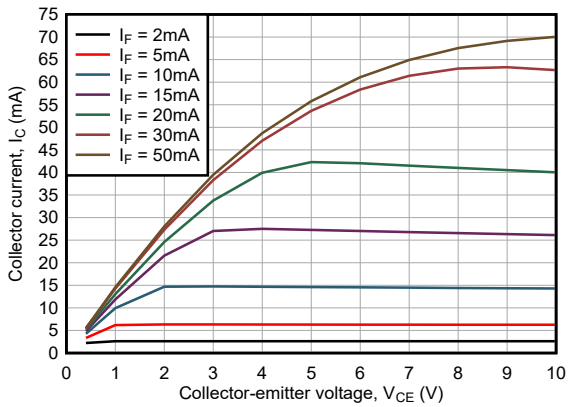
ISOM811x

図 6-2. Maximum Collector Power Dissipation vs Ambient Temperature



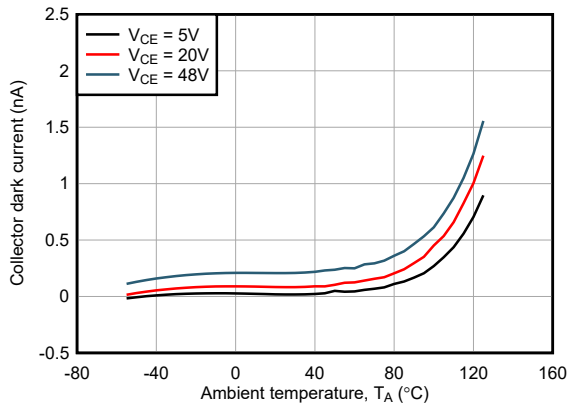
ISOM8110

図 6-3. Forward Voltage vs Forward Current



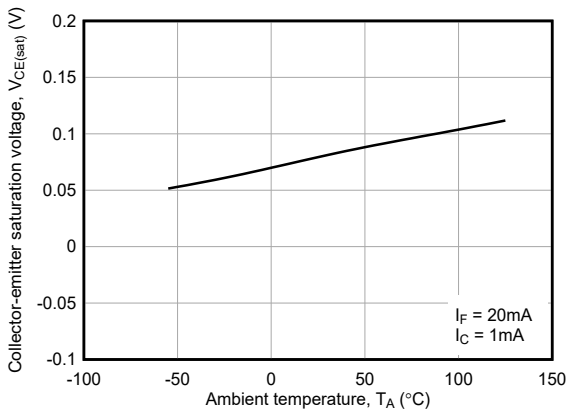
ISOM8110 T_A = 25°C

図 6-4. Collector Current vs Collector-Emitter Voltage



ISOM8110

図 6-5. Collector Dark Current vs Ambient Temperature



ISOM8110

図 6-6. Collector-Emitter Saturation Voltage vs Ambient Temperature

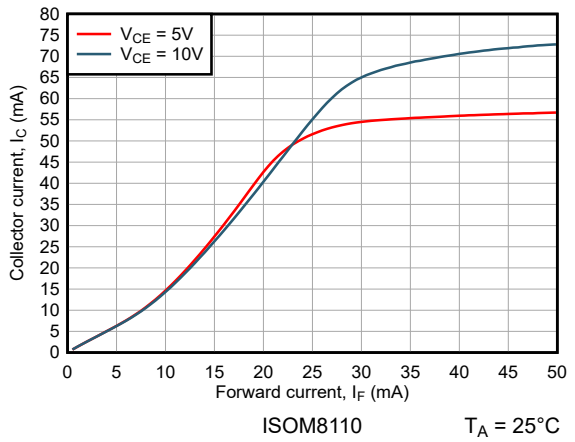


図 6-7. Collector Current vs Forward Current

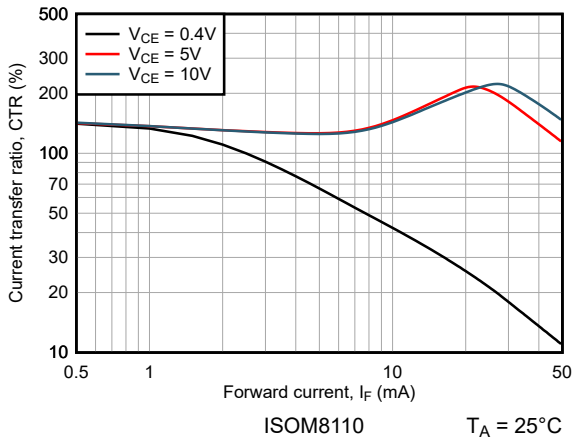


図 6-8. Current Transfer Ratio vs Forward Current

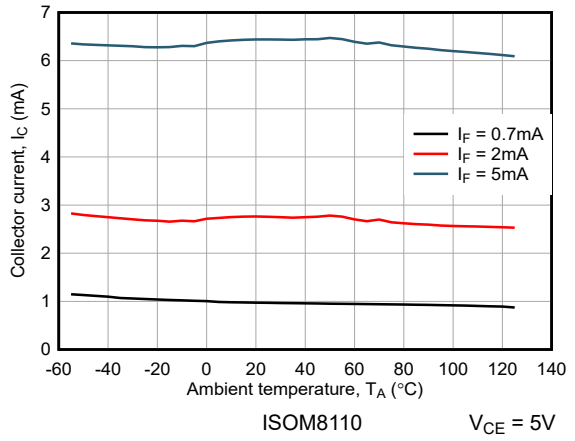


図 6-9. Collector Current vs Ambient Temperature

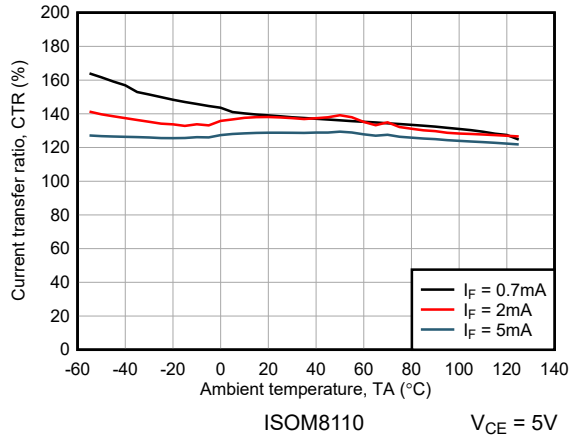


図 6-10. Current Transfer Ratio vs Ambient Temperature

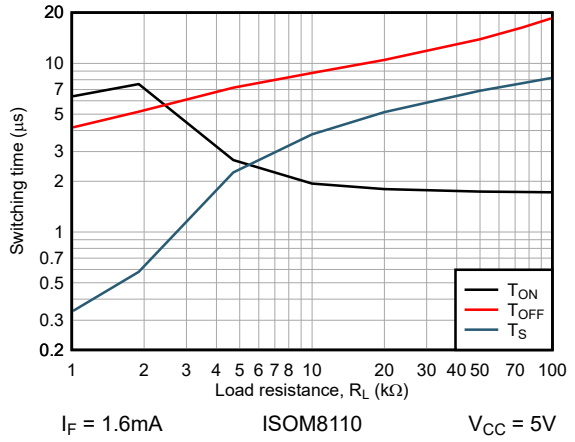


図 6-11. Switching Time vs Load Resistance

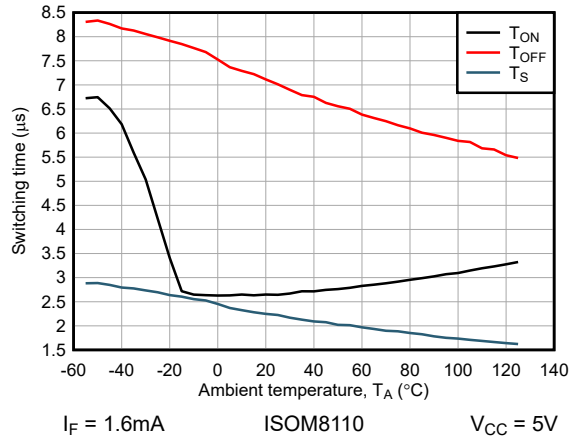


図 6-12. Switching Time vs Ambient Temperature

7 Parameter Measurement Information

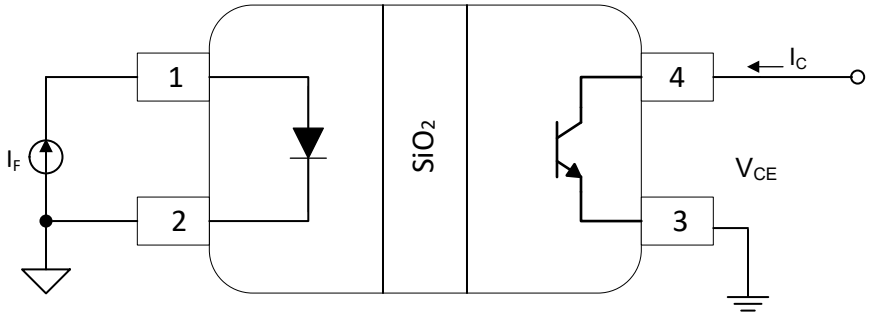


図 7-1. ISOM811x Test Circuit for CTR

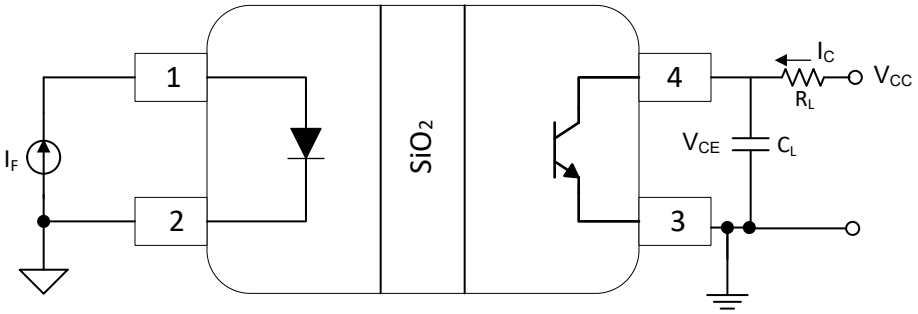


図 7-2. ISOM811x Test Circuit for Switching Timing

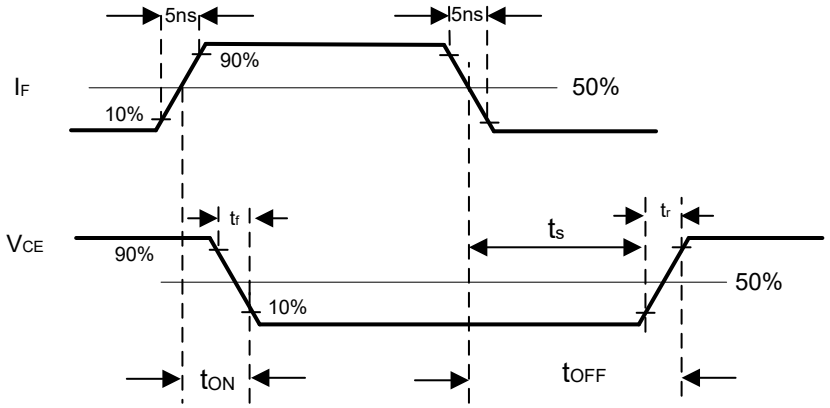


図 7-3. ISOM811x Switching Timing Waveforms

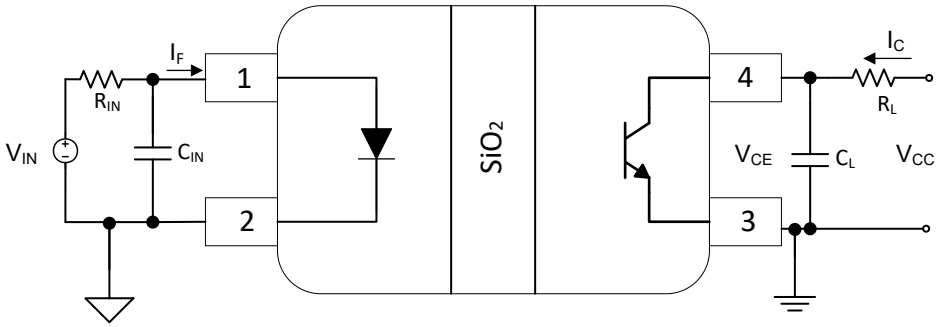


図 7-4. ISOM811(0-3) Test Circuit for Bandwidth

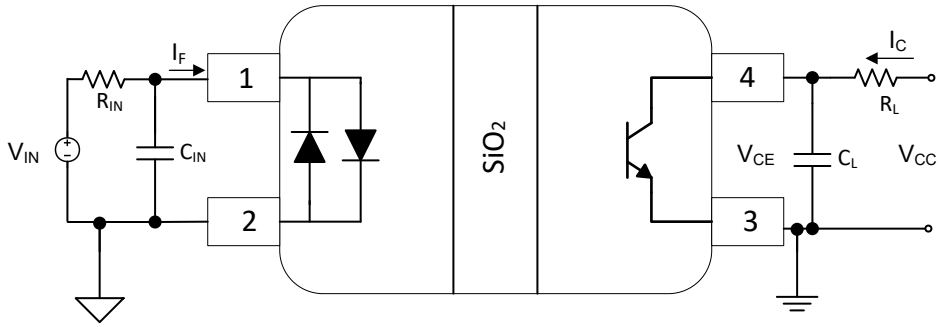


図 7-5. ISOM811(5-8) Test Circuit for Bandwidth

8 Detailed Description

8.1 Overview

The ISOM811x-Q1 opto-emulators are pin-compatible, single-channel, drop-in replacements for many traditional optocouplers. While standard optocouplers use an LED as the input stage, ISOM811x-Q1 uses an emulated LED as the input stage. The input and output stages are isolated by TI's proprietary silicon dioxide-based (SiO_2) isolation barrier. This isolation technology makes ISOM811x-Q1 resistant to the wear-out effects found in optocouplers that degrade performance with increasing temperature, forward current, and device age. Ordering options include four different ranges of current transfer ratio (CTR) and input options supporting uni-polar and bi-polar DC flow.

The ISOM811x-Q1 family of devices isolate DC and bidirectional DC signals and offer performance, reliability, and flexibility advantages not available with traditional optocouplers.

The functional block diagram of ISOM811x-Q1 devices are shown in セクション 8.2. The input signal is transmitted across the isolation barrier using an on-off keying (OOK) modulation scheme. The transmitter sends a high-frequency carrier across the barrier that contains information on how much current is flowing through the input pins. The receiver demodulates the signal after advanced signal conditioning and produces the signal through the output stage. These devices also incorporate advanced circuit techniques to maximize bandwidth and minimize radiated emissions. 図 8-3 shows conceptual details of how the OOK scheme works.

8.2 Functional Block Diagram

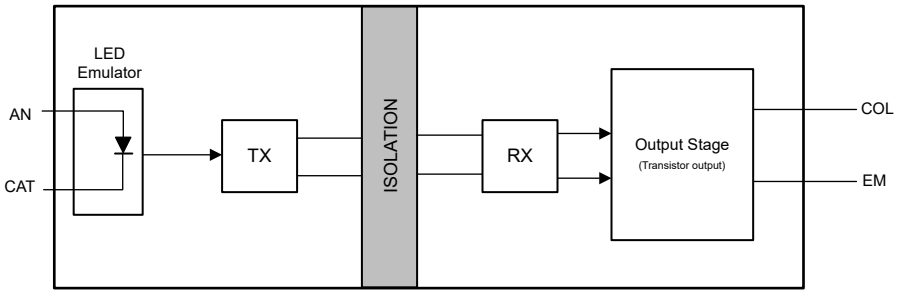


図 8-1. Conceptual Block Diagram of an Opto-emulator ISOM811(0-3)

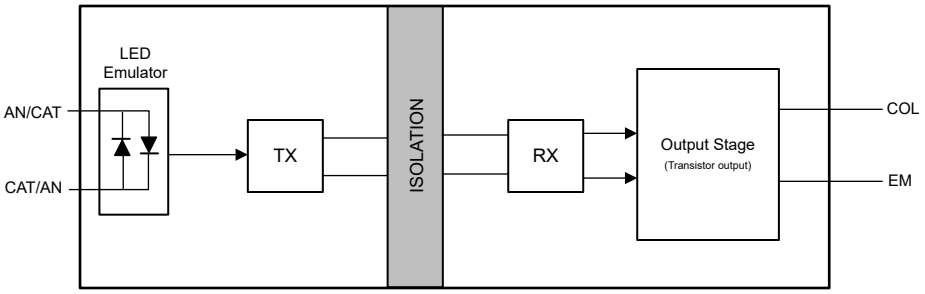


図 8-2. Conceptual Block Diagram of an Opto-emulator ISOM811(5-8)

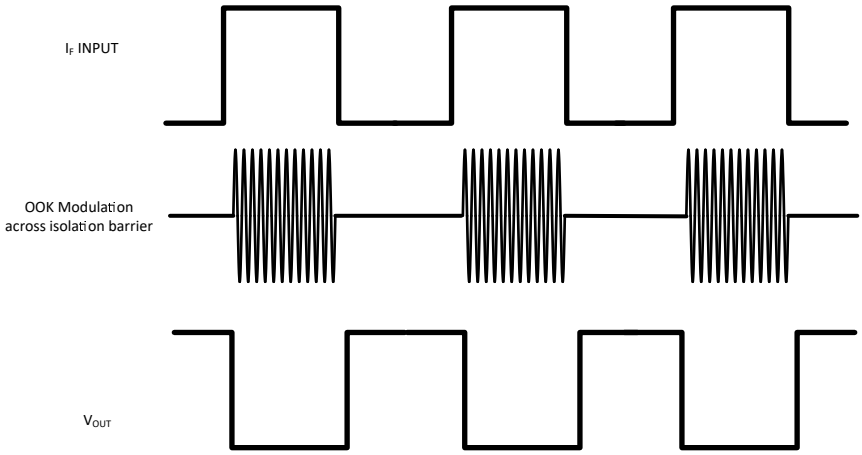


図 8-3. On-off Keying (OOK) Based Modulation Scheme

8.3 Feature Description

The ISOM811x-Q1 devices isolate DC and bidirectional DC signals. ISOM811x-Q1 has an open-collector output with multiple CTR options. All devices support an isolation withstand voltage of 3750V_{RMS} between side 1 and side 2.

8.4 Device Functional Modes

表 8-1 lists the functional modes for the ISOM811x-Q1 devices.

表 8-1. Function Table

CTR[1]	PART NUMBER	Input type
100% to 155%	ISOM8110	DC
	ISOM8115	Bidirectional DC
150% to 230%	ISOM8111	DC
	ISOM8116	Bidirectional DC
255% to 380%	ISOM8112	DC
	ISOM8117	Bidirectional DC
375% to 560%	ISOM8113	DC
	ISOM8118	Bidirectional DC

1. I_F = 5mA, T_A = 25°C, V_{CE} = 5V.

9 Application and Implementation

注

以下のアプリケーション情報は、TI の製品仕様に含まれるものではなく、TI ではその正確性または完全性を保証いたしません。個々の目的に対する製品の適合性については、お客様の責任で判断していただくことになります。お客様は自身の設計実装を検証しテストすることで、システムの機能を確認する必要があります。

9.1 Application Information

The ISOM811x-Q1 devices are single-channel opto-emulators with LED-emulator input and transistor output. The devices use on-off keying modulation to transmit data across the isolation barrier. The input stage is isolated from the driver stage by TI's proprietary silicon dioxide-based (SiO_2) isolation barrier which provides robust isolation. With wider temperature ratings than traditional optocouplers, ISOM811x-Q1 opto-emulators can provide reliable signal isolation in harsh environments.

The ISOM811x-Q1 devices are capable of sinking current when subjected to an external load being connected to the device. Like typical transistor output optocouplers, the output current depends on the input current level (I_F) and the current transfer ratio (CTR). With multiple CTR options (100% - 560%), low input current, high bandwidth, low turn-off delay, low power consumption, and wider temperature range, ISOM811x-Q1 devices are designed for use in a variety of industries such as factory automation, building automation, e-mobility, automotive, avionics, medical, and power delivery.

9.1.1 Typical Application

ISOM811x-Q1 opto-emulators are commonly used in the feedback control loops of isolated power supplies. These devices are used to solve the problem of feeding back current while isolating the primary and secondary domains to regulate the output voltage.

In power supplies, the output voltage is isolated from main input voltage using a transformer (for example: flyback converter). For analog power supply units, the controller IC is usually on the primary side of the transformer. For closed loop control, measuring the output voltage on the secondary side and feeding the voltage back to the controller on the primary is necessary. The most common method of achieving this design is using an opto-emulator such as ISOM811x-Q1, error amplifier (commonly TL431), and a voltage comparator to form a feedback loop across the isolation barrier

図 9-1 illustrates a typical isolated power supply. In this implementation, the output voltage is sensed by an error amplifier using the resistor divider (R1 and R2). Depending on the voltage level that the error amplifier senses, the TL431 can drive the current of the ISOM811x-Q1 higher or lower which is then compared to a voltage reference. The information is passed across the isolation barrier through ISOM811x-Q1 to the primary side, where the PWM control circuit modulates the power stage to regulate the output voltage. The TL431 and ISOM811x-Q1 play an important role for stable feedback and control loop.

The ISOM811x-Q1 devices enable improvements in transient response, reliability, and stability as compared to commonly used optocoupler as the CTR is stable over wide temperature range providing a small, low-cost, highly reliable, and easy-to-design implementation.

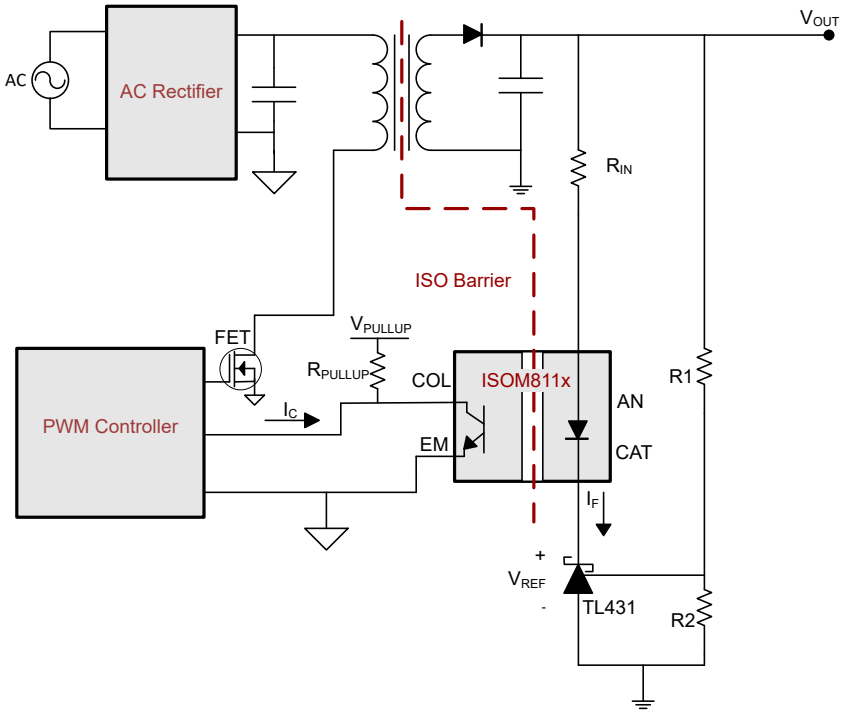


図 9-1. Typical Isolated Power Supply Application Using ISOM811x-Q1

9.1.1.1 Design Requirements

To design with ISOM811x-Q1 devices, use the parameters listed in 表 9-1.

表 9-1. Design Parameters

PARAMETER	VALUE
Input forward current range, I_F	0.7mA (min), 20mA (max)
Current transfer ratio at I_F = 5mA, CTR	100% to 155%
Collector current tolerance, I_C	50mA (max)
Collector-emitter voltage (saturation), $V_{CE(SAT)}$	0.3V (max)
Input forward voltage, V_F	1.2V (typ)

9.1.1.2 Detailed Design Procedure

This section presents the design procedure for using the ISOM811x-Q1 opto-emulators. External components must be selected to operate ISOM811x-Q1 within the *Recommended Operating Conditions*. The following recommendations on component selection focus on the design of a typical feedback control loop for an isolated flyback converter.

When using an optocoupler in a feedback control loop for an isolated power supply, many variables can affect how to properly use the optocoupler, including the output voltage of the power supply and the type of controller the feedback signal is being sent to. For this example, assume that the output voltage of this power supply, V_{OUT}, is 5V, and the PWM controller being used has an integrated error amplifier with a COMP pin that acts as the output of this amplifier.

9.1.1.2.1 Sizing R_{PULLUP}

The transistor output of ISOM811x-Q1 operates in active, saturation, reverse, and cut-off regions, just like a regular transistor. To verify that the output does not get damaged when the output is saturated, the minimum value of R_{PULLUP} can be calculated for a given pull-up voltage, V_{PULLUP}, in 式 1:

$$R_{PULLUP} > \frac{V_{PULLUP} - V_{CE(SAT)}}{I_{C(MAX)}} \quad (1)$$

For the example of a feedback loop application, we can calculate the minimum required value for R_{PULLUP} for a given V_{PULLUP} of 10V, the maximum output voltage of the error amplifier ($V_{COMP(MAX)}$) of 2.5V, and the maximum output current of the error amplifier is internally clamped at 1.6mA. The equation to calculate R_{PULLUP} is shown in 式 2:

$$R_{PULLUP} > \frac{V_{PULLUP} - V_{COMP(MAX)}}{I_{COMP(CLAMP)}} = \frac{10V - 2.5V}{1.6mA} = 4.66k\Omega \quad (2)$$

9.1.1.2.2 Sizing R_{IN}

The input side of ISOM811x-Q1 is current-driven. To limit the amount of current flowing into the AN pin, placing a series resistor, R_{IN}, in series with the input as shown in 図 9-1 is recommended.

Depending on how the ISOM811x-Q1 device is being used, the value of R_{IN} can vary quite a bit. However, at a high level, to make sure the input does not get damaged, the minimum value of R_{IN} can be calculated for a given input voltage, V_{IN}, in 式 3:

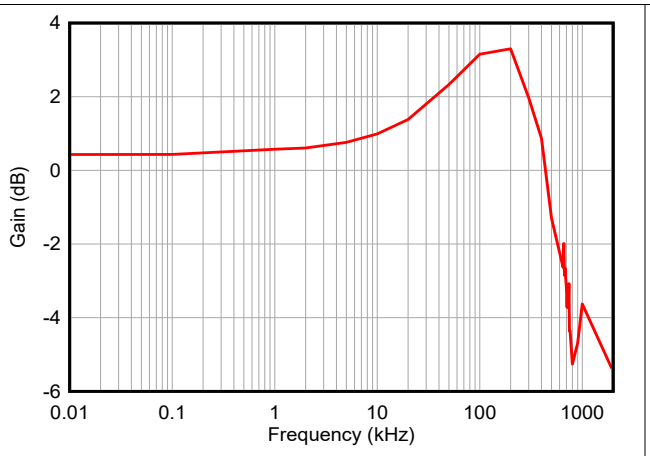
$$R_{IN} > \frac{V_{IN} - V_F}{I_{C(MAX)}} \quad (3)$$

However, in the use case of a feedback loop, R_{IN} directly affects the mid-band gain of the loop. Assuming that the TL431 has been configured to give a reference voltage, V_{REF}, of 2.5V and R_{PULLUP} is 5kΩ, 式 4 is used to calculate the maximum value of R_{IN} verifying that the V_{COMP} voltage on the primary side can be pulled to the saturation voltage of the ISOM811x-Q1, $V_{CE(SAT)}$.

$$R_{IN} < \frac{(V_{OUT} - V_{REF} - V_F) \times R_{PULLUP} \times CTR_{MIN}}{V_{PULLUP} - V_{CE(SAT)}} = \frac{(5V - 2.5V - 1.2V) \times 5k\Omega \times 100\%}{10V - 0.3V} = 670\Omega \quad (4)$$

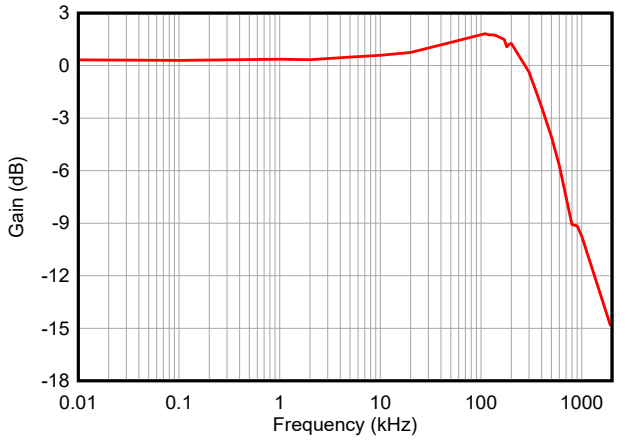
9.1.1.3 Application Curves

The following curves show ISOM8110 bandwidth performance over different loading conditions where V_{IN} = $5V_{DC}$ + $2V_{PK}$. See 図 7-4 for setup details.



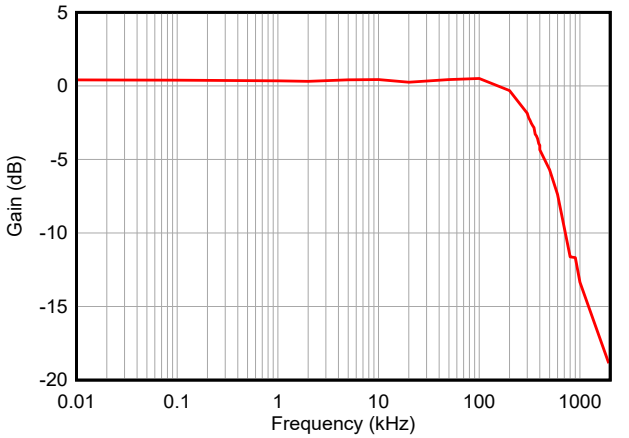
C_{IN} = 5pF R_{IN} = 2kΩ C_L = 50pF R_L = 100Ω V_{CC} = 5V

図 9-2. Bandwidth at R_L = 100Ω



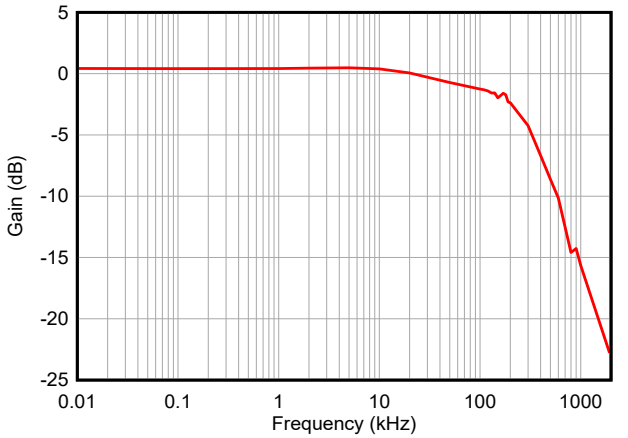
C_{IN} = 5pF R_{IN} = 2kΩ C_L = 50pF R_L = 1kΩ V_{CC} = 8V

図 9-3. Bandwidth at R_L = 1kΩ



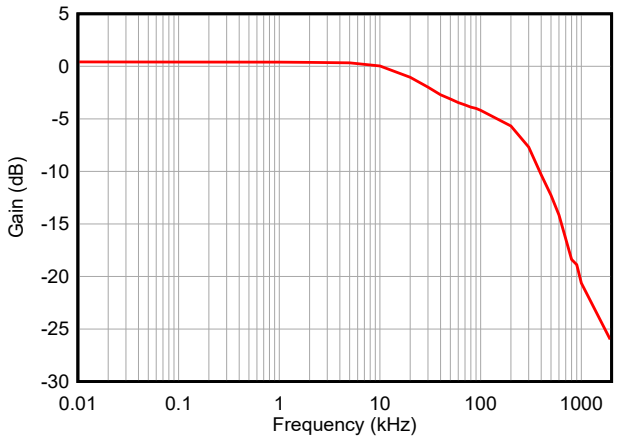
C_{IN} = 5pF R_{IN} = 2kΩ C_L = 50pF R_L = 2kΩ V_{CC} = 10V

図 9-4. Bandwidth at R_L = 2kΩ



C_{IN} = 5pF R_{IN} = 2kΩ C_L = 50pF R_L = 4.7kΩ V_{CC} = 10V

図 9-5. Bandwidth at R_L = 4.7kΩ



C_{IN} = 5pF R_{IN} = 2kΩ C_L = 50pF R_L = 10kΩ V_{CC} = 45V

図 9-6. Bandwidth at R_L = 10kΩ

9.2 Power Supply Recommendations

ISOM811x-Q1 does not require a dedicated power supply to operate since there is no supply pin. Take care to not violate recommended I/O specifications for proper device functionality.

9.3 Layout

9.3.1 Layout Guidelines

- The device connections to ground must be tied to the PCB ground plane using a direct connection or two vias to help minimize inductance.
- The connections of capacitors and other components to the PCB ground plane must use a direct connection or two vias for minimum inductance.

9.3.2 Layout Example

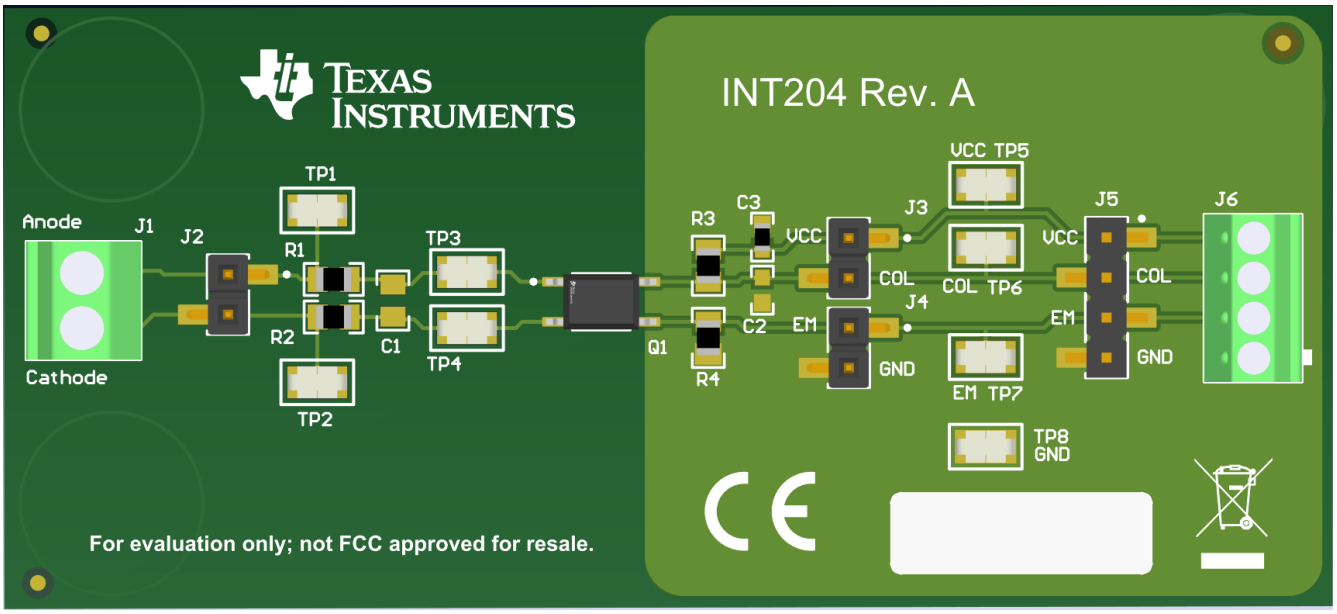


図 9-7. Layout Example of ISOM811x With a Single Layer Board

10 Device and Documentation Support

10.1 Documentation Support

10.1.1 Related Documentation

For related documentation see the following:

- Texas Instruments, *Isolation Glossary*, application note
- Texas Instruments, *Introduction to Opto-Emulators*, application note
- Texas Instruments, *ISOM8110 Single-Channel Opto-Emulator with Analog Transistor Output Evaluation Module*, EVM user's guide

10.2 ドキュメントの更新通知を受け取る方法

ドキュメントの更新についての通知を受け取るには、www.tij.co.jp のデバイス製品フォルダを開いてください。[通知] をクリックして登録すると、変更されたすべての製品情報に関するダイジェストを毎週受け取ることができます。 変更の詳細については、改訂されたドキュメントに含まれている改訂履歴をご覧ください。

10.3 サポート・リソース

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10.4 Trademarks

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10.5 静電気放電に関する注意事項

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ESD による破損は、わずかな性能低下からデバイスの完全な故障まで多岐にわたります。精密な IC の場合、パラメータがわずかに変化するだけで公表されている仕様から外れる可能性があるため、破損が発生しやすくなっています。

10.6 用語集

テキサス・インスツルメンツ用語集　　この用語集には、用語や略語の一覧および定義が記載されています。

11 Revision History

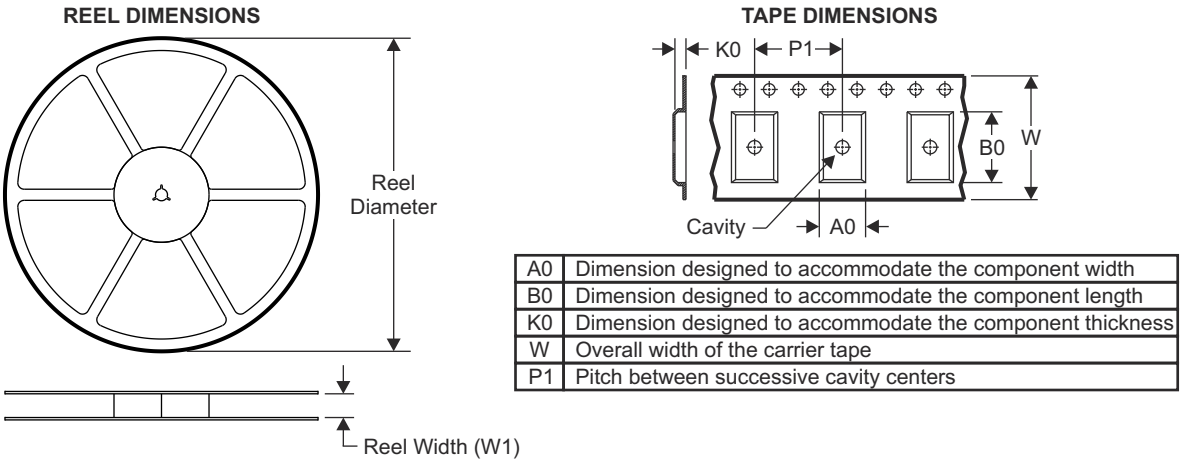
資料番号末尾の英字は改訂を表しています。その改訂履歴は英語版に準じています。

DATE	REVISION	NOTES
April 2024	*	Initial Release

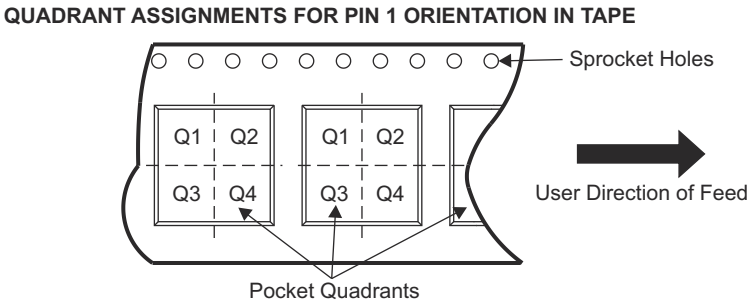
12 Mechanical, Packaging, and Orderable Information

The following pages include mechanical packaging and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

12.1 Tape and Reel Information

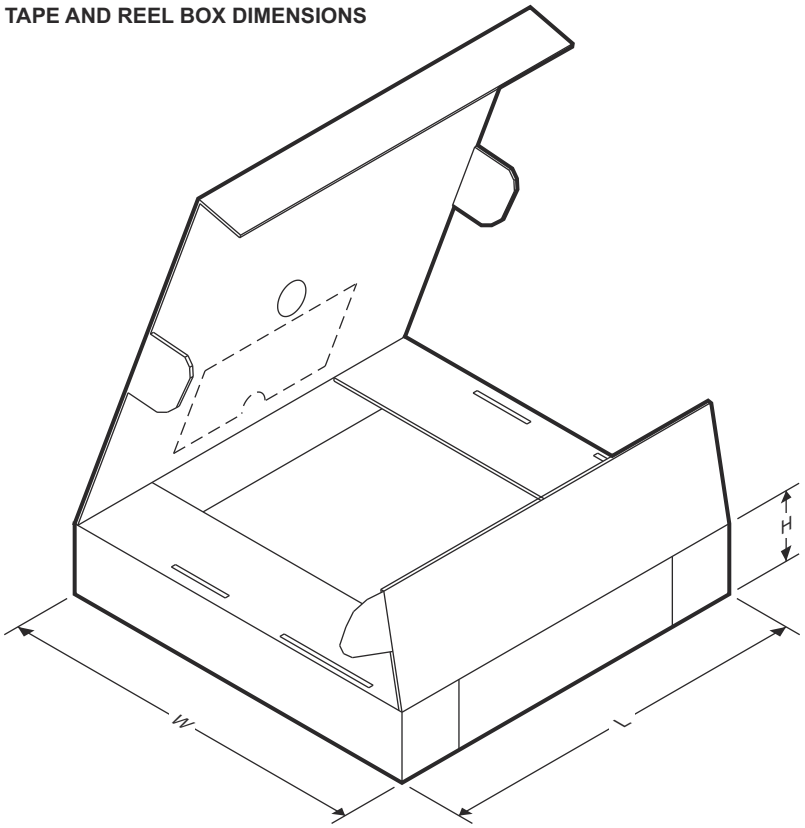


A0	Dimension designed to accommodate the component width
B0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers



Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
ISOM8110DFGRQ1	SO-4	DFG	4	2000	330.0	12.4	8.0	3.8	2.7	12.0	12.0	Q1

TAPE AND REEL BOX DIMENSIONS



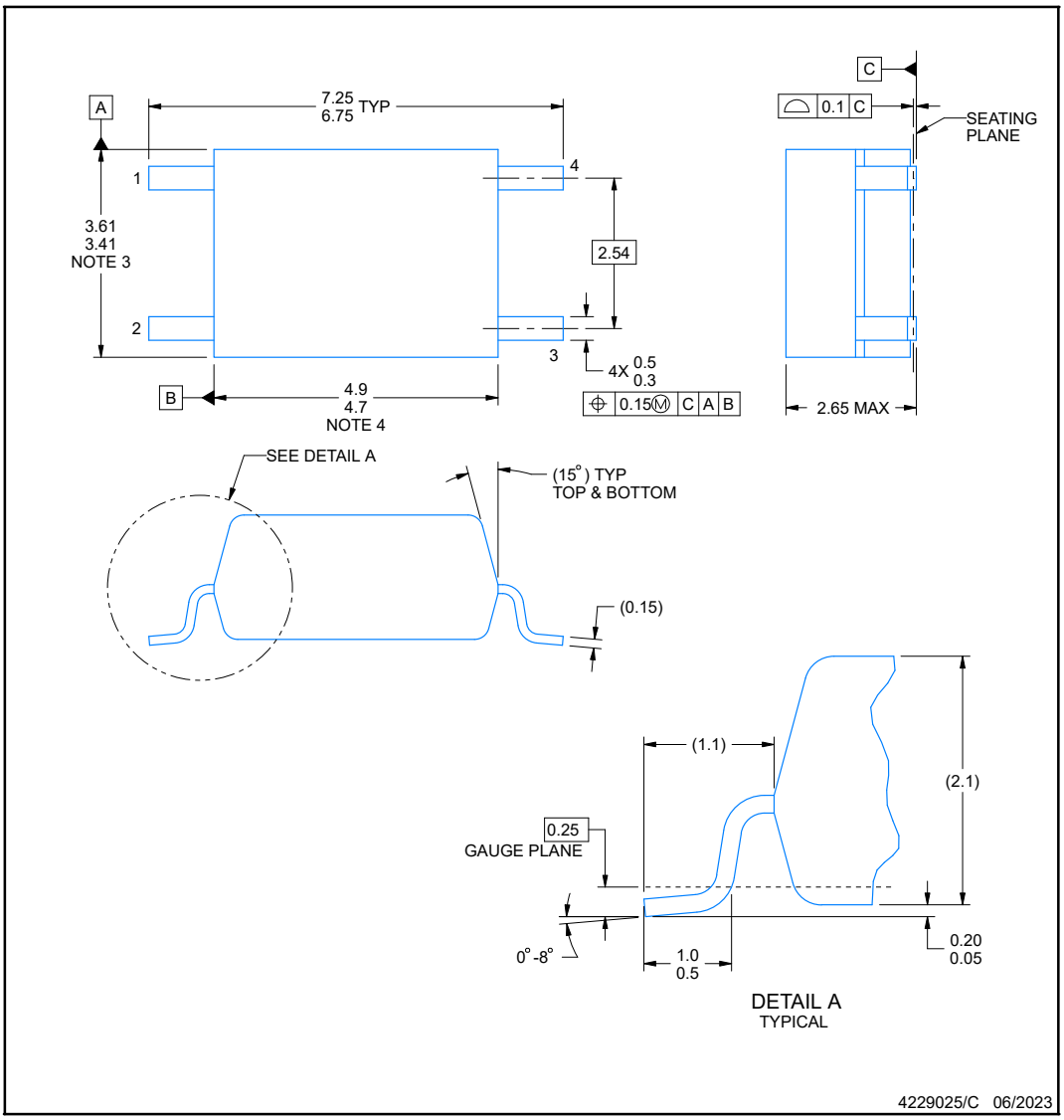
Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
ISOM8110DFGRQ1	SO-4	DFG	4	2000	356.0	356.0	35.0

PACKAGE OUTLINE

DFG0004A-C01

SOIC - 2.65 mm max height

SMALL OUTLINE INTEGRATED CIRCUIT



NOTES:

1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0,15 mm per side.
4. This dimension does not include interlead flash.

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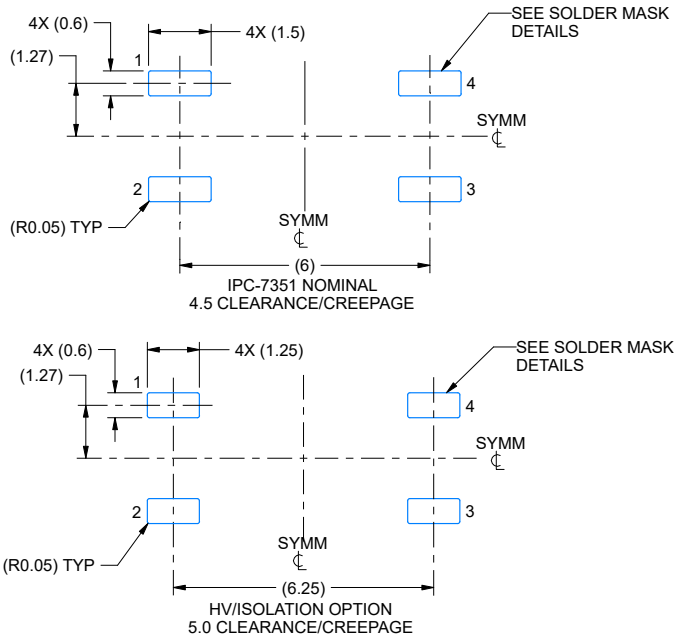
EXAMPLE BOARD LAYOUT

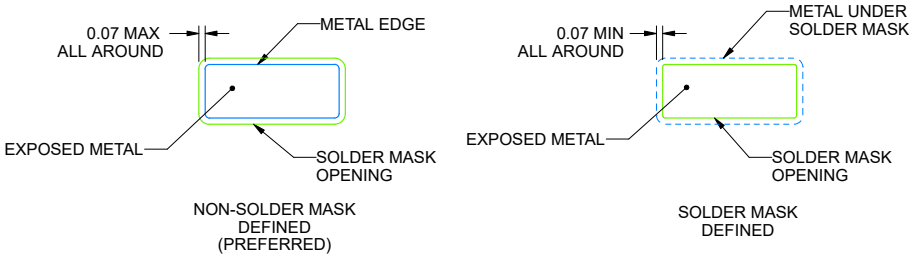
DFG0004A-C01

SOIC - 2.65 mm max height

SMALL OUTLINE INTEGRATED CIRCUIT



LAND PATTERN EXAMPLE
EXPOSED METAL SHOWN
SCALE: 7X



SOLDER MASK DETAILS

4229025/C 06/2023

NOTES: (continued)

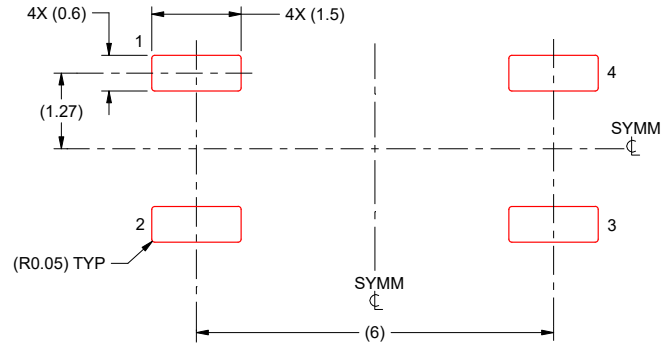
5. Publication IPC-7351 may have alternate designs.
6. Solder mask tolerances between and around signal pads can vary based on board fabrication site.

EXAMPLE STENCIL DESIGN

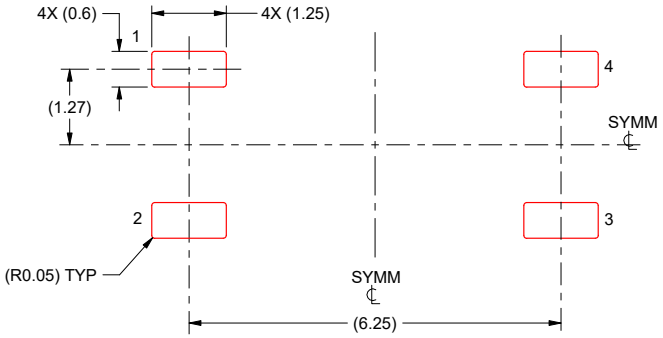
DFG0004A-C01

SOIC - 2.65 mm max height

SMALL OUTLINE INTEGRATED CIRCUIT



IPC-7351 NOMINAL
4.5 CLEARANCE/CREEPAGE



HV/ISOLATION OPTION
5.0 CLEARANCE/CREEPAGE

SOLDER PASTE EXAMPLE
BASED ON 0.125 mm THICK STENCIL
SCALE: 10X

4229025/C 06/2023

NOTES: (continued)

7. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
8. Board assembly site may have different recommendations for stencil design.

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PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead finish/ Ball material (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
ISOM8110DFGRQ1	ACTIVE	SOIC	DFG	4	2000	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	8110	Samples

(1) The marketing status values are defined as follows:
ACTIVE: Product device recommended for new designs.
LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.
NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.
PREVIEW: Device has been announced but is not in production. Samples may or may not be available.
OBSOLETE: TI has discontinued the production of the device.

(2) **RoHS:** TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".
RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.
Green: TI defines "Green" to mean the content of Chlorine (Cl) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

(6) Lead finish/Ball material - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

OTHER QUALIFIED VERSIONS OF ISOM8110-Q1 :

- Catalog : ISOM8110

NOTE: Qualified Version Definitions:

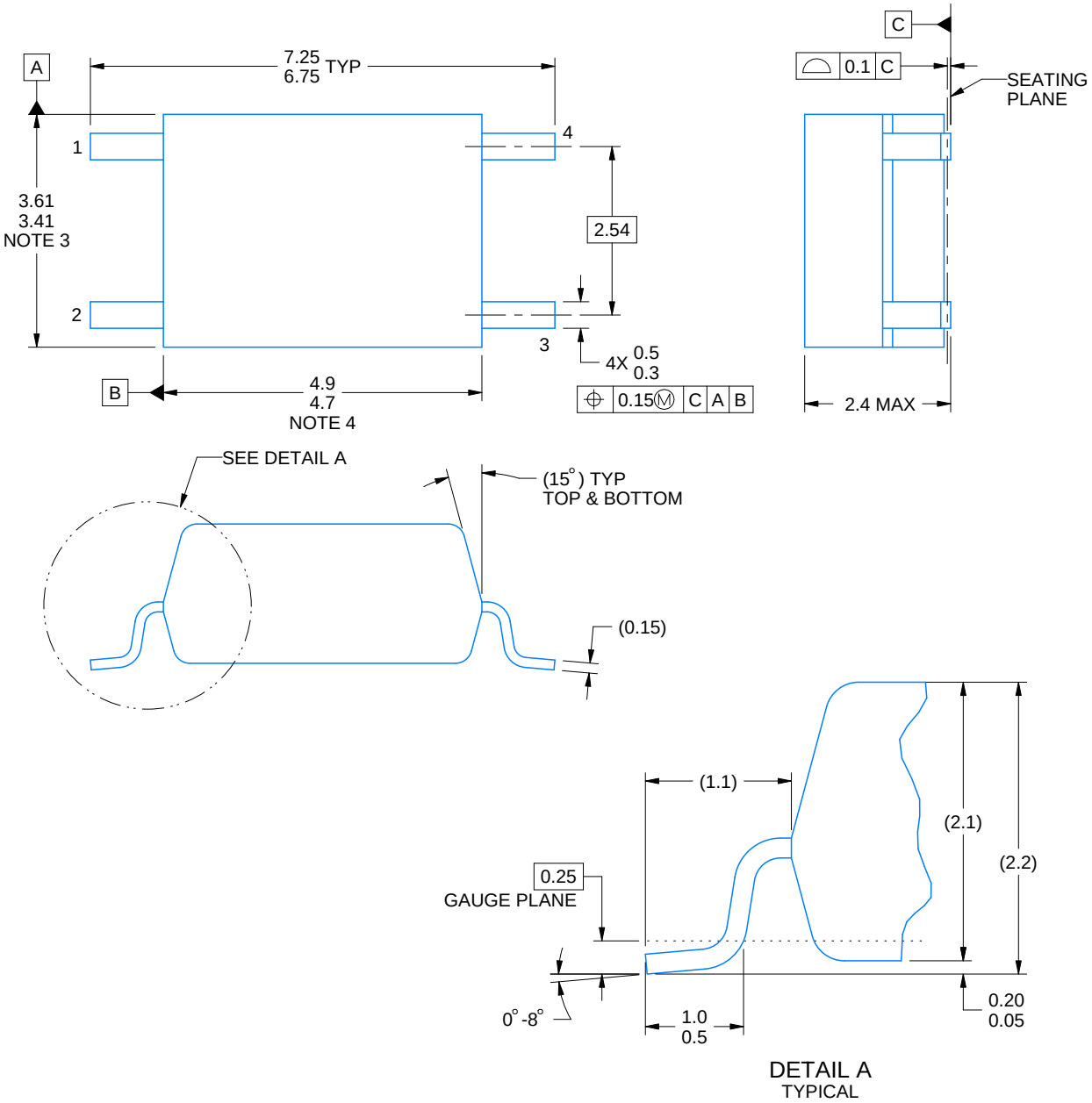
- Catalog - TI's standard catalog product

PACKAGE OUTLINE

DFG0004A

SOIC - 2.4 mm max height

SMALL OUTLINE INTEGRATED CIRCUIT



4227022/C 07/2024

NOTES:

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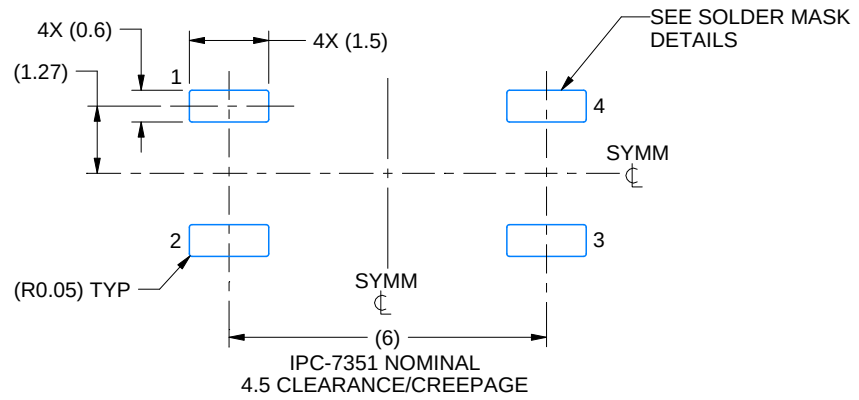
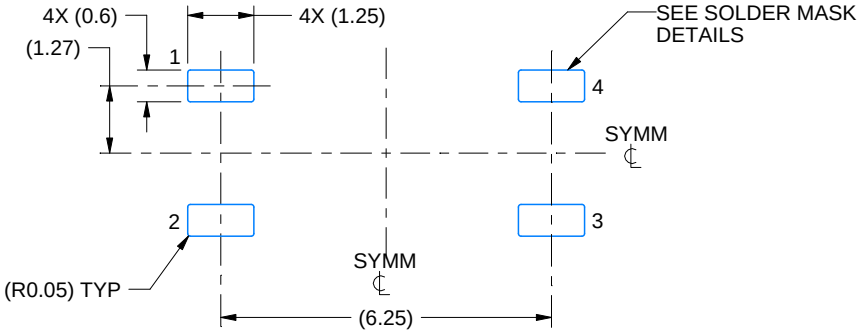
4. This dimension does not include interlead flash.

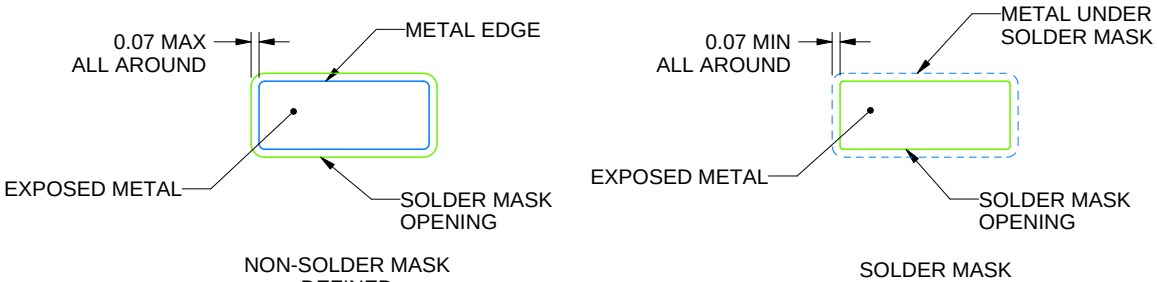


IPC-7351 NOMINAL
4.5 CLEARANCE/CREEPAGE



HV/ISOLATION OPTION
5.0 CLEARANCE/CREEPAGE

LAND PATTERN EXAMPLE
EXPOSED METAL SHOWN
SCALE: 7X

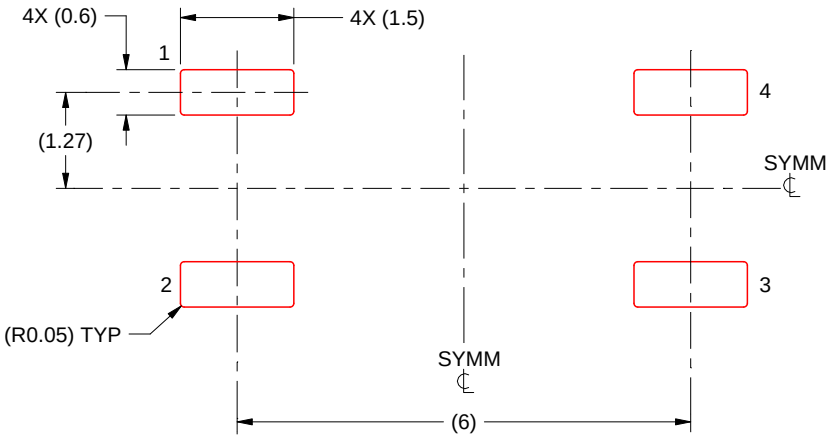


SOLDER MASK DETAILS

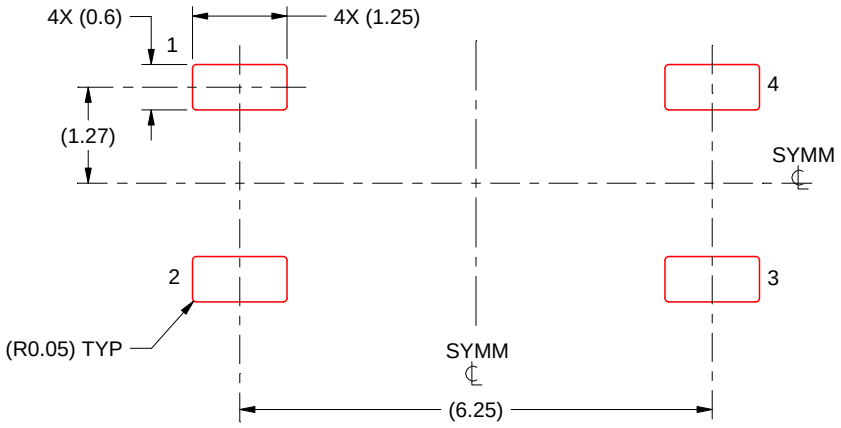
4227022/C 07/2024

NOTES: (continued)

5. Publication IPC-7351 may have alternate designs.
6. Solder mask tolerances between and around signal pads can vary based on board fabrication site.



IPC-7351 NOMINAL
4.5 CLEARANCE/CREEPAGE



HV/ISOLATION OPTION
5.0 CLEARANCE/CREEPAGE

SOLDER PASTE EXAMPLE
BASED ON 0.125 mm THICK STENCIL
SCALE: 10X

4227022/C 07/2024

NOTES: (continued)

7. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
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