

TLC555 LinCMOS™ テクノロジタイマ

1 特長

- 超低消費電力
 - $V_{DD} = 5V$ で 1mW (標準値)
- 非安定モードで動作可能
- レール・ツー・レールにスイング可能な CMOS 出力
- 高い出力電流能力
 - シンク: 標準値 100mA
 - 出典: 標準値 10mA
- 出力は CMOS、TTL、MOS と完全互換
- 電源電流が少ないため出力遷移中のスパイクが小さい
- 2V~15V の単一電源動作
- NE555 と機能的に交換可能 (同じピン配置)
- ANSI/ESDA/JEDEC JS-001 に準拠し、1000V を超える ESD 保護
- 車載対応 Q-Temp で利用可能
 - 高信頼性の車載用アプリケーション
 - 構成制御と印刷のサポート
 - 車載用規格の認定

2 アプリケーション

- 高精度のタイミング
- パルス生成
- シーケンシャル・タイミング
- 時間遅延の生成
- パルス幅変調
- パルス位置変調
- リニア・ランプ生成器

3 概要

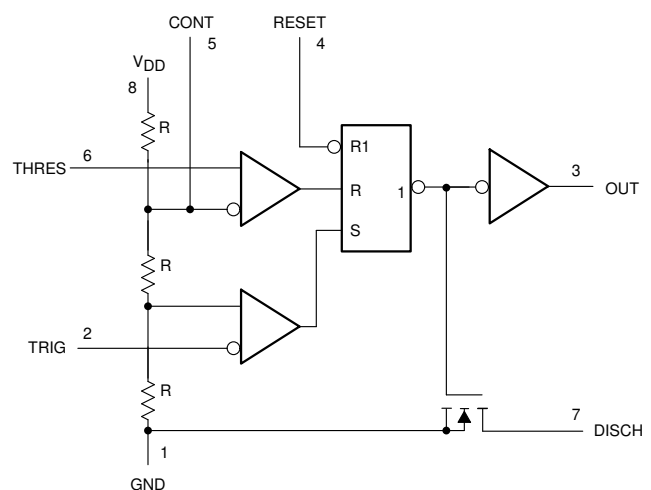
TLC555 は、テキサス・インスツルメンツの LinCMOS™ テクノロジを利用して製造されたモノリシック タイミング回路です。このタイマは CMOS、TTL、MOS ロジックと完全互換であり、最高 2 MHz の周波数で動作します。このデバイスは入力インピーダンスが高いため、NE555 または LM555 で対応しているタイミング コンデンサよりも小さな値のコンデンサに対応しています。その結果、より正確な時間遅延と発振が可能です。電源電圧の範囲全体にわたって低消費電力を実現します。

NE555 と同様、TLC555 のトリガ・レベルは電源電圧の約 1/3、スレッシュホールド・レベルは電源電圧の約 2/3 です。これらの電圧レベルは、制御電圧ピン (CONT) を使用して変更できます。トリガ入力 (TRIG) がトリガ・レベルより低くなると、フリップ・フロップがセットされ、出力は HIGH になります。TRIG がトリガ・レベルより高く、かつスレッシュホールド入力 (THRES) がスレッシュホールド・レベルより高くなると、フリップ・フロップはリセットされ、出力は LOW になります。リセット入力 (RESET) は他のすべての入力より優先され、新しいタイミング・サイクルの開始に使用できます。RESET を LOW にすると、フリップ・フロップはリセットされ、出力は LOW になります。出力が LOW のとき常に、放電ピン (DISCH) と GND との間に低インピーダンスの経路が形成されます。誤トリガを防止するため、未使用の入力はすべて、適切なロジック・レベルに接続する必要があります。

パッケージ情報

部品番号	パッケージ (1)	パッケージ サイズ (2)
TLC555C	SOIC (8)	4.9 mm × 6.0 mm
	PDIP (8)	9.81mm × 9.43mm
	SOP (8)	6.2mm × 7.8mm
	TSSOP (14)	5.0 mm × 6.4 mm
TLC555I	SOIC (8)	4.9 mm × 6.0 mm
	PDIP (8)	9.81mm × 9.43mm
TLC555M	LCCC (20)	8.89mm × 8.89mm
	CDIP (8)	9.6 mm × 9.0 mm
TLC555Q	SOIC (8)	4.9 mm × 6.0 mm

- (1) 詳細については、[セクション 10](#) を参照してください。
- (2) パッケージ サイズ (長さ×幅) は公称値であり、該当する場合はピンも含まれます。



概略回路図



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4 Pin Configuration and Functions

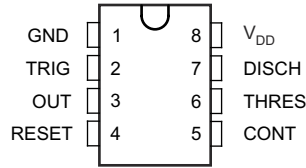


図 4-1. D, P, PS, and JG Packages, 8-Pin SOIC, PDIP, SOP, and CDIP (Top View)

表 4-1. Pin Functions: D, P, PS, and JG Packages

PIN		TYPE	DESCRIPTION
NAME	NO.		
CONT	5	Input	Controls comparator thresholds. Outputs $2/3 V_{DD}$ and allows bypass capacitor connection.
DISCH	7	Output	Open collector output to discharge timing capacitor.
GND	1	—	Ground.
NC	—	—	No internal connection.
OUT	3	Output	High current timer output signal.
RESET	4	Input	Active low reset input forces output and discharge low.
THRES	6	Input	End of timing input. $THRES > CONT$ sets output low and discharge low.
TRIG	2	Input	Start of timing input. $TRIG < 1/2 CONT$ sets output high and discharge open.
V_{DD}	8	—	Power-supply voltage.

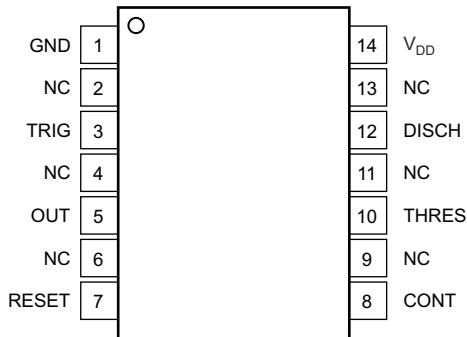


図 4-2. PW Package, 14-Pin TSSOP (Top View)

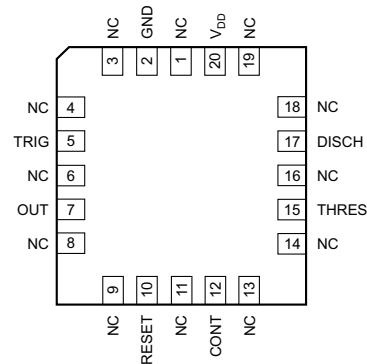


図 4-3. FK Package, 20-Pin LCCC (Top View)

表 4-2. Pin Functions: PW and FK

NAME	PIN NO.		TYPE	DESCRIPTION
	PW (TSSOP)	FK (LCCC)		
CONT	8	12	Input	Controls comparator thresholds. Outputs $2/3 V_{DD}$ and allows bypass capacitor connection.
DISCH	12	17	Output	Open-collector output to discharge timing capacitor.
GND	1	2	—	Ground.
NC	2, 4, 6, 9, 11, 13	1, 3, 4, 6, 8, 9, 11, 13, 14, 16, 18, 19	—	No internal connection.
OUT	5	7	Output	High current timer output signal.
RESET	7	10	Input	Active low reset input forces output and discharge low.
THRES	10	15	Input	End of timing input. $THRES > CONT$ sets output low and discharge low.
TRIG	3	5	Input	Start of timing input. $TRIG < 1/2 CONT$ sets output high and discharge open.

表 4-2. Pin Functions: PW and FK (続き)

NAME	PIN		TYPE	DESCRIPTION
	NO.			
	PW (TSSOP)	FK (LCCC)		
V _{DD}	14	20	—	Power-supply voltage.

5 Specifications

5.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted)⁽¹⁾

			MIN	MAX	UNIT
	Voltage	Supply, V_{DD} ⁽²⁾	-0.3	18	V
		Input, any input	-0.3	V_{DD}	
		Discharge	-0.3	18	
	Current	Sink, discharge or output		150	mA
		Source, output, I_O		15	
T_A	Operating temperature	C-suffix	0	70	°C
		I-suffix	-40	85	
		Q-suffix	-40	125	
		M-suffix	-55	125	
	Case temperature, for 60 seconds	FK package	-65	150	°C
T_{stg}	Storage temperature		-65	150	°C

- (1) Operation outside the Absolute Maximum Ratings may cause permanent device damage. Absolute Maximum Ratings do not imply functional operation of the device at these or any other conditions beyond those listed under *Recommended Operating Conditions*. If used outside the Recommended Operating Conditions but within the Absolute Maximum Ratings, the device may not be fully functional, and this may affect device reliability, functionality, performance, and shorten the device lifetime.
- (2) All voltage values are with respect to network GND.

5.2 ESD Ratings

			VALUE	UNIT
$V_{(ESD)}$	Electrostatic discharge ⁽³⁾	Human-body model (HBM), per ANSI/ESDA/JEDEC JS-001 ⁽¹⁾	±1000	V
		Charged-device model (CDM), per ANSI/ESDA/JEDEC JS-002 ⁽²⁾	±1000	
		Machine model (MM)	±200	

- (1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.
- (2) JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.
- (3) See [セクション 7.2.5](#) for application guidance on protecting the device against ESD.

5.3 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)

			MIN	MAX	UNIT
V_{DD}	Supply voltage	TLC555C	2	15	V
		TLC555I	3	15	
		TLC555M	5	15	
		TLC555Q	5	15	
T_A	Operating free-air temperature	TLC555C	0	70	°C
		TLC555I	-40	85	
		TLC555M	-55	125	
		TLC555Q	-40	125	

5.4 Thermal Information

THERMAL METRIC ⁽¹⁾		TLC555						UNIT
		D (SOIC)	FK (LCCC)	JG (CDIP)	P (PDIP)	PS (SOP)	PW (TSSOP)	
		8 PINS	20 PINS	8 PINS	8 PINS	8 PINS	14 PINS	
R _{θJA}	Junction-to-ambient thermal resistance	138.9	N/A	120	93.1	120	135	°C/W
R _{θJC(top)}	Junction-to-case (top) thermal resistance	78.8	37	81	82.5	72	61	°C/W
R _{θJB}	Junction-to-board thermal resistance	87.9	36	110	69.6	69	77	°C/W
Ψ _{JT}	Junction-to-top characterization parameter	23.2	N/A	45	52.0	32	12	°C/W
Ψ _{JB}	Junction-to-board characterization parameter	86.9	N/A	103	69.2	68	77	°C/W
R _{θJC(bot)}	Junction-to-case (bottom) thermal resistance	N/A	4.3	31	N/A	N/A	N/A	°C/W

(1) For more information about traditional and new thermal metrics, see the [Semiconductor and IC Package Thermal Metrics](#) application report.

5.5 Electrical Characteristics: V_{DD} = 2 V for TLC555C, V_{DD} = 3 V for TLC555I

over operating free-air temperature range (unless otherwise noted)

PARAMETER		TEST CONDITIONS ⁽¹⁾		MIN	TYP	MAX	UNIT
V _{IT}	Threshold voltage	25°C	TLC555C	0.95	1.33	1.65	V
			TLC555I	1.6		2.4	
		Full range	TLC555C	0.85		1.75	
			TLC555I	1.5		2.5	
I _{IT}	Threshold current	25°C	TLC555C		10		pA
			TLC555I		10		
		Max	TLC555C		75		
			TLC555I		150		
V _{I(TRIG)}	Trigger voltage	25°C	TLC555C	0.4	0.67	0.95	V
			TLC555I	0.71	1	1.29	
		Full range	TLC555C	0.3		1.05	
			TLC555I	0.61		1.39	
I _{I(TRIG)}	Trigger current	25°C	TLC555C		10		pA
			TLC555I		10		
		Max	TLC555C		75		
			TLC555I		150		
V _{I(RESET)}	Reset voltage	25°C	TLC555C	0.4	1.1	1.5	V
			TLC555I	0.4	1.1	1.5	
		Full range	TLC555C	0.3		2	
			TLC555I	0.3		1.8	
	Control voltage (open-circuit) as a percentage of supply voltage	Max	TLC555C		66.7%		
			TLC555I		66.7%		
	Discharge switch on-stage voltage	I _{OL} = 1 mA, 25°C	TLC555C		0.03	0.2	V
			TLC555I		0.03	0.2	
		I _{OL} = 1 mA, Full range	TLC555C			0.25	
			TLC555I			0.375	
	Discharge switch off-stage current	25°C	TLC555C		0.1		nA
			TLC555I		0.1		
		Max	TLC555C		0.5		
			TLC555I		120		

5.5 Electrical Characteristics: $V_{DD} = 2\text{ V}$ for TLC555C, $V_{DD} = 3\text{ V}$ for TLC555I (続き)

over operating free-air temperature range (unless otherwise noted)

PARAMETER		TEST CONDITIONS ⁽¹⁾		MIN	TYP	MAX	UNIT
V_{OH}	High-level output voltage	$I_{OH} = -300\ \mu\text{A}$, 25°C	TLC555C	1.5	1.9		V
			TLC555I	2.5	2.85		
		$I_{OH} = -300\ \mu\text{A}$, Full range	TLC555C	1.5			
			TLC555I	2.5			
V_{OL}	Low-level output voltage	$I_{OL} = 1\ \text{mA}$, 25°C	TLC555C		0.07	0.3	V
			TLC555I		0.07	0.3	
		$I_{OL} = 1\ \text{mA}$, Full range	TLC555C			0.35	
			TLC555I			0.4	
I_{DD}	Supply current ⁽²⁾	25°C	TLC555C			250	μA
			TLC555I			250	
		Full range	TLC555C			400	
			TLC555I			500	
C_{PD}	Power dissipation capacitance ⁽³⁾ ⁽⁴⁾	25°C	TLC555C		80		pF
			TLC555I		90		

(1) Full range is 0°C to 70°C the for TLC555C, and -40°C to +85°C for the TLC555I. For conditions shown as MAX, use the appropriate value specified in the [セクション 5.3](#).

(2) These values apply for the expected operating configurations in which THRES is connected directly to DISCH or to TRIG.

(3) C_{PD} is used to determine the dynamic power consumption.

(4) $P_D = V_{DD}^2 f_o (C_{PD} + C_L)$ where f_o = output frequency, C_L = output load capacitance, V_{DD} = supply voltage.

5.6 Electrical Characteristics: $V_{DD} = 5\text{ V}$

over operating free-air temperature range (unless otherwise noted)

PARAMETER		TEST CONDITIONS ⁽¹⁾		MIN	TYP	MAX	UNIT
V_{IT}	Threshold voltage	25°C	TLC555C, TLC555I, TLC555M, TLC555Q	2.8	3.3	3.8	V
		Full range	TLC555C, TLC555I, TLC555M, TLC555Q	2.7		3.9	
I_{IT}	Threshold current	25°C	TLC555C, TLC555I, TLC555M, TLC555Q		10		pA
		Max	TLC555C		75		
			TLC555I		150		
TLC555M, TLC555Q		5000					
$V_{I(TRIG)}$	Trigger voltage	25°C	TLC555C, TLC555I, TLC555M, TLC555Q	1.36	1.66	1.96	V
		Full range	TLC555C, TLC555I, TLC555M, TLC555Q	1.26		2.06	
$I_{I(TRIG)}$	Trigger current	25°C	TLC555C, TLC555I, TLC555M, TLC555Q		10		pA
		Max	TLC555C		75		
			TLC555I		150		
TLC555M, TLC555Q		5000					
C_I	Trigger, threshold capacitance (each pin)	25°C	TLC555C, TLC555I, TLC555M, TLC555Q		2.1		pF
$V_{I(RESET)}$	Reset voltage	25°C	TLC555C, TLC555I, TLC555M, TLC555Q	0.4	1.1	1.5	V
		Full range	TLC555C, TLC555I, TLC555M, TLC555Q	0.3		1.8	

5.6 Electrical Characteristics: $V_{DD} = 5\text{ V}$ (続き)

over operating free-air temperature range (unless otherwise noted)

PARAMETER		TEST CONDITIONS ⁽¹⁾		MIN	TYP	MAX	UNIT
$I_{I(\text{RESET})}$	Reset current	25°C, $V_{\text{RESET}} = 0\text{ V}$	TLC555C, TLC555I, TLC555M, TLC555Q		5.9		μA
		25°C, $V_{\text{RESET}} = V_{DD}$	TLC555C, TLC555I, TLC555M, TLC555Q		10		pA
		Max, $V_{\text{RESET}} = V_{DD}$	TLC555C		75		
			TLC555I		150		
		TLC555M, TLC555Q		5000			
	Control voltage (open circuit) as a percentage of supply voltage	Max	TLC555C, TLC555I, TLC555M, TLC555Q		66.7%		
	Discharge switch on-stage voltage	$I_{OL} = 10\text{ mA}$, 25°C	TLC555C, TLC555I, TLC555M, TLC555Q		0.14	0.5	V
		$I_{OL} = 10\text{ mA}$, Full range	TLC555C, TLC555I, TLC555M, TLC555Q			0.6	
	Discharge switch off-stage current	25°C	TLC555C, TLC555I, TLC555M, TLC555Q		0.1		nA
			TLC555C		0.5		
		Max	TLC555I		120		
			TLC555M, TLC555Q		120		
V_{OH}	High-level output voltage	$I_{OH} = -1\text{ mA}$, 25°C	TLC555C, TLC555I, TLC555M, TLC555Q	4.1	4.8		V
		$I_{OH} = -1\text{ mA}$, Full range	TLC555C, TLC555I, TLC555M, TLC555Q	4.1			
V_{OL}	Low-level output voltage	$I_{OL} = 8\text{ mA}$, 25°C	TLC555C, TLC555I, TLC555M, TLC555Q		0.21	0.4	V
			TLC555C			0.5	
		$I_{OL} = 8\text{ mA}$, Full range	TLC555I			0.5	
			TLC555M, TLC555Q			0.6	
V_{OL}	Low-level output voltage	$I_{OL} = 5\text{ mA}$, 25°C	TLC555C, TLC555I, TLC555M, TLC555Q		0.13	0.3	V
			TLC555C			0.4	
		$I_{OL} = 5\text{ mA}$, Full range	TLC555I			0.4	
			TLC555M, TLC555Q			0.45	
		$I_{OL} = 3.2\text{ mA}$, 25°C	TLC555C, TLC555I, TLC555M, TLC555Q		0.08	0.3	
			TLC555C			0.35	
		$I_{OL} = 3.2\text{ mA}$, Full range	TLC555I			0.35	
			TLC555M, TLC555Q			0.4	
I_{DD}	Supply current ⁽²⁾	25°C	TLC555C, TLC555I, TLC555M, TLC555Q		180	350	μA
			TLC555C			500	
		Full range	TLC555I			600	
			TLC555M, TLC555Q			700	
C_{PD}	Power dissipation capacitance ⁽³⁾ (4)	25°C	TLC555C, TLC555I, TLC555M, TLC555Q		115		pF

- (1) Full range is 0°C to 70°C the for TLC555C, -40°C to 85°C for the TLC555I, -40°C to 125°C for the TLC555Q, and -55°C to 125°C for the TLC555M. For conditions shown as MAX, use the appropriate value specified in the [セクション 5.3](#) table.
- (2) These values apply for the expected operating configurations in which THRES is connected directly to DISCH or to TRIG.
- (3) C_{PD} is used to determine the dynamic power consumption.
- (4) $P_D = V_{DD}^2 f_o (C_{PD} + C_L)$ where f_o = output frequency, C_L = output load capacitance, V_{DD} = supply voltage.

5.7 Electrical Characteristics: $V_{DD} = 15\text{ V}$

over operating free-air temperature range (unless otherwise noted)

PARAMETER		TEST CONDITIONS ⁽¹⁾		MIN	TYP	MAX	UNIT
V_{IT}	Threshold voltage	25°C	TLC555C, TLC555I, TLC555M, TLC555Q	9.45	10	10.55	V
		Full range	TLC555C, TLC555I, TLC555M, TLC555Q	9.35		10.65	
I_{IT}	Threshold current	25°C	TLC555C, TLC555I, TLC555M, TLC555Q		10		pA
		Max	TLC555C		75		
			TLC555I		150		
TLC555M, TLC555Q		5000					
$V_{I(TRIG)}$	Trigger voltage	25°C	TLC555C, TLC555I, TLC555M, TLC555Q	4.65	5	5.35	V
		Full range	TLC555C, TLC555I, TLC555M, TLC555Q	4.55		5.45	
$I_{I(TRIG)}$	Trigger current	25°C	TLC555C, TLC555I, TLC555M, TLC555Q		10		pA
		Max	TLC555C		75		
			TLC555I		150		
TLC555M, TLC555Q		5000					
C_I	Trigger, threshold capacitance (each pin)	25°C	TLC555C, TLC555I, TLC555M, TLC555Q		1.8		pF
$V_{I(RESET)}$	Reset voltage	25°C	TLC555C, TLC555I, TLC555M, TLC555Q	0.4	1.1	1.5	V
		Full range	TLC555C, TLC555I, TLC555M, TLC555Q	0.3		1.8	
$I_{I(RESET)}$	Reset current	25°C, $V_{RESET} = 0\text{ V}$	TLC555C, TLC555I, TLC555M, TLC555Q		17.8		μA
		25°C, $V_{RESET} = V_{DD}$	TLC555C, TLC555I, TLC555M, TLC555Q		10		pA
			TLC555C		75		
		Max, $V_{RESET} = V_{DD}$	TLC555I		150		
			TLC555M, TLC555Q		5000		
	Control voltage (open circuit) as a percentage of supply voltage	Max	TLC555C, TLC555I, TLC555M, TLC555Q		66.7%		
	Discharge switch on-stage voltage	$I_{OL} = 100\text{ mA}$, 25°C	TLC555C, TLC555I, TLC555M, TLC555Q		0.77	1.7	V
		$I_{OL} = 100\text{ mA}$, Full range	TLC555C, TLC555I, TLC555M, TLC555Q			1.8	
	Discharge switch off-stage current	25°C	TLC555C, TLC555I, TLC555M, TLC555Q		0.1		nA
		Max	TLC555C		0.5		
			TLC555I		120		
TLC555M, TLC555Q		120					
V_{OH}	High-level output voltage	$I_{OH} = -10\text{ mA}$, 25°C	TLC555C, TLC555I, TLC555M, TLC555Q	12.5	14.2		V
		$I_{OH} = -10\text{ mA}$, Full range	TLC555C, TLC555I, TLC555M, TLC555Q	12.5			
			TLC555C, TLC555I, TLC555M, TLC555Q	13.5	14.6		
		$I_{OH} = -5\text{ mA}$, Full range	TLC555C, TLC555I, TLC555M, TLC555Q	13.5			
		$I_{OH} = -1\text{ mA}$, 25°C	TLC555C, TLC555I, TLC555M, TLC555Q	14.2	14.9		
			TLC555C, TLC555I, TLC555M, TLC555Q	14.2			

5.7 Electrical Characteristics: $V_{DD} = 15\text{ V}$ (続き)

over operating free-air temperature range (unless otherwise noted)

PARAMETER		TEST CONDITIONS ⁽¹⁾		MIN	TYP	MAX	UNIT	
V_{OL}	Low-level output voltage	$I_{OL} = 100\text{ mA}$, 25°C	TLC555C, TLC555I, TLC555M, TLC555Q		1.28	3.2	V	
			TLC555C			3.6		
		$I_{OL} = 100\text{ mA}$, Full range	TLC555I			3.7		
			TLC555M, TLC555Q			3.8		
			TLC555C, TLC555I, TLC555M, TLC555Q		0.63	1		
		$I_{OL} = 50\text{ mA}$, Full range	TLC555C			1.3		
			TLC555I			1.4		
			TLC555M, TLC555Q			1.5		
		$I_{OL} = 10\text{ mA}$, 25°C	TLC555C, TLC555I, TLC555M, TLC555Q			0.12		0.3
			TLC555C					0.4
TLC555I					0.4			
$I_{OL} = 10\text{ mA}$, Full range	TLC555M, TLC555Q				0.45			
	25°C	TLC555C, TLC555I, TLC555M, TLC555Q			360	600	μA	
		TLC555C				800		
TLC555I					900			
TLC555M, TLC555Q				1000				
I_{DD}	Supply current ⁽²⁾	Full range	TLC555C, TLC555I, TLC555M, TLC555Q				μA	
			TLC555C					800
C_{PD}	Power dissipation capacitance ⁽³⁾ (4)	25°C	TLC555C, TLC555I, TLC555M, TLC555Q		140		pF	
			TLC555C, TLC555I, TLC555M, TLC555Q					

(1) Full range is 0°C to 70°C for TLC555C, –40°C to 85°C for TLC555I, –40°C to 125°C for the TLC555Q, and –55°C to 125°C for TLC555M. For conditions shown as MAX, use the appropriate value specified in the [セクション 5.3](#) table.

(2) These values apply for the expected operating configurations in which THRES is connected directly to DISCH or TRIG.

(3) C_{PD} is used to determine the dynamic power consumption.

(4) $P_D = V_{DD}^2 f_o (C_{PD} + C_L)$ where f_o = output frequency, C_L = output load capacitance, V_{DD} = supply voltage.

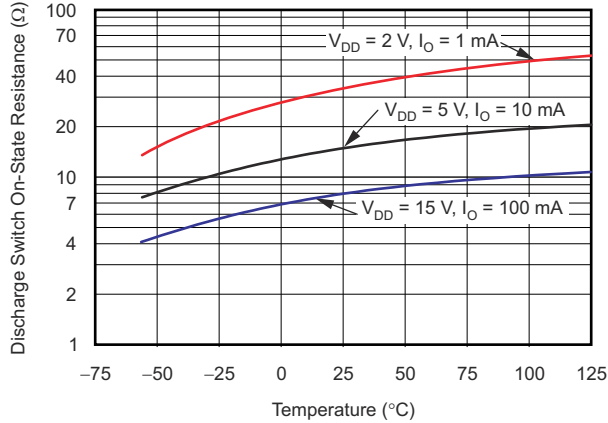
5.8 Timing Characteristics

$V_{DD} = 5\text{ V}$, $T_A = 25^\circ\text{C}$ (unless otherwise noted). Characteristic values are specified by design, characterization, or both.

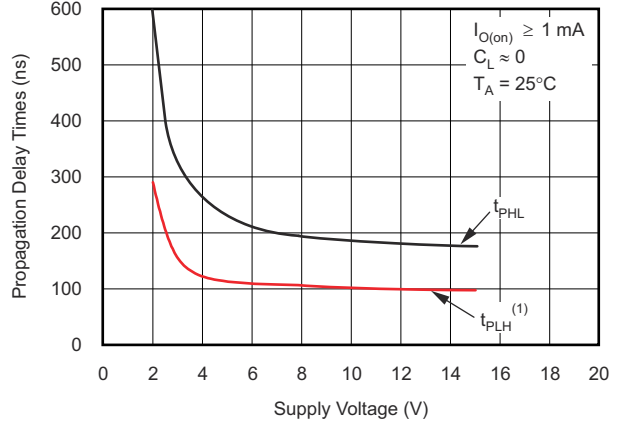
PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
	Supply voltage sensitivity of timing interval	$V_{DD} = 5\text{ V to }15\text{ V}$, $C_T = 0.1\text{ }\mu\text{F}$ $R_A = R_B = 1\text{ k}\Omega\text{ to }100\text{ k}\Omega$ ⁽¹⁾		0.1	0.5	%/V
t_r	Output pulse rise time	$R_L = 10\text{ M}\Omega$, $C_L = 10\text{ pF}$		20	75	ns
t_f	Output pulse fall time	$R_L = 10\text{ M}\Omega$, $C_L = 10\text{ pF}$		15	60	ns
f_{max}	Maximum frequency in a-stable mode	$R_A = 470\text{ }\Omega$, $C_T = 200\text{ pF}$, $R_B = 200\text{ }\Omega$ ⁽¹⁾	1.2	2.1		MHz

(1) R_A , R_B , and C_T are as defined in [図 6-5](#).

5.9 Typical Characteristics

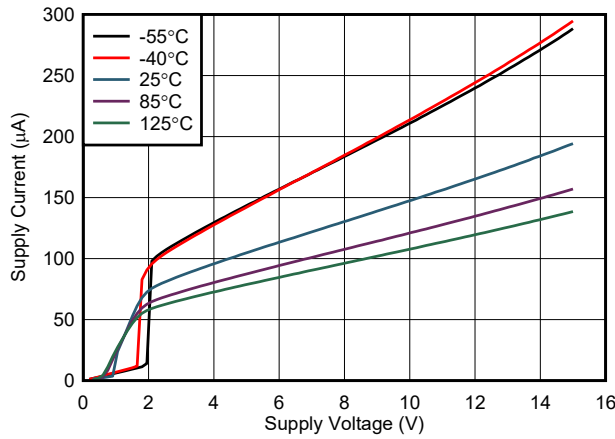


5-1. Discharge Switch On-State Resistance vs Free-Air Temperature

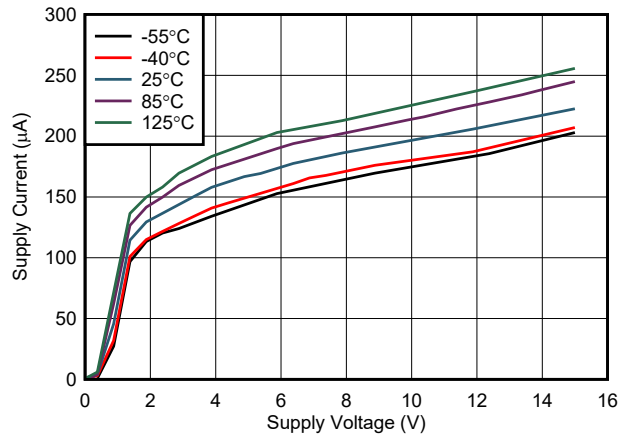


(1) The effects of the load resistance on these values must be taken into account separately.

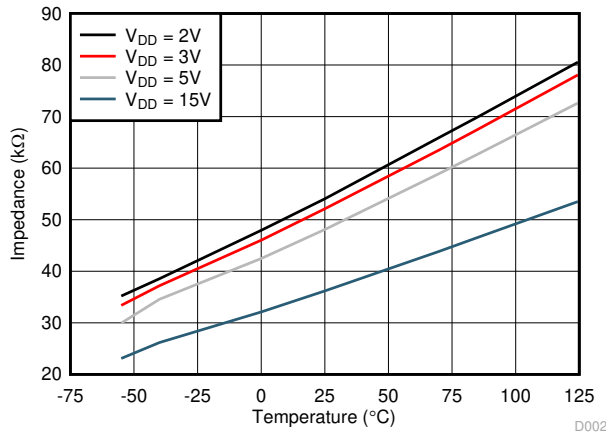
5-2. Propagation Delay Times to Discharge Output From Trigger and Threshold Shorted Together vs Supply Voltage



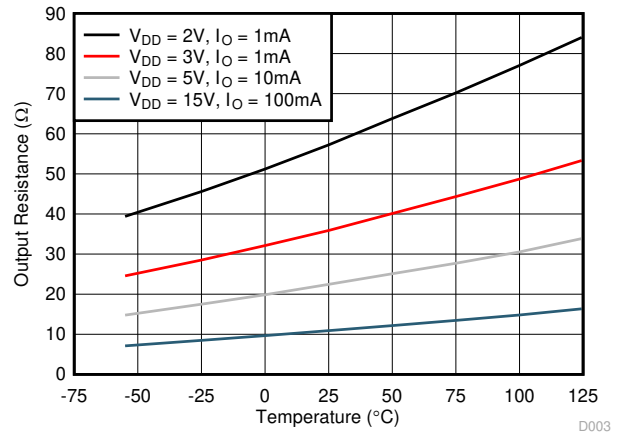
5-3. Supply Current vs Supply Voltage, Unit 1



5-4. Supply Current vs Supply Voltage, Unit 2

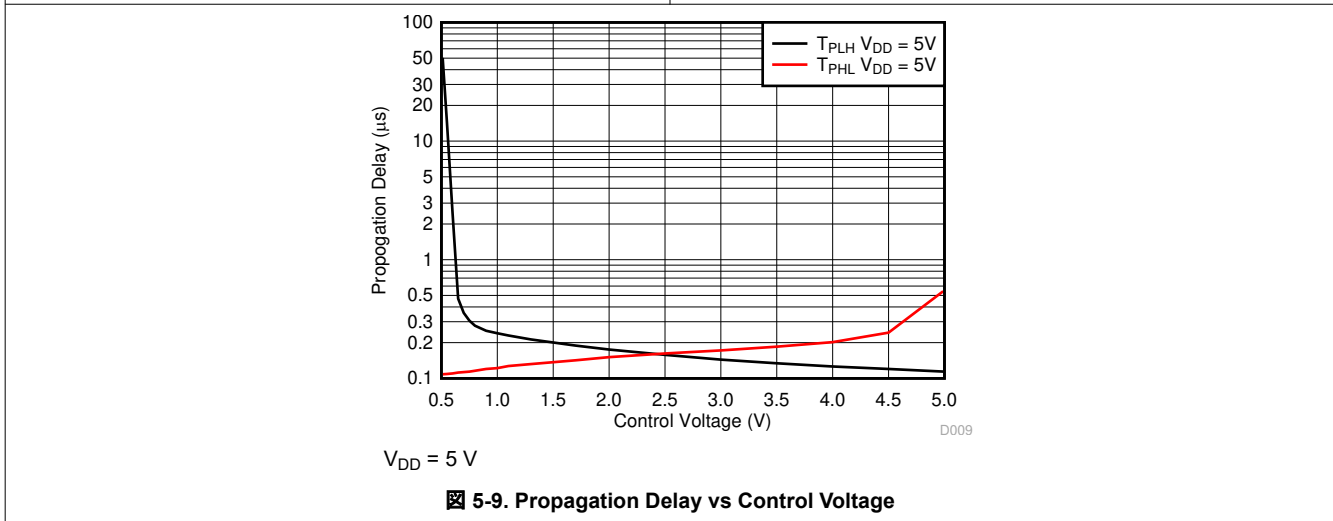
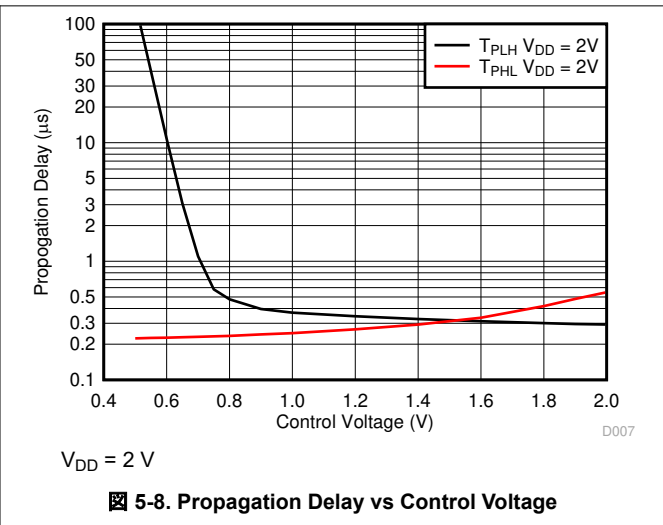
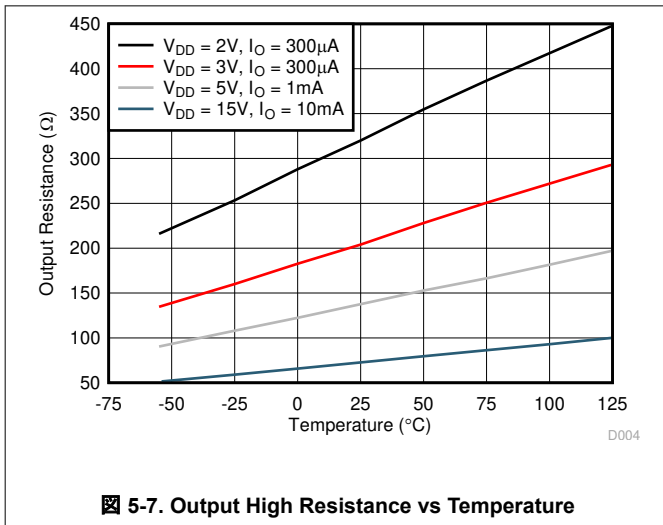


5-5. Control Impedance vs Temperature



5-6. Output Low Resistance vs Temperature

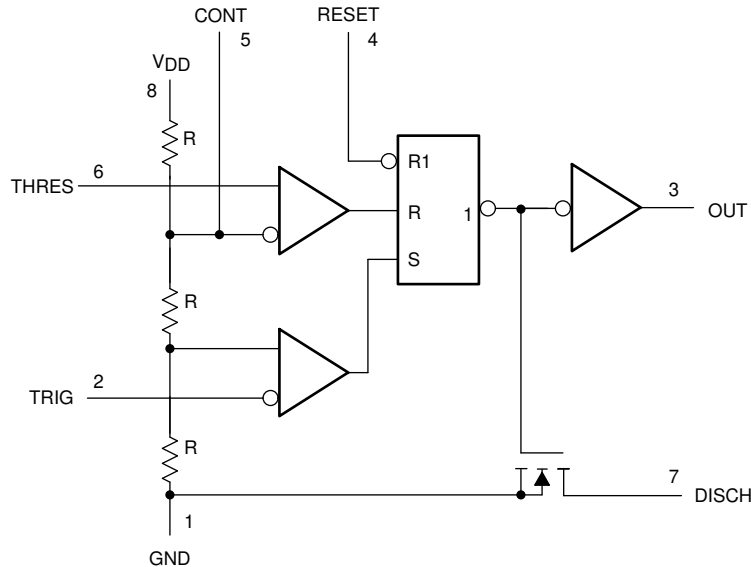
5.9 Typical Characteristics (continued)



6 Detailed Description

6.1 Overview

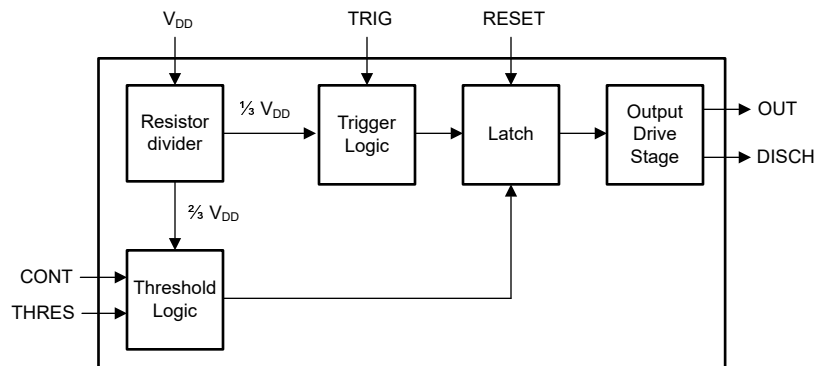
The TLC555 is a precision timing device used for general-purpose timing applications up to 2.1 MHz. All inputs are level sensitive, not edge-triggered inputs.



Pin numbers are for all packages except PW and FK. RESET overrides TRIG, which overrides THRES (when CONT pin is $2/3 V_{DD}$). The resistance of R resistors vary with V_{DD} and temperature. The resistors match each other very well across V_{DD} and temperature for a temperature-stable control-voltage ratio.

☒ 6-1. Simplified Schematic

6.2 Functional Block Diagram



6.3 Feature Description

6.3.1 Monostable Operation

For monostable operation, ☒ 6-2 shows how any of these timers can be connected. If the output is low, application of a negative-going pulse to the trigger (TRIG) sets the internal latch; the output goes high, and discharge pin (DISCH) becomes open drain. Capacitor C then is charged through R_A until the voltage across the capacitor reaches the threshold voltage of the threshold (THRES) input. If TRIG has returned to a high level, the output of the threshold comparator resets the internal latch, the output goes low, the discharge pin goes low, which quickly discharges capacitor C.

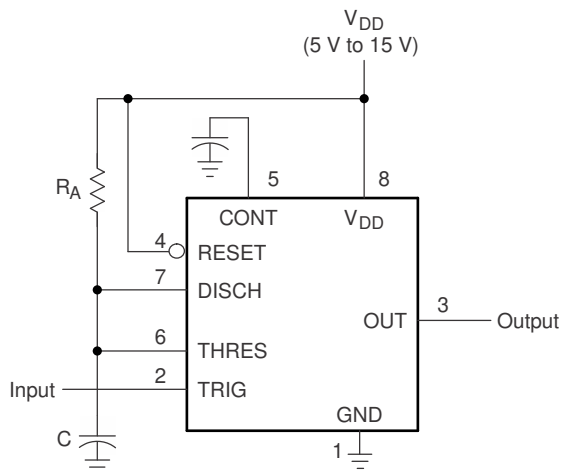


図 6-2. Circuit for Monostable Operation

Monostable operation is initiated when TRIG voltage is less than the trigger threshold. If initiated, the sequence ends only if TRIG is high for at least 1 μ s before the end of the timing interval. When the trigger is grounded, the comparator storage time can be as long as 1 μ s, which limits the minimum monostable pulse duration to 1 μ s. The output pulse duration is approximately $t_w = 1.1 \times R_A C$. 図 6-4 is a plot of the time constant for various values of R_A and C. The threshold levels and charge rates both are directly proportional to the supply voltage, V_{DD} . The timing interval is, therefore, independent of the supply voltage, so long as the supply voltage is constant during the time interval.

Applying a negative-going trigger pulse simultaneously to RESET and TRIG during the timing interval discharges capacitor C and reinitiates the cycle, commencing on the positive edge of the reset pulse. The output is held low as long as the reset pulse is low. To prevent false triggering, when RESET is not asserted low, RESET must be connected to V_{DD} . If the RESET function is required and the pin is driven by external logic or a microcontroller, use a pullup resistor to V_{DD} (such as 10 k Ω) to prevent the RESET pin from floating. If the RESET function is not required, short the RESET pin directly to the V_{DD} pin.

In monostable applications, set the trip point of the trigger input by a voltage applied to CONT. An input voltage between 10% and 80% of the supply voltage, from a resistor divider with at least 500- μ A bias, provides good results.

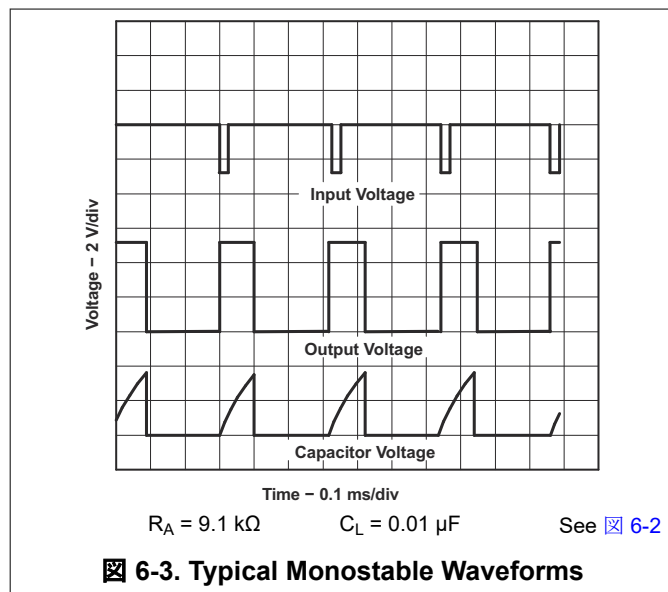


図 6-3. Typical Monostable Waveforms

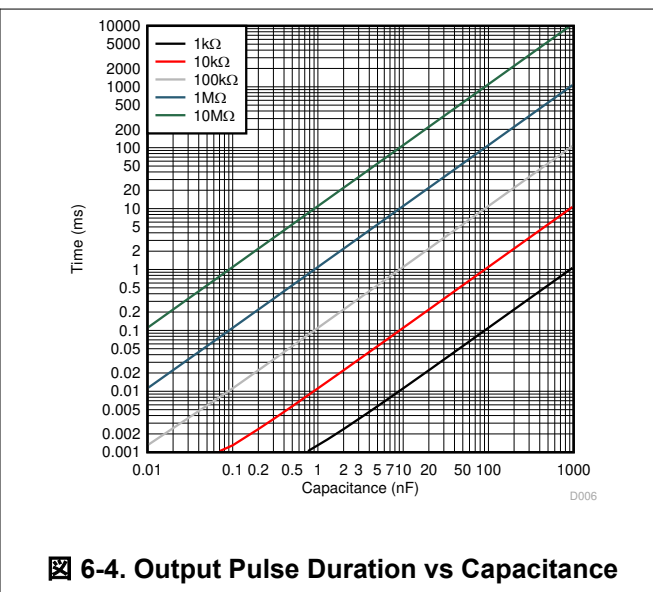
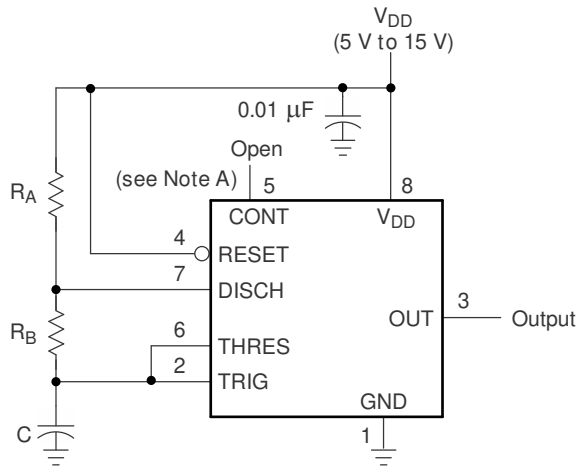


図 6-4. Output Pulse Duration vs Capacitance

6.3.2 Astable Operation

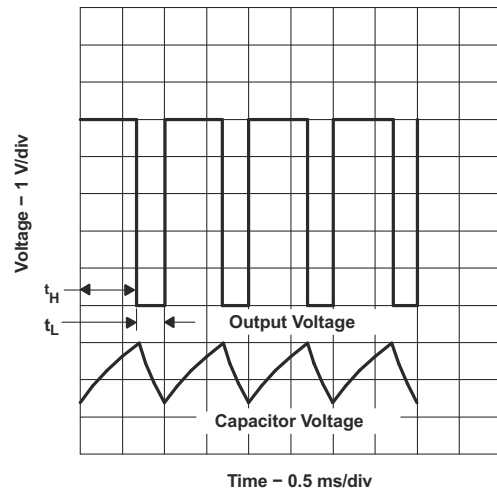
As shown in [Figure 6-5](#), adding a second resistor, R_B , to the circuit of [Figure 6-2](#) and connecting the trigger input to the threshold input causes the timer to self-trigger and run as a multivibrator. The capacitor C charges through R_A and R_B and then discharges through R_B only. Therefore, the duty cycle is controlled by the values of R_A and R_B .

This astable connection results in capacitor C charging and discharging between the threshold-voltage level ($\approx 0.67 \times V_{CC}$) and the trigger-voltage level ($\approx 0.33 \times V_{CC}$). As in the monostable circuit, charge and discharge times (and, therefore, the frequency and duty cycle) are independent of the supply voltage.



Decouple CONT voltage to ground with a capacitor to improve operation. Reevaluate for individual applications.

Figure 6-5. Circuit for Astable Operation



$R_A = 5 \text{ k}\Omega$ $R_B = 3 \text{ k}\Omega$ $C = 0.15 \text{ }\mu\text{F}$
See [Figure 6-5](#)

Figure 6-6. Typical Astable Waveforms

[Figure 6-6](#) shows typical waveforms generated during astable operation. The output high-level duration t_H and low-level duration t_L for frequencies below 100 kHz can be calculated as follows:

$$t_H = 0.693(R_A + R_B)C \tag{1}$$

$$t_L = 0.693(R_B)C \tag{2}$$

Other useful relationships are shown below:

$$\text{period} = t_H + t_L = 0.693(R_A + 2R_B)C \tag{3}$$

$$\text{frequency} \approx \frac{1.44}{(R_A + 2R_B)C} \tag{4}$$

$$\text{Output driver duty cycle} = \frac{t_L}{t_H + t_L} = \frac{R_B}{R_A + 2R_B} \tag{5}$$

$$\text{Output waveform duty cycle} = \frac{t_H}{t_H + t_L} = 1 - \frac{R_B}{R_A + 2R_B} \tag{6}$$

$$\text{Low-to-high ratio} = \frac{t_L}{t_H} = \frac{R_B}{R_A + R_B} \tag{7}$$

式 1 到 式 7 do not account for any propagation delay times from the TRIG and THRES inputs to DISCH output. These delay times add directly to the period and overcharge the capacitor, which creates differences between calculated and actual values that increase with frequency. In addition, the internal on-state resistance r_{on} during discharge adds to R_B to provide another source of timing error in the calculation when R_B is very low. The following equations provide better agreement with measured values. The formulas in 式 8 represent the actual low and high times when used at higher frequencies (beyond 100 kHz) because propagation delay and discharge on resistance is added to the formulas. The value of C_T includes both the nominal or deliberate timing capacitance, as well as parasitic capacitance on the PCB. Decoupling capacitance on CONT also affects the duty cycle, with an error contribution that depends on the capacitor leakage resistance. For additional discussion, see the [Design low-duty-cycle timer circuits article](#).

$$t_{c(H)} = C_T (R_A + R_B) \ln \left[3 - \exp \left(\frac{-t_{PLH}}{C_T (R_B + r_{on})} \right) \right] + t_{PLH}$$

$$t_{c(L)} = C_T (R_B + r_{on}) \ln \left[3 - \exp \left(\frac{-t_{PHL}}{C_T (R_A + R_B)} \right) \right] + t_{PHL}$$
(8)

These equations and those given earlier are similar in that a time constant is multiplied by the logarithm of a number or function. The limit values of the logarithmic terms must be between $\ln(2)$ at low frequencies, and $\ln(3)$ at extremely high frequencies. For a duty cycle close to 50%, an appropriate constant for the logarithmic terms can be substituted with good results. Output waveform duty cycles less than 50% require that $t_{c(H)} / t_{c(L)} < 1$ and possibly that $R_A \leq r_{on}$. These conditions can be difficult to obtain. 图 6-8 shows the nominal free-running frequency associated with various combinations of C_T and $R_A + 2 \times R_B$.

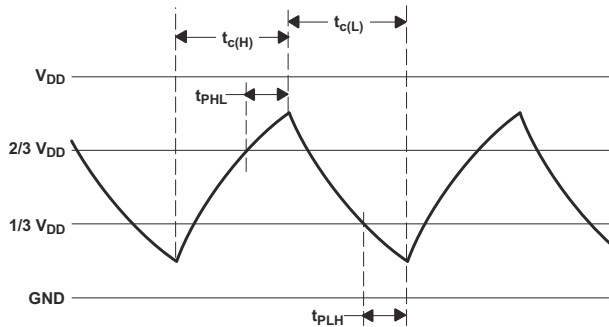


图 6-7. Trigger and Threshold Voltage Waveform

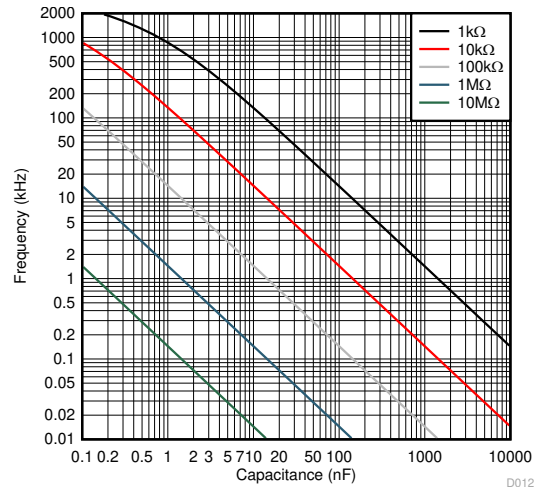


图 6-8. Nominal Free-Running Frequency vs Timing Capacitance
Resistance = $R_A + 2 \times R_B$

6.3.3 Frequency Divider

By adjusting the length of the timing cycle, the basic circuit of [Figure 6-2](#) can be made to operate as a frequency divider. [Figure 6-9](#) shows a divide-by-three circuit that makes use of the fact that retriggering cannot occur during the timing cycle.

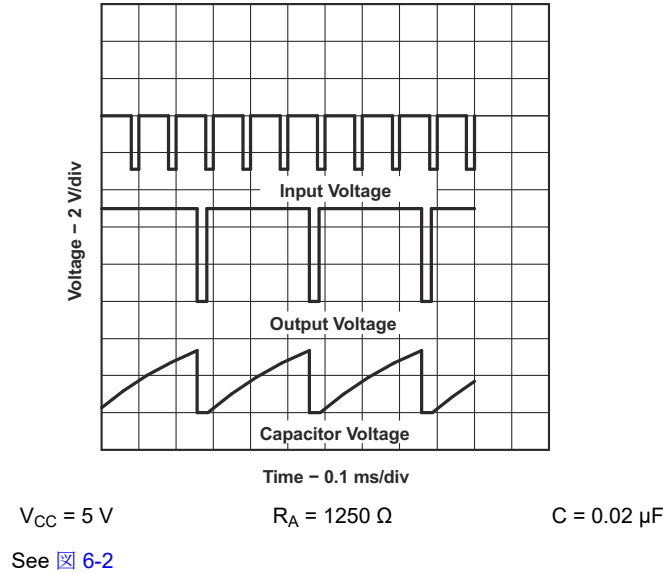


Figure 6-9. Divide-by-Three Circuit Waveforms

6.4 Device Functional Modes

[Table 6-1](#) shows the device truth table. For a valid reset voltage condition, use an external pullup resistor to V_{DD} (if using the RESET functionality), or short the RESET pin directly to V_{DD} (if the RESET functionality is not used).

Table 6-1. Function Table

RESET VOLTAGE ⁽¹⁾	TRIGGER VOLTAGE ⁽¹⁾	THRESHOLD VOLTAGE ⁽¹⁾	OUTPUT	DISCHARGE SWITCH
< MIN	Irrelevant	Irrelevant	L	On
> MAX	< MIN	Irrelevant ⁽²⁾	H	Off
> MAX	> MAX	> MAX	L	On
> MAX	> MAX	< MIN	As previously established	

- (1) For conditions shown as MIN or MAX, use the appropriate value specified under [Section 5.6](#).
 (2) CONT pin open or $2/3 V_{DD}$.

7 Application and Implementation

注

以下のアプリケーション情報は、TI の製品仕様に含まれるものではなく、TI ではその正確性または完全性を保証いたしません。個々の目的に対する製品の適合性については、お客様の責任で判断していただくこととなります。お客様は自身の設計実装を検証しテストすることで、システムの機能を確認する必要があります。

7.1 Application Information

The TLC555 timer device uses resistor and capacitor charging delay to provide a programmable time delay or operating frequency. [セクション 7.2](#) presents a simplified discussion of the design process. Reset mode forces output and discharge low and provides a small reduction in supply current.

7.2 Typical Applications

7.2.1 Missing-Pulse Detector

The circuit shown in [図 7-1](#) can be used to detect a missing pulse or abnormally long spacing between consecutive pulses in a train of pulses. The timing interval of the monostable circuit is re-triggered continuously by the input pulse train as long as the pulse spacing is less than the timing interval. A longer pulse spacing, missing pulse, or terminated pulse train permits the timing interval to be completed, thereby generating an output pulse as shown in [図 7-2](#).

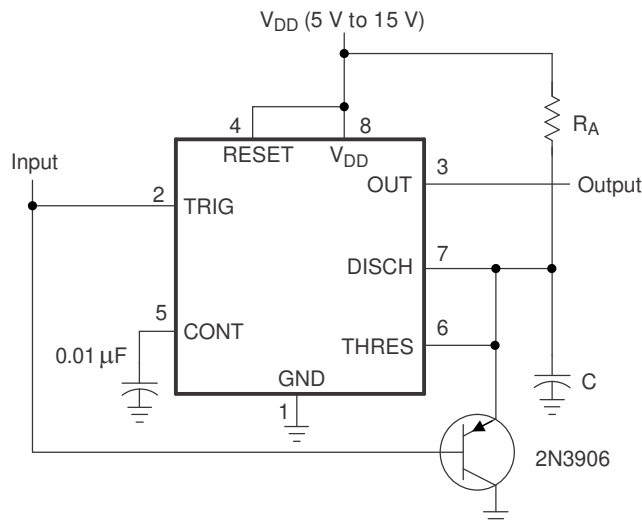


図 7-1. Circuit for Missing-Pulse Detector

7.2.1.1 Design Requirements

Input fault (missing pulses) must be input high. An input stuck low cannot be detected because the timing capacitor (C) remains discharged.

7.2.1.2 Detailed Design Procedure

Choose R_A and C so that $R_A \times C > [\text{maximum normal input high time}]$.

7.2.1.3 Application Curve

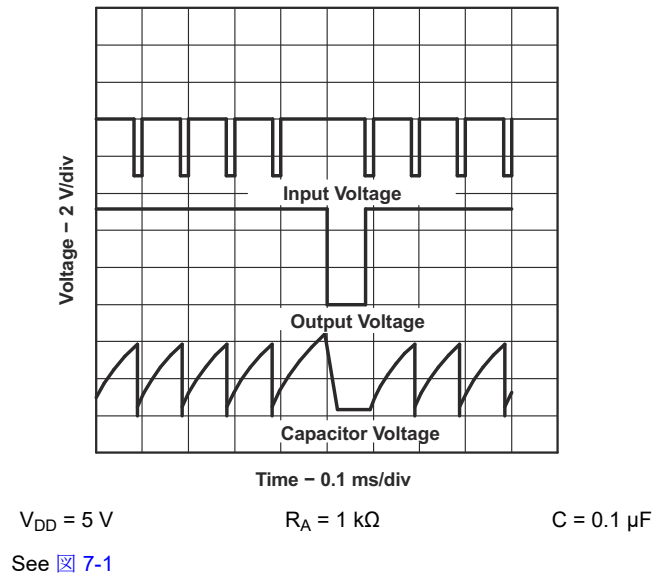
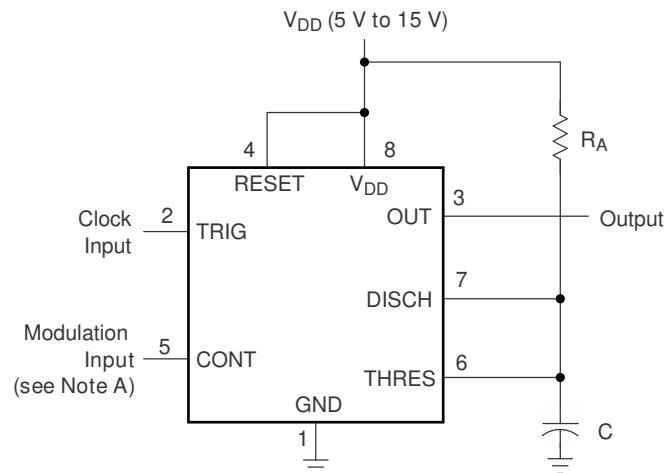


Figure 7-2. Timing Waveforms for Missing-Pulse Detector

7.2.2 Pulse-Width Modulation

To modify timer operation, apply an external voltage (or current) to CONT to modulate the internal threshold and trigger voltages. [Figure 7-3](#) shows a circuit for pulse-width modulation. A continuous input pulse train triggers the monostable circuit, and a control signal modulates the threshold voltage. [Figure 7-4](#) shows the resulting duty cycle versus control voltage transfer function. Attempting to run under 10% duty cycle can result in inconsistent output pulses. Attempting to run close to 100% duty cycle results in frequency division by 2, then 3, then 4.



- A. The modulating signal can be direct or capacitively coupled to CONT. For direct coupling, consider the effects of modulation source voltage and impedance on the bias of the timer.

Figure 7-3. Circuit for Pulse-Width Modulation

7.2.2.1 Design Requirements

The clock input must have V_{OL} and V_{OH} levels that are less than and greater than $1/3 V_{DD}$, respectively. Clock input V_{OL} time must be less than minimum output high time; therefore, a high (positive) duty cycle clock is recommended. Minimum recommended modulation voltage is 1 V. Lower CONT voltage can greatly increase threshold comparator propagation delay and storage time. The application must be tolerant of a nonlinear transfer function; the relationship between modulation input and pulse width is not linear because the capacitor charge is RC-based with an negative exponential curve.

7.2.2.2 Detailed Design Procedure

Choose R_A and C so that $R_A \times C$ is same or less than clock input period. [Fig 7-4](#) shows the non linear relationship between control voltage and output duty cycle. Duty cycle is function of control voltage and clock period relative to RC time constant.

7.2.2.3 Application Curve

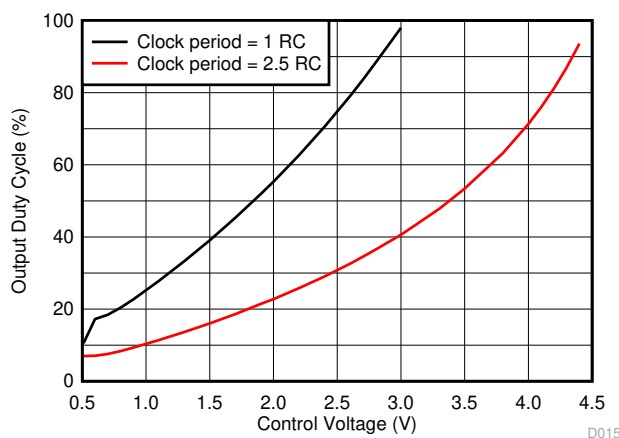
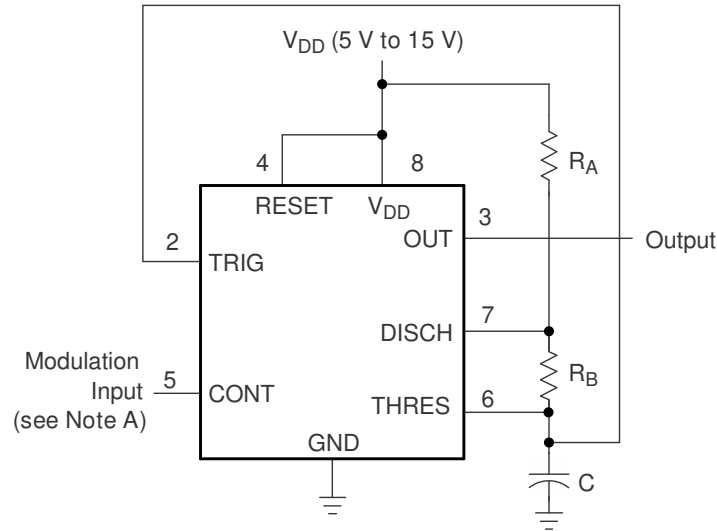


Fig 7-4. Pulse-Width-Modulation vs Control Voltage
Clock Duty Cycle 98%, $V_{DD} = 5 V$

7.2.3 Pulse-Position Modulation

As shown in [Fig 7-5](#), any of these timers can be used as a pulse-position modulator. This application modulates the threshold voltage and thereby the time delay of a free-running oscillator. [Fig 7-6](#) and [Fig 7-7](#) shows the output frequency and duty cycle versus control voltage.



- A. The modulating signal can be direct or capacitively coupled to CONT. For direct coupling, consider the effects of modulation source voltage and impedance on the bias of the timer.

$$R_A = 3 \text{ k}\Omega$$

$$R_B = 309 \text{ k}\Omega$$

$$C = 1 \text{ nF}$$

図 7-5. Circuit for Pulse-Position Modulation

7.2.3.1 Design Requirements

Both dc- and ac-coupled modulation input changes the upper and lower voltage thresholds for the timing capacitor. Both frequency and duty cycle vary with the modulation voltage. Control voltage less than 1 V can result in output glitches instead of a steady-output pulse stream

7.2.3.2 Detailed Design Procedure

The nominal output frequency and duty cycle for control voltage set to 2/3 of V_{DD} can be determined using formulas in [セクション 6.3.2](#) section.

7.2.3.3 Application Curves

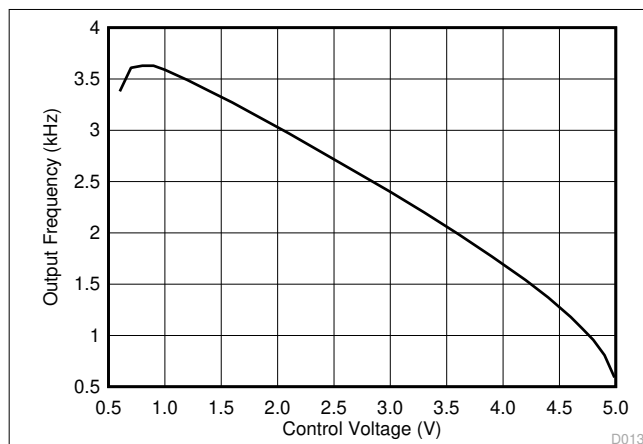


図 7-6. Pulse-Position-Modulation Frequency vs Control Voltage, $V_{DD} = 5 \text{ V}$

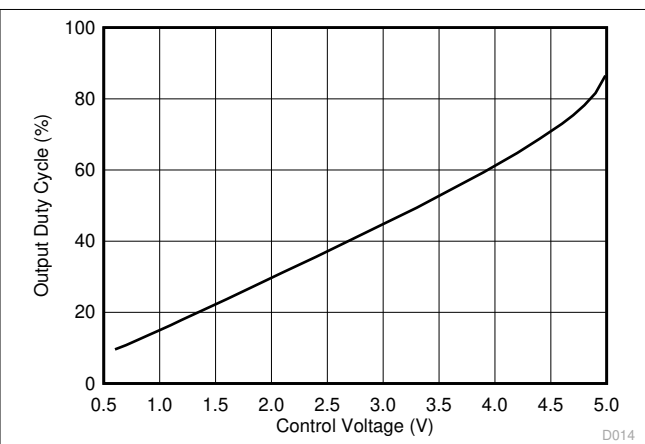
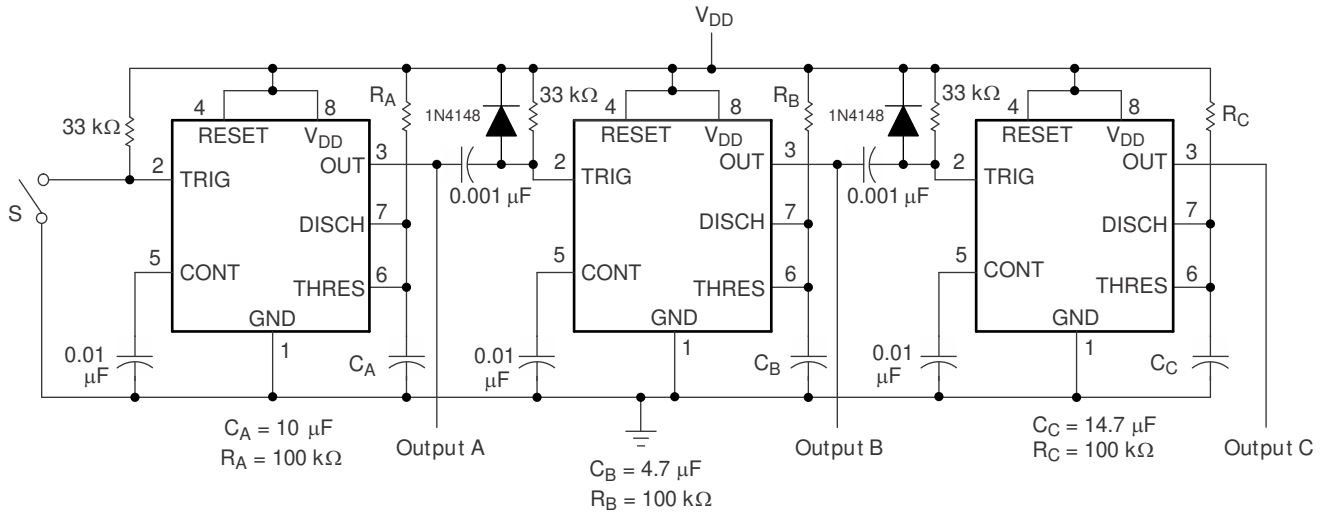


図 7-7. Pulse-Position-Modulation Duty Cycle vs Control Voltage, $V_{DD} = 5 \text{ V}$

7.2.4 Sequential Timer

Many applications, such as computers, require signals for initializing conditions during start-up. Other applications, such as test equipment, require activation of test signals in sequence. These timing circuits can be connected to provide such sequential control. The timers can be used in various combinations of astable or monostable circuit connections, with or without modulation, for extremely flexible waveform control. [Fig 7-8](#) shows a sequencer circuit with possible applications in many systems, and [Fig 7-9](#) shows the output waveforms.



S closes momentarily at $t = 0$.

Fig 7-8. Sequential Timer Circuit

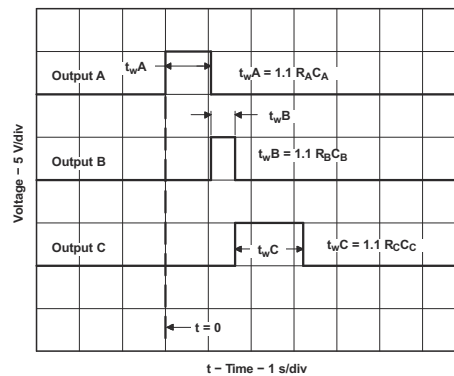
7.2.4.1 Design Requirements

The sequential-timer application chains together multiple monostable timers. The joining components are the 33-kΩ resistors and 0.001-μF capacitors. The output high-to-low edge passes a 10-μs start pulse to the next monostable. A diode is required to prevent overvoltage on the trigger input when on the previous output low-to-high edge.

7.2.4.2 Detailed Design Procedure

The timing resistors and capacitors can be chosen using this formula: $t_w = 1.1 \times R \times C$.

7.2.4.3 Application Curve



See [Fig 7-8](#)

Fig 7-9. Sequential Timer Waveforms

7.2.5 Designing for Improved ESD Performance

The TLC555 internal HBM and CDM protection allows for safe assembly in ESD-controlled environments. In applications that expose the pins of the TLC555 to ESD, additional protection is highly recommended. The following test board schematic has bypass capacitors, current-limiting resistors, and voltage-clamping TVS diodes to provide additional protection for commonly exposed pins (Reset, Trig, and Output) against ESD.

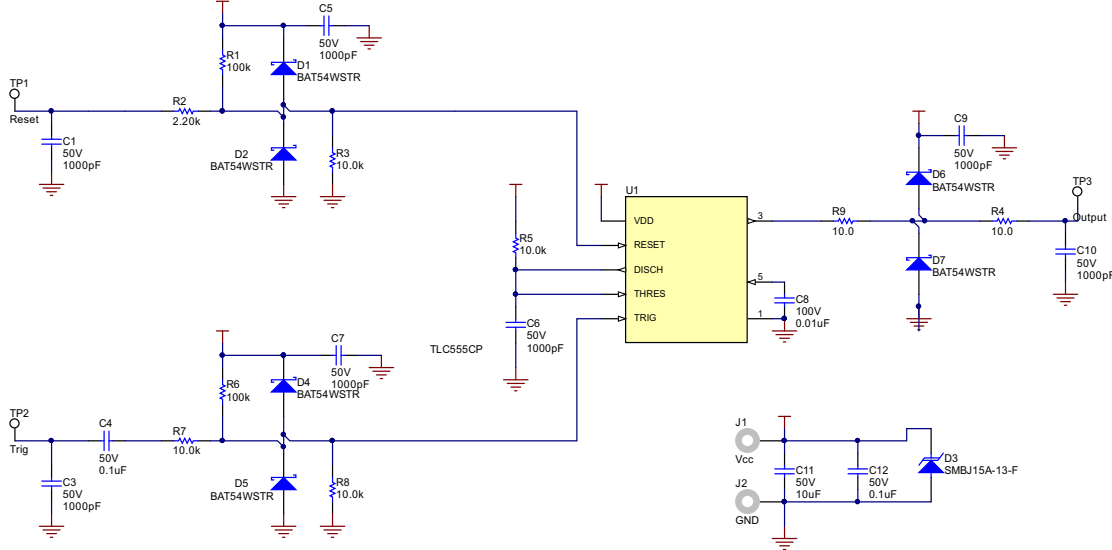


図 7-10. ESD Test Schematic

The following table gives the ESD protection levels recorded for different supply voltages and external components populated. Using only passive components to protect the TLC555 with a single 15-V supply is not recommended because the higher voltage allows for an unacceptable amount of current to flow through the device.

表 7-1. ESD Test Results

SUPPLY VOLTAGE	ONLY PASSIVE COMPONENTS POPULATED D1..D7 NOT POPULATED ⁽¹⁾	ALL COMPONENTS POPULATED ⁽¹⁾
5 V	8 kV	12 kV
15 V	Not recommended	12 kV

(1) Sample results. Results can vary with populated components, board layout, and samples used.

7.3 Power Supply Recommendations

The TLC555 requires a voltage supply greater than or equal to 2 V, 3 V, or 5 V based the coldest ambient temperature supported and a supply voltage less than or equal to 15 V. Adequate power supply bypassing is necessary to protect associated circuitry and provide stable output pulses. Minimum recommended is 0.1- μF ceramic in parallel with 1- μF electrolytic. Place the bypass capacitors as close as possible to the TLC555 and minimize the trace length.

7.4 Layout

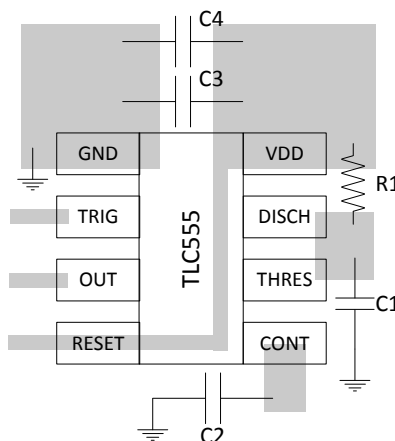
7.4.1 Layout Guidelines

Standard PCB rules apply to routing the TLC555. The 0.1- μF ceramic capacitor in parallel with a 1- μF electrolytic capacitor must be as close as possible to the TLC555. The capacitor used for the time delay must also be placed as close to the discharge pin. A ground plane on the bottom layer can be used to provide better noise immunity and signal integrity.

☒ 7-11 is the basic layout for various applications.

- C1—based on time delay calculations
- C2—0.01- μF bypass capacitor for control voltage pin
- C3—0.1- μF bypass ceramic capacitor
- C4—1- μF electrolytic bypass capacitor
- R1—based on time-delay calculations

7.4.2 Layout Example



☒ 7-11. Layout Example

8 Device and Documentation Support

8.1 ドキュメントの更新通知を受け取る方法

ドキュメントの更新についての通知を受け取るには、www.tij.co.jp のデバイス製品フォルダを開いてください。[通知] をクリックして登録すると、変更されたすべての製品情報に関するダイジェストを毎週受け取ることができます。変更の詳細については、改訂されたドキュメントに含まれている改訂履歴をご覧ください。

8.2 サポート・リソース

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8.4 静電気放電に関する注意事項



この IC は、ESD によって破損する可能性があります。テキサス・インスツルメンツは、IC を取り扱う際には常に適切な注意を払うことを推奨します。正しい取り扱いおよび設置手順に従わない場合、デバイスを破損するおそれがあります。

ESD による破損は、わずかな性能低下からデバイスの完全な故障まで多岐にわたります。精密な IC の場合、パラメータがわずかに変化するだけで公表されている仕様から外れる可能性があるため、破損が発生しやすくなっています。

8.5 用語集

[テキサス・インスツルメンツ用語集](#) この用語集には、用語や略語の一覧および定義が記載されています。

9 Revision History

資料番号末尾の英字は改訂を表しています。その改訂履歴は英語版に準じています。

Changes from Revision I (July 2019) to Revision J (November 2023)	Page
「特長」の MIL-STD-883C、手法 3015.2 に従い ESD 保護仕様を 2000V から、ANSI/ESDA/JEDEC JS-001 に従い 1000V に変更	1
「概要」セクションで「製品情報」表を「パッケージ情報」に変更し、「本体サイズ (公称)」を「パッケージ サイズ」に変更	1
Added ESD Ratings table and HBM, CDM, and MM specifications	5
Changed thermal resistance and characterization parameter values for SOIC and PDIP packages in Thermal Information table	6
Changed reset current (I_{RESET}) test conditions to $V_{\text{RESET}} = V_{\text{DD}}$, in Electrical Characteristics: $V_{\text{DD}} = 5 \text{ V}$ and Electrical Characteristics: $V_{\text{DD}} = 15 \text{ V}$	7
Added new reset current (I_{RESET}) typical specification, for test condition $V_{\text{RESET}} = 0 \text{ V}$, to Electrical Characteristics: $V_{\text{DD}} = 5 \text{ V}$ and Electrical Characteristics: $V_{\text{DD}} = 15 \text{ V}$	7
Changed supply current (I_{DD}) typical value from 170 μA to 180 μA in Electrical Characteristics: $V_{\text{DD}} = 5 \text{ V}$	7
Changed title of Operating Characteristics section to Timing Characteristics and clarified that values are specified by design or characterization	10
Deleted Initial error of timing interval specification in Timing Characteristics	10
Added Figure 5-4, Supply Current vs Supply Voltage, Unit 2	11
Changed Figure 5-3, Supply Current vs Supply Voltage, to add "Unit 1" to title, and deleted 0°C and 70°C curves	11
Changed functional block diagram to simplified schematic and moved to Overview	13
Updated Functional Block Diagram	13
Added guidance for RESET pin pullup resistance and CONT pin voltage range to Monostable Operation	13

• Added clarity regarding nominal operating frequency and parasitic terms in <i>Astable Operation</i>	15
• Deleted link to deprecated TLC555 Design Calculator in <i>Astable Operation</i>	15
• Deleted Figure 17, <i>Equivalent Schematic</i> , and added guidance concerning the RESET pin in <i>Device Functional Modes</i>	17

Changes from Revision H (August 2016) to Revision I (July 2019)	Page
• Added MIN value for input voltage in <i>Absolute Maximum Ratings</i>	5
• Added discharge pin in <i>Absolute Maximum Ratings</i>	5
• Changed MIN supply voltage based on part number in <i>Recommended Operating Conditions</i>	5
• Added power dissipation capacitance TYP value in <i>Electrical Characteristics: $V_{DD} = 2\text{ V}$ for TLC555C, $V_{DD} = 3\text{ V}$ for TLC555I</i>	6
• Added trigger, threshold capacitance TYP value in <i>Electrical Characteristics: $V_{DD} = 5\text{ V}$</i>	7
• Changed V_{OH} test condition current to -1 mA in <i>Electrical Characteristics: $V_{DD} = 5\text{ V}$</i>	7
• Added power dissipation capacitance TYP value in <i>Electrical Characteristics: $V_{DD} = 5\text{ V}$</i>	7
• Added trigger, threshold capacitance TYP value in <i>Electrical Characteristics: $V_{DD} = 15\text{ V}$</i>	9
• Added power dissipation capacitance TYP value in <i>Electrical Characteristics: $V_{DD} = 15\text{ V}$</i>	9
• Added <i>Operating Characteristics</i> to the <i>Specifications</i> section.....	10
• Added Supply Current vs Supply Voltage chart to the <i>Typical Characteristics</i> section.....	11
• Added Control Impedance vs Temperature chart to the <i>Typical Characteristics</i> section.....	11
• Added Output Low Resistance vs Temperature chart to the <i>Typical Characteristics</i> section.....	11
• Added Output High Resistance vs Temperature chart to the <i>Typical Characteristics</i> section.....	11
• Added Propagation Delay vs Control Voltage chart, $V_{DD} = 2\text{ V}$ to the <i>Typical Characteristics</i> section.....	11
• Added Propagation Delay vs Control Voltage chart, $V_{DD} = 5\text{ V}$ to the <i>Typical Characteristics</i> section.....	11
• Changed trigger high hold time to $1\text{ }\mu\text{s}$ in <i>Monostable Operation</i>	13
• Changed minimum monostable pulse width to $1\text{ }\mu\text{s}$ in <i>Monostable Operation</i>	13
• Changed Output Pulse Duration vs Capacitance chart scale down to 0.001 ms in <i>Monostable Operation</i>	13
• Added more astable frequency formulas to the <i>Astable Operation</i> section.....	15
• Changed scale on Free-Running Frequency vs Timing Capacitance chart up to 2 MHz in the <i>Astable Operation</i> section.....	15
• Added CONT pin table note to the Table 6-1, <i>Function Table</i> in the <i>Device Functional Modes</i>	17
• Changed the application curve chart in the <i>Pulse-Width Modulation</i> section.....	20
• Changed the application curve charts in the <i>Pulse-Position Modulation</i> section.....	21
• Added clamping diodes to Sequential Timer Circuit in the <i>Sequential Timer</i> section.....	22
• Added <i>Designing for Improved ESD Performance</i> section to the <i>Application Information</i> section.....	23

Changes from Revision G (November 2008) to Revision H (August 2016)	Page
• 「機能説明」セクション、「デバイスの機能モード」セクション、「アプリケーションと実装」セクション、「電源に関する推奨事項」セクション、「レイアウト」セクション、「デバイスおよびドキュメントのサポート」セクション、「メカニカル、パッケージ、および注文情報」セクションを追加	1
• Changed values in the <i>Thermal Information</i> table to align with JEDEC standards.....	6
• Deleted <i>Dissipation Ratings</i> table	6

10 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead finish/ Ball material (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
5962-89503012A	ACTIVE	LCCC	FK	20	55	Non-RoHS & Green	SNPB	N / A for Pkg Type	-55 to 125	5962-89503012A TLC555MFKB	Samples
5962-8950301PA	ACTIVE	CDIP	JG	8	50	Non-RoHS & Green	SNPB	N / A for Pkg Type	-55 to 125	8950301PA TLC555M	Samples
TLC555CD	OBSOLETE	SOIC	D	8		TBD	Call TI	Call TI	0 to 70	TL555C	
TLC555CDR	ACTIVE	SOIC	D	8	2500	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	0 to 70	TL555C	Samples
TLC555CP	ACTIVE	PDIP	P	8	50	RoHS & Green	NIPDAU	N / A for Pkg Type	0 to 70	TLC555CP	Samples
TLC555CPS	ACTIVE	SO	PS	8	80	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	0 to 70	P555	Samples
TLC555CPSR	ACTIVE	SO	PS	8	2000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	0 to 70	P555	Samples
TLC555CPW	ACTIVE	TSSOP	PW	14	90	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	0 to 70	P555	Samples
TLC555CPWR	ACTIVE	TSSOP	PW	14	2000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	0 to 70	P555	Samples
TLC555ID	OBSOLETE	SOIC	D	8		TBD	Call TI	Call TI	-40 to 85	TL555I	
TLC555IDR	ACTIVE	SOIC	D	8	2500	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 85	TL555I	Samples
TLC555IP	ACTIVE	PDIP	P	8	50	RoHS & Green	NIPDAU	N / A for Pkg Type	-40 to 85	TLC555IP	Samples
TLC555MFKB	ACTIVE	LCCC	FK	20	55	Non-RoHS & Green	SNPB	N / A for Pkg Type	-55 to 125	5962-89503012A TLC555MFKB	Samples
TLC555MJG	ACTIVE	CDIP	JG	8	50	Non-RoHS & Green	SNPB	N / A for Pkg Type	-55 to 125	TLC555MJG	Samples
TLC555MJGB	ACTIVE	CDIP	JG	8	50	Non-RoHS & Green	SNPB	N / A for Pkg Type	-55 to 125	8950301PA TLC555M	Samples
TLC555QDR	ACTIVE	SOIC	D	8	2500	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	TL555Q	Samples
TLC555QDRG4	ACTIVE	SOIC	D	8	2500	RoHS & Green	NIPDAU	Level-1-260C-UNLIM		TL555Q	Samples

(1) The marketing status values are defined as follows:
ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSELETE: TI has discontinued the production of the device.

⁽²⁾ **RoHS:** TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

Green: TI defines "Green" to mean the content of Chlorine (Cl) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

⁽³⁾ MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

⁽⁴⁾ There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

⁽⁵⁾ Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

⁽⁶⁾ Lead finish/Ball material - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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OTHER QUALIFIED VERSIONS OF TLC555, TLC555M :

- Catalog : [TLC555](#)
- Automotive : [TLC555-Q1](#), [TLC555-Q1](#)
- Military : [TLC555M](#)

NOTE: Qualified Version Definitions:

- Catalog - TI's standard catalog product
- Automotive - Q100 devices qualified for high-reliability automotive applications targeting zero defects
- Military - QML certified for Military and Defense Applications

TAPE AND REEL INFORMATION

QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE


*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TLC555CDR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
TLC555CPSR	SO	PS	8	2000	330.0	16.4	8.35	6.6	2.5	12.0	16.0	Q1
TLC555CPWR	TSSOP	PW	14	2000	330.0	12.4	6.9	5.6	1.6	8.0	12.0	Q1
TLC555IDR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
TLC555QDRG4	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1

TAPE AND REEL BOX DIMENSIONS


*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TLC555CDR	SOIC	D	8	2500	353.0	353.0	32.0
TLC555CPSR	SO	PS	8	2000	367.0	367.0	38.0
TLC555CPWR	TSSOP	PW	14	2000	356.0	356.0	35.0
TLC555IDR	SOIC	D	8	2500	356.0	356.0	35.0
TLC555QDRG4	SOIC	D	8	2500	353.0	353.0	32.0

TUBE


*All dimensions are nominal

Device	Package Name	Package Type	Pins	SPQ	L (mm)	W (mm)	T (μm)	B (mm)
5962-89503012A	FK	LCCC	20	55	506.98	12.06	2030	NA
TLC555CP	P	PDIP	8	50	506	13.97	11230	4.32
TLC555CPS	PS	SOP	8	80	530	10.5	4000	4.1
TLC555CPW	PW	TSSOP	14	90	530	10.2	3600	3.5
TLC555IP	P	PDIP	8	50	506	13.97	11230	4.32
TLC555MFKB	FK	LCCC	20	55	506.98	12.06	2030	NA

PACKAGE OUTLINE

JG0008A

CDIP - 5.08 mm max height

CERAMIC DUAL IN-LINE PACKAGE



NOTES:

1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. This package can be hermetically sealed with a ceramic lid using glass frit.
4. Index point is provided on cap for terminal identification.
5. Falls within MIL STD 1835 GDIP1-T8

EXAMPLE BOARD LAYOUT

JG0008A

CDIP - 5.08 mm max height

CERAMIC DUAL IN-LINE PACKAGE



LAND PATTERN EXAMPLE
NON SOLDER MASK DEFINED
SCALE: 9X

4230036/A 09/2023

GENERIC PACKAGE VIEW

FK 20

LCCC - 2.03 mm max height

8.89 x 8.89, 1.27 mm pitch

LEADLESS CERAMIC CHIP CARRIER

This image is a representation of the package family, actual package may vary.
Refer to the product data sheet for package details.



4229370VA\



D0008A

PACKAGE OUTLINE

SOIC - 1.75 mm max height

SMALL OUTLINE INTEGRATED CIRCUIT



4214825/C 02/2019

NOTES:

- Linear dimensions are in inches [millimeters]. Dimensions in parenthesis are for reference only. Controlling dimensions are in inches. Dimensioning and tolerancing per ASME Y14.5M.
- This drawing is subject to change without notice.
- This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed .006 [0.15] per side.
- This dimension does not include interlead flash.
- Reference JEDEC registration MS-012, variation AA.

EXAMPLE BOARD LAYOUT

D0008A

SOIC - 1.75 mm max height

SMALL OUTLINE INTEGRATED CIRCUIT



LAND PATTERN EXAMPLE
 EXPOSED METAL SHOWN
 SCALE:8X



SOLDER MASK DETAILS

4214825/C 02/2019

NOTES: (continued)

- 6. Publication IPC-7351 may have alternate designs.
- 7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.

EXAMPLE STENCIL DESIGN

D0008A

SOIC - 1.75 mm max height

SMALL OUTLINE INTEGRATED CIRCUIT



SOLDER PASTE EXAMPLE
BASED ON .005 INCH [0.125 MM] THICK STENCIL
SCALE:8X

4214825/C 02/2019

NOTES: (continued)

8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
9. Board assembly site may have different recommendations for stencil design.

MECHANICAL DATA

PS (R-PDSO-G8)

PLASTIC SMALL-OUTLINE PACKAGE



- NOTES:
- A. All linear dimensions are in millimeters.
 - B. This drawing is subject to change without notice.
 - C. Body dimensions do not include mold flash or protrusion, not to exceed 0,15.

PS (R-PDSO-G8)

PLASTIC SMALL OUTLINE



- NOTES:
- All linear dimensions are in millimeters.
 - This drawing is subject to change without notice.
 - Publication IPC-7351 is recommended for alternate designs.
 - Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC-7525 for other stencil recommendations.
 - Customers should contact their board fabrication site for solder mask tolerances between and around signal pads.

P (R-PDIP-T8)

PLASTIC DUAL-IN-LINE PACKAGE



- NOTES:
- A. All linear dimensions are in inches (millimeters).
 - B. This drawing is subject to change without notice.
 - C. Falls within JEDEC MS-001 variation BA.

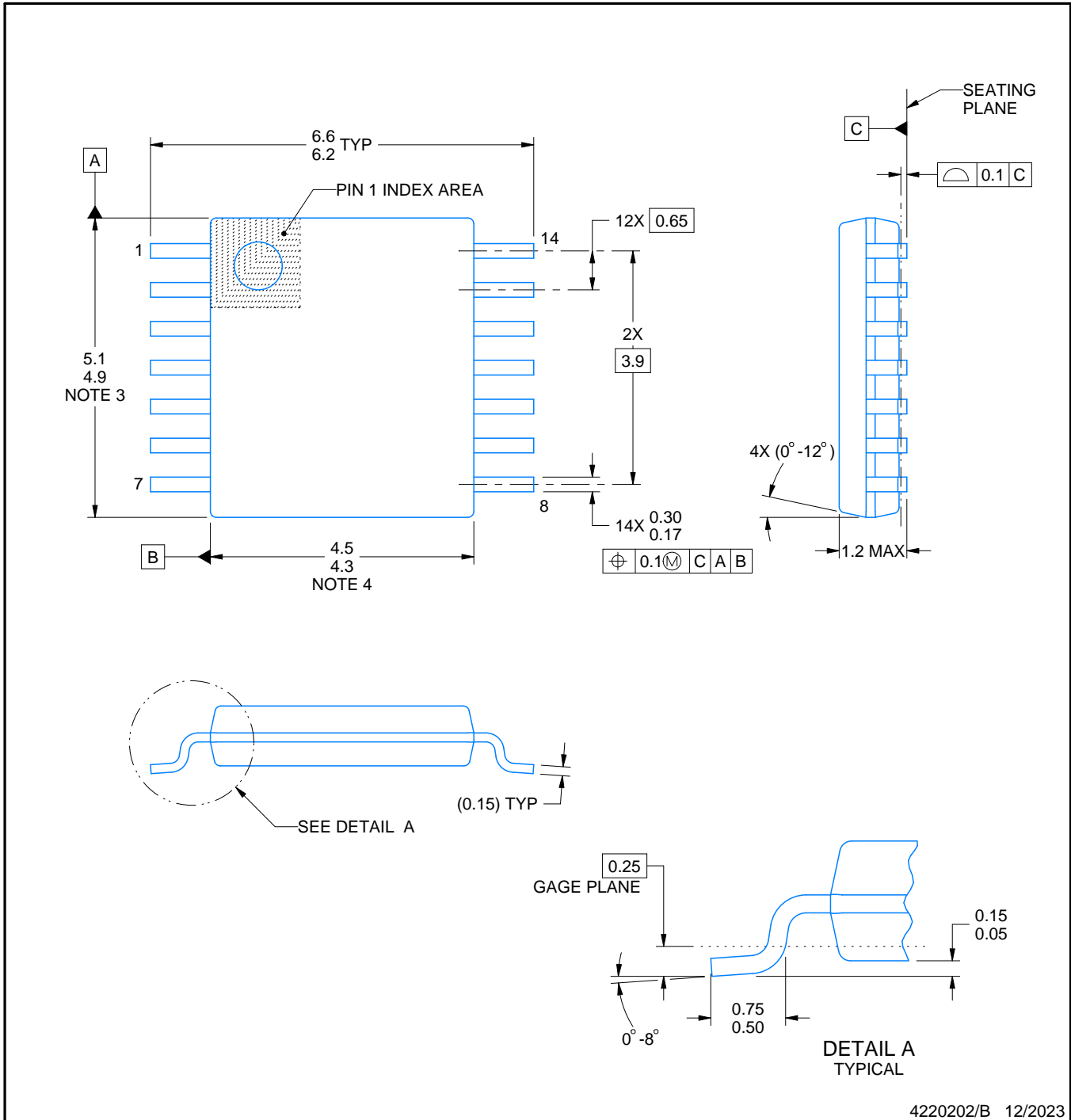
PW0014A



PACKAGE OUTLINE

TSSOP - 1.2 mm max height

SMALL OUTLINE PACKAGE



4220202/B 12/2023

NOTES:

1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.15 mm per side.
4. This dimension does not include interlead flash. Interlead flash shall not exceed 0.25 mm per side.
5. Reference JEDEC registration MO-153.

EXAMPLE BOARD LAYOUT

PW0014A

TSSOP - 1.2 mm max height

SMALL OUTLINE PACKAGE



LAND PATTERN EXAMPLE
EXPOSED METAL SHOWN
SCALE: 10X



4220202/B 12/2023

NOTES: (continued)

- 6. Publication IPC-7351 may have alternate designs.
- 7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.

EXAMPLE STENCIL DESIGN

PW0014A

TSSOP - 1.2 mm max height

SMALL OUTLINE PACKAGE



SOLDER PASTE EXAMPLE
BASED ON 0.125 mm THICK STENCIL
SCALE: 10X

4220202/B 12/2023

NOTES: (continued)

8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
9. Board assembly site may have different recommendations for stencil design.

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