

# UCC2732x/UCC3732x イネーブル機能搭載、シングル、9A 高速ローサイド MOSFET ドライバ

## 1 特長

- イネーブル機能を追加した業界標準のピン配置
- TrueDrive を使用して、ミラー プラトー領域で  $\pm 9A$  の高いピーク電流駆動能力
- 独自のバイポーラおよび CMOS 出力段を使用した効率的な定電流ソーシング
- 電源電圧から独立した TTL/CMOS 互換入力
- 標準立ち上がり時 / 立ち下がり時 20ns (10nF 負荷時)
- 標準伝播遅延時間: 25ns (入力立ち下がり時)、35ns (入力立ち上がり時)
- 電源電圧: 4V ~ 15V
- 熱的に強化された MSOP PowerPAD™ パッケージ、 $4.7^{\circ}\text{C/W } \theta_{jc}$  で供給
- $-40^{\circ}\text{C} \sim +105^{\circ}\text{C}$  で仕様を規定
- 8 ピン SOIC および PDIP パッケージでの鉛フリー仕上げ (CU NIPDAU)

## 2 アプリケーション

- スイッチ・モード電源
- DC/DC コンバータ
- モーター・コントローラ
- Class-D スwitchング アンプ
- ライン・ドライバ
- パルストランスドライバ

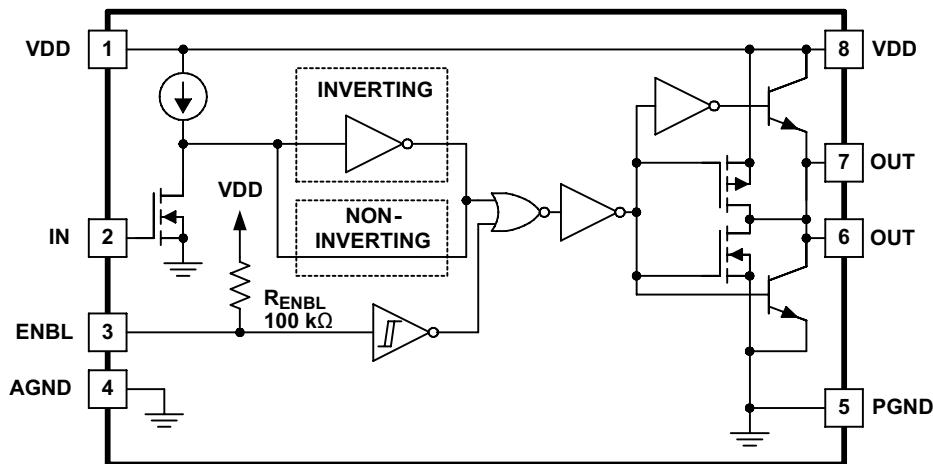
## 3 概要

UCC2732x/UCC3732x ファミリの高速ドライバは、業界標準のピン配置で 9A のピーク駆動電流を供給します。これらのドライバは、高い  $dV/dt$  遷移により極端なミラー電流を必要とするシステム向けに、最大の MOSFET を駆動できます。この結果、外部回路の追加が不要になり、複数の部品を置き換え、スペースの節減、設計の複雑さの緩和、組み立てコストの削減が可能になります。反転 (UCC37321) と非反転 (UCC37322) の 2 つの標準ロジック オプションが用意されています。

### 製品情報

部品番号	パッケージ <sup>(1)</sup>	本体サイズ (公称)
UCC2732x	MSOP-PowerPAD (8)	3.00mm × 3.00mm
UCC3732x	SOIC (8)	3.91mm × 4.90mm

(1) 利用可能なすべてのパッケージについては、データシートの末尾にある注文情報を参照してください。



ブロック図



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## 4 概要 (続き)

貫通電流を本質的に最小化する設計により、これらの出力は、MOSFET のスイッチング遷移中のミラー プラトー領域で最も必要とされる、高いゲート駆動電流を供給できます。バイポーラと MOSFET トランジスタを並列接続した独自のハイブリッド出力段 (TrueDrive) により、低い電源電圧で効率的な電流供給が可能になります。この駆動アーキテクチャにより、UCC3732x は業界標準の 6A、9A、および多くの 12A ドライバ アプリケーションで使用できます。ラッチアップおよび ESD 保護回路も搭載されています。最後に、UCC3732x にはイネーブル (ENBL) 機能があり、ドライバ アプリケーションの動作をより的確に制御できます。ENBL は、業界標準のピン配置では未使用のままにしてあったピン 3 に実装されています。このピンは、アクティブ ハイ ロジックでは内部で  $V_{DD}$  にプルアップされ、標準動作時にはオープンのままにできます。

8 ピン SOIC (D) パッケージ製品に加えて、UCC3732x は熱的に強化された小型の 8 ピン MSOP PowerPAD™ (DGN) パッケージでも供給されます。PowerPAD パッケージは熱抵抗を大幅に低減するため、温度動作範囲が拡張され、長期的な信頼性が向上します。

## 5 Related Products

PRODUCT	DESCRIPTION	PACKAGES
UCC37323/4/5	Dual 4-A Low-Side Drivers	MSOP–8 PowerPAD, SOIC–8, PDIP–8
UCC27423/4/5	Dual 4-A Low-Side Drivers with Enable	MSOP–8 PowerPAD, SOIC–8, PDIP–8
TPS2811/12/13	Dual 2-A Low-Side Drivers with Internal Regulator	TSSOP–8, SOIC–8, PDIP–8
TPS2814/15	Dual 2-A Low-Side Drivers with Two Inputs per Channel	TSSOP–8, SOIC–8, PDIP–8
TPS2816/17/18/19	Single 2-A Low-Side Driver with Internal Regulator	5-Pin SOT–23
TPS2828/29	Single 2-A Low-Side Driver	5-Pin SOT–23

## 6 Pin Configuration and Functions

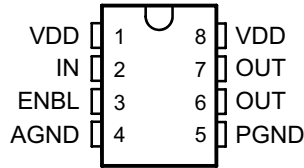


図 6-1. D, and DGN Packages 8-Pin SOIC, and MSOP With PowerPAD Top View

表 6-1. Pin Functions

PIN		I/O	DESCRIPTION
NAME	NO.		
AGND	4	—	The AGND and the PGND must be connected by a single thick trace directly under the device. There must be a low ESR, low ESL capacitor of 0.1 $\mu$ F between VDD (pin 8) and PGND and a separate 0.1- $\mu$ F capacitor between VDD (pin 1) and AGND. The power MOSFETs must be located on the PGND side of the device while the control circuit must be on the AGND side of the device. The control circuit ground must be common with the AGND while the PGND must be common with the source of the power FETs.
ENBL	3	I	Enable input for the driver with logic compatible threshold and hysteresis. The driver output can be enabled and disabled with this pin. It is internally pulled up to V <sub>DD</sub> with 100-k $\Omega$ resistor for active high operation. When the device is disabled, the output state is, low regardless of the input state.
IN	2	I	Input signal of the driver which has logic compatible threshold and hysteresis.
OUT	6, 7	O	Driver outputs that must be connected together externally. The output stage is capable of providing 9-A peak drive current to the gate of a power MOSFET.
PGND	5	—	Common ground for output stage. This ground must be connected very closely to the source of the power MOSFET which the driver is driving. Grounds are separated to minimize ringing affects due to output switching di/dt which can affect the input threshold.
VDD	1, 8	I	Supply voltage and the power input connections for this device. Two pins must be connected together externally.

## 7 Specifications

### 7.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted) <sup>(1)</sup> <sup>(2)</sup>

		MIN	MAX	UNIT
Supply voltage, $V_{DD}$		-0.3	16	V
Output current (OUT) DC, $I_{OUT\_DC}$			0.6	A
Input voltage (IN), $V_{IN}$		-0.3	6 V or $V_{DD} + 0.3^{(3)}$	V
Enable voltage (ENBL)		-0.3	6 V or $V_{DD} + 0.3^{(3)}$	V
Power dissipation at $T_A = 25^\circ\text{C}$	D package		650	mW
	DGN package		3	W
	P package		350	mW
Lead temperature (soldering, 10 s)			300	$^\circ\text{C}$
Junction operating temperature, $T_J$		-55	150	$^\circ\text{C}$
Storage temperature, $T_{stg}$		-65	150	$^\circ\text{C}$

- (1) Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under [セクション 7.3 Recommended Operating Conditions](#). Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
- (2) All voltages are with respect to GND. Currents are positive into, negative out of the specified terminal.
- (3) Whichever is larger

### 7.2 ESD Ratings

		VALUE	UNIT
$V_{(ESD)}$ Electrostatic discharge	Human-body model (HBM), per ANSI/ESDA/JEDEC JS-001 <sup>(1)</sup>	$\pm 2000$	V
	Charged-device model (CDM), per JEDEC specification JESD22-C101 <sup>(2)</sup>	$\pm 1000$	

- (1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.
- (2) JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.

### 7.3 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)

	MIN	NOM	MAX	UNIT
Supply voltage, VDD	4.5		15	V

### 7.4 Thermal Information

THERMAL METRIC <sup>(1)</sup>		UCC27322	UCC27321		UNIT
		D (SOIC)	P (PDIP)	DGN (MSOP-PowerPAD)	
		8 PINS	8 PINS	8 PINS	
$R_{\theta JA}$	Junction-to-ambient thermal resistance	56.6	55.9	56.7	$^\circ\text{C}/\text{W}$
$R_{\theta JC(top)}$	Junction-to-case (top) thermal resistance	52.8	45.3	52.9	$^\circ\text{C}/\text{W}$
$R_{\theta JB}$	Junction-to-board thermal resistance	32.6	32.6	32.7	$^\circ\text{C}/\text{W}$
$\psi_{JT}$	Junction-to-top characterization parameter	1.8	23.0	1.8	$^\circ\text{C}/\text{W}$
$\psi_{JB}$	Junction-to-board characterization parameter	32.3	32.5	32.4	$^\circ\text{C}/\text{W}$
$R_{\theta JC(bot)}$	Junction-to-case (bottom) thermal resistance	5.9	—	5.9	$^\circ\text{C}/\text{W}$

- (1) For more information about traditional and new thermal metrics, see the *Semiconductor and IC Package Thermal Metrics* application report, [SPRA953](#).

## 7.5 Electrical Characteristics

$V_{DD} = 4.5\text{ V to }15\text{ V}$ ,  $T_A = -40^\circ\text{C to }+105^\circ\text{C}$  for UCC2732x,  $T_A = 0^\circ\text{C to }70^\circ\text{C}$  for UCC3732x,  $T_A = T_J$ , unless otherwise noted.

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
<b>INPUT (IN)</b>						
$V_{IN\_H}$ , logic 1 input threshold			1.6	2.2	2.5	V
$V_{IN\_L}$ , logic 0 input threshold			0.8	1.2	1.5	V
Input current		$0\text{ V} \leq V_{IN} \leq V_{DD}$	-10	0	10	$\mu\text{A}$
<b>OUTPUT (OUT)</b>						
Peak output current <sup>(1)</sup>		$V_{DD} = 14\text{ V}$ ,		9		A
Output resistance high <sup>(2)</sup>		$I_{OUT} = -10\text{ mA}$		0.6	1.5	$\Omega$
Output resistance low <sup>(2)</sup>		$I_{OUT} = 10\text{ mA}$		0.4	1	$\Omega$
<b>OVERALL</b>						
$I_{DD}$ , static operating current	UCC37321 UCC27321	IN = LOW, EN = LOW, $V_{DD} = 15\text{ V}$		150	225	$\mu\text{A}$
		IN = HIGH, EN = LOW, $V_{DD} = 15\text{ V}$		440	650	
		IN = LOW, EN = HIGH, $V_{DD} = 15\text{ V}$		370	550	
		IN = HIGH, EN = HIGH, $V_{DD} = 15\text{ V}$		370	550	
	UCC37322 UCC27322	IN = LOW, EN = LOW, $V_{DD} = 15\text{ V}$		150	225	
		IN = HIGH, EN = LOW, $V_{DD} = 15\text{ V}$		450	650	
		IN = LOW, EN = HIGH, $V_{DD} = 15\text{ V}$		75	125	
		IN = HIGH, EN = HIGH, $V_{DD} = 15\text{ V}$		675	1000	
<b>ENABLE (ENBL)</b>						
$V_{EN\_H}$ , high-level enable voltage		LOW to HIGH transition	1.7	2.2	2.7	V
$V_{EN\_L}$ , low-level enable voltage		HIGH to LOW transition	1.1	1.6	2	V
Hysteresis			0.25	0.55	0.90	
$R_{ENBL}$ , enable impedance		$V_{DD} = 14\text{ V}$ , ENBL = GND	75	100	135	k $\Omega$

(1) Not tested in production.

(2) Output pullup resistance here is a DC measurement that measures resistance of PMOS structure only, not N-channel structure.

## 7.6 Switching Characteristics

over operating free-air temperature range (unless otherwise noted)

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
<b>ENABLE (ENBL)</b>					
$t_{D3}$ , propagation delay time <sup>(1)</sup>	$C_{LOAD} = 10 \text{ nF}$		60	90	ns
$t_{D4}$ , propagation delay time <sup>(1)</sup>	$C_{LOAD} = 10 \text{ nF}$		60	90	ns
<b>SWITCHING TIME<sup>(2)</sup></b>					
$t_R$ , rise time (OUT)	$C_{LOAD} = 10 \text{ nF}$		20	70	ns
$t_F$ , fall time (OUT)	$C_{LOAD} = 10 \text{ nF}$		20	30	ns
$t_{D1}$ , propagation delay, IN rising (IN to OUT)	$C_{LOAD} = 10 \text{ nF}$		25	70	ns
$t_{D2}$ , propagation delay, IN falling (IN to OUT)	$C_{LOAD} = 10 \text{ nF}$		35	70	ns

(1) See [Figure 7-2](#).

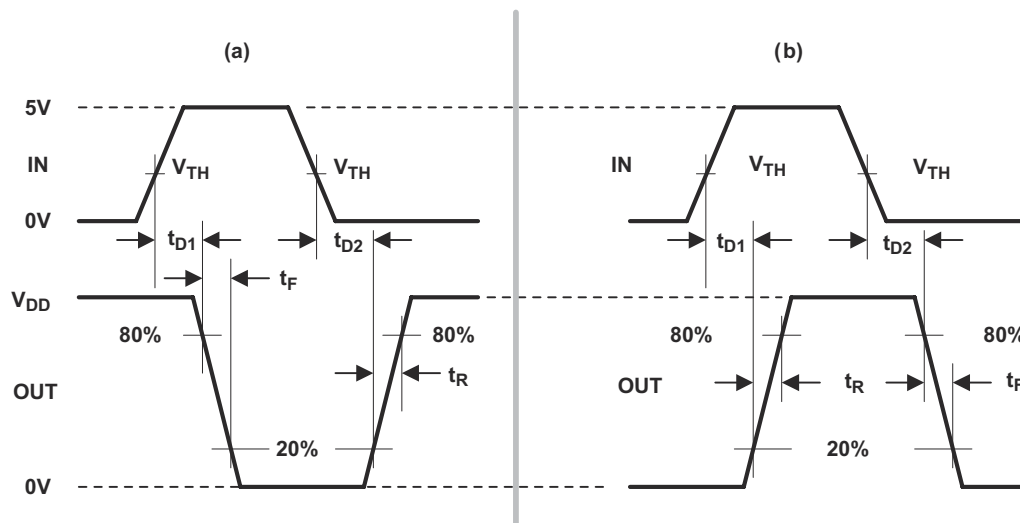
(2) See [Figure 7-1](#) for switching waveforms.

## 7.7 Power Dissipation Ratings

PACKAGE	SUFFIX	$\theta_{jc}$ (°C/W)	$\theta_{ja}$ (°C/W)	Power Rating (mW) $T_A = 70^\circ\text{C}$ <sup>(1)</sup>	Derating Factor Above $70^\circ\text{C}$ (mW/°C) <sup>(1)</sup>
SOIC-8	D	42	84 to 160 <sup>(2)</sup>	344 to 655 <sup>(2)</sup>	6.25 to 11.9 <sup>(2)</sup>
MSOP PowerPAD-8	DGN	4.7	50 to 59	1370	17.1

(1) 125°C operating junction temperature is used for power rating calculations

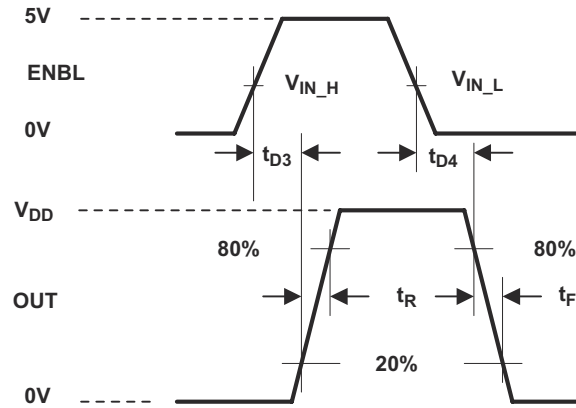
(2) The range of values indicates the effect of the printed-circuit-board. These values are intended to give the system designer an indication of the best and worst case conditions. In general, the system designer should attempt to use larger traces on the printed-circuit-board where possible to spread the heat away from the device more effectively. For additional information on device temperature management, see the *Packaging Information* section of the *Power Supply Control Products Data Book*, ([SLUD003](#)).



The 20% and 80% thresholds depict the dynamics of the BiPolar output devices that dominate the power MOSFET transition through the Miller regions of operation.

**Figure 7-1. Switching Waveforms for (a) Inverting Input to (b) Output Times**

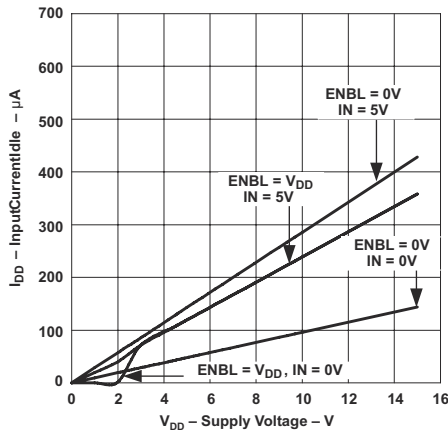




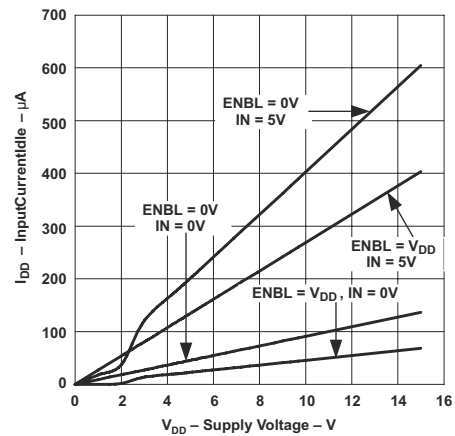
The 20% and 80% thresholds depict the dynamics of the BiPolar output devices that dominate the power MOSFET transition through the Miller regions of operation.

**☒ 7-2. Switching Waveform for Enable to Output**

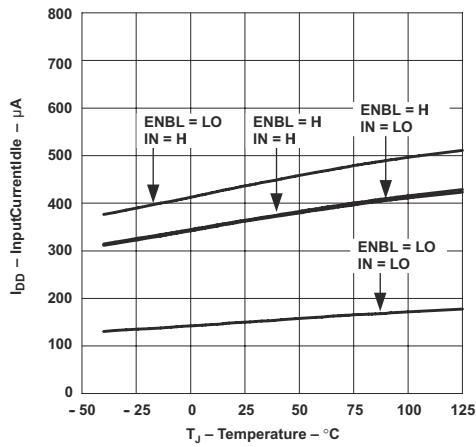
## 7.8 Typical Characteristics



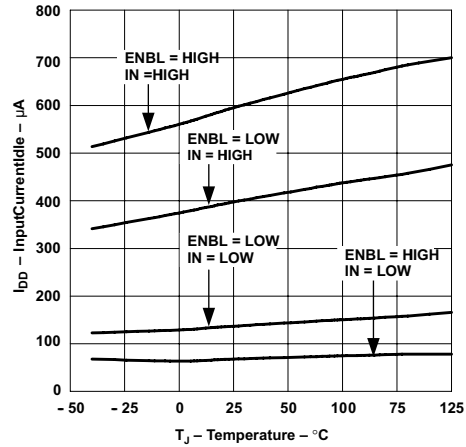
7-3. Input Current Idle vs Supply Voltage (UCCx7321)



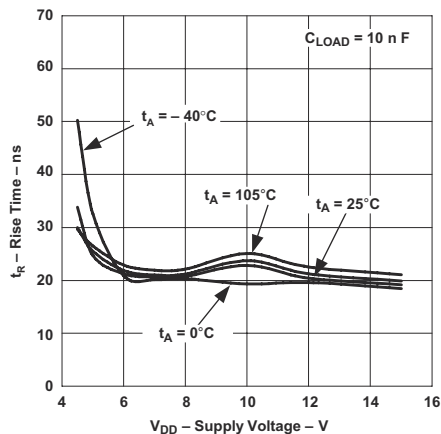
7-4. Input Current Idle vs Supply Voltage (UCCx7322)



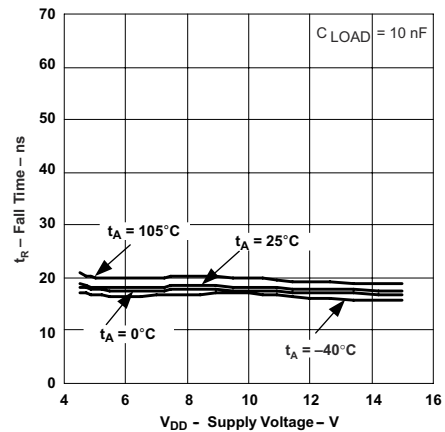
7-5. Input Current Idle vs Temperature (UCCx7321)



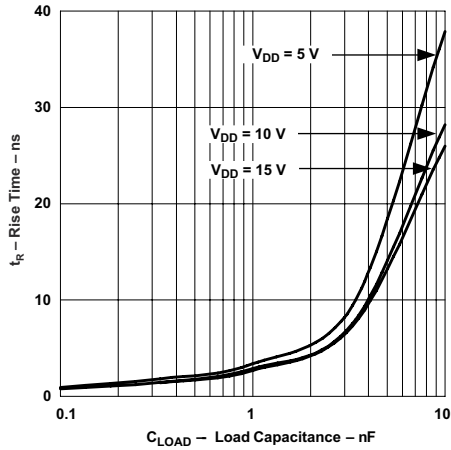
7-6. Input Current Idle vs Temperature (UCCx7322)



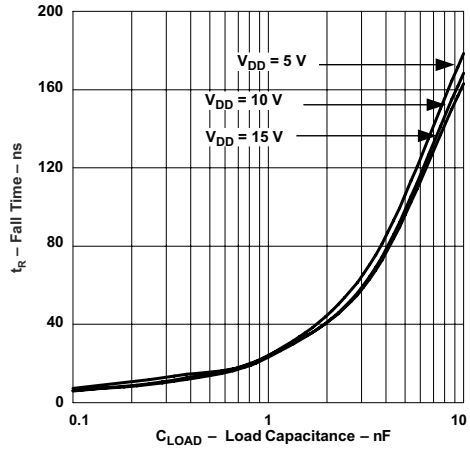
7-7. Rise Time vs Supply Voltage



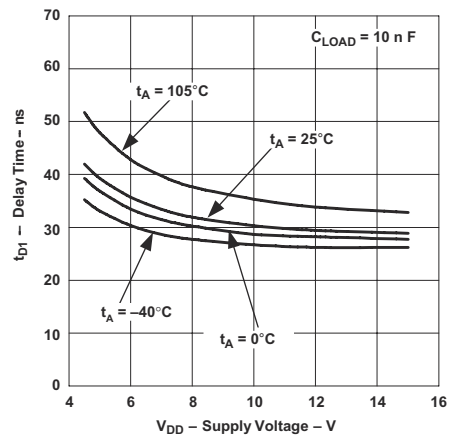
7-8. Fall Time vs Supply Voltage



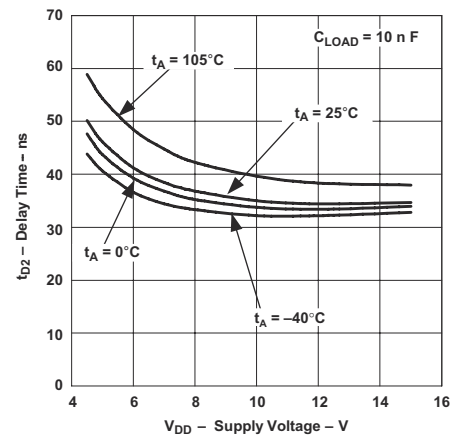
7-9. Rise Time vs Load Capacitance



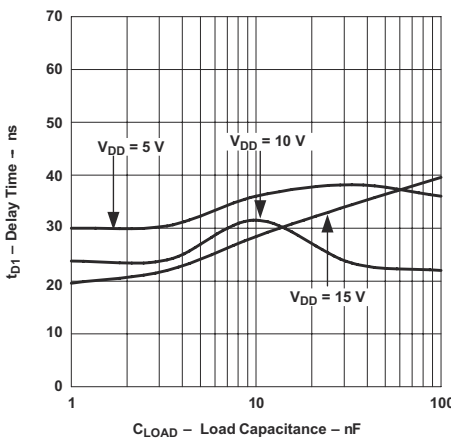
7-10. Fall Time vs Output Capacitance



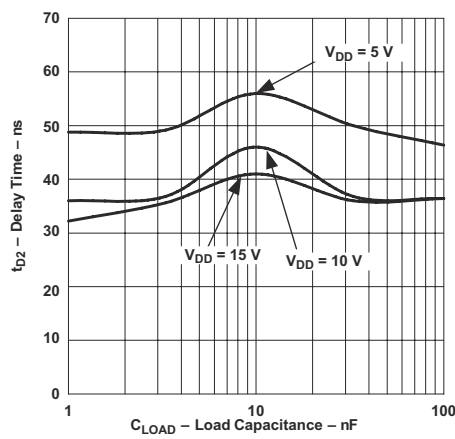
7-11.  $t_{D1}$  Delay Time vs Supply Voltage



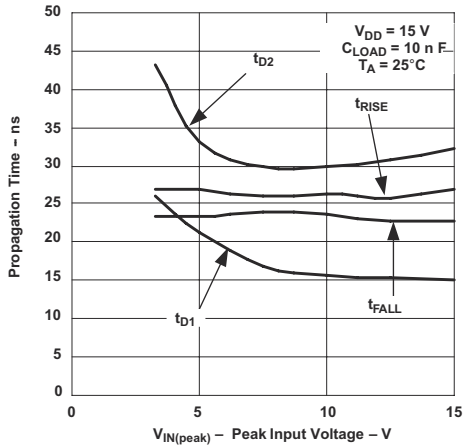
7-12.  $t_{D2}$  Delay Time vs Supply Voltage



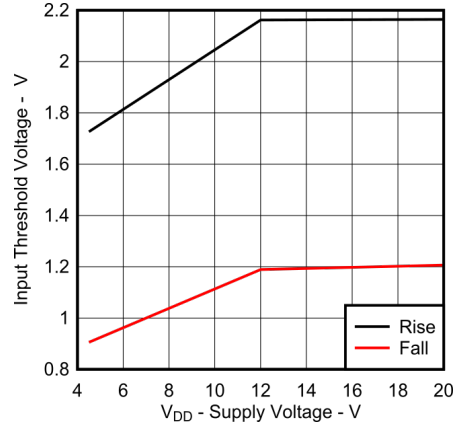
7-13.  $t_{D1}$  Delay Time vs Load Capacitance



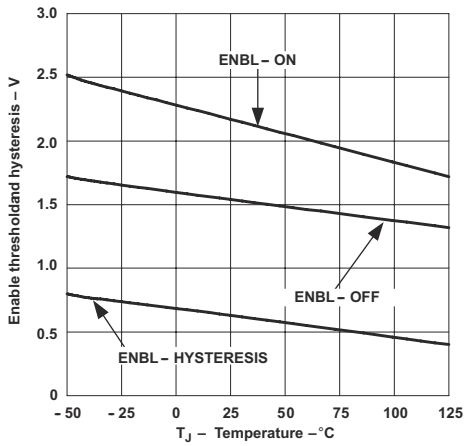
7-14.  $t_{D2}$  Delay Time vs Load Capacitance



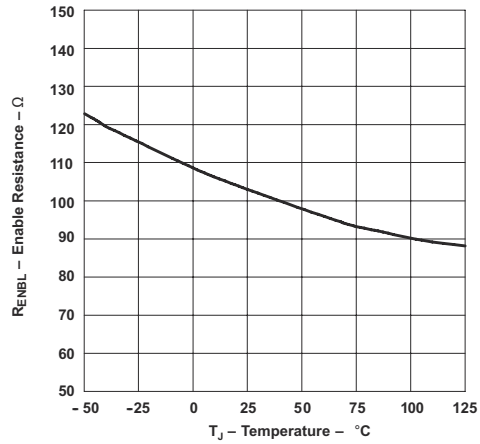
7-15. Propagation Times vs Peak Input Voltage



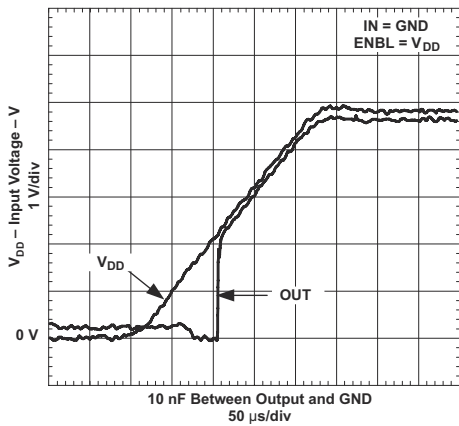
7-16. Input Threshold vs Supply Voltage



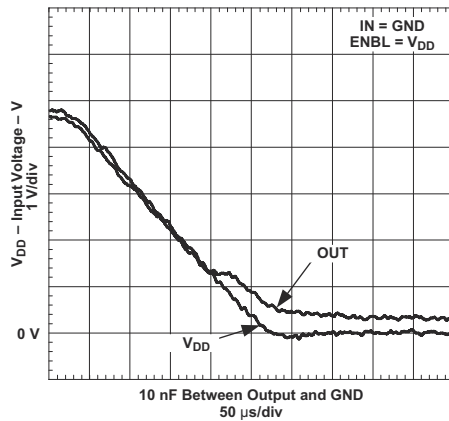
7-17. Enable Threshold and Hysteresis vs Temperature



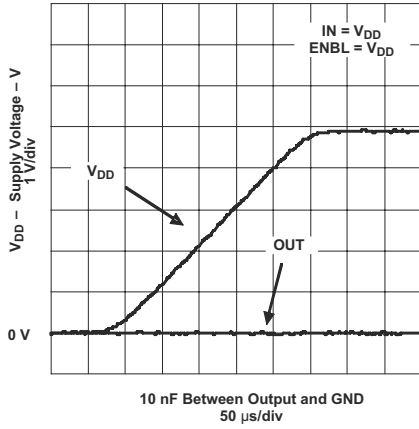
7-18. Enable Resistance vs Temperature



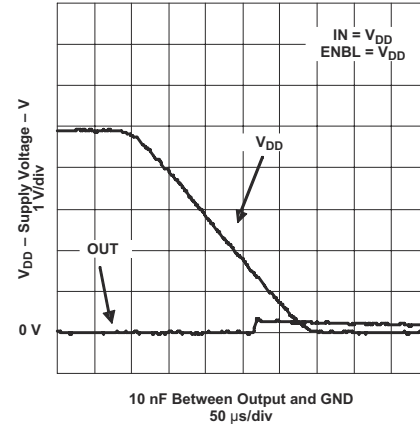
7-19. Output Behavior vs  $V_{DD}$  (UCC37321)



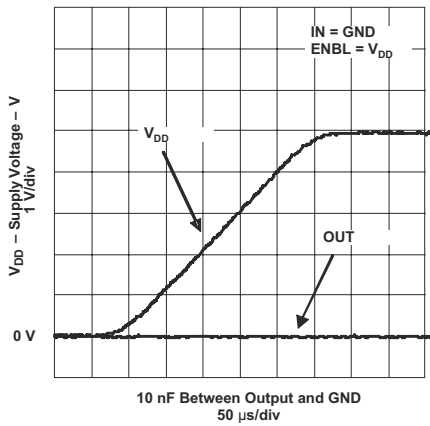
7-20. Output Behavior vs  $V_{DD}$  (UCC37321)



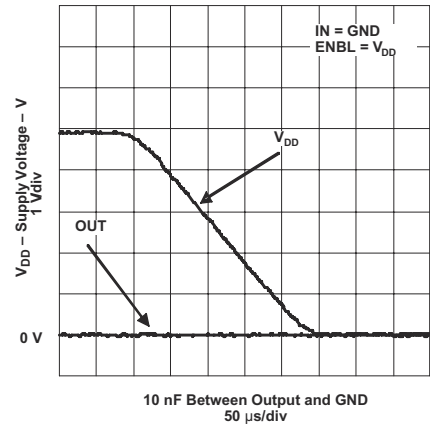
☒ 7-21. Output Behavior vs VDD (Inverting)



☒ 7-22. Output Behavior vs VDD (Inverting)



☒ 7-23. Output Behavior vs VDD (Noninverting)



☒ 7-24. Output Behavior vs VDD (Noninverting)

## 8 Detailed Description

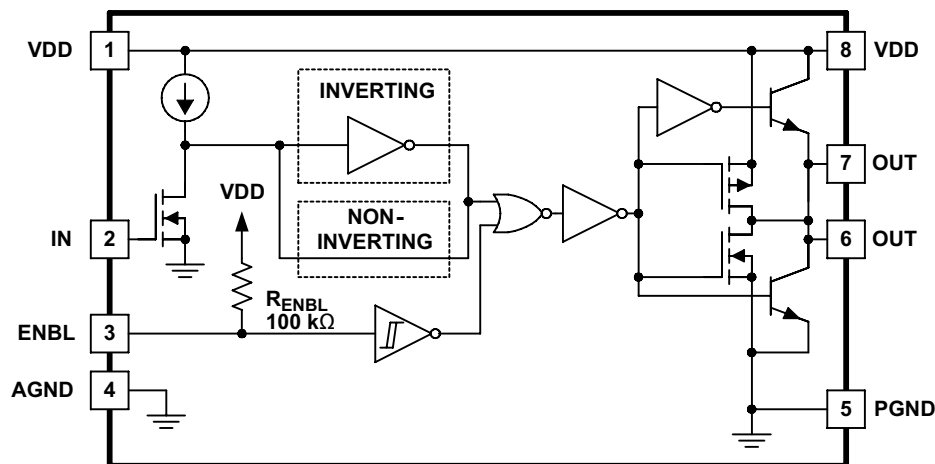
### 8.1 Overview

The UCC37321 and UCC37322 drivers serve as an interface between low-power controllers and power MOSFETs. They can also be used as an interface between DSPs and power MOSFETs. High-frequency power supplies often require high-speed, high-current drivers such as the UCC3732x family. A leading application is the need to provide a high power buffer stage between the PWM output of the control device and the gates of the primary power MOSFET or IGBT switching devices. In other cases, the device drives the power device gates through a drive transformer. Synchronous rectification supplies must simultaneously drive multiple devices which can present an extremely large load to the control circuitry.

The inverting driver (UCC37321) is useful for generating inverted gate drive signals from controllers that have only outputs of the opposite polarity. For example, this driver can provide a gate signal for ground referenced, N-channel synchronous rectifier MOSFETs in buck derived converters. This driver can also be used for generating a gate drive signal for a P-channel MOSFET from a controller that is designed for N-channel applications.

MOSFET gate drivers are generally used when it is not feasible to have the primary PWM regulator device directly drive the switching devices for one or more reasons. The PWM device may not have the brute drive capability required for the intended switching MOSFET, limiting the switching performance in the application. In other cases there may be a desire to minimize the effect of high-frequency switching noise by placing the high current driver physically close to the load. Also, newer devices that target the highest operating frequencies may not incorporate onboard gate drivers at all. Their PWM outputs are only intended to drive the high impedance input to a driver such as the UCC3732x. Finally, the control device may be under thermal stress due to power dissipation, and an external driver can help by moving the heat from the controller to an external package.

### 8.2 Functional Block Diagram



### 8.3 Feature Description

#### 8.3.1 Input Stage

The IN threshold has a 3.3-V logic sensitivity over the full range of VDD voltages; yet, it is equally compatible with 0 V to VDD signals. The inputs of UCC3732x family of drivers are designed to withstand 500-mA reverse current without either damage to the device or logic upset. In addition, the input threshold turnoff of the UCC3732x has been slightly raised for improved noise immunity. The input stage of each driver must be driven by a signal with a short rise or fall time. This condition is satisfied in typical power supply applications, where the input signals are provided by a PWM controller or logic gates with fast transition times (< 200 ns). The IN input of the driver functions as a digital gate, and it is not intended for applications where a slow changing input voltage is used to generate a switching output when the logic threshold of the input section is reached. While this may not be harmful to the driver, the output of the driver may switch repeatedly at a high frequency.

Users should not attempt to shape the input signals to the driver in an attempt to slow down (or delay) the signal at the output. If limiting the rise or fall times to the power device is desired, then an external resistance can be added between the output of the driver and the load device, which is generally a power MOSFET gate. The external resistor may also help remove power dissipation from the device package, as discussed in [セクション 11.3](#).

### 8.3.2 Output Stage

The TrueDrive output stage is capable of supplying  $\pm 9\text{-A}$  peak current pulses; it swings to both VDD and GND and can encourage even the most stubborn MOSFETs to switch. The pullup and pulldown circuits of the driver are constructed of bipolar and MOSFET transistors in parallel. The peak output current rating is the combined current from the bipolar and MOSFET transistors. The output resistance is the  $R_{DS(ON)}$  of the MOSFET transistor when the voltage on the driver output is less than the saturation voltage of the bipolar transistor. Each output stage also provides a very low impedance to overshoot and undershoot due to the body diode of the internal MOSFET. This means that in many cases, external-schottky-clamp diodes are not required.

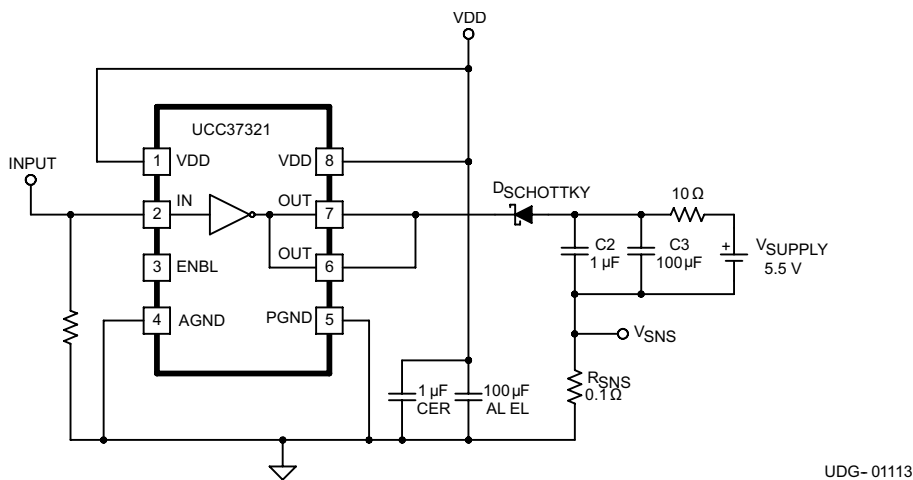
This unique BiPolar and MOSFET hybrid output architecture (TrueDrive) allows efficient current sourcing at low supply voltages. The UCC3732x family delivers 9 A of gate drive where it is most needed during the MOSFET switching transition – at the Miller plateau region – providing improved efficiency gains.

### 8.3.3 Source and Sink Capabilities during Miller Plateau

Large power MOSFETs present a significant load to the control circuitry. Proper drive is required for efficient, reliable operation. The UCC3732x drivers have been optimized to provide maximum drive to a power MOSFET during the Miller plateau region of the switching transition. This interval occurs while the drain voltage is swinging between the voltage levels dictated by the power topology, requiring the charging or discharging of the drain-gate capacitance with current supplied or removed by the driver device.

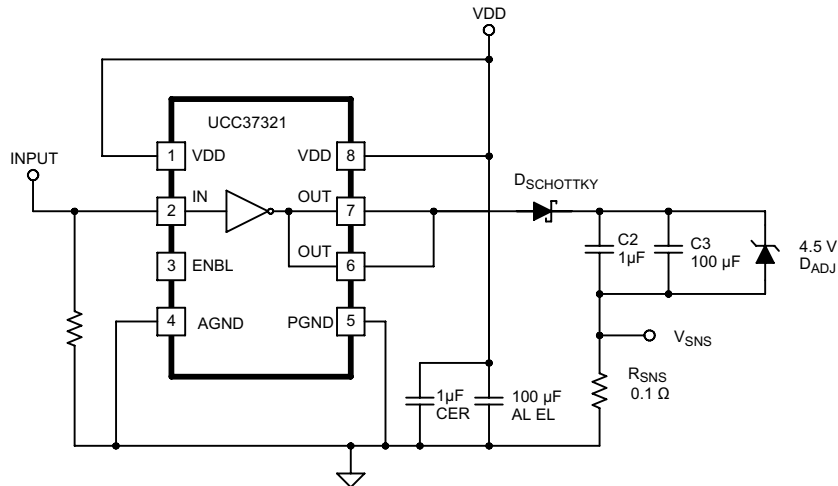
Two circuits are used to test the current capabilities of the UCC3732x driver (see Reference 1). In each case external circuitry is added to clamp the output near 5 V while the device is sinking or sourcing current. An input pulse of 250 ns is applied at a frequency of 1 kHz in the proper polarity for the respective test. In each test there is a transient period where the current peaked up and then settled down to a steady-state value. The noted current measurements are made at a time of 200 ns after the input pulse is applied, after the initial transient.

The circuit in [図 8-1](#) is used to verify the current sink capability when the output of the driver is clamped around 5 V, a typical value of gate-source voltage during the Miller plateau region. The UCC37321 is found to sink 9 A at  $V_{DD} = 15\text{ V}$ .



**図 8-1. Sink Current Test Circuit**

The circuit in [図 8-2](#) is used to test the current source capability with the output clamped to around 5 V with a string of Zener diodes. The UCC37321 is found to source 9 A at  $V_{DD} = 15\text{ V}$ .



UDG-01114

図 8-2. Source Current Test Circuit

Note that the current sink capability is slightly stronger than the current source capability at lower VDD. This is due to the differences in the structure of the bipolar-MOSFET power output section, where the current source is a P-channel MOSFET and the current sink has an N-channel MOSFET.

In most it is advantageous that the turnoff capability of a driver is stronger than the turnon capability. This helps to ensure that the MOSFET is held OFF during common power supply transients which may turn the device back ON.

### 8.3.4 Enable

The UCC37321/2 provides an enable input for improved control of the driver operation. This input also incorporates logic compatible thresholds with hysteresis. It is internally pulled up to VDD with 100-kΩ resistor for active high operation. When ENBL is high, the device is enabled and when ENBL is low, the device is disabled. The default state of the ENBL pin is to enable the device and therefore it can be left open for standard operation. The output state when the device is disabled is low regardless of the input state. See 表 8-1 for the operation using enable logic.

ENBL input is compatible with both logic signals and slow changing analog signals. It can be directly driven or a power-up delay can be programmed with a capacitor between ENBL and AGND.

### 8.4 Device Functional Modes

表 8-1 lists the logic of this device.

表 8-1. Device Logic Table

	ENBL	IN	OUT
<b>INVERTING UCC37321</b>	0	0	0
	0	1	0
	1	0	1
	1	1	0
<b>NON-- INVERTING UCC37322</b>	0	0	0
	0	1	0
	1	0	0
	1	1	1



## 9 Application and Implementation

### 注

以下のアプリケーション情報は、TI の製品仕様に含まれるものではなく、TI ではその正確性または完全性を保証いたしません。個々の目的に対する製品の適合性については、お客様の責任で判断していただくこととなります。お客様は自身の設計実装を検証しテストすることで、システムの機能を確認する必要があります。

### 9.1 Application Information

High-current gate driver devices are required in switching power applications for a variety of reasons. To enable fast switching of power devices and reduce associated power losses, a powerful gate driver can be employed between the PWM output of controllers or signal isolation devices and the gates of the power semiconductor devices. Further, gate drivers are indispensable when sometimes it is just not feasible to have the PWM controller directly drive the gates of the switching devices. The situation may be encountered because the PWM signal from a digital controller or signal isolation device is often a 3.3-V or 5-V logic signal which is not capable of effectively turning on a power switch. A level-shifting circuitry is needed to boost the logic-level signal to the gate-drive voltage to fully turn on the power device and minimize conduction losses. Traditional buffer drive circuits based on NPN/PNP bipolar, (or P- N- channel MOSFET), transistors in totem-pole arrangement, being emitter follower configurations, prove inadequate for this because they lack level-shifting capability and low-drive voltage protection. Gate drivers effectively combine both the level-shifting and buffer drive functions. Gate drivers may also minimize the effect of switching noise by locating the high-current driver physically close to the power switch, drive gate-driver transformers and control floating power device gates, reducing power dissipation and thermal stress in controllers by absorbing gate-charge power losses.

In summary gate drivers are extremely important components in switching power combining benefits of high-performance, low-cost, low component count, board-space reduction, and simplified system design.

### 9.2 Typical Application

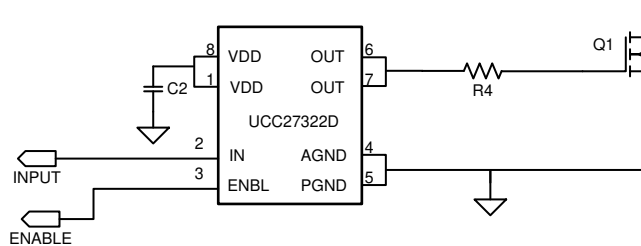


図 9-1. Typical Application Diagram of UCC27322 and UCC37322

#### 9.2.1 Design Requirements

When selecting the proper gate driver device for an end application, some design considerations must be evaluated first to make the most appropriate selection. The following design parameters should be used when selecting the proper gate driver device for an end application: input-to-output configuration, the input threshold type, bias supply voltage levels, peak source and sink currents, availability of independent enable and disable functions, propagation delay, power dissipation, and package type. See the example design parameters and requirements in 表 9-1.

表 9-1. Design Parameters

DESIGN PARAMETER	EXAMPLE VALUE
Input-to-output configuration	Noninverting
Input threshold type	CMOS
Bias supply voltage levels	12 V
dVDS/dt <sup>(1)</sup>	20 V/ns
Enable function	Yes
Propagation delay	< 50 ns
Power dissipation	< 0.45 W
Package type	SOIC (8)

(1) dVDS/dt is a typical requirement for a given design. This value can be used to find the peak source/sink currents needed as shown in [セクション 9.2.2.4](#).

## 9.2.2 Detailed Design Procedure

### 9.2.2.1 Input-to-Output Configuration

The design should specify which type of input-to-out configuration should be used. If turning on the power MOSFET or IGBT when the input signal is in high state is preferred, then a device capable of the noninverting configuration must be selected. If turning off the power MOSFET or IGBT when the input signal is in high state is preferred, then a device capable of the inverting configuration must be chosen. Based on this noninverting requirement of this application, the proper device out of the UCC27322 or UCC37322 should be selected.

### 9.2.2.2 Input Threshold Type

The type of input voltage threshold determines the type of controller that can be used with the gate driver device. The UCC2732x and UCC3732x devices feature a TTL and CMOS-compatible input threshold logic, with wide hysteresis. The threshold voltage levels are low voltage and independent of the  $V_{DD}$  supply voltage, which allows compatibility with both logic-level input signals from microcontrollers as well as higher-voltage input signals from analog controllers. See [セクション 7.5](#) for the actual input threshold voltage levels and hysteresis specifications for the UCC2732x and UCC3732x devices.

### 9.2.2.3 VDD Bias Supply Voltage

The bias supply voltage to be applied to the  $V_{DD}$  pins of the device must never exceed the values listed in [セクション 7.3](#). However, different power switches require different voltage levels to be applied at the gate. With a wide operating range from 4.5 V to 15 V, the UCC2732x and UCC3732x can be used to drive a variety of power switches, such as Si MOSFETs (for example,  $V_{gs} = 4.5$  V, 10 V, 12 V), IGBTs ( $V_{GE}=15$  V), and wide-bandgap power semiconductors (such as GaN, certain types of which allow no higher than 6 V to be applied to the gate terminals).

### 9.2.2.4 Peak Source and Sink Currents

Generally, the switching speed of the power switch during turnon and turnoff must be as fast as possible to minimize switching power losses. The gate driver device must be able to provide the required peak current for achieving the targeted switching speeds for the targeted power MOSFET.

Using the example of a power MOSFET, the system requirement for the switching speed is typically described in terms of the slew rate of the drain-to-source voltage of the power MOSFET (such as  $dV_{DS}/dt$ ). For example, the system requirement might state that a SPP20N60C3 power MOSFET must be turned on with a  $Dv_{ds}/dt$  of 20 V/ns or higher under a DC bus voltage of 400 V in a continuous-conduction-mode (CCM) boost PFC-converter application. This type of application is an inductive hard-switching application and reducing switching power loss is critical. This requirement means that the entire drain-to-source voltage swing during power MOSFET turnon event (from 400 V in the OFF state to  $V_{DS(on)}$  in ON state) must be completed in approximately 20 ns or less. When the drain-to-source voltage swing occurs, the Miller charge of the power MOSFET ( $Q_{gd}$  parameter in SPP20N60C3 power MOSFET data sheet is 33 nC typically) is supplied by the peak current of gate driver.

According to power MOSFET inductive switching mechanism, the gate-to-source voltage of the power MOSFET at this time is the Miller plateau voltage, which is typically a few volts higher than the threshold voltage of the power MOSFET,  $V_{GS(th)}$ .

To achieve the targeted  $Dv_{ds}/dt$ , the gate driver must be capable of providing the  $Q_{gd}$  charge in 20 ns or less. In other words, a peak current of 1.65 A (= 33 nC) / 20 ns or higher must be provided by the gate driver. The UCC2732x and UCC3732x devices can provide 9-A peak sourcing/sinking current which clearly exceeds the design requirement and has the capability to meet the switching speed needed. This 9-A peak sourcing/sinking current provides an extra margin against part-to-part variations in the  $Q_{gd}$  parameter of the power MOSFET along with additional flexibility to insert external gate resistors and fine tune the switching speed for efficiency versus EMI optimizations. However, in practical designs the parasitic trace in the gate driver circuit of the PCB will have a definitive role to play on the power MOSFET switching speed. The effort of this trace inductance is to limit the  $di/dt$  of the output current pulse of the gate driver. To illustrate this effect, consider output current pulse waveform from the gate driver to be approximated to a triangular profile, where the area under the triangle ( $0.5 \times I_{PEAK} \times \text{time}$ ) would equal the total gate charge of the power MOSFET ( $Q_g$  parameter in SPP20N60C3 power MOSFET data sheet= 87 nC typically). If the parasitic trace inductance limits the  $di/dt$  then a situation may occur in which the full peak current capability of the gate driver is not fully achieved in the time required to deliver the  $Q_g$  required for the power MOSFET switching. In other words, the time parameter in the equation would dominate and the  $I_{PEAK}$  value of the current pulse would be much less than the true peak current capability of the device, while the required  $Q_g$  is still delivered. Because of this, the desired switching speed may not be realized, even when theoretical calculations indicate the gate driver can achieve the targeted switching speed. Thus, placing the gate driver device very close to the power MOSFET and designing a tight gate drive-loop with minimal PCB trace inductance is important to realize the full peak-current capability of the gate driver.

### 9.2.2.5 Enable and Disable Function

Certain applications demand independent control of the output state of the driver without involving the input signal. A pin which offers enable and disable functions achieves the requirements. For these applications, the UCC2732x and UCC3732x are suitable as they feature an input pin and an Enable pin.

### 9.2.2.6 Propagation Delay

The acceptable propagation delay from the gate driver is dependent on the switching frequency at which it is used and the acceptable level of pulse distortion to the system. The UCC2732x and UCC3732x devices feature 25-ns turnon propagation delay and 35-ns turnoff propagation delay (typical), which ensure very little distortion and allow operation at higher frequencies. See [セクション 7.5](#) for the propagation and [セクション 7.6](#) of the UCC2732x and UCC3732x devices.

### 9.2.2.7 Power Dissipation

The UCC3732x family of drivers are capable of delivering 9-A of current to a MOSFET gate for a period of several hundred nanoseconds. High peak current is required to turn an N-channel device ON quickly. Then, to turn the device OFF, the driver is required to sink a similar amount of current to ground. This repeats at the operating frequency of the power device. An N-channel MOSFET is used in this discussion because it is the most common type of switching device used in high-frequency power conversion equipment.

References [1](#) and [2](#) contain detailed discussions of the drive current required to drive a power MOSFET and other capacitive-input switching devices. Much information is provided in tabular form to give a range of the current required for various devices at various frequencies. The information pertinent to calculating gate drive current requirements will be summarized here; the original document is available from the TI website.

When a driver device is tested with a discrete, capacitive load it is a fairly simple matter to calculate the power that is required from the bias supply. The energy that must be transferred from the bias supply to charge the capacitor is given by [式 1](#).

$$E = \frac{1}{2} CV^2 \tag{1}$$

where

- C is the load capacitor
- V is the bias voltage feeding the driver

There is an equal amount of energy transferred to ground when the capacitor is discharged. This leads to a power loss given by 式 2.

$$P = 2 \times \frac{1}{2} CV^2 f \quad (2)$$

where

- f is the switching frequency

This power is dissipated in the resistive elements of the circuit. Thus, with no external resistor between the driver and gate, this power is dissipated inside the driver. Half of the total power is dissipated when the capacitor is charged, and the other half is dissipated when the capacitor is discharged. An example using the conditions of the previous gate-drive waveform should help clarify this.

With  $V_{DD} = 12\text{ V}$ ,  $C_{LOAD} = 10\text{ nF}$ , and  $f = 300\text{ kHz}$ , the power loss can be calculated as shown in 式 4.

$$P = 10\text{ nF} \times (12)^2 \times (300\text{ kHz}) = 0.432\text{ W} \quad (3)$$

With a 12-V supply, this would equate, as shown in 式 4, to a current of:

$$I = \frac{P}{V} = \frac{0.432\text{ W}}{12\text{ V}} = 0.036\text{ A} \quad (4)$$

The switching load presented by a power MOSFET can be converted to an equivalent capacitance by examining the gate charge required to switch the device. This gate charge includes the effects of the input capacitance plus the added charge needed to swing the drain of the device between the ON and OFF states. Most manufacturers provide specifications that provide the typical and maximum gate charge, in nC, to switch the device under specified conditions. Using the gate charge  $Q_g$ , one can determine the power that must be dissipated when charging a capacitor. This is done by using the equivalence  $Q_g = C_{eff}V$  to provide 式 5 for power.

$$P = C \times V^2 \times f = Q_g \times V \times f \quad (5)$$

式 5 allows a power designer to calculate the bias power required to drive a specific MOSFET gate at a specific bias voltage.

### 9.2.3 Application Curves

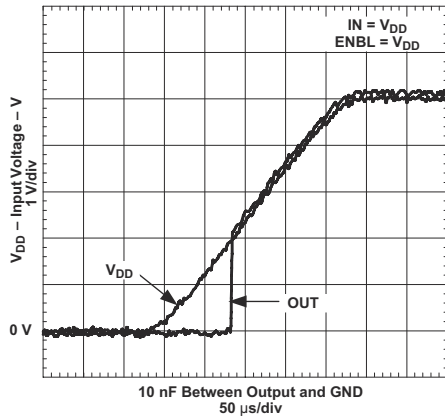


図 9-2. Output Behavior vs VDD (UCC37322)

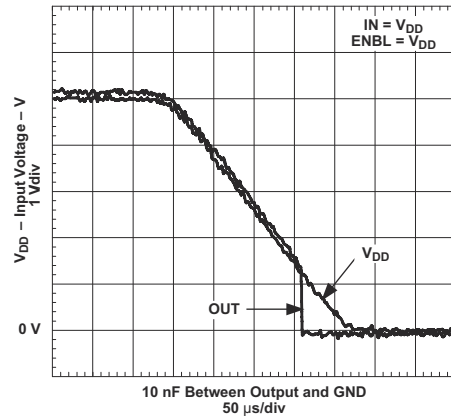


図 9-3. Output Behavior vs VDD (UCC37322)

## 10 Power Supply Recommendations

Although quiescent VDD current is very low, total supply current is higher, depending on OUTA and OUTB current and the operating frequency. Total VDD current is the sum of quiescent VDD current and the average OUT current. Knowing the operating frequency and the MOSFET gate charge (Qg), average OUT current can be calculated using 式 6.

$$I_{OUT} = Qg \times f \quad (6)$$

where

- f is frequency

For the best high-speed circuit performance, TI recommends two V<sub>DD</sub> bypass capacitors to prevent noise problems. TI also highly recommends using surface mount components. A 0.1-μF ceramic capacitor must be placed closest to the VDD to ground connection. In addition, a larger capacitor (such as 1 μF) with relatively low ESR should be connected in parallel to help deliver the high current peaks to the load. The parallel combination of capacitors presents a low impedance characteristic for the expected current levels in the driver application.

## 11 Layout

### 11.1 Layout Guidelines

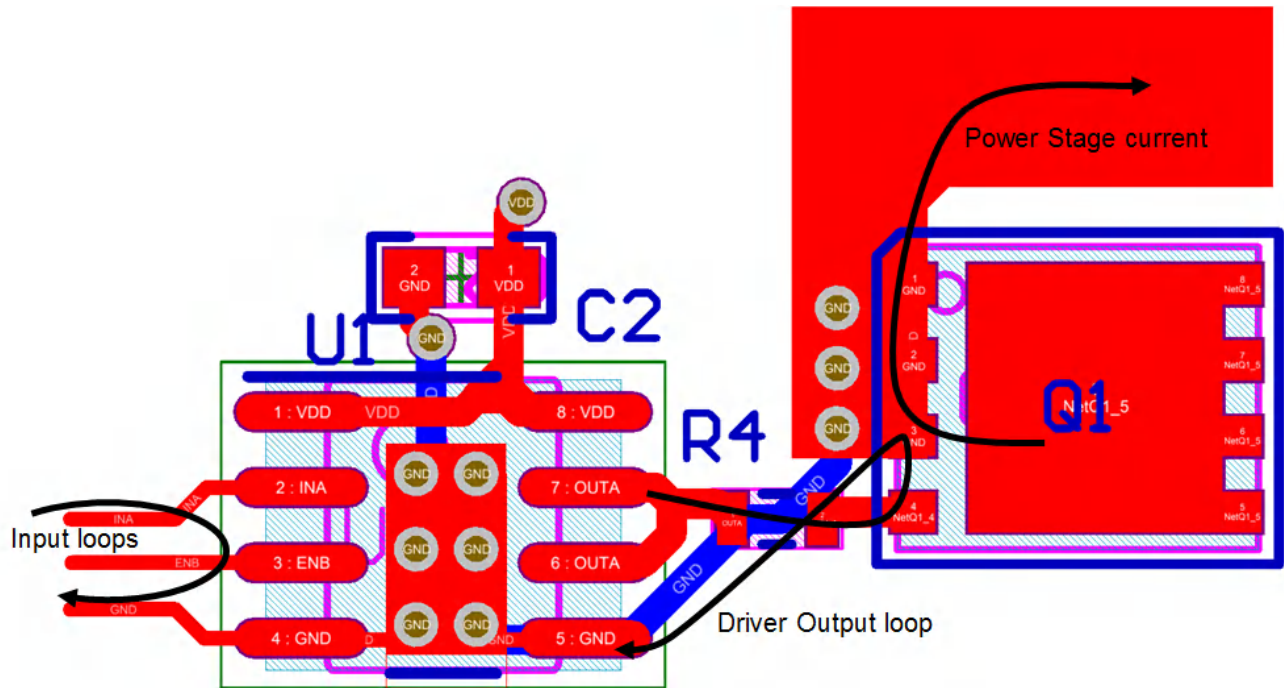
It can be a significant challenge to avoid the overshoot, undershoot, and ringing issues that can arise from circuit layout. The low impedance of these drivers and their high di/dt can induce ringing between parasitic inductances and capacitances in the circuit. Utmost care must be used in the circuit layout.

In general, position the driver physically as close to its load as possible. Place a 1-μF bypass capacitor as close to the output side of the driver as possible, connecting it to pins 1 and 8. Connect a single trace between the two VDD pins (pin 1 and pin 8); connect a single trace between PGND and AGND (pin 5 and pin 4). If a ground plane is used, it may be connected to AGND; do not extend the plane beneath the output side of the package (pins 5 – 8). Connect the load to both OUT pins (pins 7 and 6) with a single trace on the adjacent layer to the component layer; route the return current path for the output on the component side, directly over the output path.

Extreme conditions may require decoupling the input power and ground connections from the output power and ground connections. The UCCx732x has a feature that allows the user to take these extreme measures, if necessary. There is a small amount of internal impedance of about 15 Ω between the AGND and PGND pins; there is also a small amount of impedance (approximately 30 Ω) between the two VDD pins. To take advantage of this feature, connect a 1-μF bypass capacitor between VDD and PGND (pins 5 and 8) and connect a 0.1-μF bypass capacitor between VDD and AGND (pins 1 and 4). Further decoupling can be achieved by connecting between the two VDD pins with a jumper that passes through a 40-MHz ferrite bead and connect bias power only to pin 8. Even more decoupling can be achieved by connecting between AGND and PGND with a pair of anti-parallel diodes (anode connected to cathode and cathode connected to anode).



## 11.2 Layout Example



11-1. Layout Recommendation

## 11.3 Thermal Considerations

The useful range of a driver is greatly affected by the drive power requirements of the load and the thermal characteristics of the device package. For a power driver to be useful over a particular temperature range the package must allow for the efficient removal of the heat produced while keeping the junction temperature within rated limits. The UCC3732x family of drivers is available in three different packages to cover a range of application requirements.

As shown in the power dissipation rating table, the 8-pin SOIC (D) and 8-pin PDIP (P) packages each have a power rating of around 0.5 W with  $T_A = 70^\circ\text{C}$ . This limit is imposed in conjunction with the power derating factor also given in the table. The power dissipation in our earlier example is 0.432 W with a 10-nF load, 12 VDD, switched at 300 kHz. Thus, only one load of this size could be driven using the D or P packages. The difficulties with heat removal limit the drive available in the D or P packages.

The 8-pin MSOP PowerPAD (DGN) package significantly relieves this concern by offering an effective means of removing the heat from the semiconductor junction. As illustrated in Reference 3, the PowerPAD packages offer a leadframe die pad that is exposed at the base of the package. This pad is soldered to the copper on the PC board directly underneath the device package, reducing the  $\theta_{jc}$  down to  $4.7^\circ\text{C/W}$ . Data is presented in Reference 3 to show that the power dissipation can be quadrupled in the PowerPAD configuration when compared to the standard packages. The PC board must be designed with thermal lands and thermal vias to complete the heat removal subsystem, as summarized in Reference 4. This allows a significant improvement in heatsinking over that available in the D or P packages, and is shown to more than double the power capability of the D and P packages.

The PowerPAD is not directly connected to any leads of the package. However, it is electrically and thermally connected to the substrate which is the ground of the device.

## 12 Device and Documentation Support

### 12.1 Device Support

#### 12.1.1 サード・パーティ製品に関する免責事項

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### 12.2 Documentation Support

#### 12.2.1 Related Documentation

For related documentation see the following:

1. SEM-1400, Topic 2, *A Design and Application Guide for High Speed Power MOSFET Gate Drive Circuits*
2. U-137, *Practical Considerations in High Performance MOSFET, IGBT and MCT Gate Drive Circuits*, by Bill Andreyckak ([SLUA105](#))
3. Technical Brief, *PowerPad Thermally Enhanced Package* ([SLMA002](#))
4. Application Brief, *PowerPAD Made Easy* ([SLMA004](#))
5. Data Book, *Power Supply Control Products*, ([SLUD003](#))

### 12.3 ドキュメントの更新通知を受け取る方法

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### 12.7 用語集


#### テキサス・インスツルメンツ用語集

この用語集には、用語や略語の一覧および定義が記載されています。



## 13 Revision History

資料番号末尾の英字は改訂を表しています。その改訂履歴は英語版に準じています。

<b>Changes from Revision H (January 2016) to Revision I (November 2023)</b>	<b>Page</b>
• 「製品情報」表から P パッケージを削除.....	1
• 「概要 (続き)」セクションから P パッケージを削除.....	3
• Deleted P package from Pin Configuration and Functions section.....	5
• Changed ESD Ratings from $\pm 2500$ V and $\pm 1500$ V to $\pm 2000$ V and $\pm 1000$ V.....	6
• Changed input threshold voltage values, deleted $V_{OH}$ output high level and $V_{OL}$ output low level, changed output resistance high and output resistance low values and deleted Latch-up protection from Electrical Characteristics.....	7
• Deleted P package data from Power Dissipations Ratings section.....	8
• Changed  7-16 .....	10

<b>Changes from Revision G (May 2013) to Revision H (January 2016)</b>	<b>Page</b>
• 「ESD 定格」表、「機能説明」セクション、「デバイスの機能モード」セクション、「アプリケーションと実装」セクション、「電源に関する推奨事項」セクション、「レイアウト」セクション、「デバイスおよびドキュメントのサポート」セクション、「メカニカル、パッケージ、および注文情報」セクションを追加。 .....	1

<b>Changes from Revision F (March 2012) to Revision G (May 2013)</b>	<b>Page</b>
• Updated AGND pin description. ....	5
• Changed minimum value for input voltage from $-5$ to $-0.3$ V in the <i>Absolute Maximum Ratings</i> table.....	6
• Added $C_{LOAD} = 10$ nF to Fall Time vs Supply Voltage graph.....	10
• Changed Changed x-axis values from 1, 10, 100 to 0.1, 1, 10 in Rise Time vs Load Capacitance graph.....	10
• Changed Changed x-axis values from 1, 10, 100 to 0.1, 1, 10 in Fall Time vs Output Capacitance graph.....	10

## 14 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

**PACKAGING INFORMATION**

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead finish/ Ball material (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
UCC27321DGNR	ACTIVE	HVSSOP	DGN	8	2500	RoHS & Green	NIPDAU   NIPDAUAG	Level-1-260C-UNLIM	-40 to 105	27321	<a href="#">Samples</a>
UCC27321DR	ACTIVE	SOIC	D	8	2500	RoHS & Green	Call TI   NIPDAU	Level-1-260C-UNLIM	-40 to 105	27321	<a href="#">Samples</a>
UCC27321P	ACTIVE	PDIP	P	8	50	RoHS & Green	NIPDAU	N / A for Pkg Type	-40 to 105	UCC27321P	<a href="#">Samples</a>
UCC27321PE4	ACTIVE	PDIP	P	8	50	RoHS & Green	NIPDAU	N / A for Pkg Type	-40 to 105	UCC27321P	<a href="#">Samples</a>
UCC27322DGNR	ACTIVE	HVSSOP	DGN	8	2500	RoHS & Green	NIPDAU   NIPDAUAG	Level-1-260C-UNLIM	-40 to 105	27322	<a href="#">Samples</a>
UCC27322DR	ACTIVE	SOIC	D	8	2500	RoHS & Green	Call TI   NIPDAU	Level-1-260C-UNLIM	-40 to 105	27322	<a href="#">Samples</a>
UCC27322DRG4	ACTIVE	SOIC	D	8	2500	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 105	27322	<a href="#">Samples</a>
UCC27322P	ACTIVE	PDIP	P	8	50	RoHS & Green	NIPDAU	N / A for Pkg Type	-40 to 105	UCC27322P	<a href="#">Samples</a>
UCC27322PE4	ACTIVE	PDIP	P	8	50	RoHS & Green	NIPDAU	N / A for Pkg Type	-40 to 105	UCC27322P	<a href="#">Samples</a>
UCC37321DGNR	ACTIVE	HVSSOP	DGN	8	2500	RoHS & Green	NIPDAU   NIPDAUAG	Level-1-260C-UNLIM	0 to 70	37321	<a href="#">Samples</a>
UCC37321DR	ACTIVE	SOIC	D	8	2500	RoHS & Green	Call TI   NIPDAU	Level-1-260C-UNLIM	0 to 70	37321	<a href="#">Samples</a>
UCC37321P	ACTIVE	PDIP	P	8	50	RoHS & Green	NIPDAU	N / A for Pkg Type	0 to 70	UCC37321P	<a href="#">Samples</a>
UCC37322DGNR	ACTIVE	HVSSOP	DGN	8	2500	RoHS & Green	NIPDAU   NIPDAUAG	Level-1-260C-UNLIM	0 to 70	37322	<a href="#">Samples</a>
UCC37322DGNRG4	ACTIVE	HVSSOP	DGN	8	2500	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	0 to 70	37322	<a href="#">Samples</a>
UCC37322DR	ACTIVE	SOIC	D	8	2500	RoHS & Green	Call TI   NIPDAU	Level-1-260C-UNLIM	0 to 70	37322	<a href="#">Samples</a>
UCC37322P	ACTIVE	PDIP	P	8	50	RoHS & Green	NIPDAU	N / A for Pkg Type	0 to 70	UCC37322P	<a href="#">Samples</a>

(1) The marketing status values are defined as follows:

**ACTIVE:** Product device recommended for new designs.

**LIFEBUY:** TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

**NRND:** Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

**PREVIEW:** Device has been announced but is not in production. Samples may or may not be available.

**OBSELETE:** TI has discontinued the production of the device.

(2) **RoHS:** TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

**RoHS Exempt:** TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

**Green:** TI defines "Green" to mean the content of Chlorine (Cl) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of  $\leq 1000$ ppm threshold. Antimony trioxide based flame retardants must also meet the  $\leq 1000$ ppm threshold requirement.

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

(6) Lead finish/Ball material - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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**OTHER QUALIFIED VERSIONS OF UCC27321, UCC27322 :**

● Automotive : [UCC27321-Q1](#), [UCC27322-Q1](#)

● Enhanced Product : [UCC27322-EP](#)

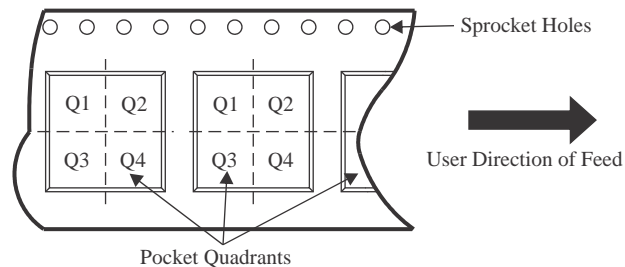
NOTE: Qualified Version Definitions:

- Automotive - Q100 devices qualified for high-reliability automotive applications targeting zero defects
- Enhanced Product - Supports Defense, Aerospace and Medical Applications

## TAPE AND REEL INFORMATION



### QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
UCC27321DGNR	HVSSOP	DGN	8	2500	330.0	12.4	5.3	3.4	1.4	8.0	12.0	Q1
UCC27321DGNR	HVSSOP	DGN	8	2500	330.0	12.4	5.3	3.4	1.4	8.0	12.0	Q1
UCC27321DR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
UCC27321DR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
UCC27322DGNR	HVSSOP	DGN	8	2500	330.0	12.4	5.3	3.4	1.4	8.0	12.0	Q1
UCC27322DR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
UCC27322DR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
UCC37321DGNR	HVSSOP	DGN	8	2500	330.0	12.4	5.3	3.4	1.4	8.0	12.0	Q1
UCC37321DGNR	HVSSOP	DGN	8	2500	330.0	12.4	5.3	3.4	1.4	8.0	12.0	Q1
UCC37321DR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
UCC37321DR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
UCC37322DGNR	HVSSOP	DGN	8	2500	330.0	12.4	5.3	3.4	1.4	8.0	12.0	Q1
UCC37322DGNR	HVSSOP	DGN	8	2500	330.0	12.4	5.3	3.4	1.4	8.0	12.0	Q1
UCC37322DR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
UCC37322DR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1

**TAPE AND REEL BOX DIMENSIONS**


\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
UCC27321DGNR	HVSSOP	DGN	8	2500	364.0	364.0	27.0
UCC27321DGNR	HVSSOP	DGN	8	2500	353.0	353.0	32.0
UCC27321DR	SOIC	D	8	2500	353.0	353.0	32.0
UCC27321DR	SOIC	D	8	2500	356.0	356.0	35.0
UCC27322DGNR	HVSSOP	DGN	8	2500	364.0	364.0	27.0
UCC27322DR	SOIC	D	8	2500	356.0	356.0	35.0
UCC27322DR	SOIC	D	8	2500	353.0	353.0	32.0
UCC37321DGNR	HVSSOP	DGN	8	2500	364.0	364.0	27.0
UCC37321DGNR	HVSSOP	DGN	8	2500	353.0	353.0	32.0
UCC37321DR	SOIC	D	8	2500	353.0	353.0	32.0
UCC37321DR	SOIC	D	8	2500	356.0	356.0	35.0
UCC37322DGNR	HVSSOP	DGN	8	2500	353.0	353.0	32.0
UCC37322DGNR	HVSSOP	DGN	8	2500	364.0	364.0	27.0
UCC37322DR	SOIC	D	8	2500	356.0	356.0	35.0
UCC37322DR	SOIC	D	8	2500	353.0	353.0	32.0

**TUBE**


\*All dimensions are nominal

Device	Package Name	Package Type	Pins	SPQ	L (mm)	W (mm)	T (μm)	B (mm)
UCC27321P	P	PDIP	8	50	506	13.97	11230	4.32
UCC27321PE4	P	PDIP	8	50	506	13.97	11230	4.32
UCC27322P	P	PDIP	8	50	506	13.97	11230	4.32
UCC27322PE4	P	PDIP	8	50	506	13.97	11230	4.32
UCC37321P	P	PDIP	8	50	506	13.97	11230	4.32
UCC37322P	P	PDIP	8	50	506	13.97	11230	4.32

## GENERIC PACKAGE VIEW

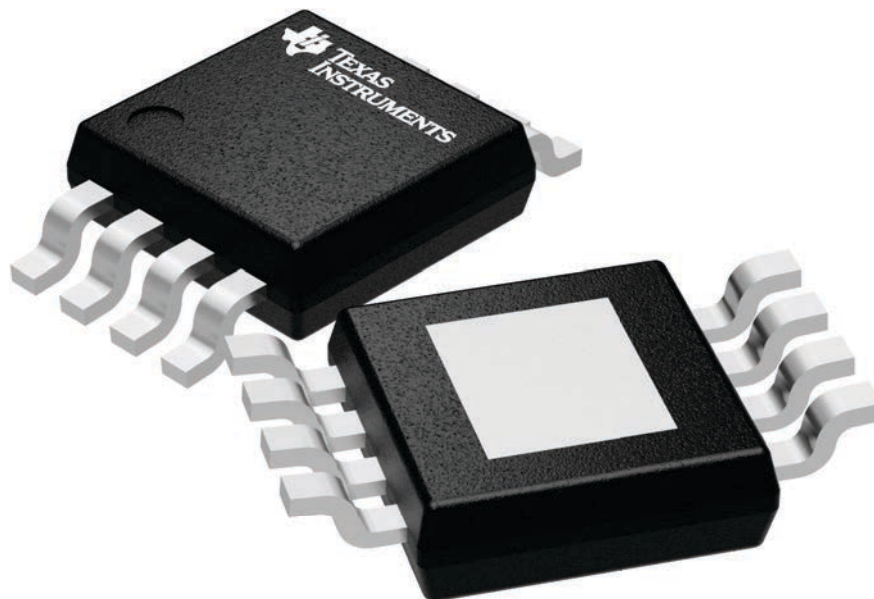
**DGN 8**

**PowerPAD VSSOP - 1.1 mm max height**

3 x 3, 0.65 mm pitch

SMALL OUTLINE PACKAGE

This image is a representation of the package family, actual package may vary.  
Refer to the product data sheet for package details.



4225482/A



D0008A

# PACKAGE OUTLINE

SOIC - 1.75 mm max height

SMALL OUTLINE INTEGRATED CIRCUIT



4214825/C 02/2019

NOTES:

1. Linear dimensions are in inches [millimeters]. Dimensions in parenthesis are for reference only. Controlling dimensions are in inches. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed  $.006$  [0.15] per side.
4. This dimension does not include interlead flash.
5. Reference JEDEC registration MS-012, variation AA.



# EXAMPLE BOARD LAYOUT

D0008A

SOIC - 1.75 mm max height

SMALL OUTLINE INTEGRATED CIRCUIT



LAND PATTERN EXAMPLE  
EXPOSED METAL SHOWN  
SCALE:8X



SOLDER MASK DETAILS

4214825/C 02/2019

NOTES: (continued)

- 6. Publication IPC-7351 may have alternate designs.
- 7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.

# EXAMPLE STENCIL DESIGN

D0008A

SOIC - 1.75 mm max height

SMALL OUTLINE INTEGRATED CIRCUIT



SOLDER PASTE EXAMPLE  
BASED ON .005 INCH [0.125 MM] THICK STENCIL  
SCALE:8X

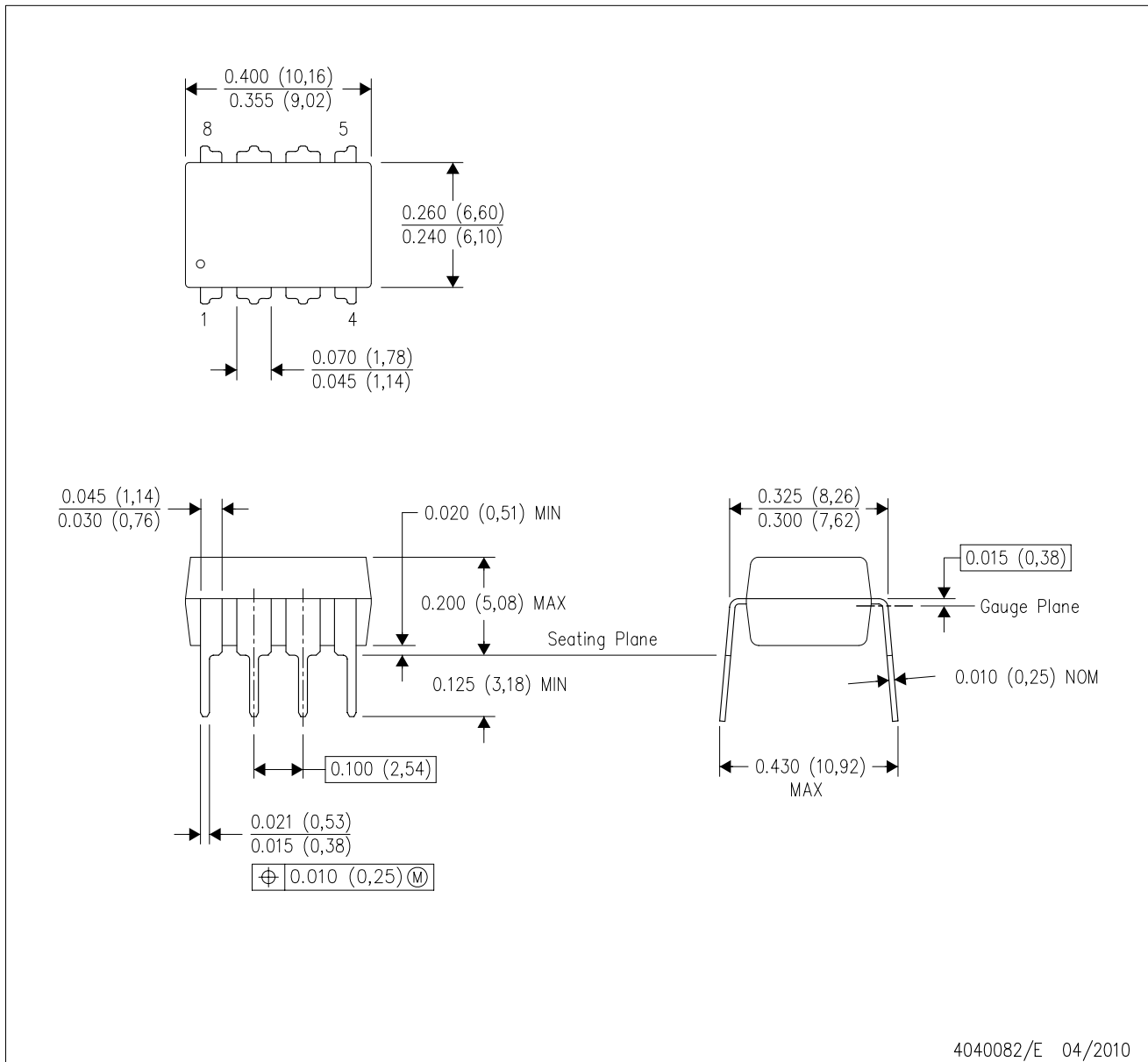
4214825/C 02/2019

NOTES: (continued)

8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
9. Board assembly site may have different recommendations for stencil design.

P (R-PDIP-T8)

PLASTIC DUAL-IN-LINE PACKAGE



- NOTES:
- A. All linear dimensions are in inches (millimeters).
  - B. This drawing is subject to change without notice.
  - C. Falls within JEDEC MS-001 variation BA.

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