

TLC070, TLC071, TLC072, TLC073, TLC074, TLC075, TLC07xA FAMILY OF WIDE-BANDWIDTH HIGH-OUTPUT-DRIVE SINGLE SUPPLY OPERATIONAL AMPLIFIERS

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- **Wide Bandwidth . . . 10 MHz**
- **High Output Drive**
 - I_{OH} . . . 57 mA at $V_{DD} - 1.5$ V
 - I_{OL} . . . 55 mA at 0.5 V
- **High Slew Rate**
 - $SR+$. . . 16 V/ μ s
 - $SR-$. . . 19 V/ μ s
- **Wide Supply Range . . . 4.5 V to 16 V**
- **Supply Current . . . 1.9 mA/Channel**
- **Ultralow Power Shutdown Mode**
 - I_{DD} . . . 125 μ A/Channel
- **Low Input Noise Voltage . . . 7 nV/ \sqrt{Hz}**
- **Input Offset Voltage . . . 60 μ V**
- **Ultra-Small Packages**
 - 8 or 10 Pin MSOP (TLC070/1/2/3)

Operational Amplifier



description

The first members of TI's new BiMOS general-purpose operational amplifier family are the TLC07x. The BiMOS family concept is simple: provide an upgrade path for BiFET users who are moving away from dual-supply to single-supply systems and demand higher AC and dc performance. With performance rated from 4.5 V to 16 V across commercial (0°C to 70°C) and an extended industrial temperature range (-40°C to 125°C), BiMOS suits a wide range of audio, automotive, industrial and instrumentation applications. Familiar features like offset nulling pins, and new features like MSOP PowerPAD™ packages and shutdown modes, enable higher levels of performance in a variety of applications.

Developed in TI's patented LBC3 BiCMOS process, the new BiMOS amplifiers combine a very high input impedance low-noise CMOS front end with a high-drive bipolar output stage, thus providing the optimum performance features of both. AC performance improvements over the TL07x BiFET predecessors include a bandwidth of 10 MHz (an increase of 300%) and voltage noise of 7 nV/ \sqrt{Hz} (an improvement of 60%). DC improvements include a factor of 4 reduction in input offset voltage down to 1.5 mV (maximum) in the standard grade, and a power supply rejection improvement of greater than 40 dB to 130 dB. Added to this list of impressive features is the ability to drive ± 50 -mA loads comfortably from an ultrasmall-footprint MSOP PowerPAD package, which positions the TLC07x as the ideal high-performance general-purpose operational amplifier family.

FAMILY PACKAGE TABLE

DEVICE	NO. OF CHANNELS	PACKAGE TYPES				SHUTDOWN	UNIVERSAL EVM BOARD
		MSOP	PDIP	SOIC	TSSOP		
TLC070	1	8	8	8	—	Yes	Refer to the EVM Selection Guide (Lit# SLOU060)
TLC071	1	8	8	8	—	—	
TLC072	2	8	8	8	—	—	
TLC073	2	10	14	14	—	Yes	
TLC074	4	—	14	14	20	—	
TLC075	4	—	16	16	20	Yes	



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TLC070 and TLC071 AVAILABLE OPTIONS

T _A	PACKAGED DEVICES			
	SMALL OUTLINE (D) [†]	SMALL OUTLINE (DGN) [†]	SYMBOL	PLASTIC DIP (P)
0°C to 70°C	TLC070CD TLC071CD	TLC070CDGN TLC071CDGN	xxTIACS xxTIACU	TLC070CP TLC071CP
-40°C to 125°C	TLC070ID TLC071ID	TLC070IDGN TLC071IDGN	xxTIACT xxTIACV	TLC070IP TLC071IP
	TLC070AID TLC071AID	— —	— —	TLC070AIP TLC071AIP

[†] This package is available taped and reeled. To order this packaging option, add an R suffix to the part number (e.g., TLC070CDR).

TLC072 and TLC073 AVAILABLE OPTIONS

T _A	PACKAGED DEVICES						
	SMALL OUTLINE (D) [†]	MSOP				PLASTIC DIP (N)	PLASTIC DIP (P)
		(DGN) [†]	SYMBOL [‡]	(DGQ) [†]	SYMBOL [‡]		
0°C to 70°C	TLC072CD TLC073CD	TLC072CDGN —	xxTIADV —	— TLC073CDGQ	— xxTIADX	— TLC073CN	TLC072CP —
-40°C to 125°C	TLC072ID TLC073ID	TLC072IDGN —	xxTIADW —	— TLC073IDGQ	— xxTIADY	— TLC073IN	TLC072IP —
	TLC072AID TLC073AID	— —	— —	— —	— —	— TLC073AIN	TLC072AIP —

[†] This package is available taped and reeled. To order this packaging option, add an R suffix to the part number (e.g., TLC072CDR).

[‡] xx represents the device date code.

TLC074 and TLC075 AVAILABLE OPTIONS

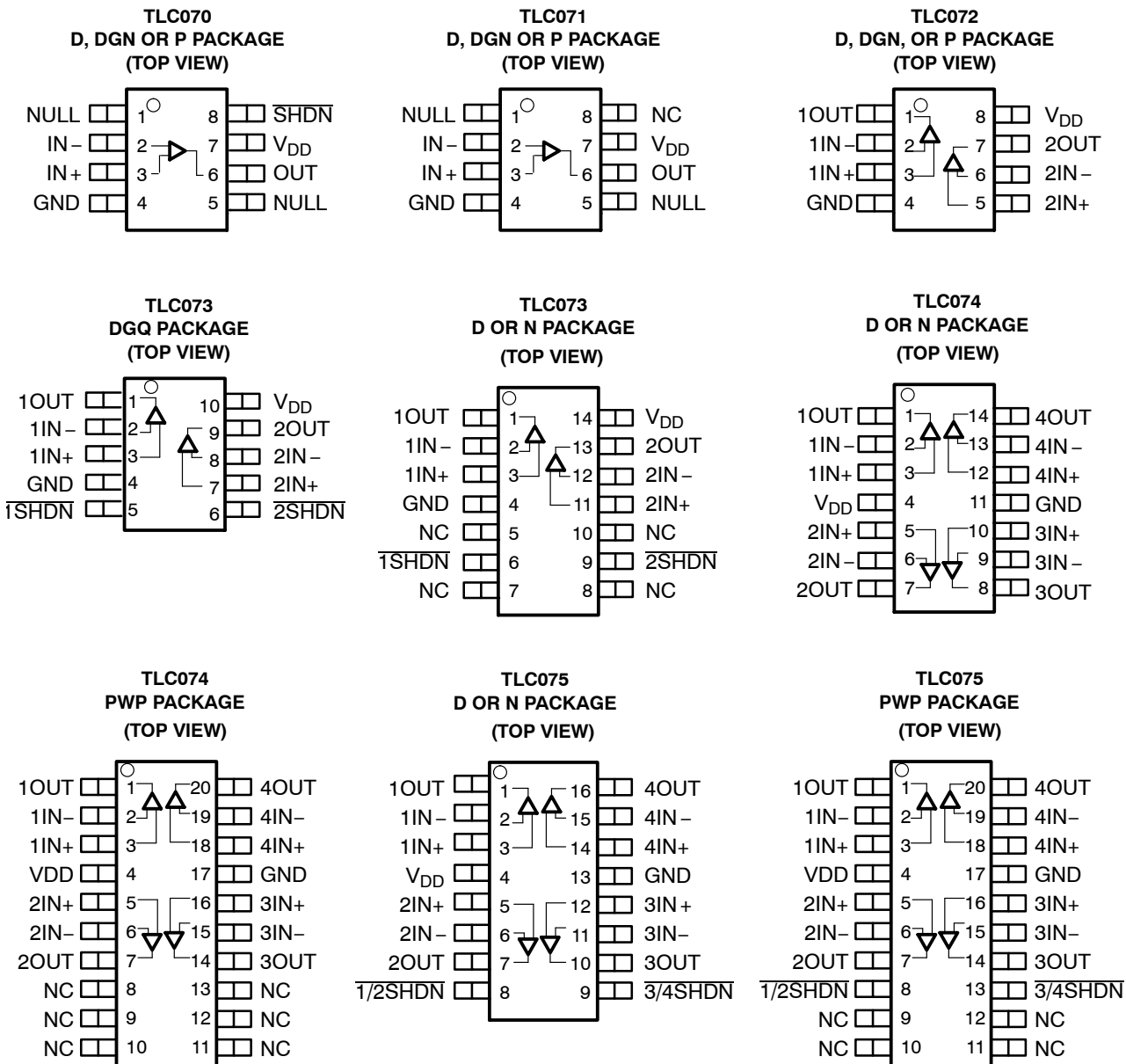
T _A	PACKAGED DEVICES		
	SMALL OUTLINE (D) [†]	PLASTIC DIP (N)	TSSOP (PWP) [†]
0°C to 70°C	TLC074CD TLC075CD	TLC074CN TLC075CN	TLC074CPWP TLC075CPWP
-40°C to 125°C	TLC074ID TLC075ID	TLC074IN TLC075IN	TLC074IPWP TLC075IPWP
	TLC074AID TLC075AID	TLC074AIN TLC075AIN	TLC074AIPWP TLC075AIPWP

[†] This package is available taped and reeled. To order this packaging option, add an R suffix to the part number (e.g., TLC074CDR).

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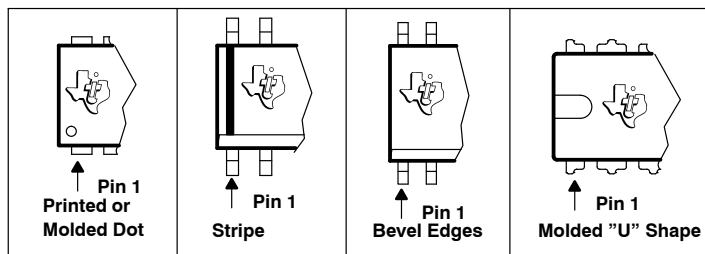
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TLC07x PACKAGE PIN OUTS



NC – No internal connection

TYPICAL PIN 1 INDICATORS



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absolute maximum ratings over operating free-air temperature range (unless otherwise noted)[†]

Supply voltage, V_{DD} (see Note 1)	17 V
Differential input voltage range, V_{ID}	$\pm V_{DD}$
Continuous total power dissipation	See Dissipation Rating Table
Operating free-air temperature range, T_A : C suffix	0°C to 70°C
I suffix	-40°C to 125°C
Maximum junction temperature, T_J	150°C
Storage temperature range, T_{stg}	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

[†] Stresses beyond those listed under “absolute maximum ratings” may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under “recommended operating conditions” is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

NOTE 1: All voltage values, except differential voltages, are with respect to GND.

DISSIPATION RATING TABLE

PACKAGE	θ_{JC} (°C/W)	θ_{JA} (°C/W)	$T_A \leq 25^\circ\text{C}$ POWER RATING
D (8)	38.3	176	710 mW
D (14)	26.9	122.3	1022 mW
D (16)	25.7	114.7	1090 mW
DGN (8)	4.7	52.7	2.37 W
DGQ (10)	4.7	52.3	2.39 W
N (14, 16)	32	78	1600 mW
P (8)	41	104	1200 mW
PWP (20)	1.40	26.1	4.79 W

recommended operating conditions

		MIN	MAX	UNIT
Supply voltage, V_{DD}	Single supply	4.5	16	V
	Split supply	± 2.25	± 8	
Common-mode input voltage, V_{ICR}		+0.5	$V_{DD}-0.8$	V
Shutdown on/off voltage level [‡]	V_{IH}	2		V
	V_{OL}	0.8		
Operating free-air temperature, T_A	C-suffix	0	70	°C
	I-suffix	-40	125	

[‡] Relative to the voltage on the GND terminal of the device.

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electrical characteristics at specified free-air temperature, $V_{DD} = 5\text{ V}$ (unless otherwise noted)
(continued)

PARAMETER		TEST CONDITIONS	T_A^\dagger	MIN	TYP	MAX	UNIT
A_{VD}	Large-signal differential voltage amplification	$V_{O(PP)} = 3\text{ V}$, $R_L = 10\text{ k}\Omega$	25°C	100	120		dB
			Full range	100			
$r_{i(d)}$	Differential input resistance		25°C		1000		$G\Omega$
C_{IC}	Common-mode input capacitance	$f = 10\text{ kHz}$	25°C		22.9		pF
z_o	Closed-loop output impedance	$f = 10\text{ kHz}$, $A_V = 10$	25°C		0.25		Ω
CMRR	Common-mode rejection ratio	$V_{IC} = 1\text{ to }3\text{ V}$, $R_S = 50\ \Omega$	25°C	80	95		dB
			Full range	80			
k_{SVR}	Supply voltage rejection ratio ($\Delta V_{DD} / \Delta V_{IO}$)	$V_{DD} = 4.5\text{ V to }16\text{ V}$, No load $V_{IC} = V_{DD}/2$	25°C	80	100		dB
			Full range	80			
I_{DD}	Supply current (per channel)	$V_O = 2.5\text{ V}$, No load	25°C		1.9	2.5	mA
			Full range			3.5	
$I_{DD(SHDN)}$	Supply current in shutdown mode (per channel) (TLC070, TLC073, TLC075)	$\overline{SHDN} \leq 0.8\text{ V}$	25°C		125	200	μA
			Full range			250	

† Full range is 0°C to 70°C for C suffix and –40°C to 125°C for I suffix. If not specified, full range is –40°C to 125°C.

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operating characteristics at specified free-air temperature, $V_{DD} = 5\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS		T_A^\dagger	MIN	TYP	MAX	UNIT
SR+	Positive slew rate at unity gain	$V_{O(PP)} = 0.8\text{ V}$, $R_L = 10\text{ k}\Omega$	$C_L = 50\text{ pF}$	25°C	10	16		V/ μs
				Full range	9.5			
SR-	Negative slew rate at unity gain	$V_{O(PP)} = 0.8\text{ V}$, $R_L = 10\text{ k}\Omega$	$C_L = 50\text{ pF}$	25°C	12.5	19		V/ μs
				Full range	10			
V_n	Equivalent input noise voltage			f = 100 Hz	25°C		12	nV/ $\sqrt{\text{Hz}}$
				f = 1 kHz	25°C		7	
I_n	Equivalent input noise current	f = 1 kHz		25°C		0.6	fA/ $\sqrt{\text{Hz}}$	
THD + N	Total harmonic distortion plus noise	$V_{O(PP)} = 3\text{ V}$, $R_L = 10\text{ k}\Omega$ and 250 Ω , f = 1 kHz	$A_V = 1$	25°C	0.002%			
			$A_V = 10$		0.012%			
			$A_V = 100$		0.085%			
$t_{(on)}$	Amplifier turn-on time [‡]	$R_L = 10\text{ k}\Omega$		25°C	0.15		μs	
$t_{(off)}$	Amplifier turn-off time [‡]			25°C	1.3		μs	
Gain-bandwidth product		f = 10 kHz,	$R_L = 10\text{ k}\Omega$	25°C		10	MHz	
t_s	Settling time	$V_{(STEP)PP} = 1\text{ V}$, $A_V = -1$, $C_L = 10\text{ pF}$, $R_L = 10\text{ k}\Omega$	0.1%	25°C	0.18		μs	
			0.01%		0.39			
		$V_{(STEP)PP} = 1\text{ V}$, $A_V = -1$, $C_L = 47\text{ pF}$, $R_L = 10\text{ k}\Omega$	0.1%		0.18			
			0.01%		0.39			
ϕ_m	Phase margin	$R_L = 10\text{ k}\Omega$,	$C_L = 50\text{ pF}$	25°C		32°		
		$R_L = 10\text{ k}\Omega$,	$C_L = 0\text{ pF}$			40°		
	Gain margin	$R_L = 10\text{ k}\Omega$,	$C_L = 50\text{ pF}$	25°C		2.2		
		$R_L = 10\text{ k}\Omega$,	$C_L = 0\text{ pF}$			3.3		

[†] Full range is 0°C to 70°C for C suffix and -40°C to 125°C for I suffix. If not specified, full range is -40°C to 125°C.

[‡] Disable time and enable time are defined as the interval between application of the logic signal to SHDN and the point at which the supply current has reached half its final value.

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electrical characteristics at specified free-air temperature, $V_{DD} = 12\text{ V}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS	T_A^\dagger	MIN	TYP	MAX	UNIT
V_{IO} Input offset voltage	$V_{DD} = 12\text{ V}$ $V_{IC} = 6\text{ V}$ $V_O = 6\text{ V}$ $R_S = 50\ \Omega$	TLC070/1/2/3, TLC074/5	25°C	390	1900	μV
			Full range		3000	
		TLC070/1/2/3A, TLC074/5A	25°C	390	1400	
			Full range		2000	
α_{VIO} Temperature coefficient of input offset voltage			1.2		$\mu\text{V}/^\circ\text{C}$	
I_{IO} Input offset current	$V_{DD} = 12\text{ V}$ $V_{IC} = 6\text{ V}$ $V_O = 6\text{ V}$ $R_S = 50\ \Omega$	TLC07xC TLC07xI	25°C	0.7	50	pA
			Full range		100 700	
I_{IB} Input bias current	$V_{DD} = 12\text{ V}$ $V_{IC} = 6\text{ V}$ $V_O = 6\text{ V}$ $R_S = 50\ \Omega$	TLC07xC TLC07xI	25°C	1.5	50	pA
			Full range		100 700	
V_{ICR} Common-mode input voltage	$R_S = 50\ \Omega$		25°C	0.5 to 11.2		V
			Full range		0.5 to 11.2	
V_{OH} High-level output voltage	$V_{IC} = 6\text{ V}$	$I_{OH} = -1\text{ mA}$	25°C	11.1	11.2	V
			Full range		11	
		$I_{OH} = -20\text{ mA}$	25°C	10.8	10.9	
			Full range		10.7	
		$I_{OH} = -35\text{ mA}$	25°C	10.6	10.7	
			Full range		10.3	
		$I_{OH} = -50\text{ mA}$	25°C	10.4	10.5	
			-40°C to 85°C		10.3	
V_{OL} Low-level output voltage	$V_{IC} = 6\text{ V}$	$I_{OL} = 1\text{ mA}$	25°C	0.17	0.25	V
			Full range		0.35	
		$I_{OL} = 20\text{ mA}$	25°C	0.35	0.45	
			Full range		0.5	
		$I_{OL} = 35\text{ mA}$	25°C	0.4	0.52	
			Full range		0.6	
		$I_{OL} = 50\text{ mA}$	25°C	0.45	0.6	
			-40°C to 85°C		0.65	
I_{OS} Short-circuit output current	Sourcing	25°C		150	mA	
	Sinking	25°C		150		
I_O Output current	$V_{OH} = 1.5\text{ V}$ from positive rail	25°C		57	mA	
	$V_{OL} = 0.5\text{ V}$ from negative rail	25°C		55		

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electrical characteristics at specified free-air temperature, $V_{DD} = 12\text{ V}$ (unless otherwise noted)
(continued)

PARAMETER		TEST CONDITIONS	T_A^\dagger	MIN	TYP	MAX	UNIT
A_{VD}	Large-signal differential voltage amplification	$V_{O(PP)} = 8\text{ V}$, $R_L = 10\text{ k}\Omega$	25°C	120	140		dB
			Full range	120			
$r_{i(d)}$	Differential input resistance		25°C		1000		G Ω
C_{IC}	Common-mode input capacitance	$f = 10\text{ kHz}$	25°C		21.6		pF
z_o	Closed-loop output impedance	$f = 10\text{ kHz}$, $A_V = 10$	25°C		0.25		Ω
CMRR	Common-mode rejection ratio	$V_{IC} = 1\text{ to }10\text{ V}$, $R_S = 50\ \Omega$	25°C	80	100		dB
			Full range	80			
k_{SVR}	Supply voltage rejection ratio ($\Delta V_{DD} / \Delta V_{IO}$)	$V_{DD} = 4.5\text{ V to }16\text{ V}$, No load $V_{IC} = V_{DD}/2$,	25°C	80	100		dB
			Full range	80			
I_{DD}	Supply current (per channel)	$V_O = 7.5\text{ V}$, No load	25°C		2.1	2.9	mA
			Full range			3.5	
$I_{DD(SHDN)}$	Supply current in shutdown mode (TLC070, TLC073, TLC075) (per channel)	$\overline{SHDN} \leq 0.8\text{ V}$	25°C		125	200	μA
			Full range			250	

† Full range is 0°C to 70°C for C suffix and –40°C to 125°C for I suffix. If not specified, full range is –40°C to 125°C.

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operating characteristics at specified free-air temperature, $V_{DD} = 12\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS		T_A^\dagger	MIN	TYP	MAX	UNIT
SR+	Positive slew rate at unity gain	$V_{O(PP)} = 2\text{ V}$, $R_L = 10\text{ k}\Omega$	$C_L = 50\text{ pF}$	25°C	10	16		V/ μs
				Full range	9.5			
SR-	Negative slew rate at unity gain	$V_{O(PP)} = 2\text{ V}$, $R_L = 10\text{ k}\Omega$	$C_L = 50\text{ pF}$	25°C	12.5	19		V/ μs
				Full range	10			
V_n	Equivalent input noise voltage	f = 100 Hz		25°C		12		nV/ $\sqrt{\text{Hz}}$
		f = 1 kHz		25°C		7		
I_n	Equivalent input noise current	f = 1 kHz		25°C		0.6		fA/ $\sqrt{\text{Hz}}$
THD + N	Total harmonic distortion plus noise	$V_{O(PP)} = 8\text{ V}$, $R_L = 10\text{ k}\Omega$ and 250 Ω , f = 1 kHz	$A_V = 1$	25°C		0.002%		
			$A_V = 10$			0.005%		
			$A_V = 100$			0.022%		
$t_{(on)}$	Amplifier turn-on time [‡]	$R_L = 10\text{ k}\Omega$		25°C		0.47		μs
$t_{(off)}$	Amplifier turn-off time [‡]	$R_L = 10\text{ k}\Omega$		25°C		2.5		μs
Gain-bandwidth product		f = 10 kHz,	$R_L = 10\text{ k}\Omega$	25°C		10		MHz
t_s	Settling time	$V_{(STEP)PP} = 1\text{ V}$, $A_V = -1$, $C_L = 10\text{ pF}$, $R_L = 10\text{ k}\Omega$	0.1%	25°C		0.17	μs	
			0.01%			0.22		
		$V_{(STEP)PP} = 1\text{ V}$, $A_V = -1$, $C_L = 47\text{ pF}$, $R_L = 10\text{ k}\Omega$	0.1%			0.17		
			0.01%			0.29		
ϕ_m	Phase margin	$R_L = 10\text{ k}\Omega$,	$C_L = 50\text{ pF}$	25°C		37°		
		$R_L = 10\text{ k}\Omega$,	$C_L = 0\text{ pF}$			42°		
	Gain margin	$R_L = 10\text{ k}\Omega$,	$C_L = 50\text{ pF}$	25°C		3.1	dB	
		$R_L = 10\text{ k}\Omega$,	$C_L = 0\text{ pF}$			4		

[†] Full range is 0°C to 70°C for C suffix and -40°C to 125°C for I suffix. If not specified, full range is -40°C to 125°C.

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TYPICAL CHARACTERISTICS

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V_{OL}	Low-level output voltage	vs Low-level output current	6, 8
Z_o	Output impedance	vs Frequency	9
I_{DD}	Supply current	vs Supply voltage	10
PSRR	Power supply rejection ratio	vs Frequency	11
CMRR	Common-mode rejection ratio	vs Frequency	12
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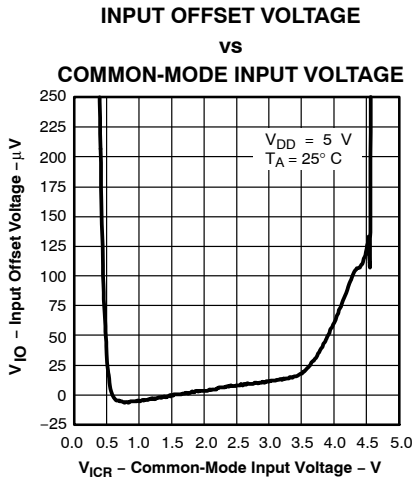


Figure 1

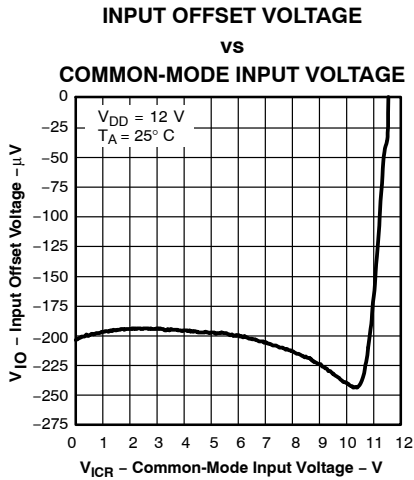


Figure 2

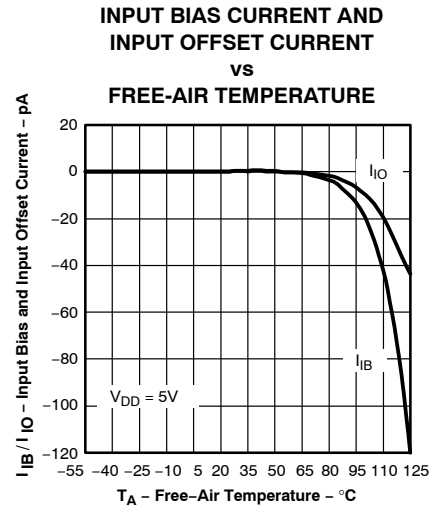


Figure 3

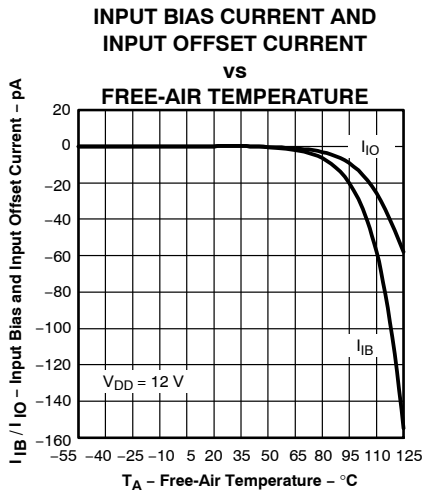


Figure 4

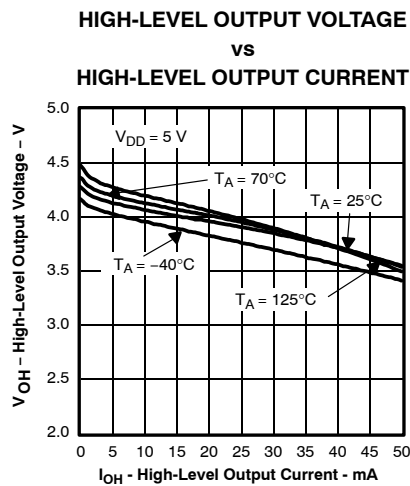


Figure 5

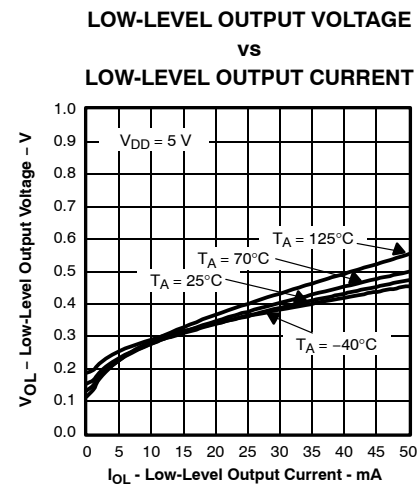


Figure 6

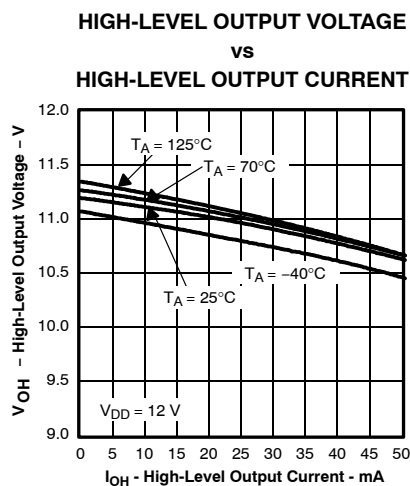


Figure 7

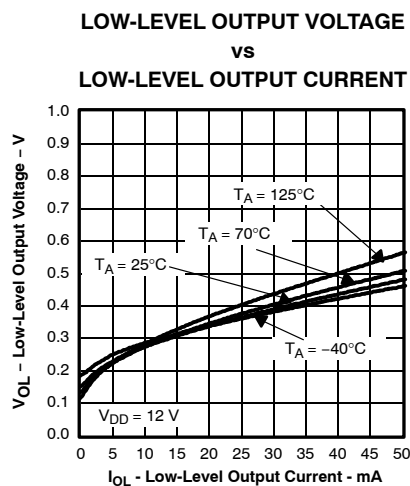


Figure 8

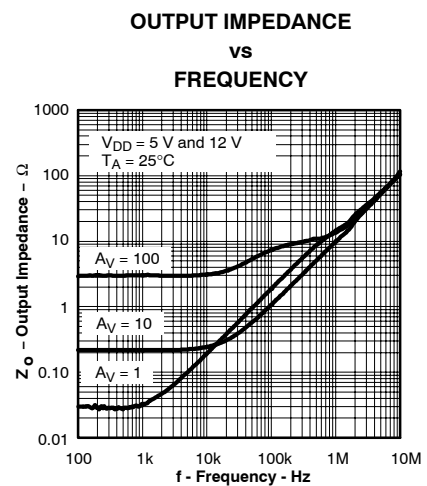
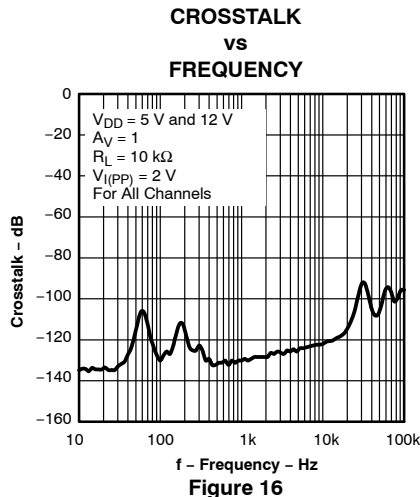
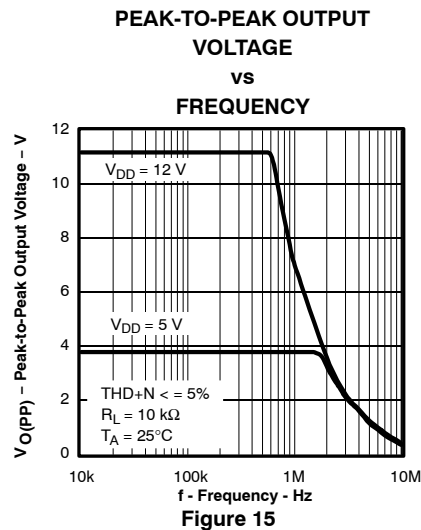
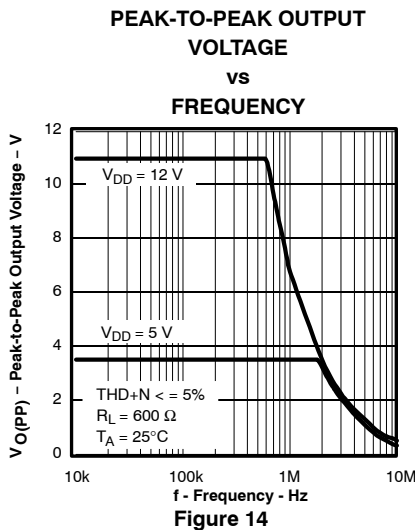
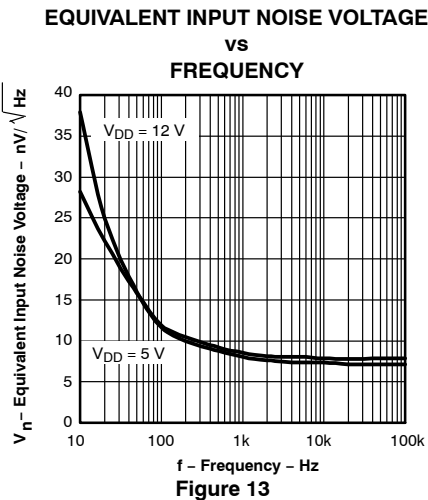
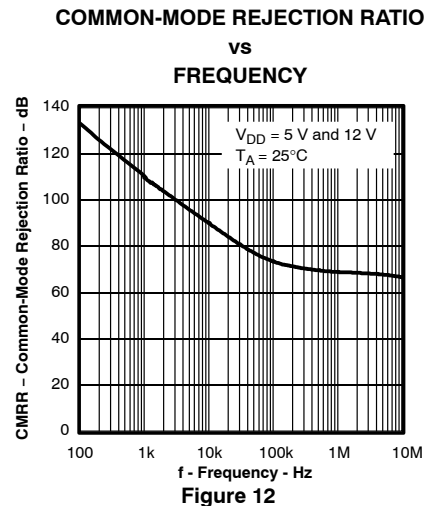
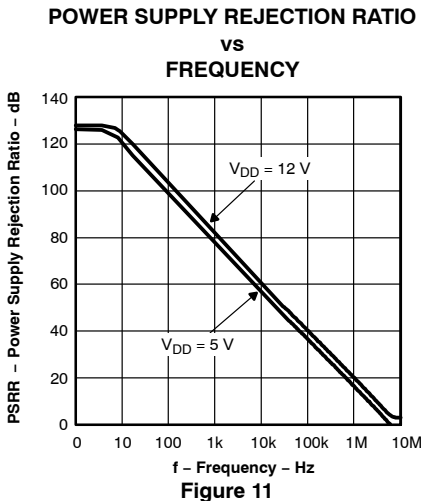
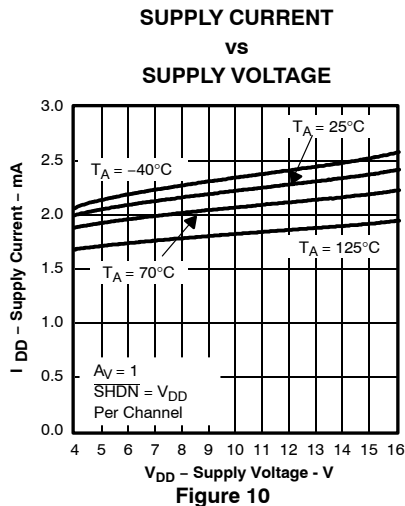


Figure 9

TLC070, TLC071, TLC072, TLC073, TLC074, TLC075, TLC07xA FAMILY OF WIDE-BANDWIDTH HIGH-OUTPUT-DRIVE SINGLE SUPPLY OPERATIONAL AMPLIFIERS

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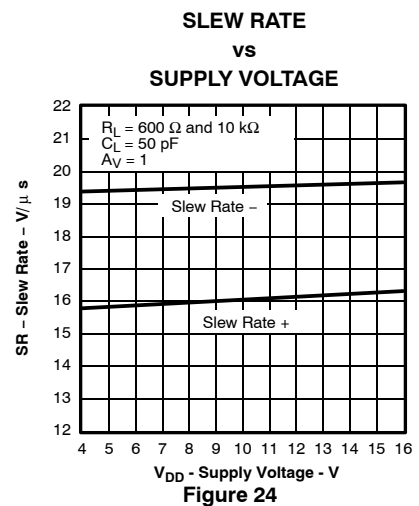
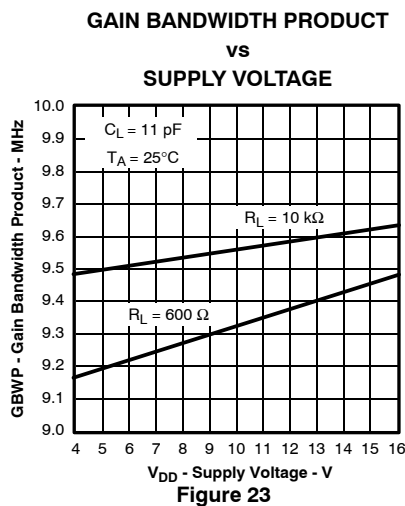
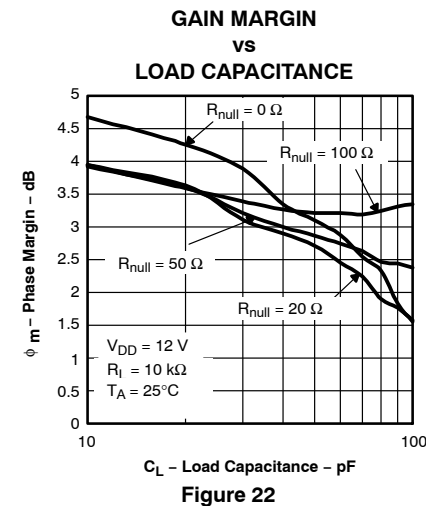
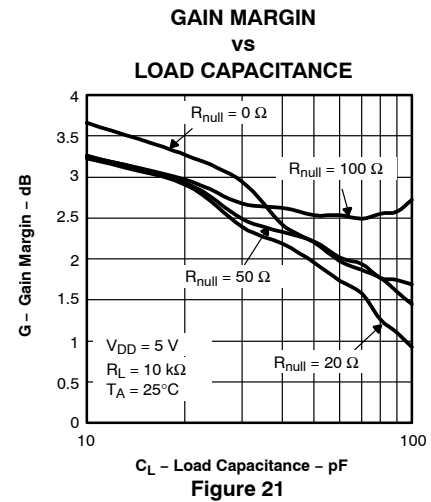
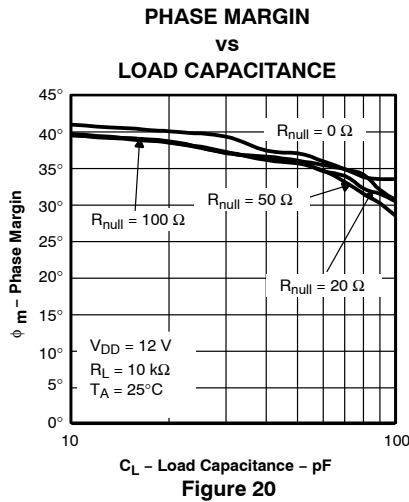
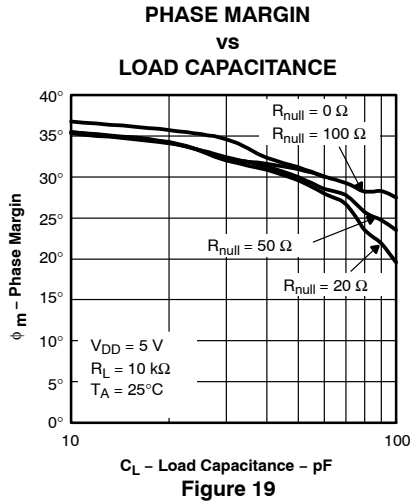
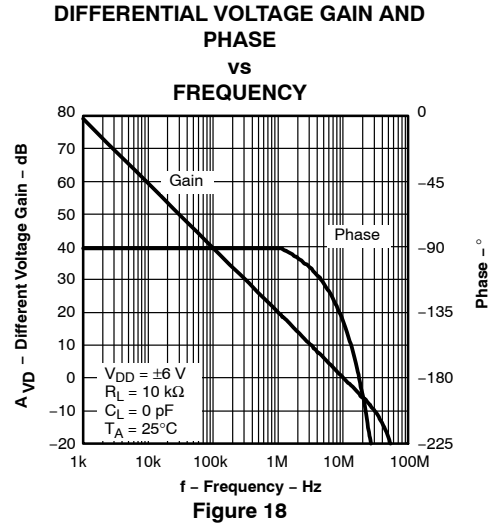
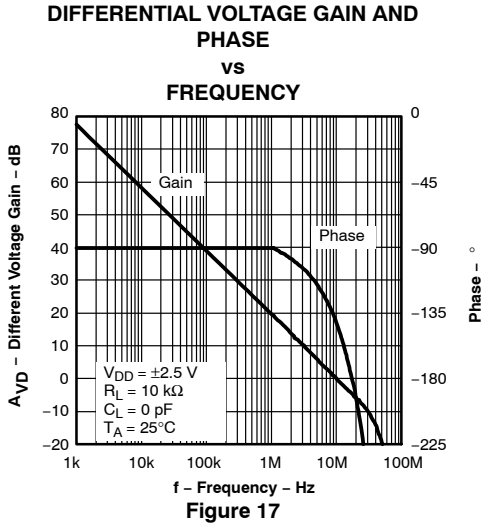
TYPICAL CHARACTERISTICS



TLC070, TLC071, TLC072, TLC073, TLC074, TLC075, TLC07xA FAMILY OF WIDE-BANDWIDTH HIGH-OUTPUT-DRIVE SINGLE SUPPLY OPERATIONAL AMPLIFIERS

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TLC070, TLC071, TLC072, TLC073, TLC074, TLC075, TLC07xA FAMILY OF WIDE-BANDWIDTH HIGH-OUTPUT-DRIVE SINGLE SUPPLY OPERATIONAL AMPLIFIERS

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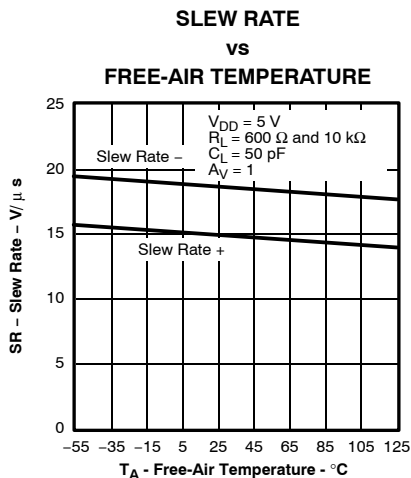


Figure 25

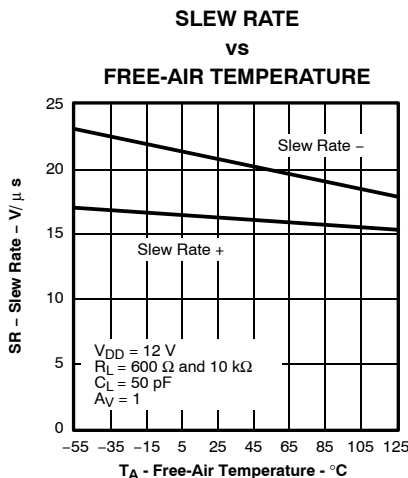


Figure 26

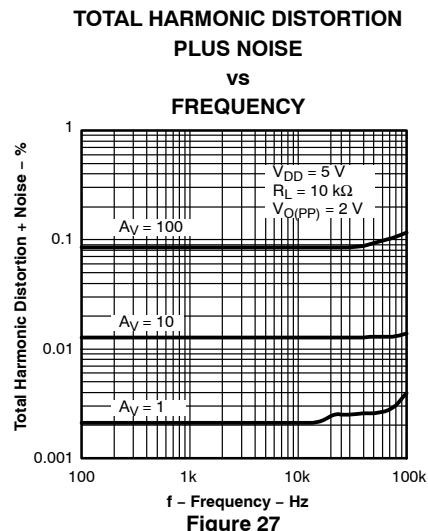


Figure 27

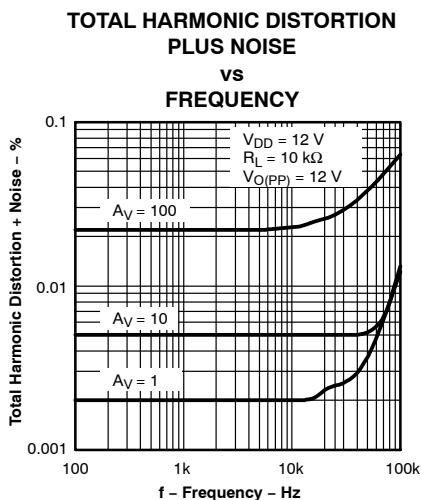


Figure 28

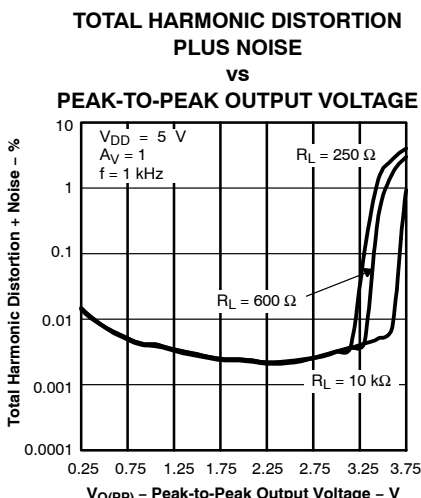


Figure 29

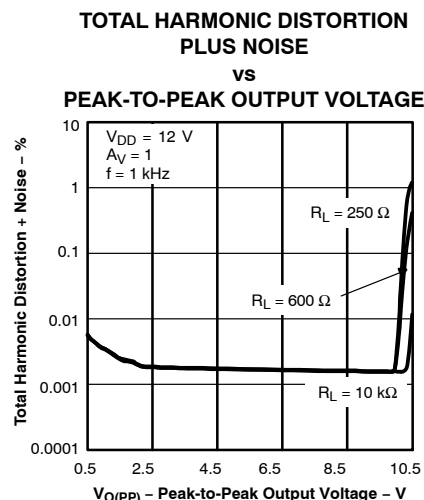


Figure 30

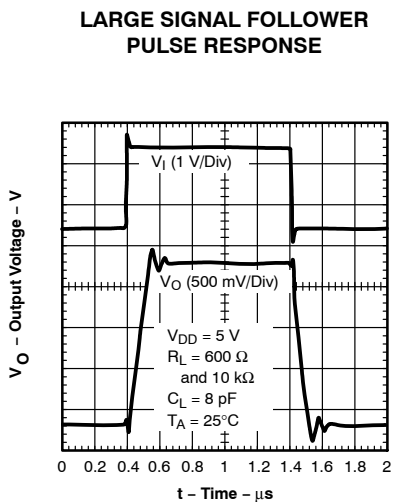


Figure 31

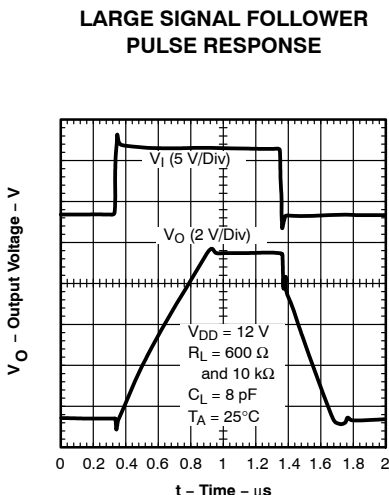


Figure 32

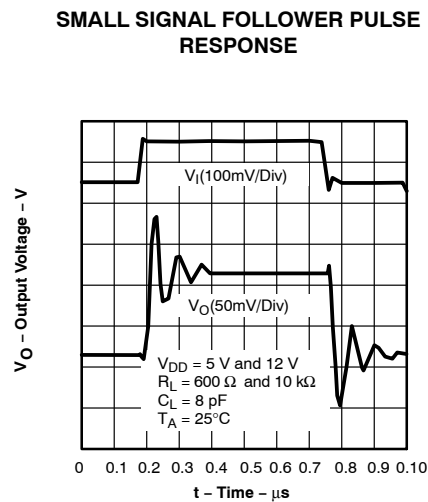


Figure 33

TLC070, TLC071, TLC072, TLC073, TLC074, TLC075, TLC07xA FAMILY OF WIDE-BANDWIDTH HIGH-OUTPUT-DRIVE SINGLE SUPPLY OPERATIONAL AMPLIFIERS

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TYPICAL CHARACTERISTICS

**LARGE SIGNAL INVERTING
PULSE RESPONSE**



Figure 34

**LARGE SIGNAL INVERTING
PULSE RESPONSE**

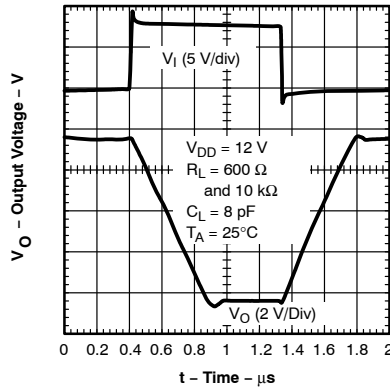


Figure 35

**SMALL SIGNAL INVERTING
PULSE RESPONSE**

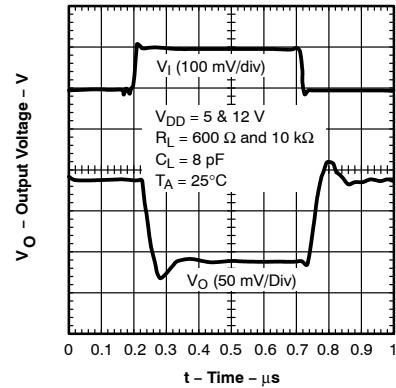


Figure 36

**SHUTDOWN FORWARD
ISOLATION
vs
FREQUENCY**

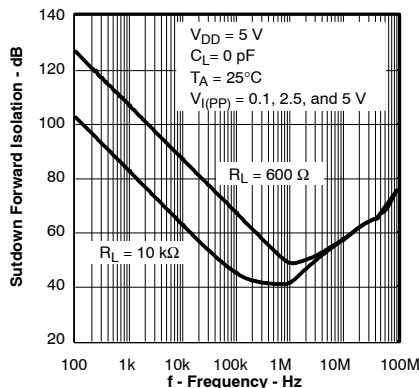


Figure 37

**SHUTDOWN FORWARD
ISOLATION
vs
FREQUENCY**

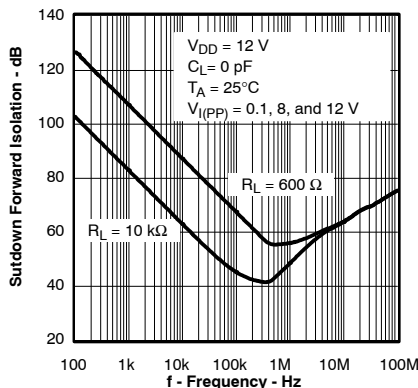


Figure 38

**SHUTDOWN REVERSE
ISOLATION
vs
FREQUENCY**

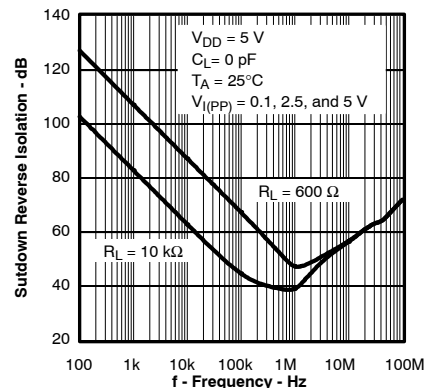


Figure 39

**SHUTDOWN REVERSE
ISOLATION
vs
FREQUENCY**

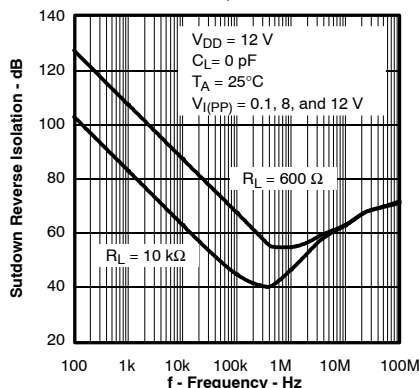


Figure 40

**SHUTDOWN SUPPLY CURRENT
vs
SUPPLY VOLTAGE**

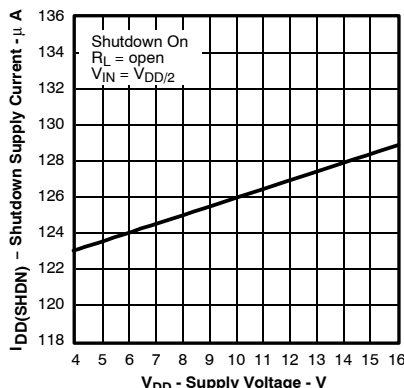


Figure 41

**SHUTDOWN SUPPLY CURRENT
vs
FREE-AIR TEMPERATURE**

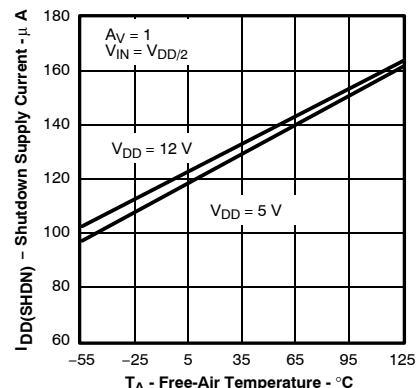
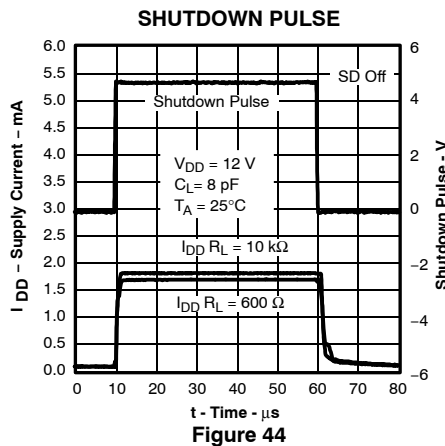
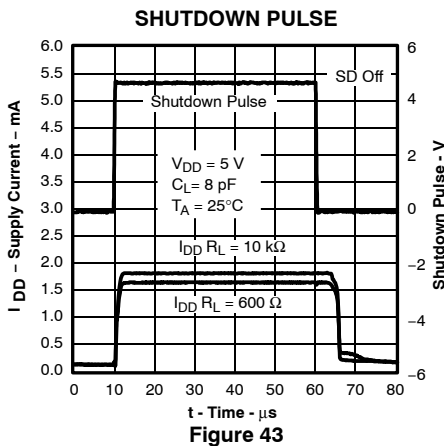


Figure 42

TLC070, TLC071, TLC072, TLC073, TLC074, TLC075, TLC07xA FAMILY OF WIDE-BANDWIDTH HIGH-OUTPUT-DRIVE SINGLE SUPPLY OPERATIONAL AMPLIFIERS

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TYPICAL CHARACTERISTICS



PARAMETER MEASUREMENT INFORMATION

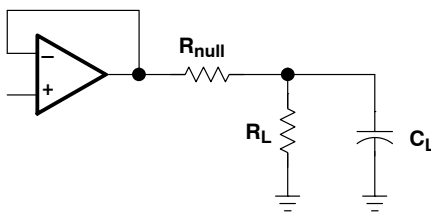
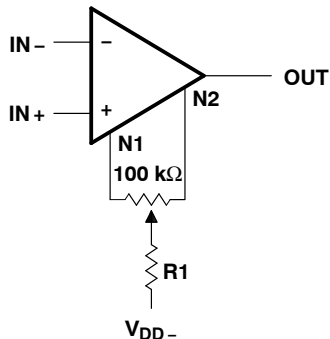


Figure 45

APPLICATION INFORMATION

input offset voltage null circuit

The TLC070 and TLC071 has an input offset nulling function. Refer to Figure 46 for the diagram.



NOTE A: R1 = 5.6 kΩ for offset voltage adjustment of ± 10 mV.
R1 = 20 kΩ for offset voltage adjustment of ± 3 mV.

Figure 46. Input Offset Voltage Null Circuit

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APPLICATION INFORMATION

driving a capacitive load

When the amplifier is configured in this manner, capacitive loading directly on the output will decrease the device's phase margin leading to high frequency ringing or oscillations. Therefore, for capacitive loads of greater than 10 pF, it is recommended that a resistor be placed in series (R_{NULL}) with the output of the amplifier, as shown in Figure 47. A minimum value of 20 Ω should work well for most applications.

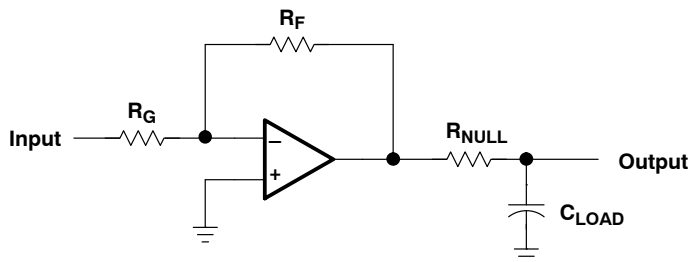


Figure 47. Driving a Capacitive Load

offset voltage

The output offset voltage, (V_{OO}) is the sum of the input offset voltage (V_{IO}) and both input bias currents (I_{IB}) times the corresponding gains. The following schematic and formula can be used to calculate the output offset voltage:

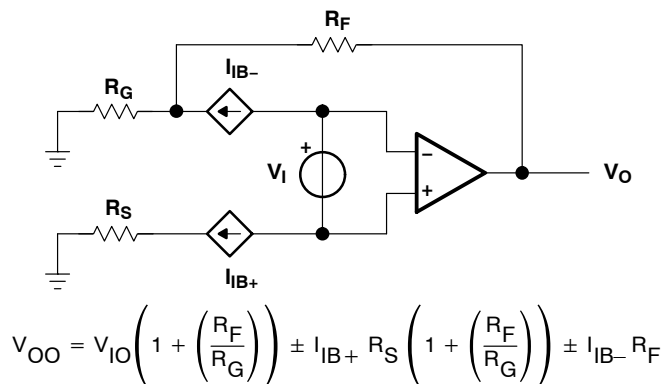


Figure 48. Output Offset Voltage Model

APPLICATION INFORMATION

high speed CMOS input amplifiers

The TLC07x is a family of high-speed low-noise CMOS input operational amplifiers that has an input capacitance of the order of 20 pF. Any resistor used in the feedback path adds a pole in the transfer function equivalent to the input capacitance multiplied by the combination of source resistance and feedback resistance. For example, a gain of -10 , a source resistance of $1\text{ k}\Omega$, and a feedback resistance of $10\text{ k}\Omega$ add an additional pole at approximately 8 MHz. This is more apparent with CMOS amplifiers than bipolar amplifiers due to their greater input capacitance.

This is of little consequence on slower CMOS amplifiers, as this pole normally occurs at frequencies above their unity-gain bandwidth. However, the TLC07x with its 10-MHz bandwidth means that this pole normally occurs at frequencies where there is on the order of 5 dB gain left and the phase shift adds considerably.

The effect of this pole is the strongest with large feedback resistances at small closed loop gains. As the feedback resistance is increased, the gain peaking increases at a lower frequency and the 180° phase shift crossover point also moves down in frequency, decreasing the phase margin.

For the TLC07x, the maximum feedback resistor recommended is $5\text{ k}\Omega$; larger resistances can be used but a capacitor in parallel with the feedback resistor is recommended to counter the effects of the input capacitance pole.

The TLC073 with a 1-V step response has an 80% overshoot with a natural frequency of 3.5 MHz when configured as a unity gain buffer and with a $10\text{-k}\Omega$ feedback resistor. By adding a 10-pF capacitor in parallel with the feedback resistor, the overshoot is reduced to 40% and eliminates the natural frequency, resulting in a much faster settling time (see Figure 49). The 10-pF capacitor was chosen for convenience only.

Load capacitance had little effect on these measurements due to the excellent output drive capability of the TLC07x.

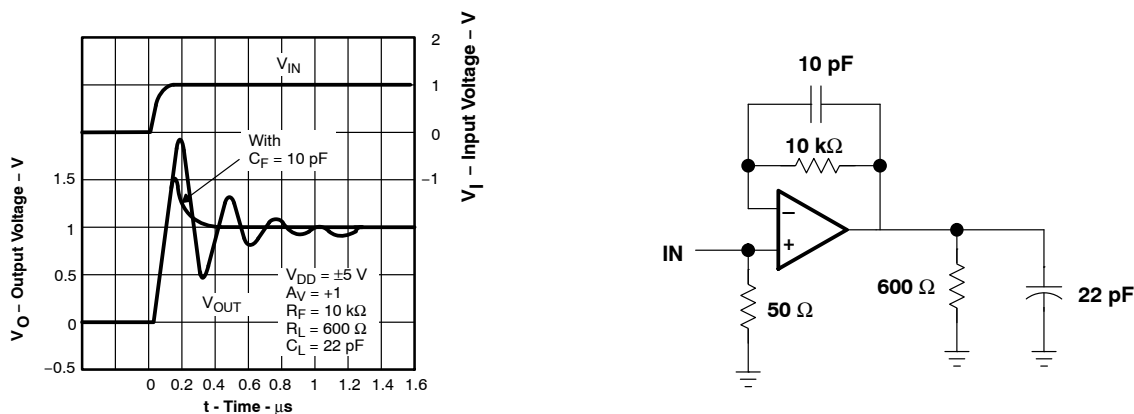


Figure 49. 1-V Step Response

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APPLICATION INFORMATION

general configurations

When receiving low-level signals, limiting the bandwidth of the incoming signals into the system is often required. The simplest way to accomplish this is to place an RC filter at the noninverting terminal of the amplifier (see Figure 50).

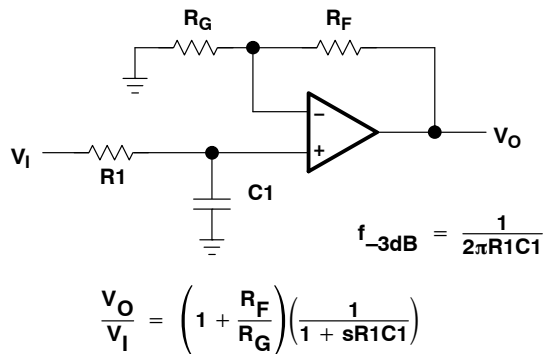


Figure 50. Single-Pole Low-Pass Filter

If even more attenuation is needed, a multiple pole filter is required. The Sallen-Key filter can be used for this task. For best results, the amplifier should have a bandwidth that is 8 to 10 times the filter frequency bandwidth. Failure to do this can result in phase shift of the amplifier.

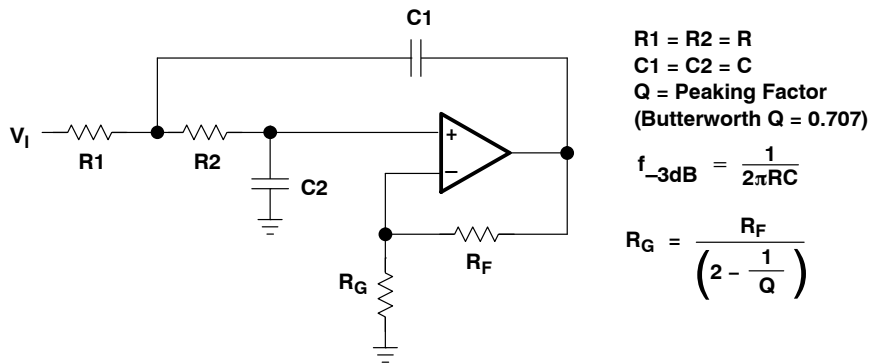


Figure 51. 2-Pole Low-Pass Sallen-Key Filter

APPLICATION INFORMATION

shutdown function

Three members of the TLC07x family (TLC070/3/5) have a shutdown terminal (**SHDN**) for conserving battery life in portable applications. When the shutdown terminal is tied low, the supply current is reduced to 125 $\mu\text{A}/\text{channel}$, the amplifier is disabled, and the outputs are placed in a high-impedance mode. To enable the amplifier, the shutdown terminal can either be left floating or pulled high. When the shutdown terminal is left floating, care should be taken to ensure that parasitic leakage current at the shutdown terminal does not inadvertently place the operational amplifier into shutdown. The shutdown terminal threshold is always referenced to the voltage on the GND terminal of the device. Therefore, when operating the device with split supply voltages (e.g. $\pm 2.5\text{ V}$), the shutdown terminal needs to be pulled to $V_{\text{DD-}}$ (not system ground) to disable the operational amplifier.

The amplifier's output with a shutdown pulse is shown in Figures 43 and 44. The amplifier is powered with a single 5-V supply and is configured as noninverting with a gain of 5. The amplifier turn-on and turn-off times are measured from the 50% point of the shutdown pulse to the 50% point of the output waveform. The times for the single, dual, and quad are listed in the data tables.

Figures 37, 38, 39, and 40 show the amplifier's forward and reverse isolation in shutdown. The operational amplifier is configured as a voltage follower ($A_V = 1$). The isolation performance is plotted across frequency using 0.1 V_{PP} , 2.5 V_{PP} , and 5 V_{PP} input signals at $\pm 2.5\text{ V}$ supplies and 0.1 V_{PP} , 8 V_{PP} , and 12 V_{PP} input signals at $\pm 6\text{ V}$ supplies.

circuit layout considerations

To achieve the levels of high performance of the TLC07x, follow proper printed-circuit board design techniques. A general set of guidelines is given in the following.

- Ground planes – It is highly recommended that a ground plane be used on the board to provide all components with a low inductive ground connection. However, in the areas of the amplifier inputs and output, the ground plane can be removed to minimize the stray capacitance.
- Proper power supply decoupling – Use a 6.8- μF tantalum capacitor in parallel with a 0.1- μF ceramic capacitor on each supply terminal. It may be possible to share the tantalum among several amplifiers depending on the application, but a 0.1- μF ceramic capacitor should always be used on the supply terminal of every amplifier. In addition, the 0.1- μF capacitor should be placed as close as possible to the supply terminal. As this distance increases, the inductance in the connecting trace makes the capacitor less effective. The designer should strive for distances of less than 0.1 inches between the device power terminals and the ceramic capacitors.
- Sockets – Sockets can be used but are not recommended. The additional lead inductance in the socket pins will often lead to stability problems. Surface-mount packages soldered directly to the printed-circuit board is the best implementation.
- Short trace runs/compact part placements – Optimum high performance is achieved when stray series inductance has been minimized. To realize this, the circuit layout should be made as compact as possible, thereby minimizing the length of all trace runs. Particular attention should be paid to the inverting input of the amplifier. Its length should be kept as short as possible. This will help to minimize stray capacitance at the input of the amplifier.
- Surface-mount passive components – Using surface-mount passive components is recommended for high performance amplifier circuits for several reasons. First, because of the extremely low lead inductance of surface-mount components, the problem with stray series inductance is greatly reduced. Second, the small size of surface-mount components naturally leads to a more compact layout thereby minimizing both stray inductance and capacitance. If leaded components are used, it is recommended that the lead lengths be kept as short as possible.

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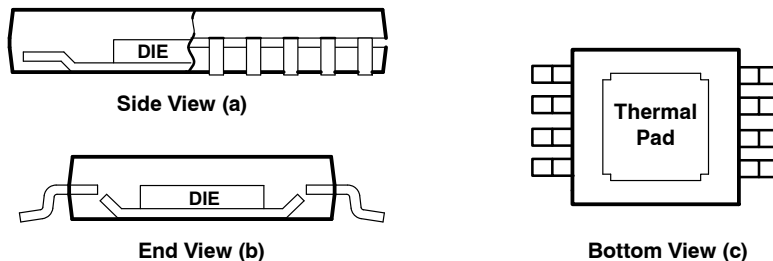
general PowerPAD design considerations

The TLC07x is available in a thermally-enhanced PowerPAD family of packages. These packages are constructed using a downset leadframe upon which the die is mounted [see Figure 52(a) and Figure 52(b)]. This arrangement results in the lead frame being exposed as a thermal pad on the underside of the package [see Figure 52(c)]. Because this thermal pad has direct thermal contact with the die, excellent thermal performance can be achieved by providing a good thermal path away from the thermal pad.

The PowerPAD package allows for both assembly and thermal management in one manufacturing operation. During the surface-mount solder operation (when the leads are being soldered), the thermal pad must be soldered to a copper area underneath the package. Through the use of thermal paths within this copper area, heat can be conducted away from the package into either a ground plane or other heat dissipating device.

Soldering the PowerPAD to the PCB is always required, even with applications that have low-power dissipation. This provides the necessary thermal and mechanical connection between the lead frame die pad and the PCB.

The PowerPAD package represents a breakthrough in combining the small area and ease of assembly of surface mount with mechanical methods of heatsinking.



NOTE A: The thermal pad is electrically isolated from all terminals in the package.

Figure 52. Views of Thermally-Enhanced DGN Package

APPLICATION INFORMATION

Although there are many ways to properly heatsink the PowerPAD package, the following steps illustrate the recommended approach.

general PowerPAD design considerations (continued)

1. The thermal pad must be connected to the same voltage potential as the GND pin.
2. Prepare the PCB with a top side etch pattern as illustrated in the thermal land pattern mechanical drawing at the end of this document. There should be etch for the leads as well as etch for the thermal pad.
3. Place five holes (single and dual) or nine holes (quad) in the area of the thermal pad. These holes should be 13 mils in diameter. Keep them small so that solder wicking through the holes is not a problem during reflow.
4. Additional vias may be placed anywhere along the thermal plane outside of the thermal pad area. This helps dissipate the heat generated by the TLC07x IC. These additional vias may be larger than the 13-mil diameter vias directly under the thermal pad. They can be larger because they are not in the thermal pad area to be soldered so that wicking is not a problem.
5. Connect all holes to the internal ground plane that is the same potential as the device GND pin.
6. When connecting these holes to the ground plane, do not use the typical web or spoke via connection methodology. Web connections have a high thermal resistance connection that is useful for slowing the heat transfer during soldering operations. This makes the soldering of vias that have plane connections easier. In this application, however, low thermal resistance is desired for the most efficient heat transfer. Therefore, the holes under the TLC07x PowerPAD package should make their connection to the internal ground plane with a complete connection around the entire circumference of the plated-through hole.
7. The top-side solder mask should leave the terminals of the package and the thermal pad area with its five holes (dual) or nine holes (quad) exposed. The bottom-side solder mask should cover the five or nine holes of the thermal pad area. This prevents solder from being pulled away from the thermal pad area during the reflow process.
8. Apply solder paste to the exposed thermal pad area and all of the IC terminals.
9. With these preparatory steps in place, the TLC07x IC is simply placed in position and run through the solder reflow operation as any standard surface-mount component. This results in a part that is properly installed.

For a given θ_{JA} , the maximum power dissipation is shown in Figure 54 and is calculated by the following formula:

$$P_D = \left(\frac{T_{MAX} - T_A}{\theta_{JA}} \right)$$

Where:

P_D = Maximum power dissipation of TLC07x IC (watts)

T_{MAX} = Absolute maximum junction temperature (150°C)

T_A = Free-ambient air temperature (°C)

θ_{JA} = $\theta_{JC} + \theta_{CA}$

θ_{JC} = Thermal coefficient from junction to case

θ_{CA} = Thermal coefficient from case to ambient air (°C/W)

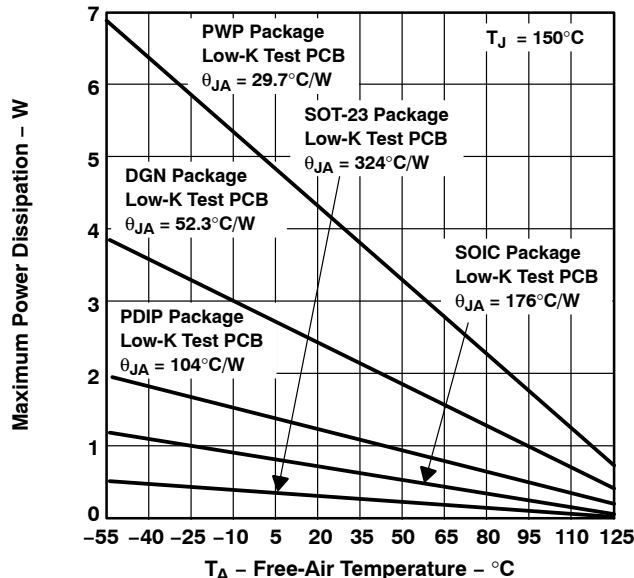
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APPLICATION INFORMATION

general PowerPAD design considerations (continued)

MAXIMUM POWER DISSIPATION
VS
FREE-AIR TEMPERATURE



NOTE A: Results are with no air flow and using JEDEC Standard Low-K test PCB.

Figure 53. Maximum Power Dissipation vs Free-Air Temperature

The next consideration is the package constraints. The two sources of heat within an amplifier are quiescent power and output power. The designer should never forget about the quiescent heat generated within the device, especially multi-amplifier devices. Because these devices have linear output stages (Class A-B), most of the heat dissipation is at low output voltages with high output currents.

The other key factor when dealing with power dissipation is how the devices are mounted on the PCB. The PowerPAD devices are extremely useful for heat dissipation. But, the device should always be soldered to a copper plane to fully use the heat dissipation properties of the PowerPAD. The SOIC package, on the other hand, is highly dependent on how it is mounted on the PCB. As more trace and copper area is placed around the device, θ_{JA} decreases and the heat dissipation capability increases. The currents and voltages shown in these graphs are for the total package. For the dual or quad amplifier packages, the sum of the RMS output currents and voltages should be used to choose the proper package.

TLC070, TLC071, TLC072, TLC073, TLC074, TLC075, TLC07xA FAMILY OF WIDE-BANDWIDTH HIGH-OUTPUT-DRIVE SINGLE SUPPLY OPERATIONAL AMPLIFIERS

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APPLICATION INFORMATION

macromodel information

Macromodel information provided was derived using Microsim *Parts*[™], the model generation software used with Microsim *PSpice*[™]. The Boyle macromodel (see Note 1) and subcircuit in Figure 55 are generated using the TLC07x typical electrical and operating characteristics at $T_A = 25^\circ\text{C}$. Using this information, output simulations of the following key parameters can be generated to a tolerance of 20% (in most cases):

- Maximum positive output voltage swing
- Maximum negative output voltage swing
- Slew rate
- Quiescent power dissipation
- Input bias current
- Open-loop voltage amplification
- Unity-gain frequency
- Common-mode rejection ratio
- Phase margin
- DC output resistance
- AC output resistance
- Short-circuit output current limit

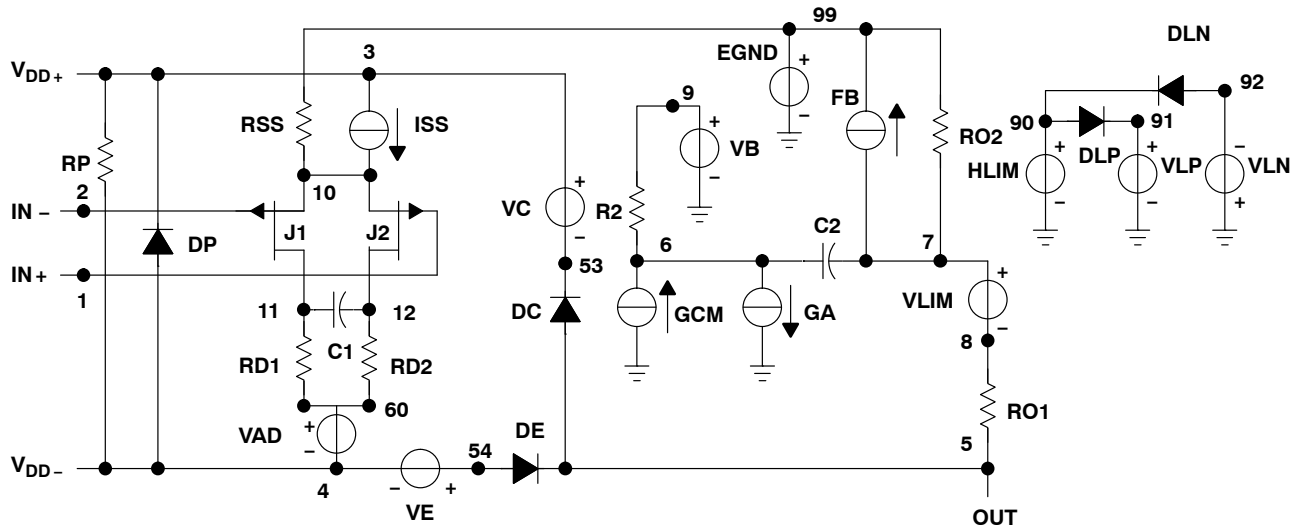
NOTE 2: G. R. Boyle, B. M. Cohn, D. O. Pederson, and J. E. Solomon, "Macromodeling of Integrated Circuit Operational Amplifiers," *IEEE Journal of Solid-State Circuits*, SC-9, 353 (1974).

PSpice and *Parts* are trademarks of MicroSim Corporation.

TLC070, TLC071, TLC072, TLC073, TLC074, TLC075, TLC07xA FAMILY OF WIDE-BANDWIDTH HIGH-OUTPUT-DRIVE SINGLE SUPPLY OPERATIONAL AMPLIFIERS

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APPLICATION INFORMATION



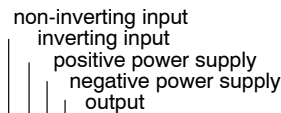
*DEVICE=TLC07X_5V, OPAMP, PJF, INT

* TLC07X – 5V operational amplifier "macromodel" subcircuit

* created using Parts release 8.0 on 12/16/99 at 08:38

* Parts is a MicroSim product.

* connections:



.subckt TLC07X_5V 1 2 3 4 5

```

*
c1 11 12 4.8697E-12
c2 6 7 8.0000E-12
css 10 99 4.0063E-12
dc 5 53 dy
de 54 5 dy
dip 90 91 dx
dln 92 90 dx
dp 4 3 dx
egnd 99 0 poly(2) (3,0) (4,0) 0 .5 .5
fb 7 99 poly(5) vb vc ve vlp vln 0 6.9132E6 -1E3 1E3
6E6 -6E6

```

```

ga 6 0 11 12 457.42E-6
gcm 0 6 10 99 1.1293E-6
iss 3 10 dc 183.67E-6
ioff 0 6 dc .806E-6
hlim 90 0 vlim 1K
j1 11 2 10 jx1
j2 12 1 10 jx2
r2 6 9 100.00E3
rd1 4 11 2.1862E3
rd2 4 12 2.1862E3
ro1 8 5 10
ro2 7 99 10
rp 3 4 2.4728E3
rss 10 99 1.0889E6
vb 9 0 dc 0
vc 3 53 dc 1.5410
ve 54 4 dc .84403
vlim 7 8 dc 0
vlp 91 0 dc 119
vln 0 92 dc 119
.model dx D(Is=800.00E-18)
.model dy D(Is=800.00E-18 Rs=1m Cjo=10p)
.model jx1 PJF(Is=117.50E-15 Beta=1.1391E-3 Vto=-1)
.model jx2 PJF(Is=117.50E-15 Beta=1.1391E-3 Vto=-1)
.ends

```

Figure 54. Boyle Macromodel and Subcircuit

PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead finish/ Ball material (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
TLC070AIP	ACTIVE	PDIP	P	8	50	RoHS & Green	NIPDAU	N / A for Pkg Type	-40 to 125	TLC070AI	Samples
TLC070CD	ACTIVE	SOIC	D	8	75	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	0 to 70	C070C	Samples
TLC070CDR	ACTIVE	SOIC	D	8	2500	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	0 to 70	C070C	Samples
TLC070IDR	ACTIVE	SOIC	D	8	2500	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	C070I	Samples
TLC070IP	ACTIVE	PDIP	P	8	50	RoHS & Green	NIPDAU	N / A for Pkg Type	-40 to 125	TLC070I	Samples
TLC071AID	ACTIVE	SOIC	D	8	75	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	C071AI	Samples
TLC071AIP	ACTIVE	PDIP	P	8	50	RoHS & Green	NIPDAU	N / A for Pkg Type	-40 to 125	TLC071AI	Samples
TLC071CD	ACTIVE	SOIC	D	8	75	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	0 to 70	C071C	Samples
TLC071CDG4	ACTIVE	SOIC	D	8	75	TBD	Call TI	Call TI	0 to 70		Samples
TLC071CDGN	ACTIVE	HVSSOP	DGN	8	80	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	0 to 70	ACU	Samples
TLC071CDGNR	ACTIVE	HVSSOP	DGN	8	2500	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	0 to 70	ACU	Samples
TLC071CDR	ACTIVE	SOIC	D	8	2500	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	0 to 70	C071C	Samples
TLC071CP	ACTIVE	PDIP	P	8	50	RoHS & Green	NIPDAU	N / A for Pkg Type	0 to 70	TLC071C	Samples
TLC071ID	ACTIVE	SOIC	D	8	75	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	C071I	Samples
TLC071IDGNR	ACTIVE	HVSSOP	DGN	8	2500	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	ACV	Samples
TLC071IDR	ACTIVE	SOIC	D	8	2500	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	C071I	Samples
TLC071IP	ACTIVE	PDIP	P	8	50	RoHS & Green	NIPDAU	N / A for Pkg Type	-40 to 125	TLC071I	Samples
TLC072AID	ACTIVE	SOIC	D	8	75	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	C072AI	Samples
TLC072AIDG4	ACTIVE	SOIC	D	8	75	TBD	Call TI	Call TI	-40 to 125		Samples
TLC072AIDR	ACTIVE	SOIC	D	8	2500	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	C072AI	Samples

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead finish/ Ball material (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
TLC072AIP	ACTIVE	PDIP	P	8	50	RoHS & Green	NIPDAU	N / A for Pkg Type	-40 to 125	C072AI	Samples
TLC072CD	ACTIVE	SOIC	D	8	75	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	0 to 70	C072C	Samples
TLC072CDG4	ACTIVE	SOIC	D	8	75	TBD	Call TI	Call TI	0 to 70		Samples
TLC072CDGN	ACTIVE	HVSSOP	DGN	8	80	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	0 to 70	ADV	Samples
TLC072CDGNR	ACTIVE	HVSSOP	DGN	8	2500	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	0 to 70	ADV	Samples
TLC072CDGNRG4	ACTIVE	HVSSOP	DGN	8	2500	TBD	Call TI	Call TI	0 to 70		Samples
TLC072CDR	ACTIVE	SOIC	D	8	2500	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	0 to 70	C072C	Samples
TLC072CDRG4	ACTIVE	SOIC	D	8	2500	TBD	Call TI	Call TI	0 to 70		Samples
TLC072CP	ACTIVE	PDIP	P	8	50	RoHS & Green	NIPDAU	N / A for Pkg Type	0 to 70	C072C	Samples
TLC072ID	ACTIVE	SOIC	D	8	75	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	C072I	Samples
TLC072IDGN	ACTIVE	HVSSOP	DGN	8	80	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	ADW	Samples
TLC072IDGNR	ACTIVE	HVSSOP	DGN	8	2500	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	ADW	Samples
TLC072IDR	ACTIVE	SOIC	D	8	2500	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	C072I	Samples
TLC072IP	ACTIVE	PDIP	P	8	50	RoHS & Green	NIPDAU	N / A for Pkg Type	-40 to 125	C072I	Samples
TLC073AID	ACTIVE	SOIC	D	14	50	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	CO73AI	Samples
TLC073CDGQ	ACTIVE	HVSSOP	DGQ	10	80	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	0 to 70	ADX	Samples
TLC073CDR	ACTIVE	SOIC	D	14	2500	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	0 to 70	C073C	Samples
TLC073IDGQ	ACTIVE	HVSSOP	DGQ	10	80	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	ADY	Samples
TLC073IDGQR	ACTIVE	HVSSOP	DGQ	10	2500	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	ADY	Samples
TLC073IN	ACTIVE	PDIP	N	14	25	RoHS & Green	NIPDAU	N / A for Pkg Type	-40 to 125	C073I	Samples
TLC074AID	ACTIVE	SOIC	D	14	50	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	TLC074AI	Samples

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead finish/ Ball material (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
TLC074AIDR	ACTIVE	SOIC	D	14	2500	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	TLC074AI	Samples
TLC074AIN	ACTIVE	PDIP	N	14	25	RoHS & Green	NIPDAU	N / A for Pkg Type	-40 to 125	TLC074AI	Samples
TLC074CD	ACTIVE	SOIC	D	14	50	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	0 to 70	TLC074C	Samples
TLC074CDR	ACTIVE	SOIC	D	14	2500	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	0 to 70	TLC074C	Samples
TLC074CN	ACTIVE	PDIP	N	14	25	RoHS & Green	NIPDAU	N / A for Pkg Type	0 to 70	TLC074C	Samples
TLC074CPWP	ACTIVE	HTSSOP	PWP	20	70	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	0 to 70	TLC074C	Samples
TLC074CPWPR	ACTIVE	HTSSOP	PWP	20	2000	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	0 to 70	TLC074C	Samples
TLC074ID	ACTIVE	SOIC	D	14	50	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	TLC074I	Samples
TLC074IDR	ACTIVE	SOIC	D	14	2500	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	TLC074I	Samples
TLC074IN	ACTIVE	PDIP	N	14	25	RoHS & Green	NIPDAU	N / A for Pkg Type	-40 to 125	TLC074I	Samples
TLC074IPWP	ACTIVE	HTSSOP	PWP	20	70	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	TLC074I	Samples
TLC075AID	ACTIVE	SOIC	D	16	40	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	TLC075AI	Samples
TLC075AIN	ACTIVE	PDIP	N	16	25	RoHS & Green	NIPDAU	N / A for Pkg Type	-40 to 125	TLC075AI	Samples
TLC075AIPWP	ACTIVE	HTSSOP	PWP	20	70	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	TLC075AI	Samples
TLC075CPWP	ACTIVE	HTSSOP	PWP	20	70	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	0 to 70	TLC075C	Samples
TLC075IPWP	ACTIVE	HTSSOP	PWP	20	70	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	TLC075I	Samples

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSELETE: TI has discontinued the production of the device.

(2) **RoHS:** TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

Green: TI defines "Green" to mean the content of Chlorine (Cl) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of ≤ 1000 ppm threshold. Antimony trioxide based flame retardants must also meet the ≤ 1000 ppm threshold requirement.

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

(6) Lead finish/Ball material - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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OTHER QUALIFIED VERSIONS OF TLC072 :

- Automotive : [TLC072-Q1](#)

NOTE: Qualified Version Definitions:

- Automotive - Q100 devices qualified for high-reliability automotive applications targeting zero defects

TAPE AND REEL INFORMATION

QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE


*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TLC070CDR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
TLC070IDR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
TLC071CDGNR	HVSSOP	DGN	8	2500	330.0	12.4	5.3	3.4	1.4	8.0	12.0	Q1
TLC071CDR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
TLC071IDGNR	HVSSOP	DGN	8	2500	330.0	12.4	5.3	3.4	1.4	8.0	12.0	Q1
TLC071IDR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
TLC072AIDR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
TLC072CDGNR	HVSSOP	DGN	8	2500	330.0	12.4	5.3	3.4	1.4	8.0	12.0	Q1
TLC072CDR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
TLC072IDGNR	HVSSOP	DGN	8	2500	330.0	12.4	5.3	3.4	1.4	8.0	12.0	Q1
TLC072IDR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
TLC073CDR	SOIC	D	14	2500	330.0	16.4	6.5	9.0	2.1	8.0	16.0	Q1
TLC073IDGQR	HVSSOP	DGQ	10	2500	330.0	12.4	5.3	3.4	1.4	8.0	12.0	Q1
TLC074AIDR	SOIC	D	14	2500	330.0	16.4	6.5	9.0	2.1	8.0	16.0	Q1
TLC074CDR	SOIC	D	14	2500	330.0	16.4	6.5	9.0	2.1	8.0	16.0	Q1
TLC074CPWPR	HTSSOP	PWP	20	2000	330.0	16.4	6.95	7.1	1.6	8.0	16.0	Q1

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TLC074IDR	SOIC	D	14	2500	330.0	16.4	6.5	9.0	2.1	8.0	16.0	Q1

TAPE AND REEL BOX DIMENSIONS


*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TLC070CDR	SOIC	D	8	2500	340.5	338.1	20.6
TLC070IDR	SOIC	D	8	2500	340.5	338.1	20.6
TLC071CDGNR	HVSSOP	DGN	8	2500	358.0	335.0	35.0
TLC071CDR	SOIC	D	8	2500	353.0	353.0	32.0
TLC071IDGNR	HVSSOP	DGN	8	2500	358.0	335.0	35.0
TLC071IDR	SOIC	D	8	2500	353.0	353.0	32.0
TLC072AIDR	SOIC	D	8	2500	353.0	353.0	32.0
TLC072CDGNR	HVSSOP	DGN	8	2500	358.0	335.0	35.0
TLC072CDR	SOIC	D	8	2500	353.0	353.0	32.0
TLC072IDGNR	HVSSOP	DGN	8	2500	358.0	335.0	35.0
TLC072IDR	SOIC	D	8	2500	353.0	353.0	32.0
TLC073CDR	SOIC	D	14	2500	350.0	350.0	43.0
TLC073IDGQR	HVSSOP	DGQ	10	2500	358.0	335.0	35.0
TLC074AIDR	SOIC	D	14	2500	350.0	350.0	43.0
TLC074CDR	SOIC	D	14	2500	353.0	353.0	32.0
TLC074CPWPR	HTSSOP	PWP	20	2000	350.0	350.0	43.0
TLC074IDR	SOIC	D	14	2500	350.0	350.0	43.0

TUBE


*All dimensions are nominal

Device	Package Name	Package Type	Pins	SPQ	L (mm)	W (mm)	T (µm)	B (mm)
TLC070AIP	P	PDIP	8	50	506	13.97	11230	4.32
TLC070CD	D	SOIC	8	75	505.46	6.76	3810	4
TLC070CD	D	SOIC	8	75	507	8	3940	4.32
TLC070IP	P	PDIP	8	50	506	13.97	11230	4.32
TLC071AID	D	SOIC	8	75	505.46	6.76	3810	4
TLC071AID	D	SOIC	8	75	507	8	3940	4.32
TLC071AIP	P	PDIP	8	50	506	13.97	11230	4.32
TLC071CD	D	SOIC	8	75	505.46	6.76	3810	4
TLC071CD	D	SOIC	8	75	507	8	3940	4.32
TLC071CP	P	PDIP	8	50	506	13.97	11230	4.32
TLC071ID	D	SOIC	8	75	507	8	3940	4.32
TLC071ID	D	SOIC	8	75	505.46	6.76	3810	4
TLC071IP	P	PDIP	8	50	506	13.97	11230	4.32
TLC072AID	D	SOIC	8	75	507	8	3940	4.32
TLC072AID	D	SOIC	8	75	505.46	6.76	3810	4
TLC072AIP	P	PDIP	8	50	506	13.97	11230	4.32
TLC072CD	D	SOIC	8	75	505.46	6.76	3810	4
TLC072CD	D	SOIC	8	75	507	8	3940	4.32
TLC072CP	P	PDIP	8	50	506	13.97	11230	4.32
TLC072ID	D	SOIC	8	75	507	8	3940	4.32
TLC072ID	D	SOIC	8	75	505.46	6.76	3810	4
TLC072IP	P	PDIP	8	50	506	13.97	11230	4.32
TLC073AID	D	SOIC	14	50	505.46	6.76	3810	4
TLC073IN	N	PDIP	14	25	506	13.97	11230	4.32
TLC074AID	D	SOIC	14	50	505.46	6.76	3810	4
TLC074AIN	N	PDIP	14	25	506	13.97	11230	4.32
TLC074CD	D	SOIC	14	50	507	8	3940	4.32
TLC074CD	D	SOIC	14	50	505.46	6.76	3810	4
TLC074CN	N	PDIP	14	25	506	13.97	11230	4.32

Device	Package Name	Package Type	Pins	SPQ	L (mm)	W (mm)	T (μm)	B (mm)
TLC074CPWP	PWP	HTSSOP	20	70	530	10.2	3600	3.5
TLC074ID	D	SOIC	14	50	505.46	6.76	3810	4
TLC074IN	N	PDIP	14	25	506	13.97	11230	4.32
TLC074IPWP	PWP	HTSSOP	20	70	530	10.2	3600	3.5
TLC075AID	D	SOIC	16	40	505.46	6.76	3810	4
TLC075AIN	N	PDIP	16	25	506	13.97	11230	4.32
TLC075AIPWP	PWP	HTSSOP	20	70	530	10.2	3600	3.5
TLC075CPWP	PWP	HTSSOP	20	70	530	10.2	3600	3.5
TLC075IPWP	PWP	HTSSOP	20	70	530	10.2	3600	3.5

D0014A



PACKAGE OUTLINE

SOIC - 1.75 mm max height

SMALL OUTLINE INTEGRATED CIRCUIT



NOTES:

1. All linear dimensions are in millimeters. Dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.15 mm, per side.
4. This dimension does not include interlead flash. Interlead flash shall not exceed 0.43 mm, per side.
5. Reference JEDEC registration MS-012, variation AB.

EXAMPLE BOARD LAYOUT

D0014A

SOIC - 1.75 mm max height

SMALL OUTLINE INTEGRATED CIRCUIT



LAND PATTERN EXAMPLE
SCALE:8X



SOLDER MASK DETAILS

4220718/A 09/2016

NOTES: (continued)

- 6. Publication IPC-7351 may have alternate designs.
- 7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.

EXAMPLE STENCIL DESIGN

D0014A

SOIC - 1.75 mm max height

SMALL OUTLINE INTEGRATED CIRCUIT



SOLDER PASTE EXAMPLE
BASED ON 0.125 mm THICK STENCIL
SCALE:8X

4220718/A 09/2016

NOTES: (continued)

8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
9. Board assembly site may have different recommendations for stencil design.

D (R-PDSO-G16)

PLASTIC SMALL OUTLINE



4040047-6/M 06/11

- NOTES:
- A. All linear dimensions are in inches (millimeters).
 - B. This drawing is subject to change without notice.
 -  C. Body length does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.006 (0,15) each side.
 -  D. Body width does not include interlead flash. Interlead flash shall not exceed 0.017 (0,43) each side.
 - E. Reference JEDEC MS-012 variation AC.

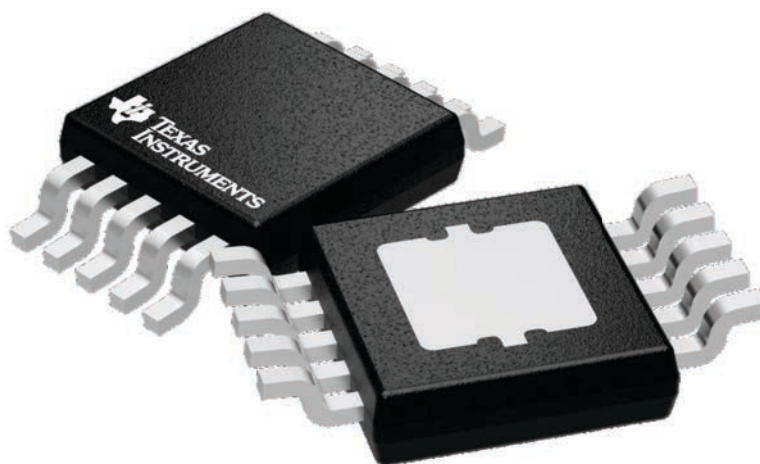
GENERIC PACKAGE VIEW

DGQ 10

PowerPAD™ HVSSOP - 1.1 mm max height

3 x 3, 0.5 mm pitch

PLASTIC SMALL OUTLINE



Images above are just a representation of the package family, actual package may vary.
Refer to the product data sheet for package details.

4224775/A

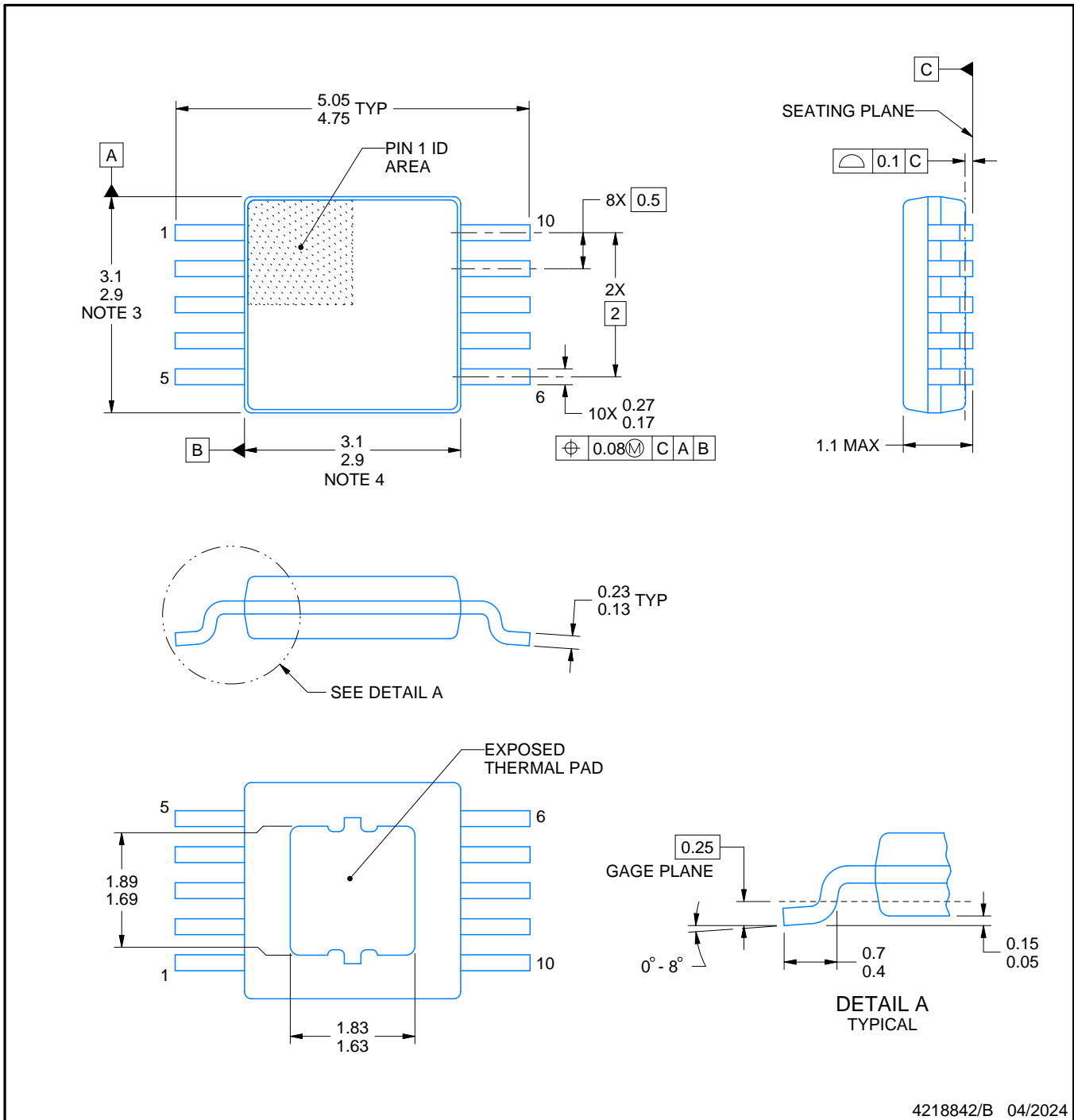
DGQ0010D



PACKAGE OUTLINE

PowerPAD™ - 1.1 mm max height

PLASTIC SMALL OUTLINE



4218842/B 04/2024

PowerPAD is a trademark of Texas Instruments.

NOTES:

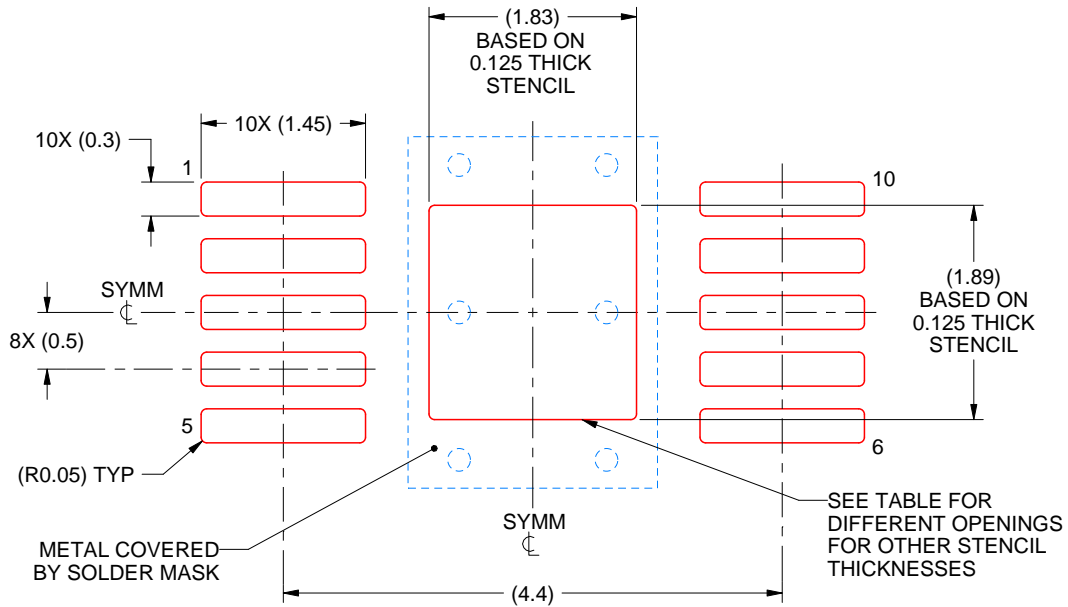
1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.15 mm per side.
4. This dimension does not include interlead flash. Interlead flash shall not exceed 0.25 mm per side.
5. Reference JEDEC registration MO-187, variation BA-T.

EXAMPLE STENCIL DESIGN

DGQ0010D

PowerPAD™ - 1.1 mm max height

PLASTIC SMALL OUTLINE



SOLDER PASTE EXAMPLE
EXPOSED PAD
100% PRINTED SOLDER COVERAGE BY AREA
SCALE:15X

STENCIL THICKNESS	SOLDER STENCIL OPENING
0.1	2.05 X 2.11
0.125	1.83 X 1.89 (SHOWN)
0.150	1.67 X 1.73
0.175	1.55 X 1.60

4218842/B 04/2024

NOTES: (continued)

10. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
11. Board assembly site may have different recommendations for stencil design.

GENERIC PACKAGE VIEW

DGN 8

PowerPAD VSSOP - 1.1 mm max height

3 x 3, 0.65 mm pitch

SMALL OUTLINE PACKAGE

This image is a representation of the package family, actual package may vary.
Refer to the product data sheet for package details.



4225482/A



4225481/A 11/2019

NOTES:

PowerPAD is a trademark of Texas Instruments.

1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.15 mm per side.
4. This dimension does not include interlead flash. Interlead flash shall not exceed 0.25 mm per side.
5. Reference JEDEC registration MO-187.

EXAMPLE BOARD LAYOUT

DGN0008D

PowerPAD™ VSSOP - 1.1 mm max height

SMALL OUTLINE PACKAGE



LAND PATTERN EXAMPLE
EXPOSED METAL SHOWN
SCALE: 15X



4225481/A 11/2019

NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.
7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.
8. Vias are optional depending on application, refer to device data sheet. If any vias are implemented, refer to their locations shown on this view. It is recommended that vias under paste be filled, plugged or tented.
9. Size of metal pad may vary due to creepage requirement.

EXAMPLE STENCIL DESIGN

DGN0008D

PowerPAD™ VSSOP - 1.1 mm max height

SMALL OUTLINE PACKAGE



SOLDER PASTE EXAMPLE
EXPOSED PAD 9:
100% PRINTED SOLDER COVERAGE BY AREA
SCALE: 15X

STENCIL THICKNESS	SOLDER STENCIL OPENING
0.1	1.76 X 2.11
0.125	1.57 X 1.89 (SHOWN)
0.15	1.43 X 1.73
0.175	1.33 X 1.60

4225481/A 11/2019

NOTES: (continued)

10. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
11. Board assembly site may have different recommendations for stencil design.



D0008A

PACKAGE OUTLINE

SOIC - 1.75 mm max height

SMALL OUTLINE INTEGRATED CIRCUIT



4214825/C 02/2019

NOTES:

1. Linear dimensions are in inches [millimeters]. Dimensions in parenthesis are for reference only. Controlling dimensions are in inches. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed .006 [0.15] per side.
4. This dimension does not include interlead flash.
5. Reference JEDEC registration MS-012, variation AA.

EXAMPLE BOARD LAYOUT

D0008A

SOIC - 1.75 mm max height

SMALL OUTLINE INTEGRATED CIRCUIT



LAND PATTERN EXAMPLE
EXPOSED METAL SHOWN
SCALE:8X



SOLDER MASK DETAILS

4214825/C 02/2019

NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.

7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.

EXAMPLE STENCIL DESIGN

D0008A

SOIC - 1.75 mm max height

SMALL OUTLINE INTEGRATED CIRCUIT



SOLDER PASTE EXAMPLE
BASED ON .005 INCH [0.125 MM] THICK STENCIL
SCALE:8X

4214825/C 02/2019

NOTES: (continued)

8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
9. Board assembly site may have different recommendations for stencil design.

THERMAL PAD MECHANICAL DATA

PWP (R-PDSO-G20) PowerPAD™ SMALL PLASTIC OUTLINE

THERMAL INFORMATION

This PowerPAD™ package incorporates an exposed thermal pad that is designed to be attached to a printed circuit board (PCB). The thermal pad must be soldered directly to the PCB. After soldering, the PCB can be used as a heatsink. In addition, through the use of thermal vias, the thermal pad can be attached directly to the appropriate copper plane shown in the electrical schematic for the device, or alternatively, can be attached to a special heatsink structure designed into the PCB. This design optimizes the heat transfer from the integrated circuit (IC).

For additional information on the PowerPAD package and how to take advantage of its heat dissipating abilities, refer to Technical Brief, PowerPAD Thermally Enhanced Package, Texas Instruments Literature No. SLMA002 and Application Brief, PowerPAD Made Easy, Texas Instruments Literature No. SLMA004. Both documents are available at www.ti.com.

The exposed thermal pad dimensions for this package are shown in the following illustration.



Exposed Thermal Pad Dimensions

4206332-15/AO 01/16

NOTE: A. All linear dimensions are in millimeters

 Exposed tie strap features may not be present.

PowerPAD is a trademark of Texas Instruments

PWP (R-PDSO-G20)

PowerPAD™ PLASTIC SMALL OUTLINE



- NOTES:
- A. All linear dimensions are in millimeters.
 - B. This drawing is subject to change without notice.
 - C. Customers should place a note on the circuit board fabrication drawing not to alter the center solder mask defined pad.
 - D. This package is designed to be soldered to a thermal pad on the board. Refer to Technical Brief, PowerPad Thermally Enhanced Package, Texas Instruments Literature No. SLMA002, SLMA004, and also the Product Data Sheets for specific thermal information, via requirements, and recommended board layout. These documents are available at www.ti.com <<http://www.ti.com>>. Publication IPC-7351 is recommended for alternate designs.
 - E. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Example stencil design based on a 50% volumetric metal load solder paste. Refer to IPC-7525 for other stencil recommendations.
 - F. Customers should contact their board fabrication site for solder mask tolerances between and around signal pads.

P (R-PDIP-T8)

PLASTIC DUAL-IN-LINE PACKAGE



- NOTES:
- A. All linear dimensions are in inches (millimeters).
 - B. This drawing is subject to change without notice.
 - C. Falls within JEDEC MS-001 variation BA.

N (R-PDIP-T**)

PLASTIC DUAL-IN-LINE PACKAGE

16 PINS SHOWN



- NOTES: A. All linear dimensions are in inches (millimeters).
 B. This drawing is subject to change without notice.
 (C) Falls within JEDEC MS-001, except 18 and 20 pin minimum body length (Dim A).
 (D) The 20 pin end lead shoulder width is a vendor option, either half or full width.

4040049/E 12/2002

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