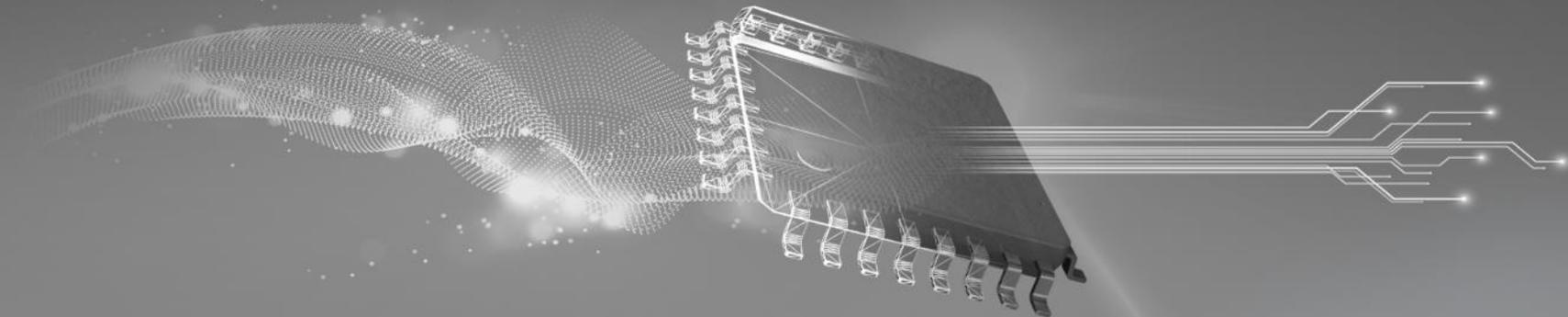


TI TECH DAYS



Design Challenges for Automotive USB Type C® in CarPlay Application

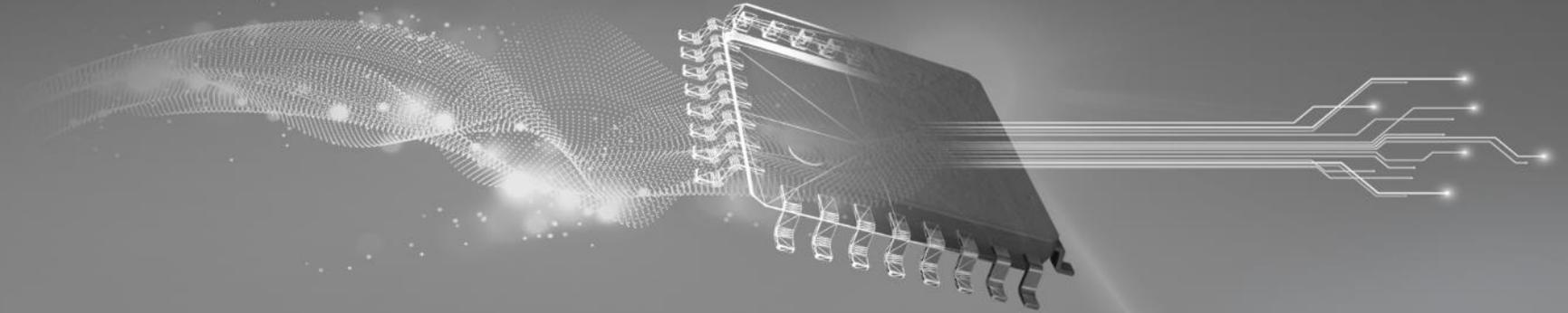
Michael Tan

TI-APP-PSIL

Agenda

- **What is CarPlay/MFi and what is USB's role in the application?**
- **How to pass MFi VBUS requirement with TPS2583x/4x**
 - TPS2583X/4x introduction
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 - Cable compensation introduction
 - Cable compensation design to pass MFi
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 - How to pass MFi OCP with TPS2583x/4x
- **How to improve eye diagram performance with TPS2583x/4x**
 - Eye diagram introduction and challenges in long cables
 - How to improve eye diagram performance with TPS2583x/4x
- **How to improve EMI with TPS2583x/4x**
- **Q&A**

TI TECH DAYS



What is CarPlay/MFi and what is USB's role in the application?

On-the-Go (OTG)

- On-the-go:
 - Allows USB **devices** to act as **host**
 - In automotive: Interoperability between automotive infotainment systems (head unit) and USB devices (phone)
 - Phone becomes the host and control phone App on the Head Unit
- Major players:
 - Apple CarPlay, Google Android Auto,

What is Apple CarPlay?

* **“Apple CarPlay** is a smarter, safer way to use your iPhone while you drive. With an all-new CarPlay Dashboard that helps you with *tunes, turns, and Siri suggestions*, a *Calendar* app that lets you view and manage your day, and *maps* for navigation.”

Apple Carplay will *swap the host* between the *Head Unit* and *Apple phone*, you can control the App on the actual Head Unit.

*Text source : <https://www.apple.com/ios/carplay/>

Apple CarPlay handshake process

1. Phone connect to head unit and is recognized by head unit
2. If CarPlay feature enabled in phone, the head unit will send some packets indicating made for iPhone (MFi) information.
3. Head unit sends role switch request
4. If phone supports CarPlay, it will disconnect from the D+ and D-
5. Head unit turns on and pull up D+ (per OTG supplement)
6. Head unit waits for Apple device to enumerate, communication success

Apple MFi



Apple MFi ----Made For iPhone/iPod/iPad

It is a licensing program for developers of hardware and software peripherals that work with Apple's iPod, iPad and iPhone.

MFi has strict requirement for USB port voltage, current and bandwidth.

What is USB's role in the Apple CarPlay/MFi?

1. USB port VBUS voltage range
2. USB Port bandwidth and signal integrity
3. USB Port Overcurrent and short circuit protection



What is USB's role in the Apple CarPlay/MFi?

- USB port VBUS voltage range

Power for non-lightning accessories			
Supply Rating	Min Output	Max Output	
1 A	4.90	5.25	V
2.1 A delivering up to 1 A	4.90	5.25	
2.1 A delivering up to 1 to 2.1 A	4.97	5.25	
2.4 A delivering 1 to 2.4 A	4.97	5.25	

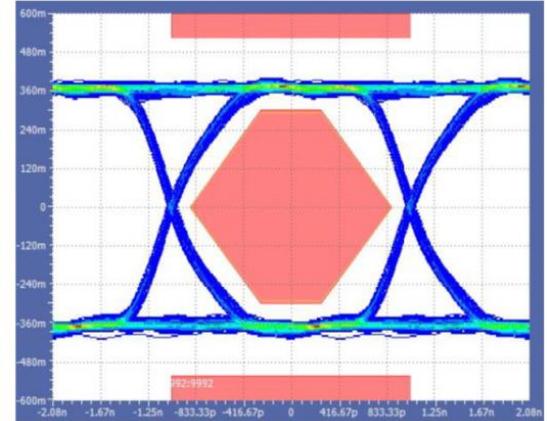
*Data from MFi

USB port voltage range should always within apple MFi VBUS windows.

What is USB's role in the Apple CarPlay/MFi?

- USB Port bandwidth and signal integrity

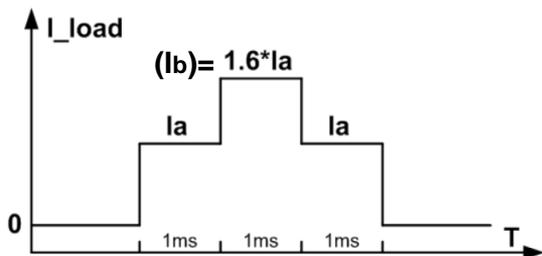
Lightning cables transfer data at **USB 2.0 speeds**.



USB 2.0 need to supports 480Mb/sec data rate transfer, Apple Carplay will require the good signal integrity quality for the USB Port.

What is USB's role in the Apple CarPlay/MFi?

- USB Port overcurrent and short circuit protection



Data from MFi

Threshold	Definition
I_a	Nominal accessory output current (i.e. 1000 mA, 2100 mA, 2400 mA, 3000 mA).
I_b	$I_a + 60\%$.
I_c	Lowest device current draw that will cause accessory output voltage (measured at Lightning Device Power) to drop below 2 V.

Table 12-3 Overcurrent/Short Circuit Protection Behaviors

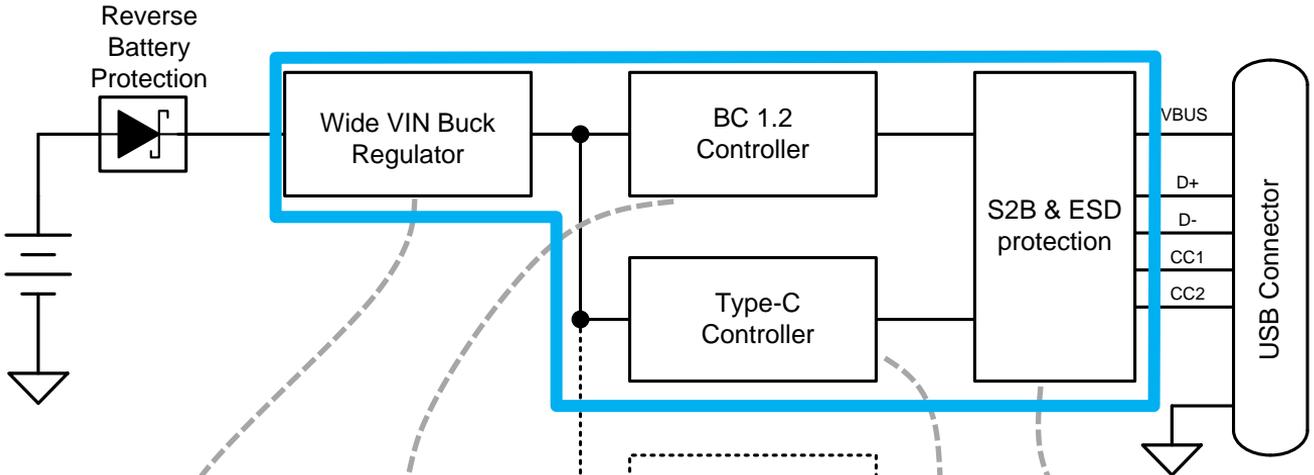
Region	Name	Accessory Behavior
A	Normal Operation	Accessory must not limit or shutdown output current.
B	Overcurrent Transient	Accessory must not shutdown output current. Accessory may limit output current to I_a or higher.
C	Overcurrent	Accessory must shutdown output current.
D	Potential Overcurrent	Accessory may shutdown output current.
E	Potential Short Circuit	If Lightning Device Power voltage drops below 2 V, the accessory may trigger short circuit protection. Accessories must not trigger short circuit protection on device current draw.

The USB port need to sustain 160% I_a for 1mS without VBUS drop to Zero.

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- **How to improve EMI with TPS2583x/4x**
- **Q&A**

Automotive USB power



Buck (36V / 2.5A+)
 LM14030-Q
 LM53635-Q
 LMS3635-Q
 LMR33630-Q

BC 1.2
 TPS2546-Q
 TPS2549/900-Q
 TPS2514A-Q

Type-C
 TPS25810/A-Q
 TUSB319-Q

ESD & S2B
 TPDxS300-Q
 TPDxEyyy-Q



USB All-in-One
 TPS25830/1-Q

TPS25830/1: Single USB Port Controllers

Common Features:

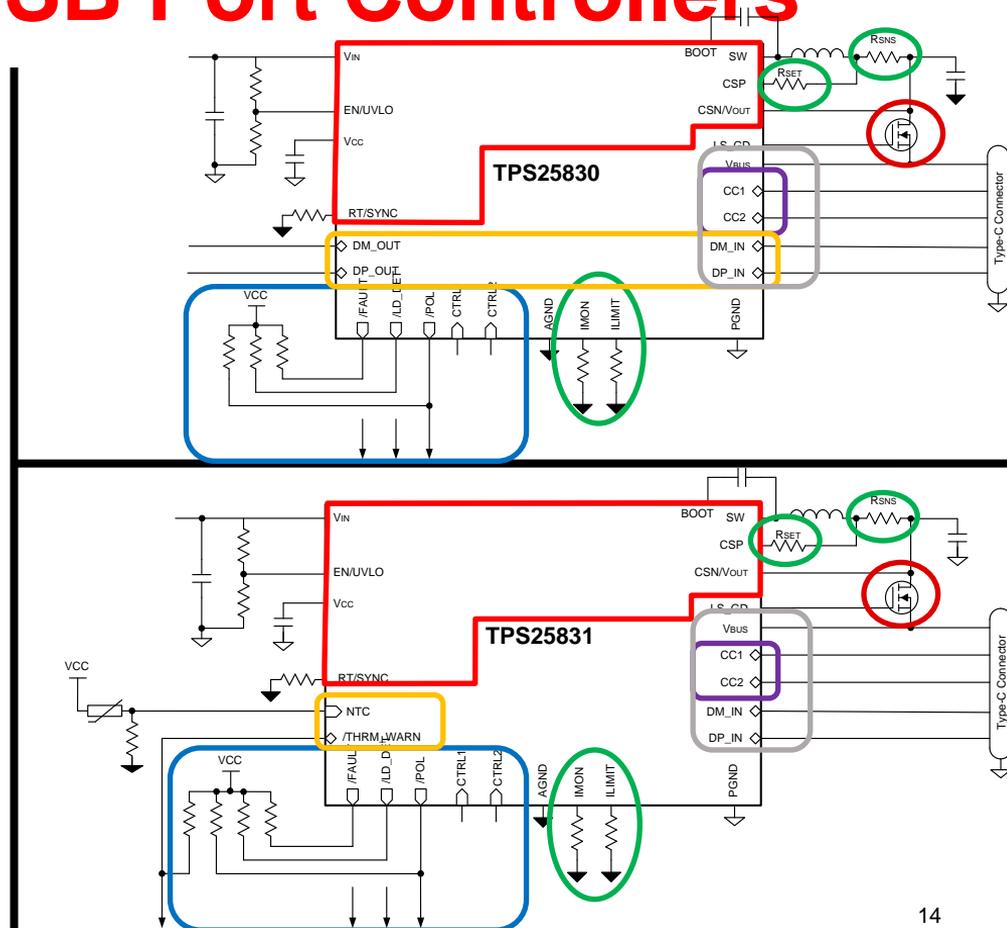
- 6-36V VIN, 3.5A Buck with Spread-Spectrum
- USB Type-C & BC1.2 port control with DCP, CDP, SDP
- $\pm 7\%$ Acc. Current Sense / Current Limit
- Linear VBUS Droop Compensation
- External FET for current limited
- VCONN support: 5V @ 250mA
- Simple Control & Status flags
- Fault protection:
 - VBUS: short to VBAT (OV) or GND (OC)
 - CC: short to VBAT (OV) or GND (OC)
 - Dx: Short to VBAT (OV)
 - IC Over-temperature

TPS25830:

- >800MHz USB HS pass-through switch with IEC61000-4-2 ESD protection

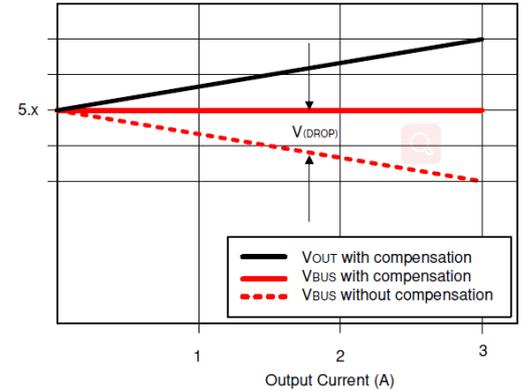
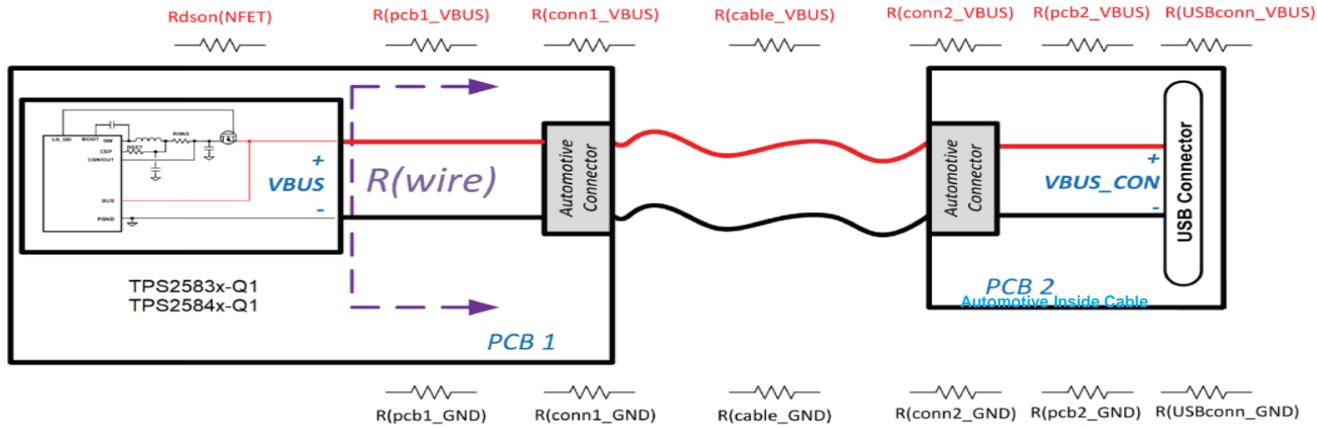
TPS25831:

- Thermistor input and thermal warning flag for monitoring PCB temperature

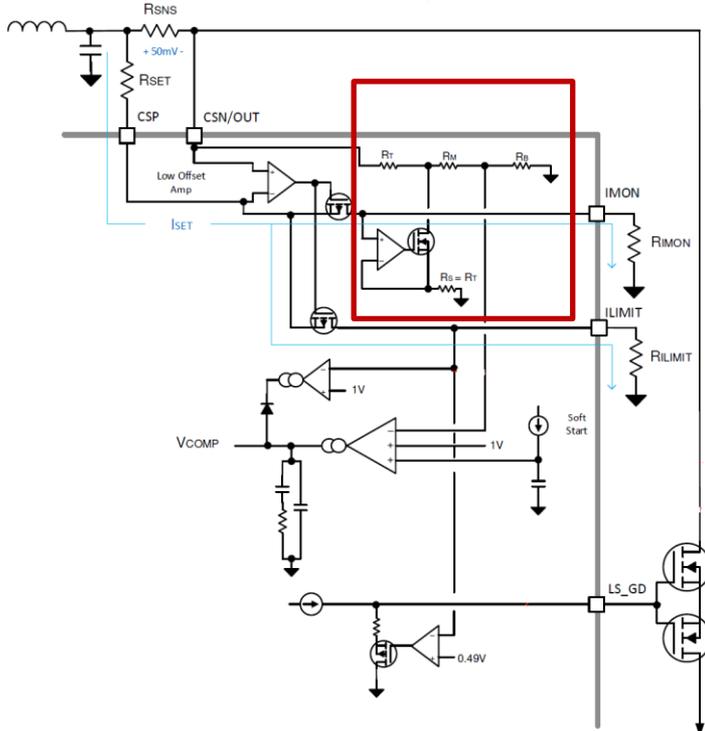


MFI output current and VBUS voltage range

SUPPLY RATING	MINIMUM OUTPUT	MAXIMUM OUTPUT
1 A	4.90 V	5.25 V
2.1 A delivering up to 1.0 A	4.90 V	5.25 V
2.1 A delivering 1.0 to 2.1 A	4.97 V	5.25 V
2.4 A delivering 1.0 to 2.4 A	4.97 V	5.25 V



TPS2583X/4X block diagram for cable compensation



The current that passes the RSET can be calculated by equation:

$$I_{SET} = RSNS \times I_{LOAD} / RSET$$

The voltage on the IMON can be calculated by equation :

$$V_{IMON} = (I_{SET} / 2) \times R_{IMON} = RSNS \times I_{LOAD} \times R_{IMON} / (2 \times RSET)$$

Cable compensation voltage as equation:

$$V_{COMP} = V_{IMON}$$

RIMON can be deduced by equation:

$$R_{IMON} = V_{COMP} \times 2 \times R_{SET} / (I_{LOAD} \times RSNS)$$

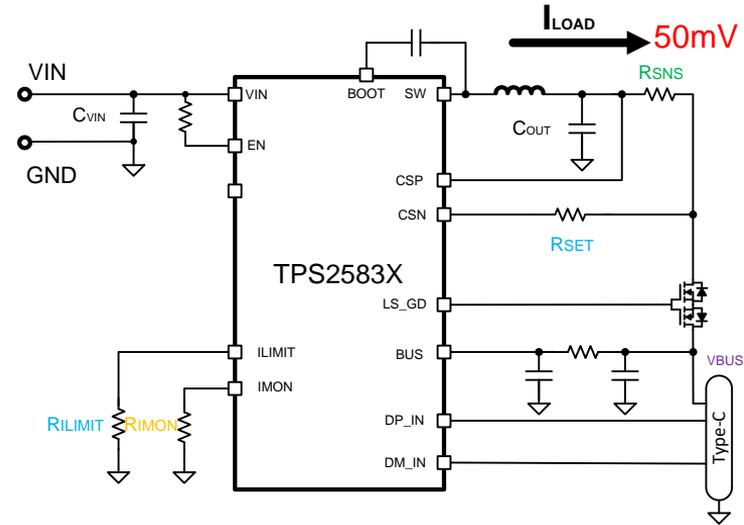
VBUS accuracy estimation for MFi certification

Cable compensation voltage

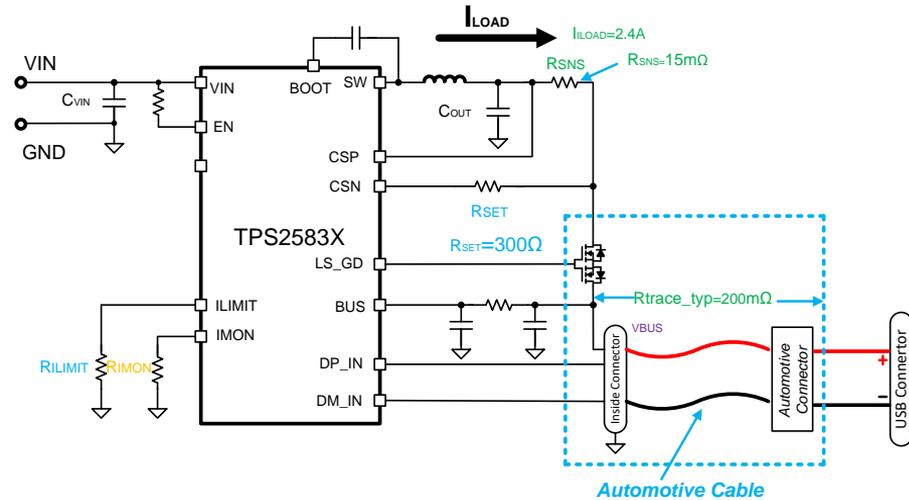
PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT	Compensation Accuracy
V _{IMON}	(VCSP - VCSN) = 46 mV, RSET= 300 Ω, RILIMIT = 13 kΩ, RIMON = 13 kΩ	0.935	1	1.065	V	±6.5%
V _{IMON}	(VCSP - VCSN) = 26 mV, RSET = 300 Ω, RILIMIT = 13 kΩ, RIMON = 13 kΩ	0.435	0.5	0.565	V	±13%

TI Recommends: Choosing the suitable R_{SNS} , which has an approximate 50 mV voltage drop under desired maximum output current.

$$\begin{aligned}
 V_{BUS_typ} &= V_{OUT_typ} + V_{COMP_typ} - V_{DROP_typ} \\
 V_{BUS_min} &= V_{OUT_min} + V_{COMP_min} - V_{DROP_max} \\
 V_{BUS_max} &= V_{OUT_max} + V_{COMP_max} - V_{DROP_min}
 \end{aligned}$$



VBUS cable compensation design example



1. $V_{DROP_typ} = I_{LOAD_typ} \times R_{trace_typ} = 2.4 \text{ A} \times 200 \text{ m}\Omega = 480 \text{ mV}$

2. $I_{LOAD_min} = I_{LOAD_max} = I_{LOAD_typ} = 2.4 \text{ A}; V_{DROP_max} = V_{DROP_min} = V_{DROP_typ} = 480 \text{ mV}.$

3. Choose $R_{SNS} = 15 \text{ m}\Omega$ (choose reason)

4. Choose $R_{SET} = 300 \Omega$

5. $V_{COMP} = V_{DROP} = 480 \text{ mV}$

6. $R_{IMON} = V_{COMP} \times 2 \times R_{SET} / (R_{SNS} \times I_{LOAD}) = 480 \text{ mV} \times 2 \times 300 \Omega / (15 \text{ m}\Omega \times 2.4 \text{ A}) = 8 \text{ K}\Omega$

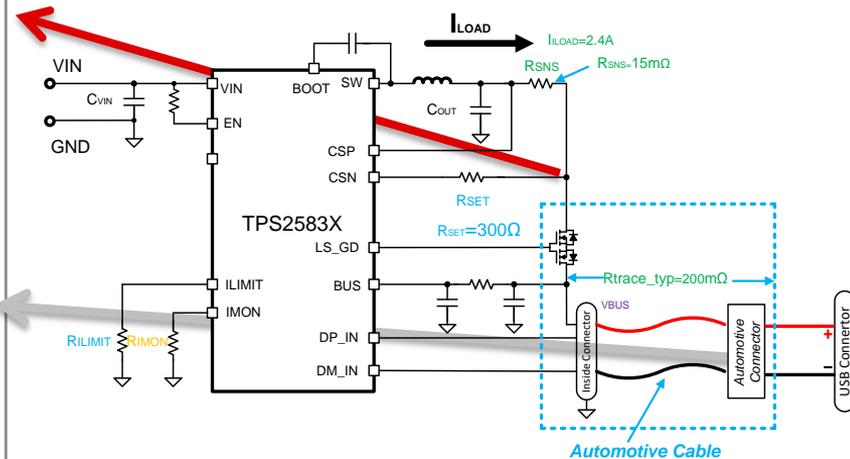
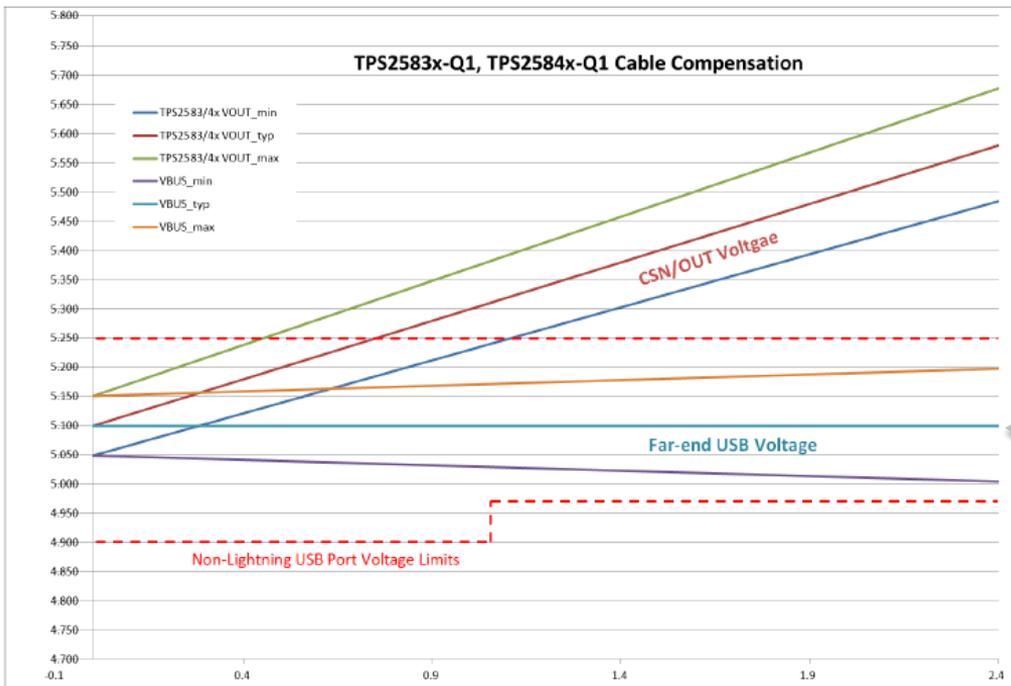
7. Assume the resistor accuracy is $\pm 1\%$, that is $R_{SNS_min/max} = 15 \text{ m}\Omega \times (1 \pm 1\%)$, $R_{IMON_min/max} = 8 \text{ K}\Omega \times (1 \pm 1\%)$, and $R_{SET} = 300 \Omega \times (1 \pm 1\%)$.

8. $V_{COMP_min} = 15 \text{ m}\Omega \times 99\% \times 2.4 \text{ A} \times 8 \text{ K}\Omega \times 99\% / (2 \times 300 \Omega \times 101\%) \times 93.5\% \approx 435.5 \text{ mV}$. Choose $V_{IMON_acc} = \pm 6.5\%$ according.
 $V_{COMP_max} = 15 \text{ m}\Omega \times 101\% \times 2.4 \text{ A} \times 8 \text{ K}\Omega \times 101\% / (2 \times 300 \Omega \times 99\%) \times 106.5\% \approx 526.74 \text{ mV}$

9. $V_{BUS_min} = 5.1 \text{ V} \times 99\% + 435.5 \text{ mV} - 480 \text{ mV} \approx 5.005 \text{ V}$

$V_{BUS_max} = 5.1 \text{ V} \times 101\% + 526.74 \text{ mV} - 480 \text{ mV} \approx 5.198 \text{ V}$

TPS2583x-Q1 far-end USB voltage

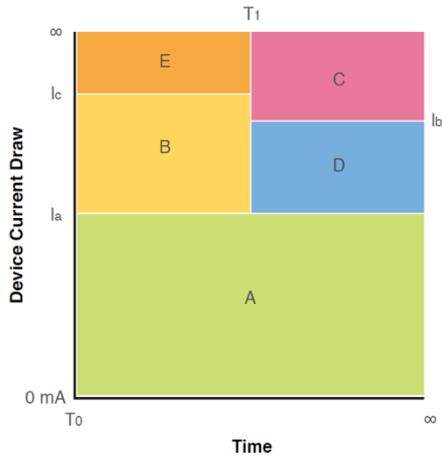


Application Note Link: <http://www.ti.com/lit/an/slvaeb8/slvaeb8.pdf?&ts=1589098739222>

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MFi OCP test standard



standard load waveform for OCP test

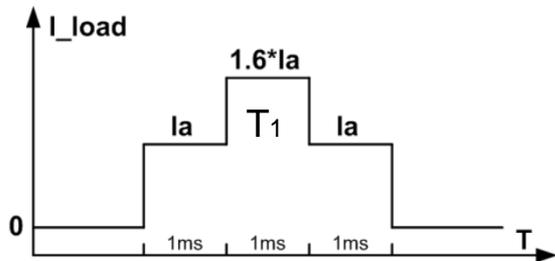


Table 12-2 Overcurrent/Short Circuit Protection Time Thresholds

Threshold	Definition
T_0	Start of any device current draw transient.
T_1	Accessory overcurrent/short circuit deglitch/debounce time, must $\geq T_0 + 1$ ms.

Threshold	Definition
I_a	Nominal accessory output current (i.e. 1000 mA, 2100 mA, 2400 mA, 3000 mA).
I_b	$I_a + 60\%$.
I_c	Lowest device current draw that will cause accessory output voltage (measured at Lightning Device Power) to drop below 2 V.

Table 12-3 Overcurrent/Short Circuit Protection Behaviors

Region	Name	Accessory Behavior
A	Normal Operation	Accessory must not limit or shutdown output current.
B	Overcurrent Transient	Accessory must not shutdown output current. Accessory may limit output current to I_a or higher.
C	Overcurrent	Accessory must shutdown output current.
D	Potential Overcurrent	Accessory may shutdown output current.
E	Potential Short Circuit	If Lightning Device Power voltage drops below 2 V, the accessory may trigger short circuit protection. Accessories must not trigger short circuit protection on device current draw.

Required equipment for MFi test

N6791A 100W Electronic Load



KEYSIGHT N6705C Mainframe



MFi lightning tester fixture

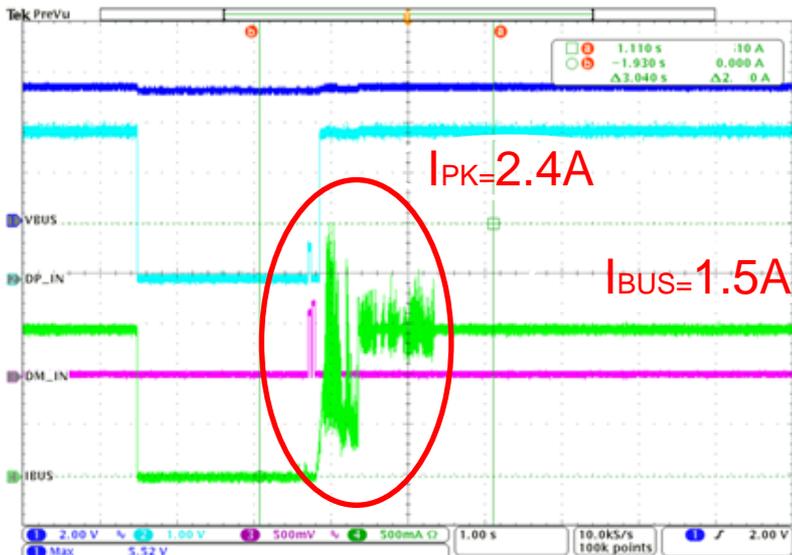
Performance Specifications	N6791A	
Input Ratings:		
Current	0 - 20 A	
Voltage	0 - 60 V	
Maximum Power @ 40 °C	100 W	
Specified Current @ Low Voltage Operation:		
1.6 V	20 A	
1 V	12.5 A	
0.5 V	6.25 A	
0.1 V	1.25 A	
Programming Accuracy:		
Current high range	20 A/40 A	0.04 % + 2.6 mA
Current low range	2 A/4 A	0.04 % + 0.46 mA
Voltage	60 V	0.03 % + 7.2 mV
Resistance high range	8 kΩ	± (0.1 % + 0.0014) S
Resistance medium range	100 Ω	± (0.1 % + 0.014) S
Resistance low range	3 Ω	± (0.1 % + 0.38) S
Power high range	100 W/200 W	0.06 % + 180 mW
Power low range	10 W/20 W	0.06 % + 30 mW
Measurement Accuracy:		
Current high range	20 A/40 A	0.04 % + 2.4 mA
Current low range	2 A/4 A	0.04 % + 0.40 mA
Voltage	60 V	0.03 % + 7.2 mV
Power high range	100 W/200 W	0.06 % + 160 mW
Power low range	10 W/20 W	0.06 % + 25 mW



Apple charging test

CURRENT LIMIT - BUCK REGULATOR PEAK CURRENT LIMIT						
I _{L-SC-HS}	High-side current limit		4.6	5.4	6.2	A
I _{L-SC-LS}	Low-side current limit		3.5	4	4.5	A
I _{L-NEG-LS}	Low-side negative current limit		-3.1	-2.1	-1.3	A

Apple SE charging waveform



Apple MFi OCP Transient Current

I _a (A)	I _b (A)=160%*I _a
1.5	2.4
2.1	3.36
2.4	3.84
3	4.8

Current solution based on TPS2583x/4x-Q1

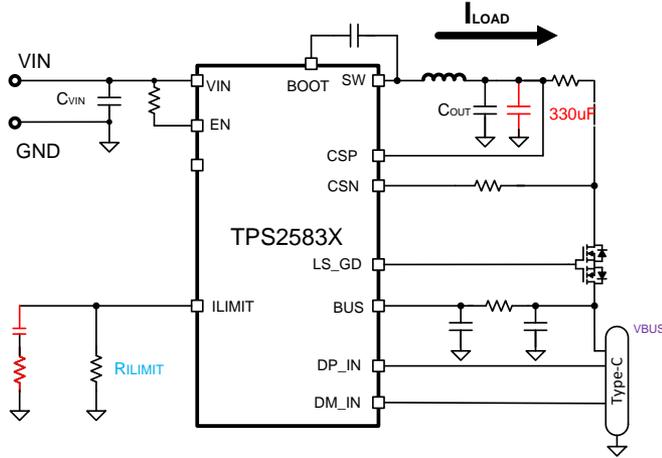
Solution 1. Set current limit $>1.6 \cdot I_a$

- 1.5A port, set current limit $> 2.4A$
- 2.1A port, set current limit $> 3.36A$
- 2.4A port, set current limit $> 3.84A$
- 3A port: set current limit $>4.8A$ and add **330uf** cap on CSP (**High-side current limit=4.6A**)

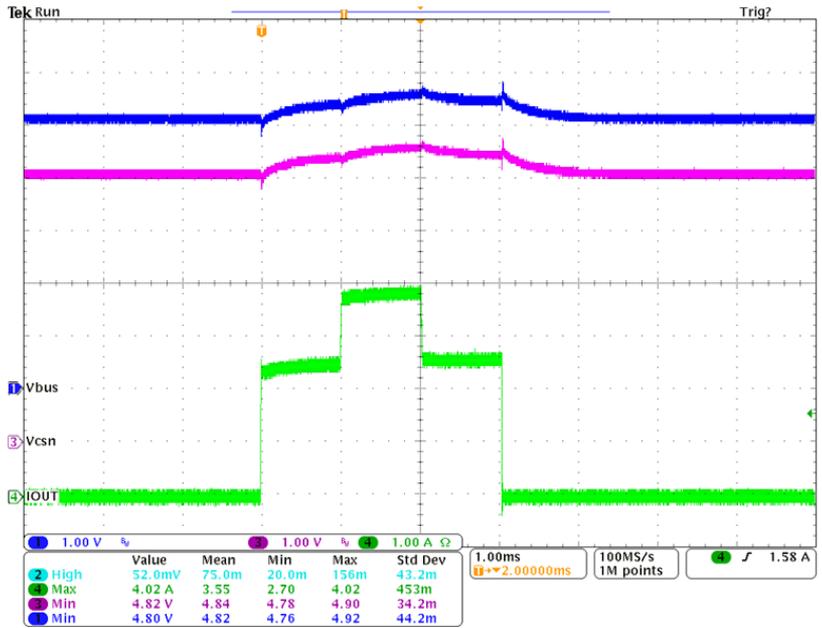
Solution 2. Parallel Cap+R with Rlimit to delay the response of current limit.

Solution and result	With ex FET Rlimit//RC	Without ex FET Rlimit//RC
1.5A port (1.8A current limit)	10.7kΩ//10.7k+82nF Vbus no drop	21.5kΩ//21.5kΩ+82nf Vcsn no drop
2.1A port(2.52A current limit)	7.68kΩ//7.68kΩ+82nf Vbus no drop	15.8kΩ//15.8kΩ+82nf Vcsn no drop
2.4A port(2.88A current limit)	Recommend using Solution 1	13.7kΩ//13.7kΩ+82nf Vcsn no drop
3A port(3.5A currnt limit)	Recommend using Solution 1	11.3kΩ//11.3kΩ+82nf add 330uf on CSP Vcsn drop to 4.02V

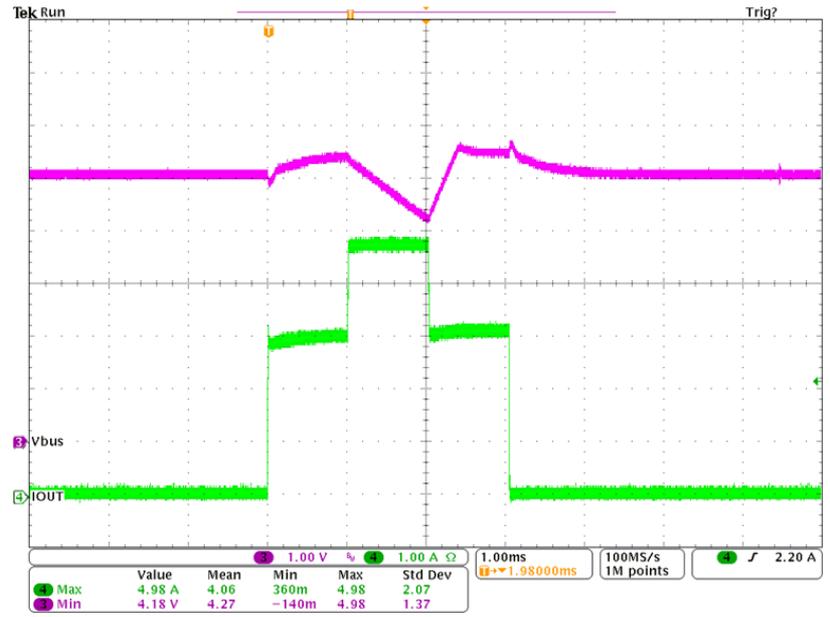
Solution 3 . Next generation products



Solution 1: 2.4A and 3A MFI OCP test results (with external FET)

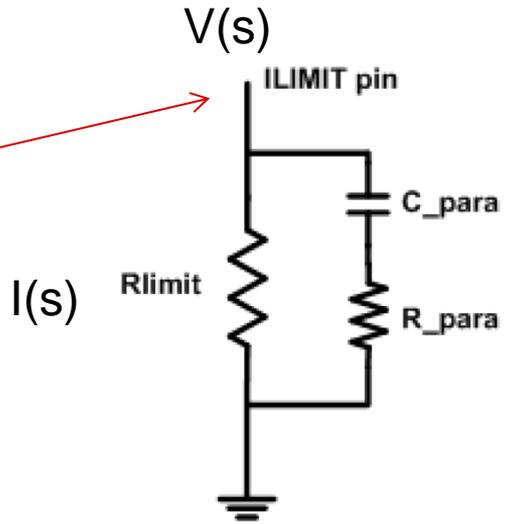
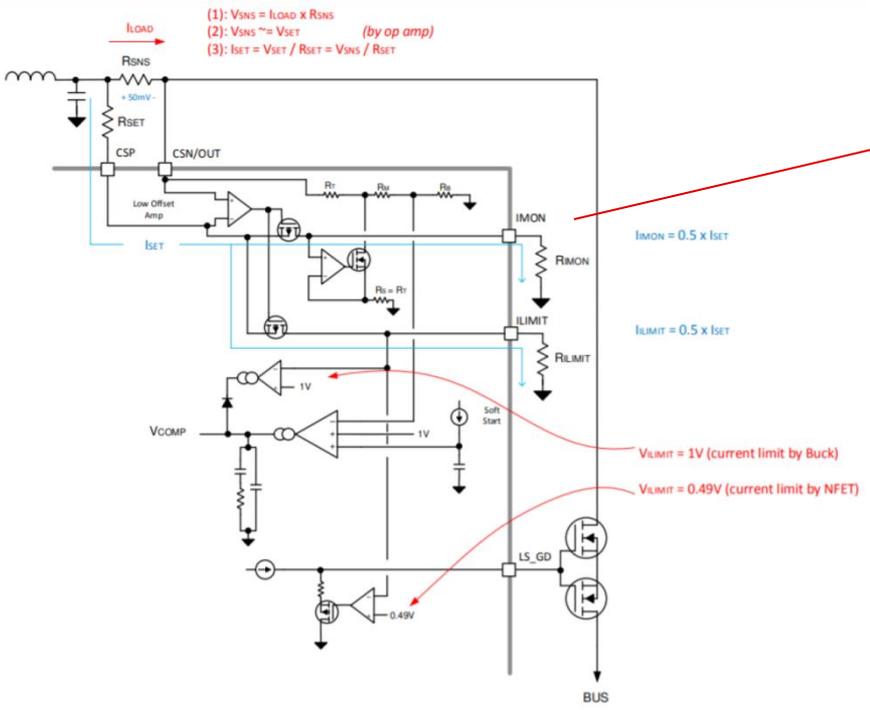


Set 3.96A Current Limit for 2.4A Port



Set 5A Current Limit, and Add 330 μf Cap on CSP for 3A Port

Solution 2: Paralleling RC with Rlimit to delay the current limit response

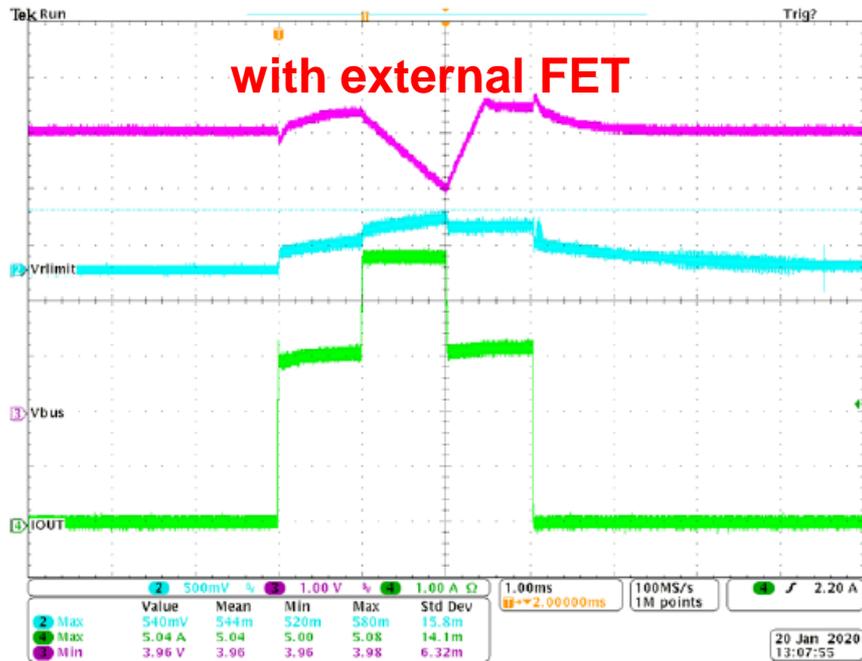


$$\frac{V(s)}{I(s)} = \frac{R_{limit} \times (1 + sC_{para}R_{para})}{1 + sC_{para}(R_{limit} + R_{para})}$$

Current limit function diagrams

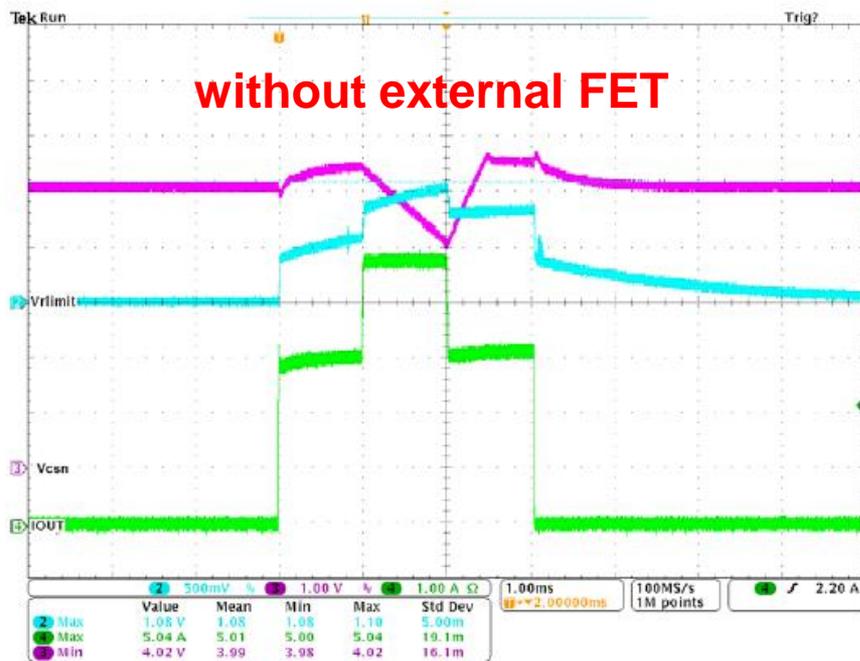
Solution 2 :Paralleling RC with Rlimit to Delay the current limit Response

5.49k Ω Parallel 3.5k Ω +220nF with RC Rlimit



Set 3.5A current limit for 3A port

11.3k Ω Parallel 11.3k Ω + 82 nF with RC Rlimit



Set 3.5A current limit for 3A port

TPS2583x/4x- MFi OCP summary of solution

Solution and result	with external FET-- Rlimit//(Rpara+Cpara)	without external FET-- Rlimit//(Rpara+Cpara)
1.5A port (1.8A current limit)	10.7k Ω /(10.7k Ω + 82nF) Result: Vbus no drop	21.5k Ω /(21.5K Ω + 82nF) Result : Vcsn no drop
2.1A port (2.52A current limit)	7.68k Ω /(7.68k Ω + 82nF) Result: Vbus no drop	15.8k Ω /(15.8k Ω + 82nF) Result: Vcsn no drop
2.4A port (2.88A current limit)	6.65k Ω /(6.65k Ω + 82nF) Result: Vbus no drop	13.7k Ω /(13.7k Ω + 82nF) Result: Vcsn no drop
3A port (3.5A current limit)	5.45k Ω /(3.5k Ω + 220nF) add 330 μ f on CSP Result: Vbus drop to 3.96V	11.3k Ω /(11.3k Ω + 82nF) add 330 μ f on CSP Result: Vcsn drop to 4.02 V

Application Note Link: <http://www.ti.com/lit/an/slvaeq2/slvaeq2.pdf?&ts=1589098829411>

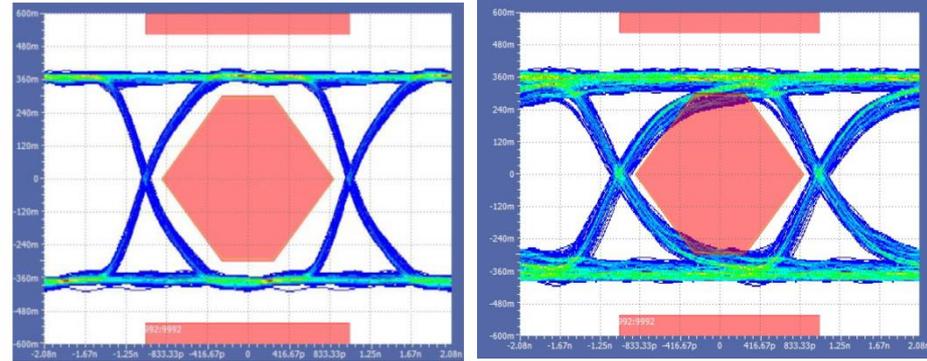
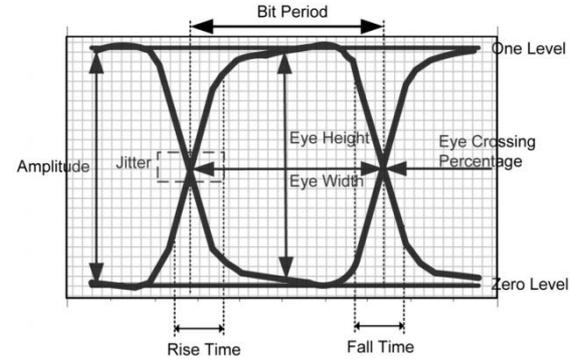
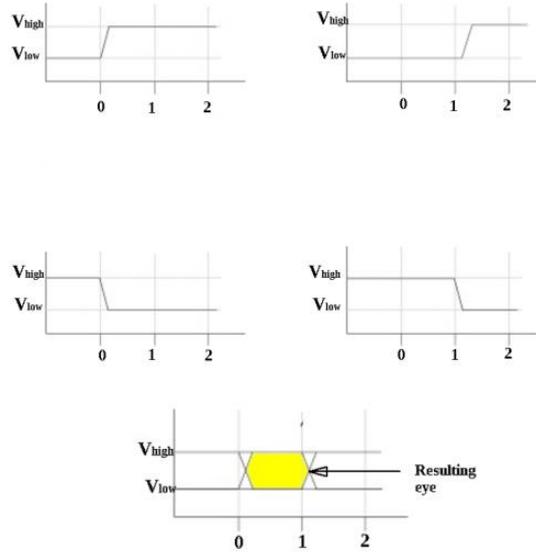
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Eye diagram introduction

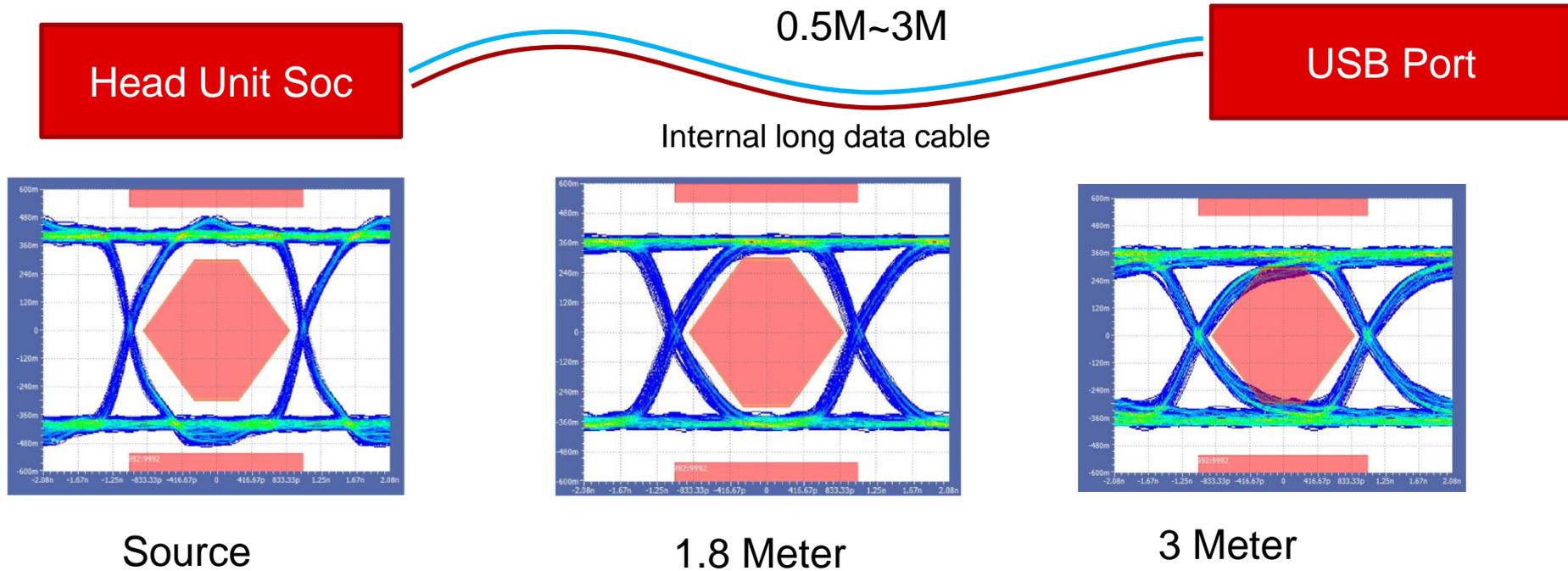
An eye diagram is evaluation method the signal quality in system data communication.

Eye makeup



The bit sequences 011, 001, 100, and 110 are superimposed over one another to obtain the example eye diagram.

Challenge: Eye diagram in long cables



The long cable will create the equivalent capacitance and resistance which will make the system bandwidth become narrow. It is challenging to design good eye diagrams in long data cable applications.

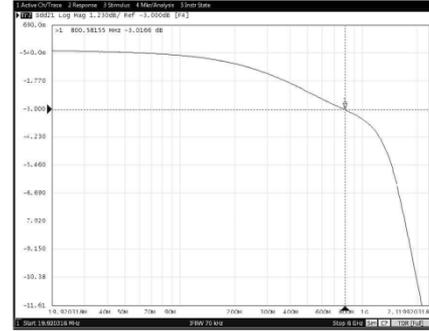
How to improve eye diagram performance with TPS2583x/4x

1. TPS2583x/4x eye diagram performance
2. PCB layout
3. Using LC network
4. Using signal re-driver (TPS25840 + TUSB217)

TPS2583x/4x eye diagram performance



TPS25840Q1EVM-079



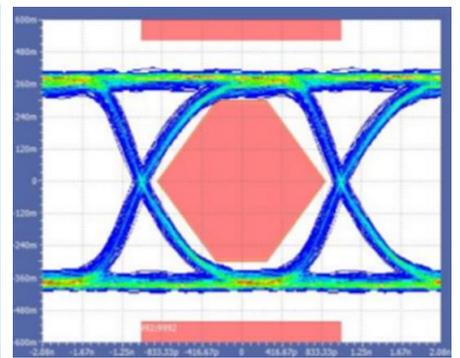
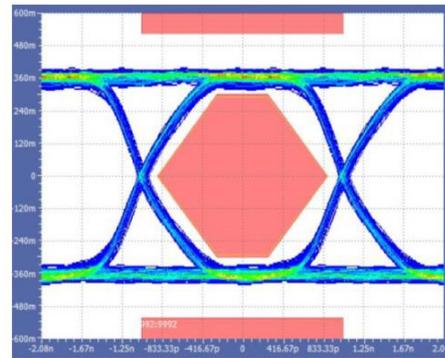
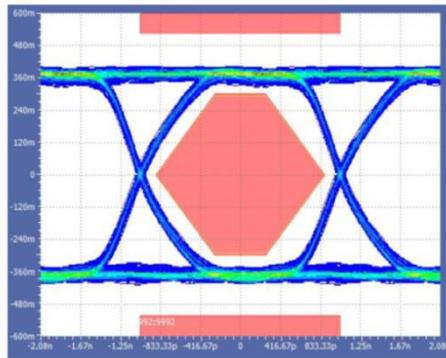
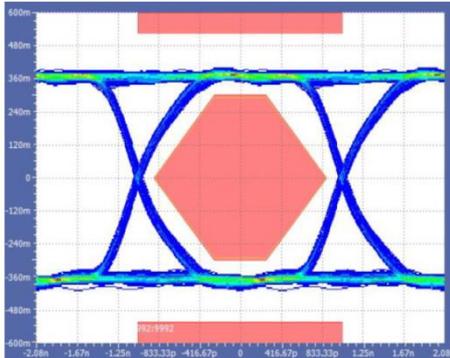
Bandwidth (-3 dB)=800Mhz

30cm Cable+ EVM TPS25840

1m Cable+ EVM TPS25840

1.5m Cable+ EVM TPS25840

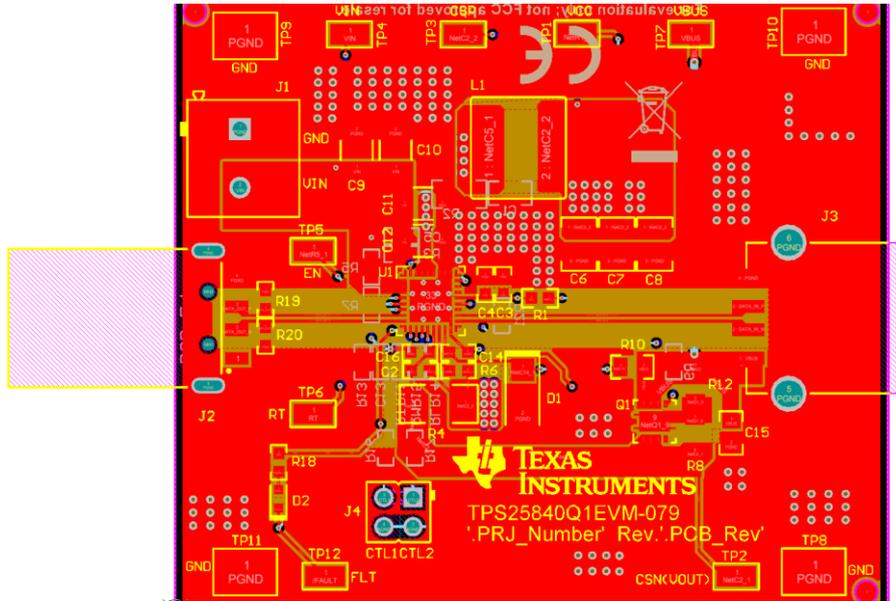
1.8m Cable +EVM TPS25840



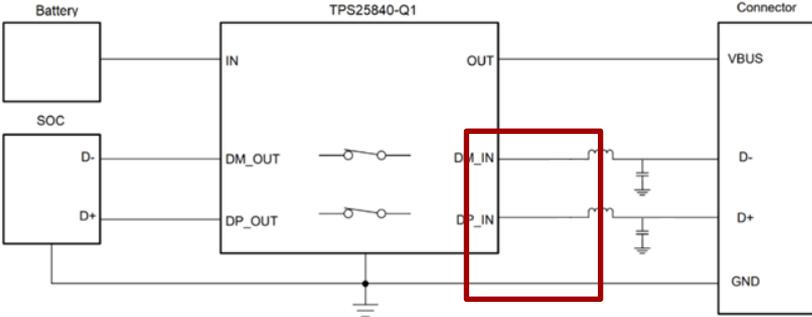
The TPS2583x have 800Mhz bandwidth help system pass 1.5Meter long cable eye diagram test.

PCB Layout

1. Both the DP and the DM signals should travel the same distance.
2. Control the DP and DM trace differential impedance to 90 ohm.
3. DP and DM signals should consistently be routed over a signal ground layer.
4. DP and DM keep away from noisy power signals and clock circuitry components.



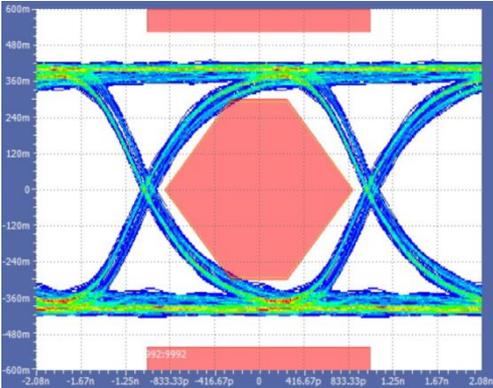
How to improve eye diagram with TPS2583x/4x-Q1



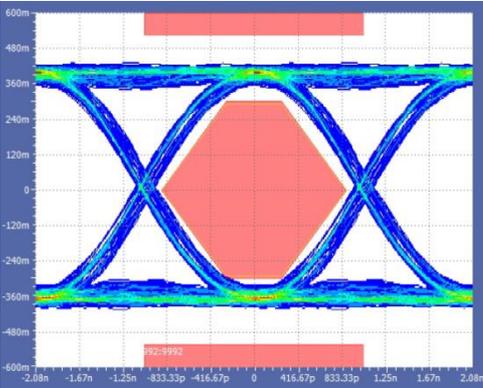
1.8m cable customer board

LC=16nH + 10pF

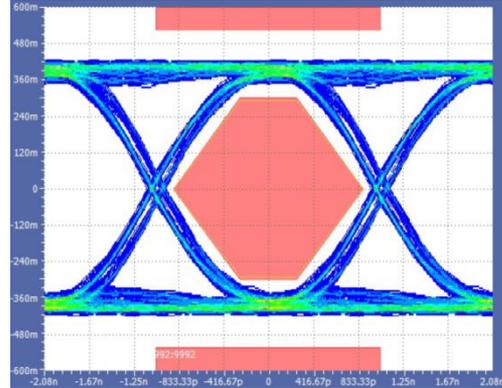
LC=16nH + 4.7pF



Fail

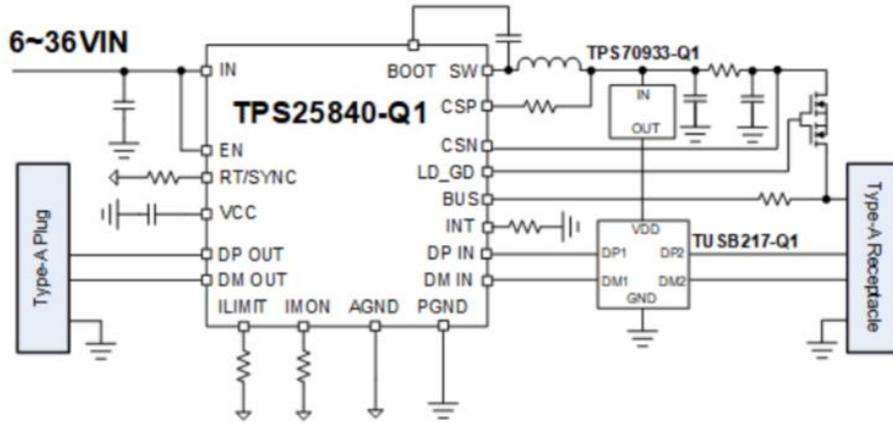


Fail



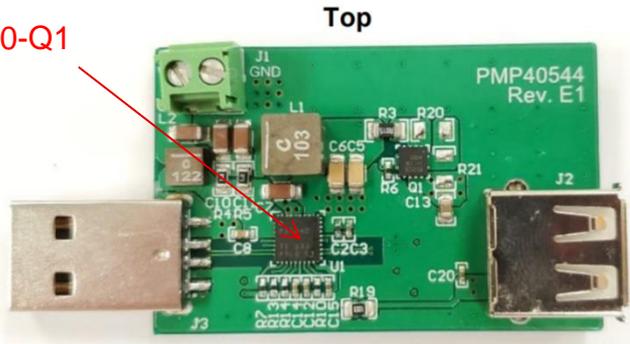
Pass

TPS25840 + TUSB217 eye diagram performance

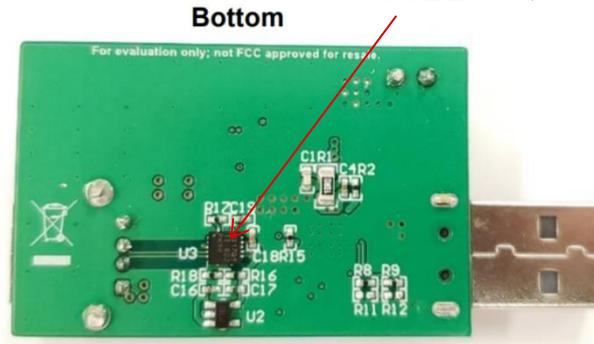


Reference Design:
PMP40544

TPS25840-Q1

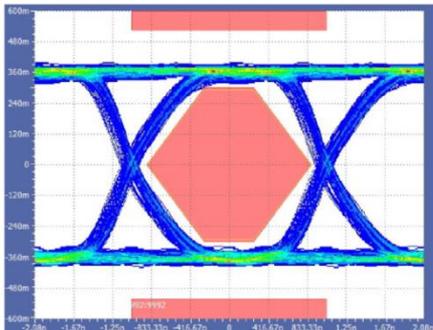


TUSB217-Q1

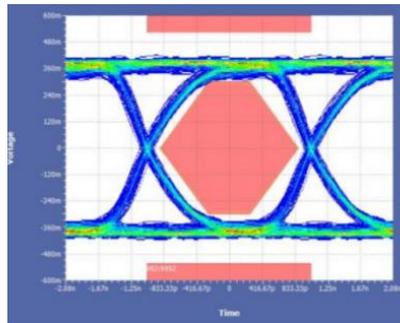


TPS25840 + TUSB217 eye diagram performance

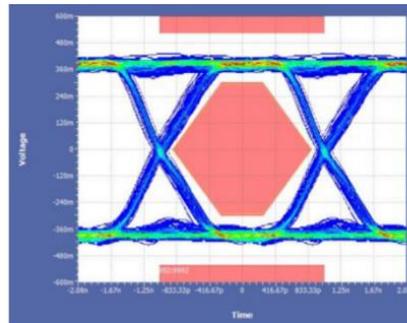
1.8M source



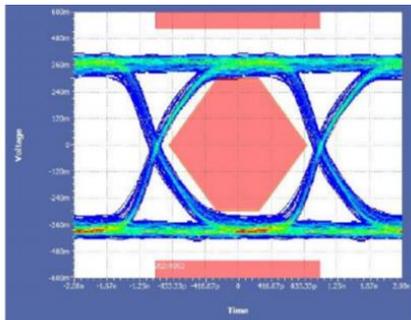
Add TPS25840



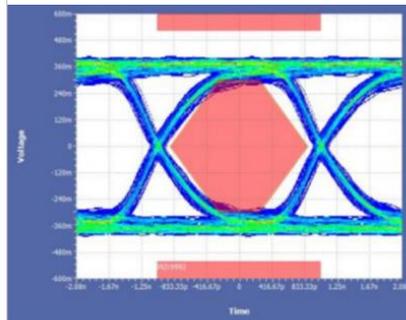
TPS25840+TUSB217



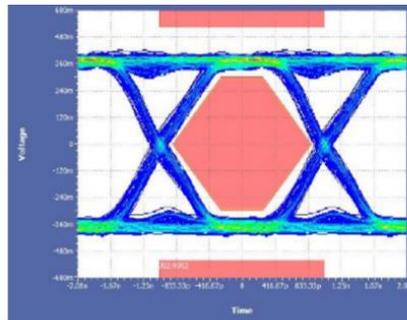
3M source



Add TPS25840



TPS25840+TUSB217



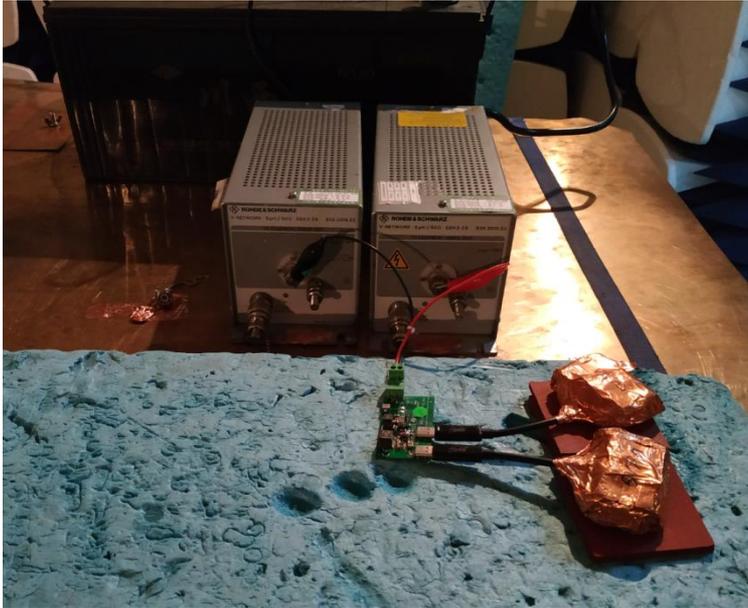
Agenda

- **What is CarPlay/MFi and what is USB's role in the application?**
- **How to pass MFi VBUS requirement with TPS2583x/4x**
 - TPS2583X/4x introduction
 - MFi VBUS spec and test set up
 - Cable compensation introduction
 - Cable compensation design to pass MFi
- **How to pass MFi OCP requirement with TPS2583x/4x**
 - MFi OCP spec and test set up
 - How to pass MFi OCP with TPS2583x/4x
- **How to improve eye diagram performance with TPS2583x/4x**
 - Eye diagram introduction and challenges in long cables
 - How to improve eye diagram performance with TPS2583x/4x
- **How to improve EMI with TPS2583x/4x**
- **Q&A**

TPS2583X/4X EMC test setup

CONDUCTED EMISSION STANDARD: CISPR25, CLASS 5

Test Site: Accurate technology Co., Ltd(ATC)

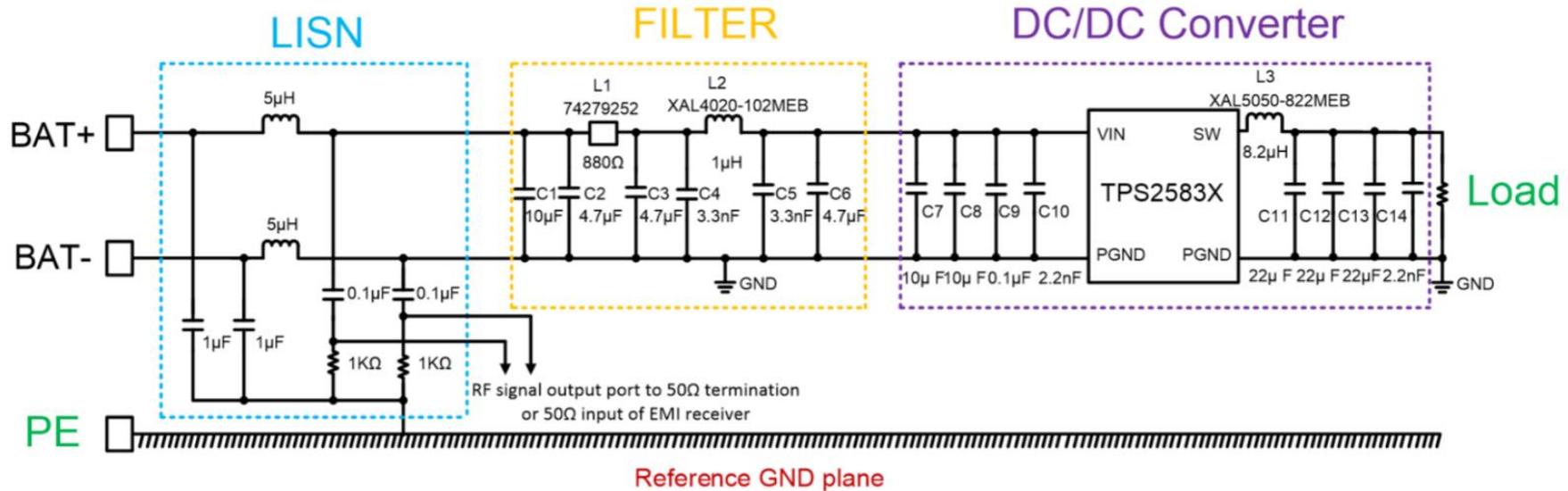


CE Setup



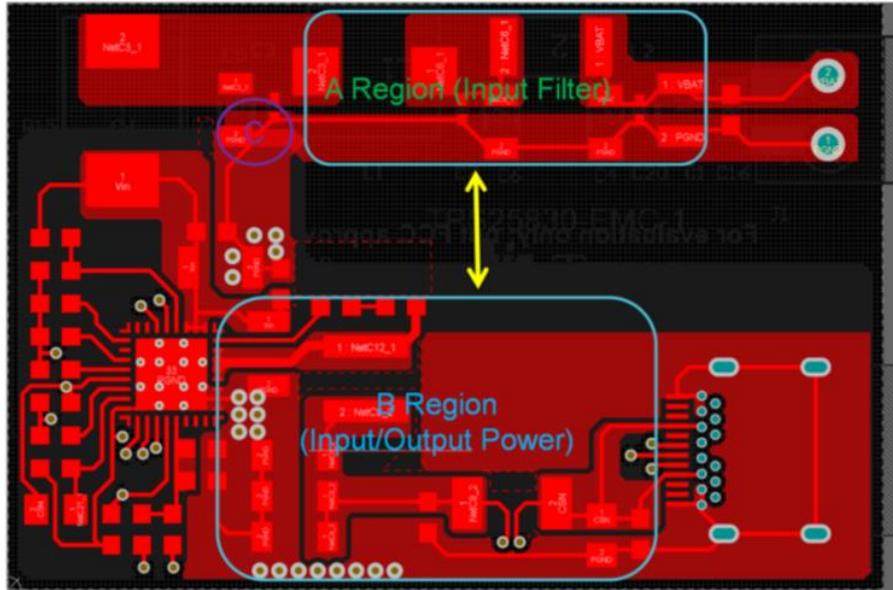
RE Setup

Recommended parameters for input filter

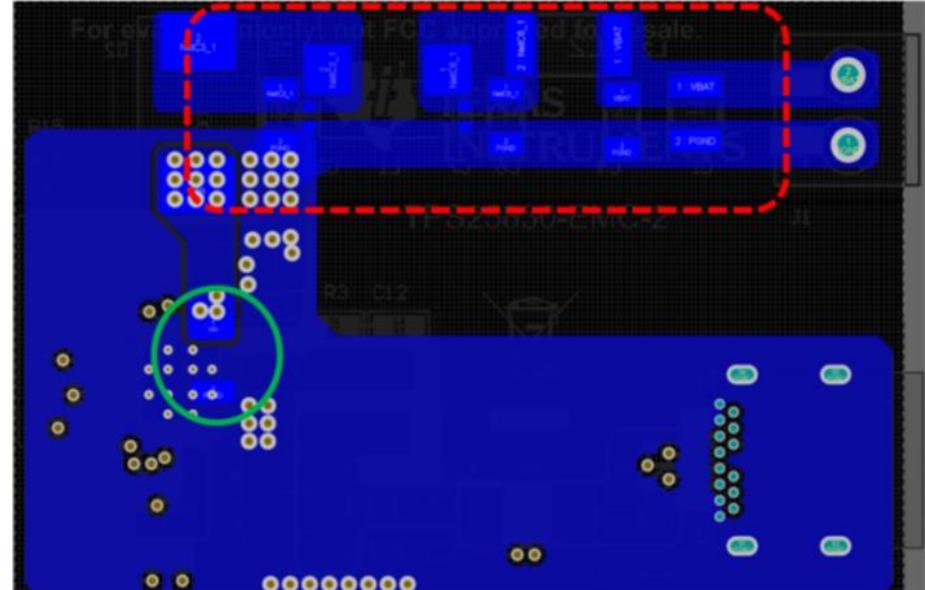


Input filter

Distance between **input filter** and **DCDC controller** should be more than 15 mm in order to prevent the switching noise bypassing the input filter.



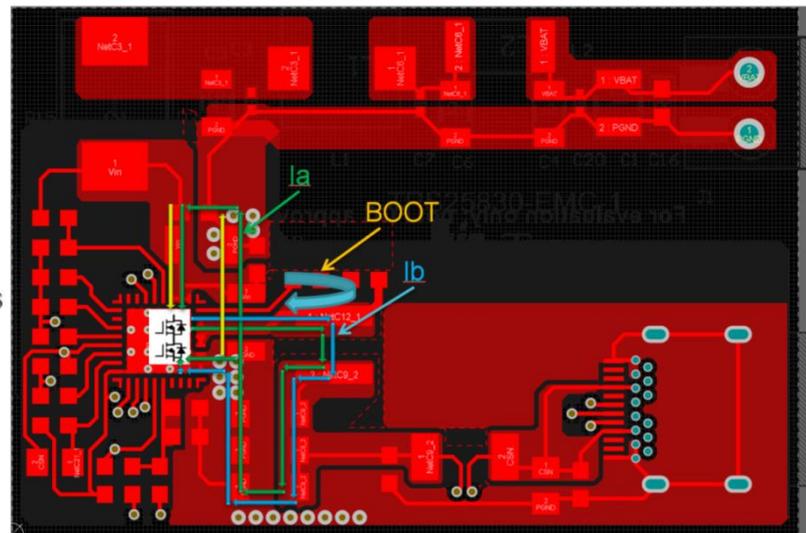
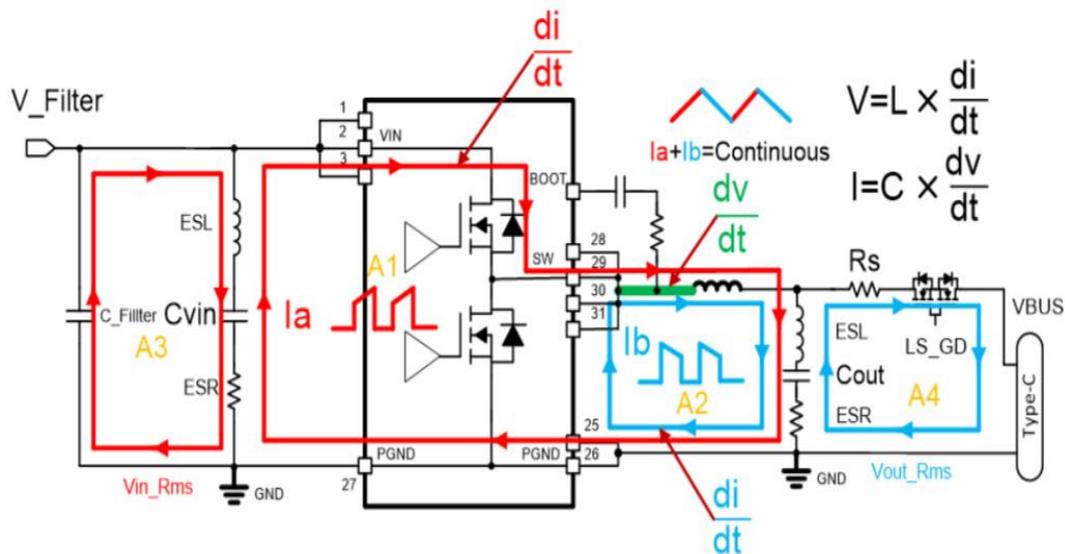
Top-layer of the PCB layout for input filter



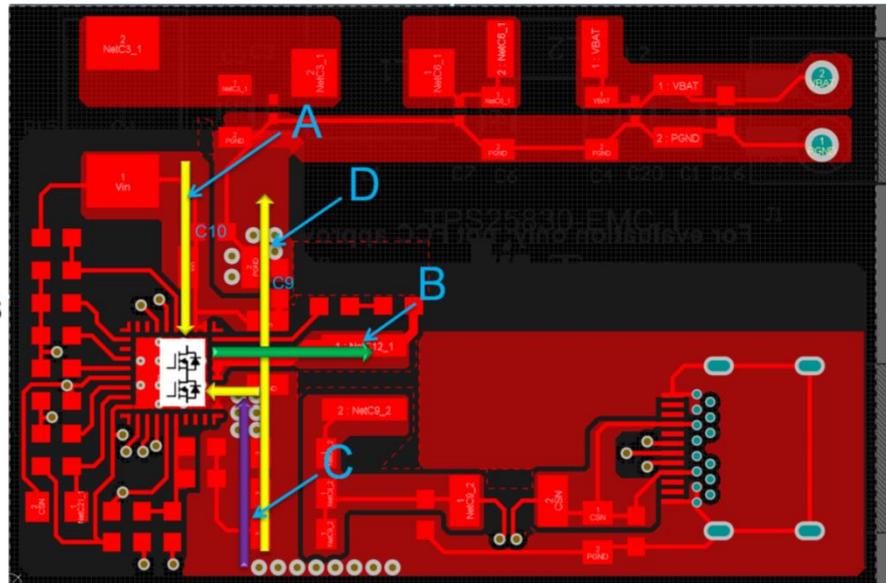
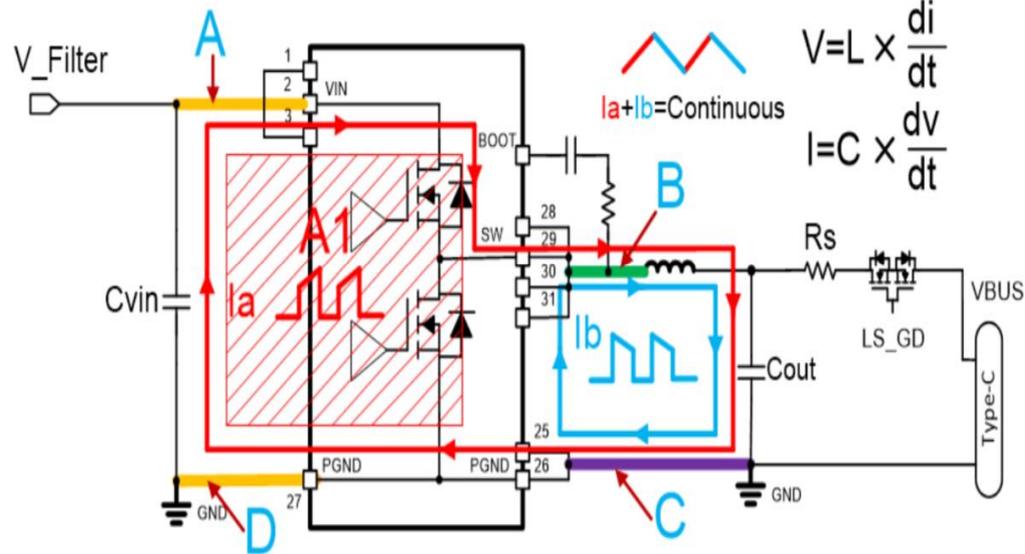
Bottom Layer of the PCB layout for input filter

DC/DC power loop

***la** and **lb** have high di/dt . These two loops should be kept as small as possible*

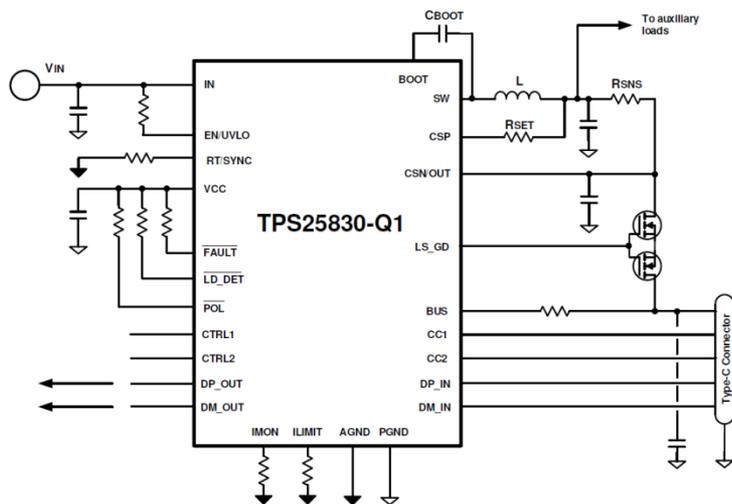


DC/DC power loop



line A, line B, line C, and line D trace length should be as short as possible.

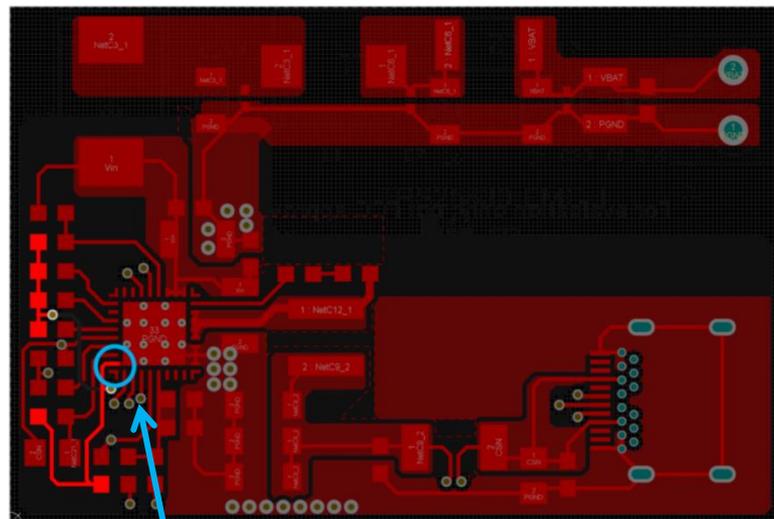
PGND and AGND



The signal ground AGND and the power ground PGND need to be separated in the actual PCB layout to reduce the power ground PGND causing oscillating to interfere with the signal ground AGND.

PGND includes: the ground of the input capacitor, output capacitor, VCC capacitor, input port, output port, and so forth.

AGND includes: the ground of RT, IMON, ILMIT, CC1, CC2, FAULT, LD_DET, POL, CTRL1/2, DP, DM resistors, capacitors and ground lines.

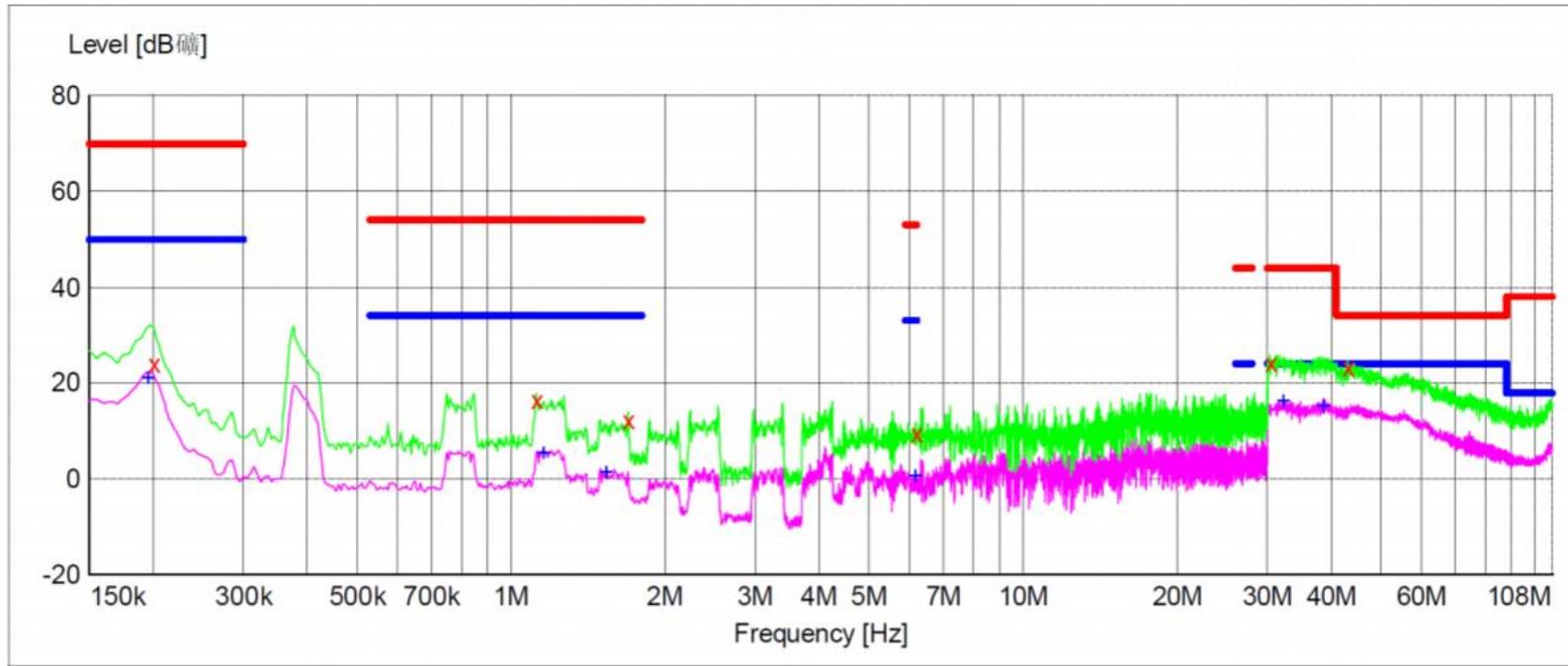


AGND and PGND connect point

TPS2583X EMI test result

CONDUCTED EMISSION STANDARD: CISPR25, CLASS 5

Test Condition: $V_{in}=12v$ $V_{BUS}=5V/3A$ $F=400KHz$



Summary

1. The input and output capacitors should be as close as possible to the input and output pins of the device.
2. Do not put Via on SW lines and keep the SW area as small as possible.
3. The input and output power loop should be as small as possible.
4. The BOOT pin trace loop circuit should as small as possible.
5. Power ground PGND and signal ground AGND should be separated.
6. The input and output capacitors better selected different value and package ,such as 10 μ F(1210) + 0.1 μ F(0603) + 2.2 nF(0603).

Application Note Link: <http://www.ti.com/lit/an/slvaen5/slvaen5.pdf?&ts=1589098906250>

Thank you for listening!



Do you have any questions?



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