







DRV5055-Q1

SBAS639D - OCTOBER 2017 - REVISED JUNE 2024

# DRV5055-Q1 Automotive Ratiometric Linear Hall Effect Sensor

#### 1 Features

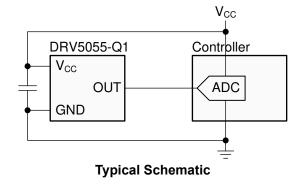
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- Ratiometric Linear Hall Effect Magnetic Sensor
- Operates From 3.3V and 5V Power Supplies
- Analog Output With V<sub>CC</sub> / 2 Quiescent Offset
- Magnetic Sensitivity Options (At  $V_{CC} = 5V$ ):
  - A1: 100mV/mT, ±21mT Range
  - A2/Z2: 50 mV/mT, ±42-mT Range
  - A3: 25mV/mT, ±85mT Range
  - A4: 12.5mV/mT, ±169mT Range
  - A5: -100mV/mT, ±21mT Range
- Fast 20kHz Sensing Bandwidth
- Low-Noise Output With ±1mA Drive
- Compensation for magnet temperature drift for A versions and none for Z versions
- AEC-Q100 Qualified for Automotive Applications:
  - Temperature Grade 0: -40°C to 150°C
- Standard Industry Packages:
  - Surface-Mount SOT-23
  - Through-Hole TO-92

### 2 Applications

- **Automotive Position Sensing**
- Brake, Acceleration, Clutch Pedals
- **Torque Sensors**
- **Gear Shifters**
- Throttle Position
- Height Leveling
- **Powertrain and Transmission Components**
- Absolute Angle Encoding
- Current Sensing



### 3 Description

The DRV5055-Q1 is a linear Hall effect sensor that responds proportionally to magnetic flux density. The device can be used for accurate position sensing in a wide range of applications.

The device operates from 3.3V or 5V power supplies. When no magnetic field is present, the analog output drives half of V<sub>CC</sub>. The output changes linearly with the applied magnetic flux density, and various sensitivity options enable maximal output voltage swing based on the required sensing range. North and south magnetic poles produce unique voltages.

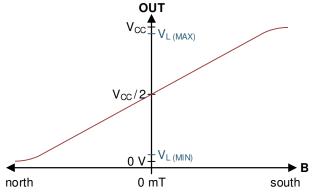
Magnetic flux perpendicular to the top of the package is sensed, and the two package options provide different sensing directions.

The device uses a ratiometric architecture that can eliminate error from V<sub>CC</sub> tolerance when the external analog-to-digital converter (ADC) uses the same V<sub>CC</sub> for its reference. Additionally, the device features magnet temperature compensation to counteract how magnets drift for linear performance across a wide -40°C to +150°C temperature range.

#### **Package Information**

<u> </u>						
PART NUMBER	PACKAGE <sup>(1)</sup>	PACKAGE SIZE <sup>(2)</sup>				
DRV5055-Q1	DBZ (SOT-23, 3)	2.92mm × 2.37mm				
DI(V3030-Q1	LPG (TO-92, 3)	4mm × 1.52mm				

- (1) For all available packages, see the orderable addendum at the end of the data sheet.
- (2)The package size (length × width) is a nominal value and includes pins, where applicable



Magnetic Response (A1, A2, A3, A4, Z2 Versions)



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# 4 Pin Configuration and Functions

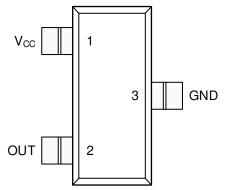


Figure 4-1. DBZ Package 3-Pin SOT-23 Top View

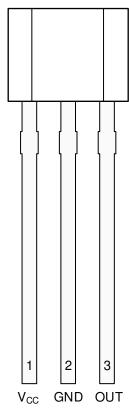


Figure 4-2. LPG Package 3-Pin TO-92 Top View

**Table 4-1. Pin Functions** 

	PIN		I/O	DESCRIPTION	
NAME	SOT-23	TO-92	1/0	DESCRIPTION	
V <sub>CC</sub>	1	1	_	Power supply. TI recommends connecting this pin to a ceramic capacitor to ground with a value of at least 0.01 µF.	
OUT	2	3	0	Analog output	
GND	3	2	_	Ground reference	



### **5 Specifications**

### 5.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted)(1)

1 3 1 3 1	,			
		MIN	MAX	UNIT
Power supply voltage	V <sub>CC</sub>	-0.3	7	V
Output voltage	OUT	-0.3	V <sub>CC</sub> + 0.3	V
Magnetic flux density, B <sub>MAX</sub>	Unlimited		Т	
Operating junction temperature, T <sub>J</sub>		-40	170	°C
Storage temperature, T <sub>stg</sub>		-65	150	°C

<sup>(1)</sup> Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under Recommended Operating Conditions. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

#### 5.2 ESD Ratings

			VALUE	UNIT
V		Human body model (HBM), per AEC Q100-002 <sup>(1)</sup>	±2500	V
V <sub>(ESD)</sub>	Electrostatic discharge	Charged device model (CDM), per AEC Q100-011	±750	V

<sup>(1)</sup> AEC Q100-002 indicates that HBM stressing shall be in accordance with the ANSI/ESDA/JEDEC JS-001 specification.

### 5.3 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)

		MIN	MAX	UNIT
V	Power supply voltage(1)	3	3.63	V
V <sub>CC</sub> Power supply voltage <sup>(1)</sup>	Power suppry voltage (*)	4.5	5.5	V
Io	Output continuous current	-1	1	mA
T <sub>A</sub>	Operating ambient temperature <sup>(2)</sup>	-40	150	°C

<sup>(1)</sup> There are two isolated operating V<sub>CC</sub> ranges. For more information see the Operating V<sub>CC</sub> Ranges section.

### 5.4 Thermal Information

		DRV50		
	THERMAL METRIC <sup>(1)</sup>	SOT-23 (DBZ)	TO-92 (LPG)	UNIT
		3 PINS	3 PINS	
$R_{\theta JA}$	Junction-to-ambient thermal resistance	170	121	°C/W
R <sub>0</sub> JC(top)	Junction-to-case (top) thermal resistance	66	67	°C/W
$R_{\theta JB}$	Junction-to-board thermal resistance	49	97	°C/W
Y <sub>JT</sub>	Junction-to-top characterization parameter	1.7	7.6	°C/W
$Y_{JB}$	Junction-to-board characterization parameter	48	97	°C/W

<sup>(1)</sup> For more information about traditional and new thermal metrics, see the Semiconductor and IC Package Thermal Metrics application note.

Product Folder Links: DRV5055-Q1

<sup>(2)</sup> Power dissipation and thermal limits must be observed.



#### 5.5 Electrical Characteristics

for  $V_{CC}$  = 3V to 3.63V and 4.5V to 5.5V, over operating free-air temperature range (unless otherwise noted)

	PARAMETER	TEST CONDI	TEST CONDITIONS <sup>(1)</sup>		TYP	MAX	UNIT
I <sub>CC</sub>	Operating supply current				6	10	mA
t <sub>ON</sub>	Power-on time (see Figure 6-4)				175	330	μs
f <sub>BW</sub>	Sensing bandwidth				20		kHz
t <sub>d</sub>	Propagation delay time	From change in B to c	hange in OUT		10		μs
D	Input referred PMS paige density	V <sub>CC</sub> = 5V			130		nT/√ <del>Hz</del>
B <sub>ND</sub>	Input-referred RMS noise density	V <sub>CC</sub> = 3.3V			215		III/VIIZ
D	Input referred pains	B <sub>ND</sub> × 6.6 × √20 kHz	V <sub>CC</sub> = 5V		0.12		m.T
B <sub>N</sub>	Input-referred noise	D <sub>ND</sub> × 0.0 × √20 kH2	V <sub>CC</sub> = 3.3V		0.2		$mT_PP$
			DRV5055A1, DRV5055A5		12		${\sf mV_{PP}}$
V <sub>N</sub>	√ <sub>N</sub> Output-referred noise <sup>(2)</sup>	B <sub>N</sub> × S	DRV5055A2, DRV5055Z2		6		${\rm mV_{PP}}$
			DRV5055A3		3		${\rm mV_{PP}}$
			DRV5055A4		1.5		${\rm mV_{PP}}$

 <sup>(1)</sup> B is the applied magnetic flux density.
 (2) V<sub>N</sub> describes voltage noise on the device output. If the full device bandwidth is not needed, noise can be reduced with an RC filter.



#### **5.6 Magnetic Characteristics**

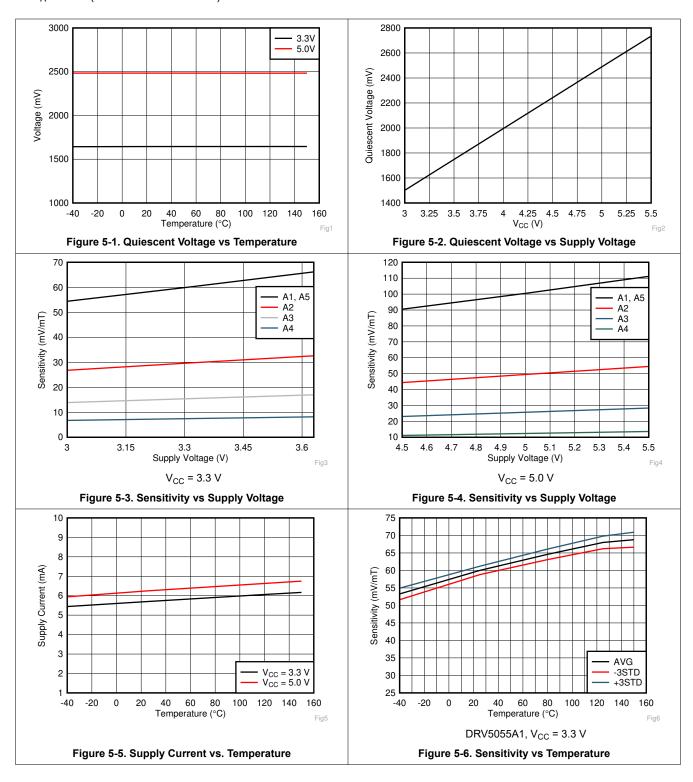
for V<sub>CC</sub> = 3V to 3.63V and 4.5V to 5.5V, over operating free-air temperature range (unless otherwise noted)

	PARAMETER	TEST COND	ITIONS <sup>(1)</sup>	MIN	TYP	MAX	UNIT
. ,	0	D 0 T T 0500	V <sub>CC</sub> = 5V	2.43	2.5	2.57	.,
√ <sub>Q</sub>	Quiescent voltage	B = 0mT, T <sub>A</sub> = 25°C	V <sub>CC</sub> = 3.3V	1.59	1.65	1.71	V
$V_{Q\Delta T}$	Quiescent voltage temperature drift	B = 0 mT, T <sub>A</sub> = -40°C to 150°C	versus 25°C	±1	% × V <sub>CC</sub>		V
V <sub>QRE</sub>	Quiescent voltage ratiometry error <sup>(2)</sup>				±0.2		%
$V_{Q\Delta L}$	Quiescent voltage lifetime drift	High-temperature ope	erating stress for		<0.5		%
			DRV5055A1	95	100	105	
			DRV5055A2/Z2	47.5	50	52.5	
S Sensitivity	V <sub>CC</sub> = 5V, T <sub>A</sub> = 25°C	DRV5055A3	23.8	25	26.2		
		1 A 20 0	DRV5055A4	11.9	12.5	13.2	
	Concitivity		DRV5055A5	-105	-100	-95	m)//m/T
	Sensitivity		DRV5055A1	57	60	63	mV/mT
		V <sub>CC</sub> = 3.3V, T <sub>A</sub> = 25°C	DRV5055A2/Z2	28.5	30	31.5	
			DRV5055A3	14.3	15	15.8	
			DRV5055A4	7.1	7.5	7.9	
			DRV5055A5	-63	-60	-57	
		V <sub>CC</sub> = 5V, T <sub>A</sub> = 25°C	DRV5055A1, DRV5055A5	±21			mT
			DRV5055A2/Z2	±42			mT
			DRV5055A3	±85			mT
_	l in a a marking a main man a (3) (4)		DRV5055A4	±169			mT
B <sub>L</sub>	Linear magnetic sensing range <sup>(3) (4)</sup>	V <sub>CC</sub> = 3.3V, T <sub>A</sub> = 25°C	DRV5055A1, DRV5055A5	±22			mT
			DRV5055A2/Z2	±44			mT
			DRV5055A3	±88			mT
			DRV5055A4	±176			mT
٧L	Linear range of output voltage <sup>(4)</sup>			0.2		V <sub>CC</sub> - 0.2	V
c	Sensitivity temperature compensation	DRV5055A1/A2/A3/A	4		0.12		%/°C
S <sub>TC</sub>	for magnets <sup>(5)</sup>	DRV5055Z2			0		%/°C
S <sub>LE</sub>	Sensitivity linearity error <sup>(4)</sup>	V <sub>OUT</sub> is within V <sub>L</sub>			±1		%
S <sub>SE</sub>	Sensitivity symmetry error <sup>(4)</sup>	V <sub>OUT</sub> is within V <sub>L</sub>			±1		%
S <sub>RE</sub>	Sensitivity ratiometry error <sup>(2)</sup>	T <sub>A</sub> = 25°C, with respect to V <sub>CC</sub> = 3.3V or 5V		-2.5		2.5	%
S <sub>ΔL</sub>	Sensitivity lifetime drift	High-temperature operating stress for 1000 hours			<0.5		%

- B is the applied magnetic flux density.
- (1) (2) See the Ratiometric Architecture section.
- (3) B<sub>L</sub> describes the minimum linear sensing range at 25°C taking into account the maximum V<sub>Q</sub> and Sensitivity tolerances.
- (4) See the Sensitivity Linearity section.
- S<sub>TC</sub> describes the rate the device increases Sensitivity with temperature. For more information, see the Sensitivity Temperature Compensation for Magnets section.

### **5.7 Typical Characteristics**

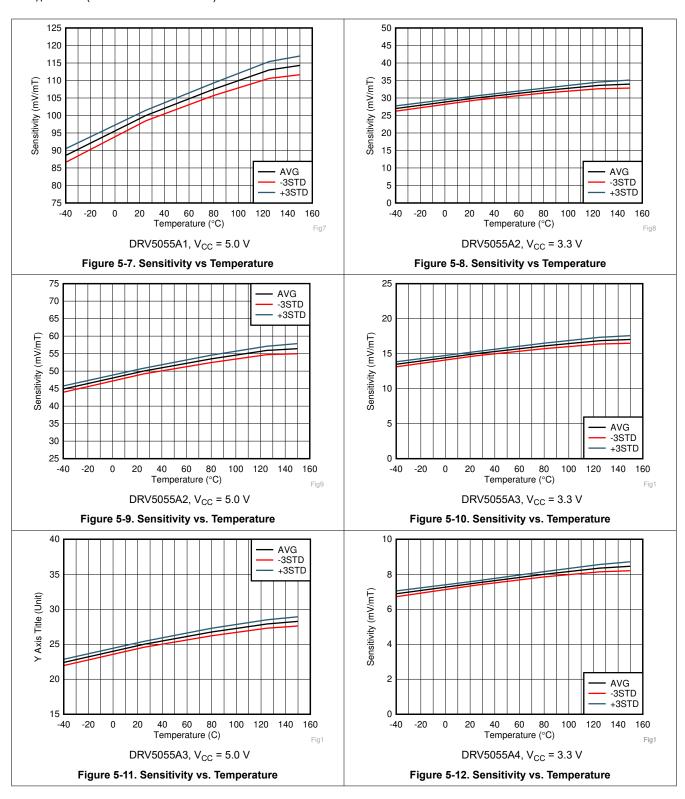
for  $T_A = 25$ °C (unless otherwise noted)





### 5.7 Typical Characteristics (continued)

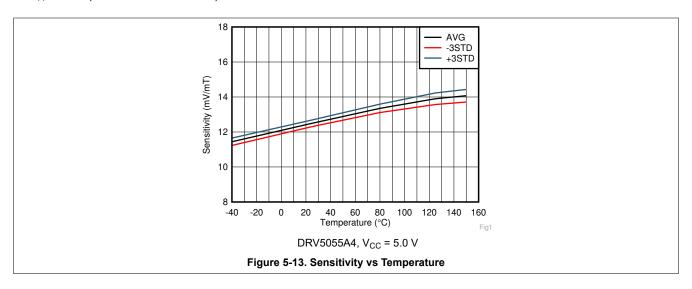
for T<sub>A</sub> = 25°C (unless otherwise noted)





# **5.7 Typical Characteristics (continued)**

for  $T_A = 25$ °C (unless otherwise noted)



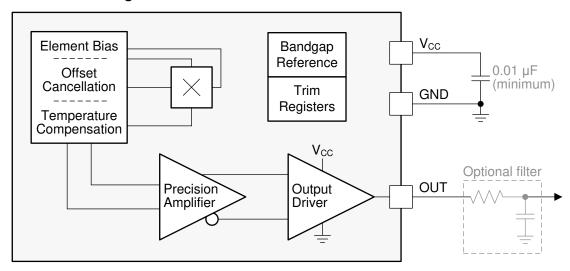


# **6 Detailed Description**

#### 6.1 Overview

The DRV5055-Q1 is a 3-pin linear Hall effect sensor with fully integrated signal conditioning, temperature compensation circuits, mechanical stress cancellation, and amplifiers. The device operates from 3.3-V and 5-V ( $\pm 10\%$ ) power supplies, measures magnetic flux density, and outputs a proportional analog voltage that is referenced to  $V_{CC}$ .

### 6.2 Functional Block Diagram



### **6.3 Feature Description**

### 6.3.1 Magnetic Flux Direction

As shown in Figure 6-1, the DRV5055-Q1 is sensitive to the magnetic field component that is perpendicular to the top of the package.

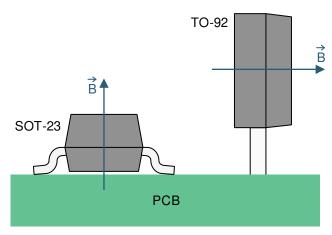


Figure 6-1. Direction of Sensitivity

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Magnetic flux that travels from the bottom to the top of the package is considered positive in this document. This condition exists when a south magnetic pole is near the top (marked-side) of the package. Magnetic flux that travels from the top to the bottom of the package results in negative millitesla values.

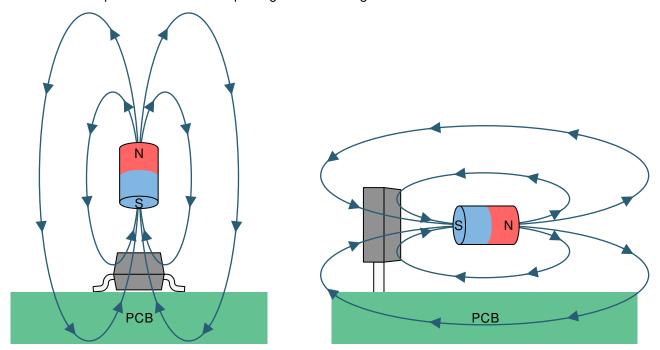


Figure 6-2. The Flux Direction for Positive B

#### 6.3.2 Magnetic Response

When the DRV5055-Q1 is powered, the DRV5055-Q1 outputs an analog voltage according to Equation 1:

$$V_{OUT} = V_Q + B \times \left( Sensitivity_{(25^{\circ}C)} \times (1 + S_{TC} \times (T_A - 25^{\circ}C)) \right)$$
 (1)

#### where

- V<sub>Q</sub> is typically half of V<sub>CC</sub>
- · B is the applied magnetic flux density
- Sensitivity $_{(25^{\circ}\text{C})}$  depends on the device option and  $V_{\text{CC}}$
- S<sub>TC</sub> is typically 0.12%/°C for device options DRV5055A1 DRV5055A4 and is 0%/°C for DRV5055Z2
- · T<sub>A</sub> is the ambient temperature
- V<sub>OUT</sub> is within the V<sub>L</sub> range

As an example, consider the DRV5055A3 with  $V_{CC}$  = 3.3 V, a temperature of 50°C, and 67 mT applied. Excluding tolerances,  $V_{OUT}$  = 1650 mV + 67 mT × (15 mV/mT × (1 + 0.0012/°C × (50°C – 25°C))) = 2685 mV.

#### 6.3.3 Sensitivity Linearity

The device produces a linear response when the output voltage is within the specified  $V_L$  range. Outside this range, sensitivity is reduced and nonlinear. Figure 6-3 graphs the magnetic response.

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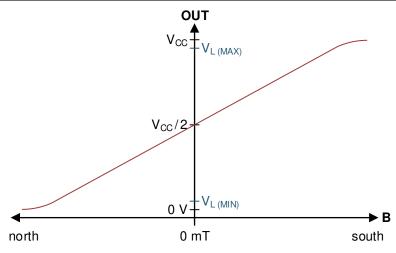


Figure 6-3. Magnetic Response

Equation 2 calculates parameter  $B_L$ , the minimum linear sensing range at 25°C taking into account the maximum quiescent voltage and sensitivity tolerances.

$$B_{L(MIN)} = \frac{V_{L(MAX)} - V_{Q(MAX)}}{S_{(MAX)}}$$
(2)

The parameter  $S_{LE}$  defines linearity error as the difference in sensitivity between any two positive B values, and any two negative B values, while the output is within the  $V_L$  range.

The parameter  $S_{SE}$  defines symmetry error as the difference in sensitivity between any positive B value and the negative B value of the same magnitude, while the output voltage is within the  $V_L$  range.

#### 6.3.4 Ratiometric Architecture

The DRV5055-Q1 has a ratiometric analog architecture that scales the quiescent voltage and sensitivity linearly with the power-supply voltage. For example, the quiescent voltage and sensitivity are 5% higher when  $V_{CC}$  = 5.25 V compared to  $V_{CC}$  = 5 V. This behavior enables external ADCs to digitize a consistent value regardless of the power-supply voltage tolerance, when the ADC uses  $V_{CC}$  as its reference.

Equation 3 calculates the sensitivity ratiometry error:

$$S_{RE} = 1 - \frac{S_{(VCC)} / S_{(5V)}}{V_{CC} / 5V} \text{ for } V_{CC} = 4.5 \text{ V to } 5.5 \text{ V}, \qquad S_{RE} = 1 - \frac{S_{(VCC)} / S_{(3.3V)}}{V_{CC} / 3.3V} \text{ for } V_{CC} = 3 \text{ V to } 3.63 \text{ V}$$

$$(3)$$

#### where

- $S_{(VCC)}$  is the sensitivity at the current  $V_{CC}$  voltage
- $S_{(5V)}$  or  $S_{(3.3V)}$  is the sensitivity when  $V_{CC} = 5 \text{ V}$  or 3.3 V
- V<sub>CC</sub> is the current V<sub>CC</sub> voltage

Equation 4 calculates the quiescent voltage ratiometry error:

$$V_{QRE} = 1 - \frac{V_{Q(VCC)} / V_{Q(5V)}}{V_{CC} / 5V} \text{ for } V_{CC} = 4.5 \text{ V to } 5.5 \text{ V}, \qquad V_{QRE} = 1 - \frac{V_{Q(VCC)} / V_{Q(3.3V)}}{V_{CC} / 3.3V} \text{ for } V_{CC} = 3 \text{ V to } 3.63 \text{ V}$$

#### where

- V<sub>Q(VCC)</sub> is the quiescent voltage at the current V<sub>CC</sub> voltage
- V<sub>Q(5V)</sub> or V<sub>Q(3.3V)</sub> is the quiescent voltage when V<sub>CC</sub> = 5 V or 3.3 V
- V<sub>CC</sub> is the current V<sub>CC</sub> voltage

#### 6.3.5 Operating V<sub>CC</sub> Ranges

The DRV5055-Q1 has two recommended operating  $V_{CC}$  ranges: 3 V to 3.63 V and 4.5 V to 5.5 V. When  $V_{CC}$  is in the middle region between 3.63 V to 4.5 V, the device continues to function, but sensitivity is less known because there is a crossover threshold near 4 V that adjusts device characteristics.

#### 6.3.6 Sensitivity Temperature Compensation for Magnets

Magnets generally produce weaker fields as temperature increases. The DRV5055-Q1 compensates by increasing sensitivity with temperature, as defined by the parameter  $S_{TC}$ . For device options DRV5055A1 - DRV5055A4, the sensitivity at  $T_A$  = 150°C is typically 12% higher than at  $T_A$  = 25°C. For device options DRV5055Z2, the sensitivity at  $T_A$  = 150°C is typically same as the value at  $T_A$  = 25°C.

#### 6.3.7 Power-On Time

After the  $V_{CC}$  voltage is applied, the DRV5055-Q1 requires a short initialization time before the output is set. The parameter  $t_{ON}$  describes the time from when  $V_{CC}$  crosses 3 V until OUT is within 5% of  $V_{Q}$ , with 0 mT applied and no load attached to OUT. Figure 6-4 shows this timing diagram.

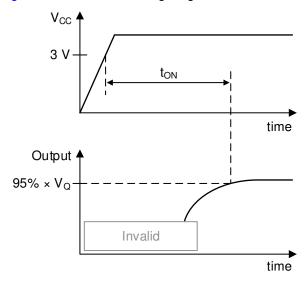


Figure 6-4. t<sub>ON</sub> Definition



#### 6.3.8 Hall Element Location

Figure 6-5 shows the location of the sensing element inside each package option.

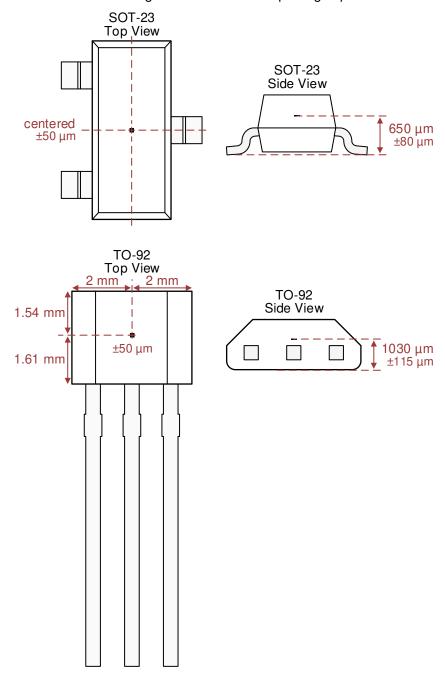


Figure 6-5. Hall Element Location

### **6.4 Device Functional Modes**

The DRV5055-Q1 has one mode of operation that applies when the *Recommended Operating Conditions* are met.

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# 7 Application and Implementation

#### Note

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes, as well as validating and testing their design implementation to confirm system functionality.

#### 7.1 Application Information

# 7.1.1 Selecting the Sensitivity Option

Select the highest DRV5055-Q1 sensitivity option that can measure the required range of magnetic flux density, so that the output voltage swing is maximized.

Larger-sized magnets and farther sensing distances can generally enable better positional accuracy than very small magnets at close distances, because magnetic flux density increases exponentially with the proximity to a magnet. TI created an online tool to help with simple magnet calculations at <a href="https://www.ti.com/product/drv5013">https://www.ti.com/product/drv5013</a>.

#### 7.1.2 Temperature Compensation for Magnets

The DRV5055-Q1 temperature compensation is designed to directly compensate the average drift of neodymium (NdFeB) magnets and partially compensate ferrite magnets. The residual induction ( $B_r$ ) of a magnet typically reduces by 0.12%/°C for NdFeB, and 0.20%/°C for ferrite. When the operating temperature of a system is reduced, temperature drift errors are also reduced.

#### 7.1.3 Adding a Low-Pass Filter

As shown in *Functional Block Diagram*, an RC low-pass filter can be added to the device output for the purpose of minimizing voltage noise when the full 20-kHz bandwidth is not needed. This filter can improve the signal-to-noise ratio (SNR) and overall accuracy. Do not connect a capacitor directly to the device output without a resistor in between because doing so can make the output unstable.

#### 7.1.4 Designing for Wire Break Detection

Some systems must detect if interconnect wires become open or shorted. The DRV5055-Q1 can support this function.

First, select a sensitivity option that causes the output voltage to stay within the  $V_L$  range during normal operation. Second, add a pullup resistor between OUT and  $V_{CC}$ . TI recommends a value between 20 k $\Omega$  to 100 k $\Omega$ , and the current through OUT must not exceed the  $I_O$  specification, including current going into an external ADC. Then, if the output voltage is ever measured to be within 150 mV of  $V_{CC}$  or GND, a fault condition exists. Figure 7-1 shows the circuit, and Table 7-1 describes fault scenarios.

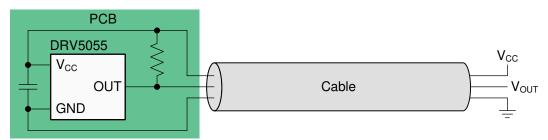


Figure 7-1. Wire Fault Detection Circuit



Table 7-1. Fault Scenarios and the Resulting V <sub>0</sub>	OUT
---	-----

FAULT SCENARIO	V <sub>OUT</sub>
V <sub>CC</sub> disconnects	Close to GND
GND disconnects	Close to V <sub>CC</sub>
V <sub>CC</sub> shorts to OUT	Close to V <sub>CC</sub>
GND shorts to OUT	Close to GND

### 7.2 Typical Application

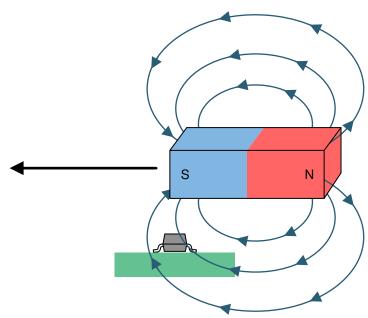


Figure 7-2. Common Magnet Orientation

#### 7.2.1 Design Requirements

Use the parameters listed in Table 7-2 for this design example.

**Table 7-2. Design Parameters** 

DESIGN PARAMETER	EXAMPLE VALUE
V <sub>CC</sub>	5 V
Magnet	15 × 5 × 5 mm NdFeB
Travel distance	12 mm
Maximum B at the sensor at 25°C	±75 mT
Device option	DRV5055A3

#### 7.2.2 Detailed Design Procedure

Linear Hall effect sensors provide flexibility in mechanical design, because many possible magnet orientations and movements produce a usable response from the sensor. Figure 7-2 shows one of the most common orientations, which uses the full north to south range of the sensor and causes a close-to-linear change in magnetic flux density as the magnet moves across.

When designing a linear magnetic sensing system, always consider these three variables: the magnet, sensing distance, and the range of the sensor. Select the DRV5055-Q1 with the highest sensitivity that has a  $B_L$  (linear magnetic sensing range) that is larger than the maximum magnetic flux density in the application. To determine the magnetic flux density the sensor receives, TI recommends using magnetic field simulation software, referring to magnet specifications, and testing.

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#### 7.2.3 Application Curve

Figure 7-3 shows the simulated magnetic flux from a NdFeB magnet.

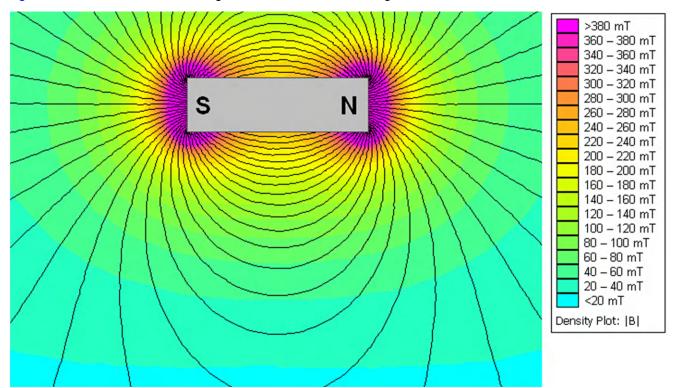


Figure 7-3. Simulated Magnetic Flux

# 7.3 Best Design Practices

The Hall element is sensitive to magnetic fields that are perpendicular to the top of the package, therefore a correct magnet approach must be used for the sensor to detect the field. Figure 7-4 shows correct and incorrect approaches.

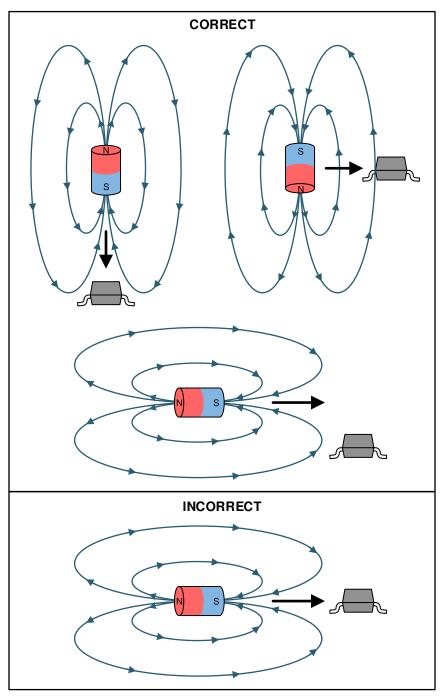


Figure 7-4. Correct and Incorrect Magnet Approaches

### 7.4 Power Supply Recommendations

A decoupling capacitor close to the device must be used to provide local energy with minimal inductance. TI recommends using a ceramic capacitor with a value of at least  $0.01 \, \mu F$ .

#### 7.5 Layout

#### 7.5.1 Layout Guidelines

Magnetic fields pass through most nonferromagnetic materials with no significant disturbance. Embedding Hall effect sensors within plastic or aluminum enclosures and sensing magnets on the outside is common practice. Magnetic fields also easily pass through most printed circuit boards, which makes placing the magnet on the opposite side possible.

### 7.5.2 Layout Examples

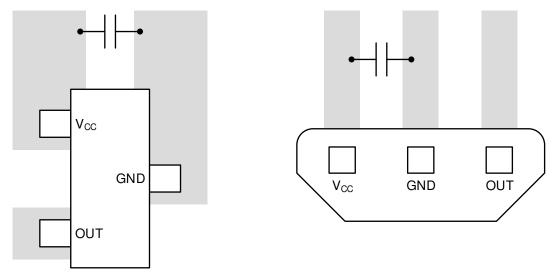


Figure 7-5. Layout Examples



### 8 Device and Documentation Support

#### **8.1 Documentation Support**

#### 8.1.1 Related Documentation

For related documentation see the following:

- Texas Instruments, Overview Using Linear Hall Effect Sensors to Measure Angle application brief
- Texas Instruments, Incremental Rotary Encoder Design Considerations application brief

### 8.2 Receiving Notification of Documentation Updates

To receive notification of documentation updates, navigate to the device product folder on ti.com. Click on Notifications to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

#### 8.3 Support Resources

TI E2E<sup>™</sup> support forums are an engineer's go-to source for fast, verified answers and design help — straight from the experts. Search existing answers or ask your own question to get the quick design help you need.

Linked content is provided "AS IS" by the respective contributors. They do not constitute TI specifications and do not necessarily reflect TI's views; see TI's Terms of Use.

#### 8.4 Trademarks

TI E2E<sup>™</sup> is a trademark of Texas Instruments.

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### 8.5 Electrostatic Discharge Caution



This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

#### 8.6 Glossary

TI Glossary

This glossary lists and explains terms, acronyms, and definitions.

#### 9 Revision History

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

С	hanges from Revision C (July 2018) to Revision D (June 2024)	Page
•	Updated the numbering format for tables, figures, and cross-references throughout the document Changed Device Information table to Package Information	1
•	Changed number of sensitivity options from five to various in <i>Description</i>	
_	Changed Dos and Donts table to Best Design Practices	18
С	hanges from Revision B (January 2018) to Revision C (July 2018)	Page
•	Released to production	1

# 10 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

Product Folder Links: DRV5055-Q1

www.ti.com

16-Jun-2024

#### PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan	Lead finish/ Ball material	MSL Peak Temp	Op Temp (°C)	Device Marking (4/5)	Samples
DRV5055A1EDBZRQ1	ACTIVE	SOT-23	DBZ	3	3000	RoHS & Green	SN	Level-3-260C-168 HR	-40 to 150	55A1Z	Samples
DRV5055A1ELPGMQ1	ACTIVE	TO-92	LPG	3	3000	RoHS & Green	SN	N / A for Pkg Type	-40 to 150	55A1Z	Samples
DRV5055A1ELPGQ1	ACTIVE	TO-92	LPG	3	1000	RoHS & Green	SN	N / A for Pkg Type	-40 to 150	55A1Z	Samples
DRV5055A2EDBZRQ1	ACTIVE	SOT-23	DBZ	3	3000	RoHS & Green	SN	Level-3-260C-168 HR	-40 to 150	55A2Z	Samples
DRV5055A2ELPGMQ1	ACTIVE	TO-92	LPG	3	3000	RoHS & Green	SN	N / A for Pkg Type	-40 to 150	55A2Z	Samples
DRV5055A2ELPGQ1	ACTIVE	TO-92	LPG	3	1000	RoHS & Green	SN	N / A for Pkg Type	-40 to 150	55A2Z	Samples
DRV5055A3EDBZRQ1	ACTIVE	SOT-23	DBZ	3	3000	RoHS & Green	SN	Level-3-260C-168 HR	-40 to 150	55A3Z	Samples
DRV5055A3ELPGMQ1	ACTIVE	TO-92	LPG	3	3000	RoHS & Green	SN	N / A for Pkg Type	-40 to 150	55A3Z	Samples
DRV5055A3ELPGQ1	ACTIVE	TO-92	LPG	3	1000	RoHS & Green	SN	N / A for Pkg Type	-40 to 150	55A3Z	Samples
DRV5055A4EDBZRQ1	ACTIVE	SOT-23	DBZ	3	3000	RoHS & Green	SN	Level-3-260C-168 HR	-40 to 150	55A4Z	Samples
DRV5055A4ELPGMQ1	ACTIVE	TO-92	LPG	3	3000	RoHS & Green	SN	N / A for Pkg Type	-40 to 150	55A4Z	Samples
DRV5055A4ELPGQ1	ACTIVE	TO-92	LPG	3	1000	RoHS & Green	SN	N / A for Pkg Type	-40 to 150	55A4Z	Samples
DRV5055Z2EDBZRQ1	ACTIVE	SOT-23	DBZ	3	3000	RoHS & Green	SN	Level-1-260C-UNLIM	-40 to 150	55Z2	Samples

<sup>(1)</sup> The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

Green: TI defines "Green" to mean the content of Chlorine (CI) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

<sup>(2)</sup> **RoHS**: TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

# PACKAGE OPTION ADDENDUM

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- (3) MSL, Peak Temp. The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.
- (4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.
- (5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.
- (6) Lead finish/Ball material Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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#### OTHER QUALIFIED VERSIONS OF DRV5055-Q1:

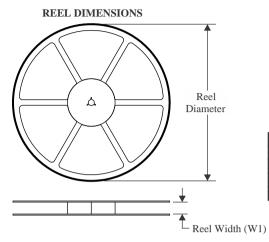
Catalog: DRV5055

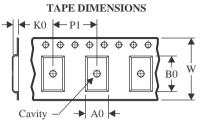
NOTE: Qualified Version Definitions:

Catalog - TI's standard catalog product

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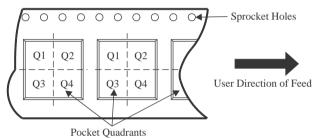
### TAPE AND REEL INFORMATION





A0	Dimension designed to accommodate the component width
В0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

#### QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



#### \*All dimensions are nominal

Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
DRV5055A1EDBZRQ1	SOT-23	DBZ	3	3000	180.0	8.4	3.15	2.77	1.22	4.0	8.0	Q3
DRV5055A2EDBZRQ1	SOT-23	DBZ	3	3000	180.0	8.4	3.15	2.77	1.22	4.0	8.0	Q3
DRV5055A3EDBZRQ1	SOT-23	DBZ	3	3000	180.0	8.4	3.15	2.77	1.22	4.0	8.0	Q3
DRV5055A4EDBZRQ1	SOT-23	DBZ	3	3000	180.0	8.4	3.15	2.77	1.22	4.0	8.0	Q3
DRV5055Z2EDBZRQ1	SOT-23	DBZ	3	3000	180.0	8.4	3.2	2.85	1.3	4.0	8.0	Q3



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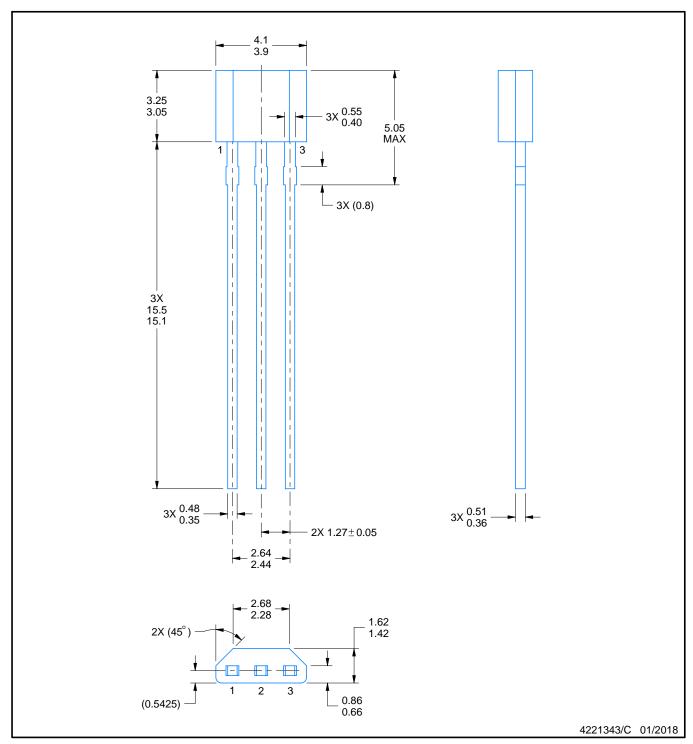


#### \*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
DRV5055A1EDBZRQ1	SOT-23	DBZ	3	3000	213.0	191.0	35.0
DRV5055A2EDBZRQ1	SOT-23	DBZ	3	3000	213.0	191.0	35.0
DRV5055A3EDBZRQ1	SOT-23	DBZ	3	3000	213.0	191.0	35.0
DRV5055A4EDBZRQ1	SOT-23	DBZ	3	3000	213.0	191.0	35.0
DRV5055Z2EDBZRQ1	SOT-23	DBZ	3	3000	210.0	185.0	35.0



TRANSISTOR OUTLINE



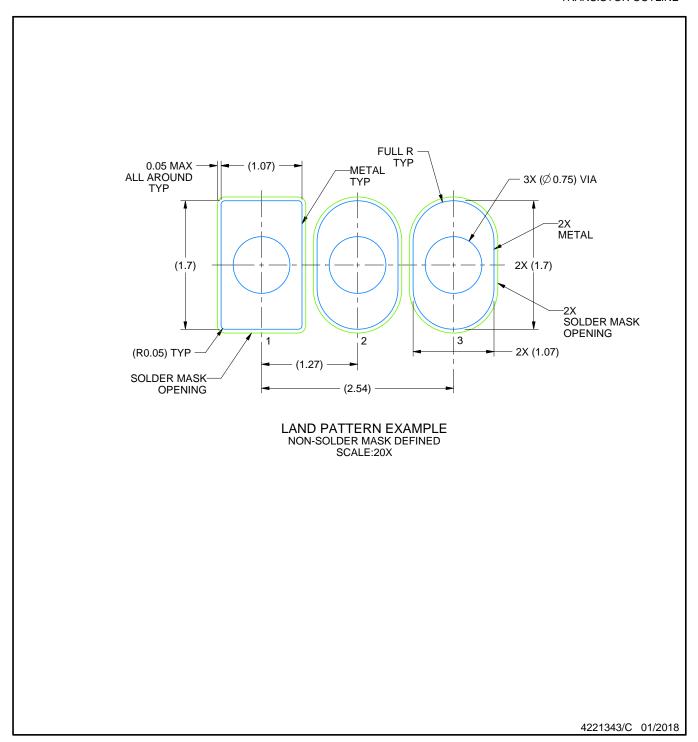
#### NOTES:

- 1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.

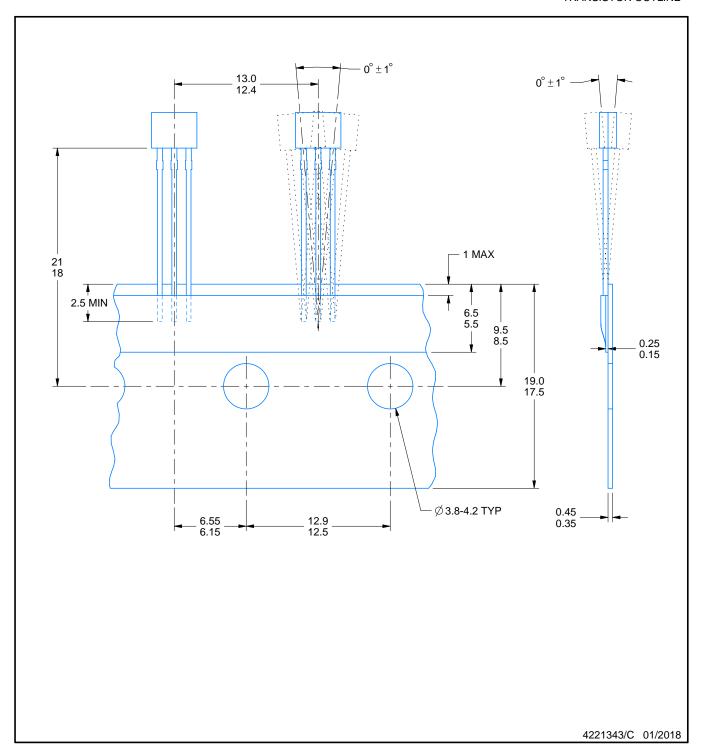
  2. This drawing is subject to change without notice.



TRANSISTOR OUTLINE

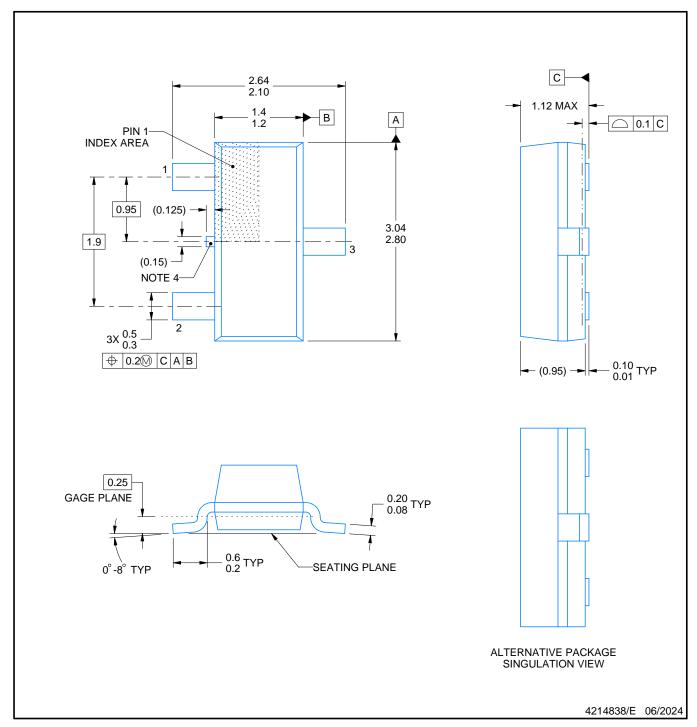


TRANSISTOR OUTLINE





SMALL OUTLINE TRANSISTOR



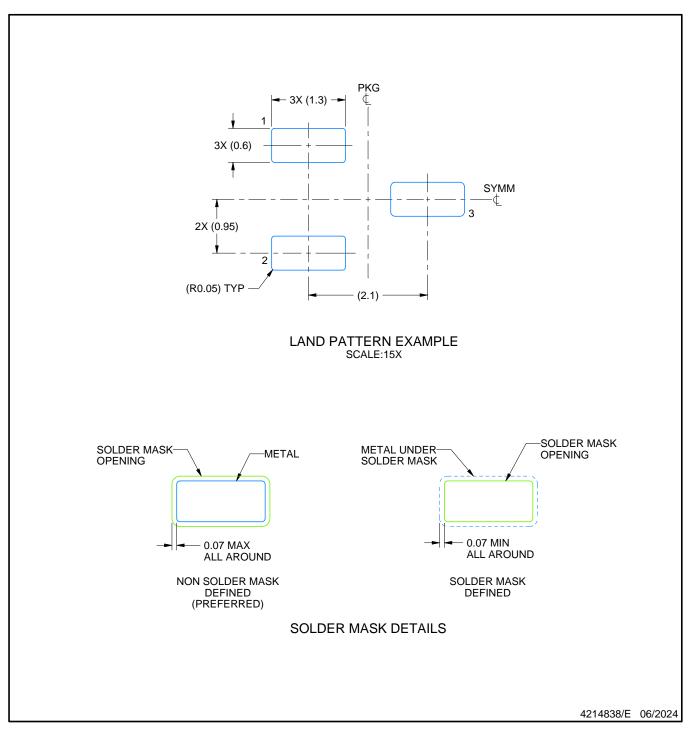
### NOTES:

- All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
   This drawing is subject to change without notice.
   Reference JEDEC registration TO-236, except minimum foot length.

- 4. Support pin may differ or may not be present.
- 5. Body dimensions do not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.25mm per side



SMALL OUTLINE TRANSISTOR

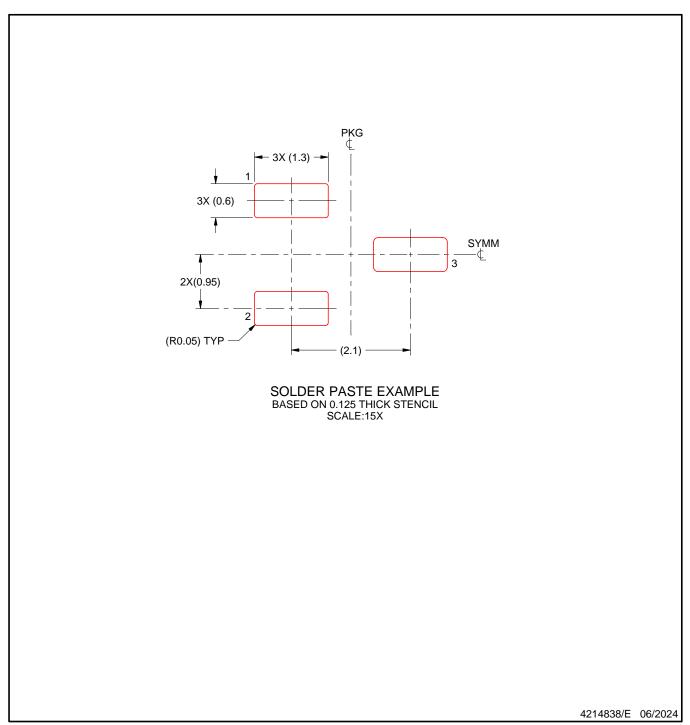


NOTES: (continued)

- 5. Publication IPC-7351 may have alternate designs.6. Solder mask tolerances between and around signal pads can vary based on board fabrication site.



SMALL OUTLINE TRANSISTOR



NOTES: (continued)

- 7. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
- 8. Board assembly site may have different recommendations for stencil design.



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