

ISO722x-Q1 Dual-Channel Digital Isolators

1 Features

- 1 and 25Mbps Signaling Rate Options
 - Low Channel-to-Channel Output Skew; 1ns Max
 - Low Pulse-Width Distortion (PWD); 1ns Max
 - Low Jitter Content; 1ns Typ at 25Mbps
- 50kV/μs Typical Transient Immunity
- Operates with 2.8V (C-Grade), 3.3V, or 5V Supplies
- 4kV ESD Protection
- –40°C to +125°C Operating Range
- Typical 28-Year Life at Rated Voltage (see [Isolation Lifetime Projection](#))
- [Safety-Related Certifications](#)
 - DIN EN IEC 60747-17 (VDE 0884-17)
 - UL 1577 component recognition program
 - IEC 61010-1, IEC 62368-1 certifications

2 Applications

- [Factory Automation](#)
 - [Modbus](#)
 - [Profibus™](#)
 - [DeviceNet™ Data Buses](#)
- [Computer Peripheral Interface](#)
- [Servo Control Interface](#)
- [Data Acquisition](#)

3 Description

The ISO7220x-Q1 and ISO7221x-Q1 family devices are dual-channel digital isolators. To facilitate PCB layout, the channels are oriented in the same direction in the ISO7220x-Q1 and in opposite directions in the ISO7221x-Q1. These devices have a logic input and output buffer separated by TI's silicon-dioxide (SiO₂) isolation barrier, providing galvanic isolation of up to 4000V_{PK} per VDE. Used in conjunction with isolated power supplies, these devices block high voltage and isolate grounds, as well as prevent noise currents on a data bus or other circuits from entering the local ground and interfering with or damaging sensitive circuitry.

A binary input signal is conditioned, translated to a balanced signal, then differentiated by the capacitive isolation barrier. Across the isolation barrier, a differential comparator receives the logic transition information, then sets or resets a flip-flop and the output circuit accordingly. A periodic update pulse is sent across the barrier to verify that the proper dc level of the output. If this dc-refresh pulse is not received every 4μs, the input is assumed to

be unpowered or not being actively driven, and the failsafe circuit drives the output to a logic high state.

The small capacitance and resulting time constant provide fast operation with signaling rates available from 0Mbps (DC) to 25Mbps (The signaling rate of a line is the number of voltage transitions that are made per second expressed in the units bps). The A-option, and C-option devices have TTL input thresholds and a noise filter at the input that prevents transient pulses from being passed to the output of the device. The M-option devices have CMOS $V_{CC}/2$ input thresholds and do not have the input noise filter and the additional propagation delay.

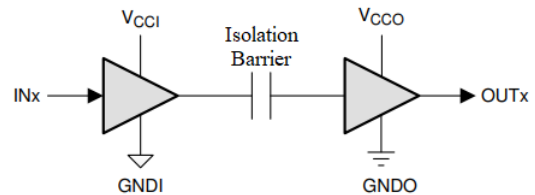
The ISO7220x-Q1 and ISO7221x-Q1 family of devices require two supply voltages of 2.8V (C-Grade), 3.3V, 5V, or any combination. All inputs are 5V tolerant when supplied from a 2.8V or 3.3V supply and all outputs are 4mA CMOS.

The ISO7220x-Q1 and ISO7221x-Q1 family of devices are characterized for operation over the ambient temperature range of –40°C to +125°C.

Package Information

PART NUMBER	PACKAGE ⁽¹⁾	BODY SIZE (NOM)	PACKAGE SIZE ⁽²⁾
ISO7220x-Q1	D (SOIC, 8)	4.90mm × 3.91mm	4.9mm × 6mm
ISO7221x-Q1			

- (1) For all available packages, see the orderable addendum at the end of the data sheet.
- (2) The package size (length × width) is a nominal value and includes pins, where applicable.



V_{CCI} and $GNDI$ are supply and ground connections respectively for the input channels.

V_{CCO} and $GNDO$ are supply and ground connections respectively for the output channels.

Simplified Schematic

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4 Pin Configuration and Functions

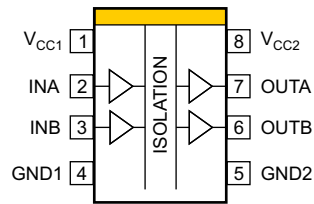


Figure 4-1. ISO7220x-Q1 D Package 8-Pin SOIC Top View

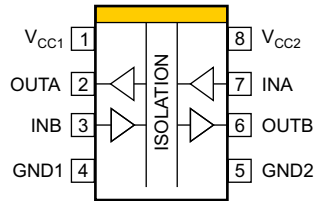


Figure 4-2. ISO7221x-Q1 D Package 8-Pin SOIC Top View

Table 4-1. Pin Functions

NAME	PIN		Type ⁽¹⁾	DESCRIPTION
	ISO7220x-Q1	ISO7221x-Q1		
INA	2	7	I	Input, channel A
INB	3	3	I	Input, channel B
GND1	4	4	—	Ground connection for V _{CC1}
GND2	5	5	—	Ground connection for V _{CC2}
OUTA	7	2	O	Output, channel A
OUTB	6	6	O	Output, channel B
V _{CC1}	1	1	—	Power supply, V _{CC1}
V _{CC2}	8	8	—	Power supply, V _{CC2}

(1) I = Input; O = Output

5 Specifications

5.1 Absolute Maximum Ratings

Parameter		Value
V _{CC}	Supply voltage ⁽¹⁾ , V _{CC1} , V _{CC2}	–0.5 V to 6 V
V _I	Voltage at IN, OUT	–0.5 V to V _{CC} + 0.5 V ⁽²⁾
I _O	Output current	±15 mA
T _J	Maximum junction temperature	150°C
T _{stg}	Storage temperature	–65°C to 150°C

- (1) All voltage values except differential I/O bus voltages are with respect to the local ground terminal (GND1 or GND2) and are peak voltage values.
 (2) Maximum voltage must not exceed 6 V.

5.2 Recommended Operating Conditions

			MIN	TYP	MAX	UNIT
V _{CC}	Supply voltage ⁽¹⁾	V _{CC1} , V _{CC2}	3		5.5	V
I _{OH}	High-level output current				4	mA
I _{OL}	Low-level output current		–4			mA
t _{ui}	Input pulse width	ISO722xA-Q1	1			µs
		ISO722xC-Q1	40			ns
1/t _{ui}	Signaling rate	ISO722xA-Q1	0		1000	kbps
		ISO722xC-Q1	0		25	Mbps
V _{IH}	High-level input voltage		2		V _{CC}	V
V _{IL}	Low-level input voltage		0		0.8	V
T _A	Ambient temperature		–40		125	°C
T _J	Operating virtual-junction temperature		–40		150	°C
H	External magnetic field-strength immunity per IEC 61000-4-8 and IEC 61000-4-9 certification				1000	A/m

- (1) For the 5-V operation, V_{CC1} or V_{CC2} is specified from 4.5 V to 5.5 V.
 For the 3.3-V operation, V_{CC1} or V_{CC2} is specified from 3 V to 3.6 V.

5.3 Safety-Related Certifications

VDE	CSA	UL
Plan to certify according to DIN EN IEC 60747-17 (VDE 0884-17)	Plan to certify according to IEC 62368-1	Plan to certify according to UL 1577 Component Recognition Program
Certificate planned	Certificate planned	Certificate planned

5.4 Thermal Information

THERMAL METRIC ⁽¹⁾		ISO7220x-Q1 ISO7221x-Q1		UNIT
		D (SOIC)		
		8 PINS		
R _{θJA}	Junction-to-ambient thermal resistance	Low-K Thermal Resistance ⁽¹⁾	212	°C/W
		High-K Thermal Resistance	122	
R _{θJC(top)}	Junction-to-case (top) thermal resistance		69.1	°C/W
R _{θJB}	Junction-to-board thermal resistance		47.7	°C/W
ψ _{JT}	Junction-to-top characterization parameter		15.2	°C/W
ψ _{JB}	Junction-to-board characterization parameter		47.2	°C/W
R _{θJC(bot)}	Junction-to-case (bottom) thermal resistance		—	°C/W

- (1) Tested in accordance with the Low-K or High-K thermal metric definitions of EIA/JESD51-3 for leaded surface mount packages.

5.5 Safety Limiting Values

Safety limiting⁽¹⁾ intends to minimize potential damage to the isolation barrier upon failure of input or output circuitry. A failure of the I/O can allow low resistance to ground or the supply and, without current limiting, dissipate sufficient power to overheat the die and damage the isolation barrier, potentially leading to secondary system failures.

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
I _S	Safety input, output, or supply current	R _{θJA} = 212°C/W, V _I = 5.5 V, T _J = 170°C, T _A = 25°C, see Thermal Information			124	mA
		R _{θJA} = 212°C/W, V _I = 3.6 V, T _J = 170°C, T _A = 25°C, see Thermal Information			190	
T _S	Safety temperature				150	°C

- (1) The safety-limiting constraint is the maximum junction temperature specified in the data sheet. The power dissipation and junction-to-air [thermal impedance](#) of the device installed in the application hardware determines the junction temperature. The assumed junction-to-air thermal resistance in the table is that of a device installed on a high-K test board for leaded surface-mount packages. The power is the recommended maximum input voltage times the current. The junction temperature is then the ambient temperature plus the power times the junction-to-air thermal resistance.

5.6 Insulation Specifications

PARAMETER		TEST CONDITIONS	VALUE	UNIT
GENERAL				
CLR	External clearance ⁽¹⁾	Shortest terminal-to-terminal distance through air	4	mm
CPG	External creepage ⁽¹⁾	Shortest terminal-to-terminal distance across the package surface	4	mm
DTI	Distance through the insulation	Minimum internal gap (internal clearance)	0.008	mm
CTI	Comparative tracking index	DIN EN 60112 (VDE 0303-11); IEC 60112	400	V
	Material group		II	
	Overvoltage category	Rated mains voltage ≤150 V _{RMS}	I-IV	
		Rated mains voltage ≤300 V _{RMS}	I-III	
		Rated mains voltage ≤400 V _{RMS}	I-II	
DIN EN IEC 60747-17 (VDE 0884-17):⁽²⁾				
V _{IORM}	Maximum repetitive peak isolation voltage	AC voltage (bipolar)	560	V _{PK}
V _{IOTM}	Maximum transient isolation voltage	V _{TEST} = V _{IOTM} t = 60 s (qualification), t = 1 s (100% production)	4000	V _{PK}
q _{pd}	Apparent charge ⁽³⁾	Method a: After I/O safety test subgroup 2/3, V _{ini} = V _{IOTM} , t _{ini} = 60 s; V _{pd(m)} = 1.2 × V _{IORM} , t _m = 10 s	≤5	pC
		Method a: After environmental tests subgroup 1, V _{ini} = V _{IOTM} , t _{ini} = 60 s; V _{pd(m)} = 1.3 × V _{IORM} , t _m = 10 s	≤5	
		Method b: At routine test (100% production); V _{ini} = 1.2 × V _{IOTM} , t _{ini} = 1 s; V _{pd(m)} = 1.5 × V _{IORM} , t _m = 1 s (method b1) or V _{pd(m)} = V _{ini} , t _m = t _{ini} (method b2)	≤5	
C _{IO}	Barrier capacitance, input to output ⁽⁴⁾	V _{IO} = 0.4 × sin(2πft), f = 1 MHz	1	pF
R _{IO}	Isolation resistance, input to output ⁽⁴⁾	V _{IO} = 500 V, T _A = 25°C	>10 ¹²	Ω
		V _{IO} = 500 V, 100°C ≤ T _A ≤ 125°C	>10 ¹¹	
		V _{IO} = 500 V at T _S = 150°C	>10 ⁹	
	Pollution degree		2	
	Climatic category		40/125/21	
UL 1577				
V _{ISO}	Withstand isolation voltage	V _{TEST} = V _{ISO} = 2500 V _{RMS} , t = 60 s (qualification); V _{TEST} = 1.2 × V _{ISO} = 3000 V _{RMS} , t = 1 s (100% production)	2500	V _{RMS}

- (1) Creepage and clearance requirements should be applied according to the specific equipment isolation standards of an application. Care should be taken to maintain the creepage and clearance distance of a board design to ensure that the mounting pads of the isolator on the printed-circuit board do not reduce this distance. Creepage and clearance on a printed-circuit board become equal in certain cases. Techniques such as inserting grooves and/or ribs on a printed circuit board are used to help increase these specifications.
- (2) This coupler is suitable for *basic electrical insulation* only within the maximum operating ratings. Compliance with the safety ratings shall be ensured by means of suitable protective circuits.
- (3) Apparent charge is electrical discharge caused by a partial discharge (pd).

(4) All pins on each side of the barrier tied together creating a two-terminal device

5.7 Electrical Characteristics

$V_{CC1} = 3.3\text{ V}$, $V_{CC2} = 5\text{ V}$ ⁽¹⁾, over recommended operating conditions (unless otherwise noted)

PARAMETER			TEST CONDITIONS	MIN	TYP	MAX	UNIT
I_{CC1}	Supply current, V_{CC1}	ISO7220x-Q1	Quiescent	$V_I = V_{CC}$ or 0 V, no load	0.6	1	mA
		ISO7221x-Q1			4.3	9.5	
		ISO7220A-Q1	1 Mbps		1	2	
		ISO7221A-Q1			5	11	
		ISO7221C-Q1	25 Mbps		6	12	
I_{CC2}	Supply current, V_{CC2}	ISO7220x-Q1	Quiescent	$V_I = V_{CC}$ or 0 V, no load	16	31	mA
		ISO7221x-Q1			8.5	17	
		ISO7220A-Q1	1 Mbps		18	32	
		ISO7221A-Q1			10	18	
		ISO7221C-Q1	25 Mbps		12	22	
V_{OH}	High-level output voltage	ISO7220x-Q1	$I_{OH} = -4\text{ mA}$	$V_{CC} - 0.8$		V	
		ISO7221x-Q1 (3.3-V side)		$V_{CC} - 0.4$			
				$V_{CC} - 0.1$			
V_{OL}	Low-level output voltage		$I_{OL} = 4\text{ mA}$	0.4		V	
			$I_{OL} = 20\text{ }\mu\text{A}$	0	0.1		
$V_{I(HYS)}$	Input threshold voltage hysteresis			150		mV	
I_{IH}	High-level input current		IN from 0 V or V_{CC}			10	μA
I_{IL}	Low-level input current		IN from 0 V or V_{CC}	-10			μA
C_I	Input capacitance to ground		IN at V_{CC} , $V_I = 0.4 \sin(2\pi ft)$, $f=2\text{MHz}$	1			pF
CMTI	Common-mode transient immunity		$V_I = V_{CC}$ or 0 V, See Figure 6-3	15	40		kV/ μs

(1) For the 5-V operation, V_{CC1} or V_{CC2} is specified from 4.5 V to 5.5 V.
 For the 3.3-V operation, V_{CC1} or V_{CC2} is specified from 3 V to 3.6 V.

5.8 Electrical Characteristics

$V_{CC1} = V_{CC2} = 3.3\text{ V}^{(1)}$, over recommended operating conditions (unless otherwise noted)

PARAMETER			TEST CONDITIONS	MIN	TYP	MAX	UNIT		
I_{CC1}	Supply current, V_{CC1}	ISO7220x-Q1	$V_I = V_{CC}$ or 0 V, no load				mA		
		Quiescent						0.6	1
		ISO7221x-Q1						4.3	9.5
		1 Mbps						1	2
		25 Mbps						5	11
I_{CC2}	Supply current, V_{CC2}	ISO7220x-Q1	$V_I = V_{CC}$ or 0 V, no load				mA		
		Quiescent						8	18
		ISO7221x-Q1						4.3	9.5
		1 Mbps						9	19
		25 Mbps						5	11
V_{OH}	High-level output voltage		$I_{OH} = -4\text{ mA}$	$V_{CC} - 0.4$	3		V		
			$I_{OH} = -20\text{ }\mu\text{A}$	$V_{CC} - 0.1$	3.3				
V_{OL}	Low-level output voltage		$I_{OL} = 4\text{ mA}$		0.2	0.4	V		
			$I_{OL} = 20\text{ }\mu\text{A}$		0	0.1			
$V_{I(HYS)}$	Input voltage hysteresis				150		mV		
I_{IH}	High-level input current		IN from 0 V or V_{CC}			10	μA		
I_{IL}	Low-level input current		IN from 0 V or V_{CC}		-10		μA		
C_I	Input capacitance to ground		IN at V_{CC} , $V_I = 0.4\text{ sin}(2\pi ft)$, $f=2\text{MHz}$		1		pF		
CMTI	Common-mode transient immunity		$V_I = V_{CC}$ or 0 V, See Figure 6-3		15	40	kV/ μs		

- (1) For the 5-V operation, V_{CC1} or V_{CC2} is specified from 4.5 V to 5.5 V.
For the 3.3-V operation, V_{CC1} or V_{CC2} is specified from 3 V to 3.6 V.

5.9 Electrical Characteristics

$V_{CC1} = 5\text{ V}$, $V_{CC2} = 3.3\text{ V}$ ⁽¹⁾, over recommended operating conditions (unless otherwise noted)

PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT
I_{CC1}	Supply current, V_{CC1}	ISO7220x-Q1	Quiescent	$V_I = V_{CC}$ or 0 V, no load	1	2	mA
		ISO7221x-Q1			8.5	17	
		ISO7220A-Q1	1 Mbps	$V_I = V_{CC}$ or 0 V, no load	2	3	
		ISO7221A-Q1			10	18	
		ISO7221C-Q1	25 Mbps	$V_I = V_{CC}$ or 0 V, no load	12	22	
I_{CC2}	Supply current, V_{CC2}	ISO7220x-Q1	Quiescent	$V_I = V_{CC}$ or 0 V, no load	8	18	mA
		ISO7221x-Q1			4.3	9.5	
		ISO7220A-Q1	1 Mbps	$V_I = V_{CC}$ or 0 V, no load	9	19	
		ISO7221A-Q1			5	11	
		ISO7221C-Q1	25 Mbps	$V_I = V_{CC}$ or 0 V, no load	6	12	
V_{OH}	High-level output voltage	ISO7220x-Q1	$I_{OH} = -4\text{ mA}$	$V_{CC} - 0.4$		V	
		ISO7221x-Q1 (5-V side)		$V_{CC} - 0.8$			
				$V_{CC} - 0.1$			
V_{OL}	Low-level output voltage		$I_{OL} = 4\text{ mA}$	0.4		V	
			$I_{OL} = 20\text{ }\mu\text{A}$	0.1			
$V_{I(HYS)}$	Input voltage hysteresis			150		mV	
I_{IH}	High-level input current		IN from 0 V to V_{CC}			10	μA
I_{IL}	Low-level input current		IN from 0 V to V_{CC}	-10			μA
C_i	Input capacitance to ground		IN at V_{CC} , $V_I = 0.4\text{ sin}(2\pi ft)$, $f=2\text{MHz}$	1			pF
CMTI	Common-mode transient immunity		$V_I = V_{CC}$ or 0 V, See Figure 6-3	15	40		kV/ μs

- (1) For the 5-V operation, V_{CC1} or V_{CC2} is specified from 4.5 V to 5.5 V.
 For the 3.3-V operation, V_{CC1} or V_{CC2} is specified from 3 V to 3.6 V.

5.10 Electrical Characteristics

V_{CC1} and V_{CC2} at 5 V⁽¹⁾, over recommended operating conditions (unless otherwise noted)

PARAMETER			TEST CONDITIONS	MIN	TYP	MAX	UNIT		
I_{CC1}	Supply current, V_{CC1}	ISO7220x-Q1	$V_I = V_{CC}$ or 0 V, no load				mA		
		Quiescent						1	2
		ISO7221x-Q1						8.5	17
		1 Mbps						2	3
		25 Mbps						10	18
ISO7221C-Q1	12	22							
I_{CC2}	Supply current, V_{CC2}	ISO7220x-Q1	$V_I = V_{CC}$ or 0 V, no load				mA		
		Quiescent						16	31
		ISO7221x-Q1						8.5	17
		1 Mbps						17	32
		25 Mbps						10	18
ISO7221C-Q1	12	22							
V_{OH}	High-level output voltage		$I_{OH} = -4$ mA	$V_{CC} - 0.8$	4.6		V		
			$I_{OH} = -20$ μ A	$V_{CC} - 0.1$	5				
V_{OL}	Low-level output voltage		$I_{OL} = 4$ mA		0.2	0.4	V		
			$I_{OL} = 20$ μ A		0	0.1			
$V_{I(HYS)}$	Input voltage hysteresis				150		mV		
I_{IH}	High-level input current		IN from 0 V to V_{CC}			10	μ A		
I_{IL}	Low-level input current		IN from 0 V to V_{CC}		-10		μ A		
C_I	Input capacitance to ground		IN at V_{CC} , $V_I = 0.4 \sin(2\pi ft)$, $f=2$ MHz		1		pF		
CMTI	Common-mode transient immunity		$V_I = V_{CC}$ or 0 V, See Figure 6-3		25	50	kV/ μ s		

- (1) For the 5-V operation, V_{CC1} or V_{CC2} is specified from 4.5 V to 5.5 V.
For the 3.3-V operation, V_{CC1} or V_{CC2} is specified from 3 V to 3.6 V.

5.11 Switching Characteristics

$V_{CC1} = 3.3$ V \pm 10%, $V_{CC2} = 5$ V \pm 10%, over recommended operating conditions (unless otherwise noted)

PARAMETER			TEST CONDITIONS	MIN	TYP	MAX	UNIT
t_{pLH} , t_{pHL}	Propagation delay	ISO722xA-Q1	See Figure 6-1	268	395	605	ns
PWD	Pulse-width distortion $ t_{pHL} - t_{pLH} $ ⁽¹⁾						
t_{pLH} , t_{pHL}	Propagation delay	ISO722xC-Q1	See Figure 6-1	21	36	48	ns
PWD	Pulse-width distortion $ t_{pHL} - t_{pLH} $ ⁽¹⁾						
$t_{sk(pp)}$	Part-to-part skew ⁽²⁾	ISO722xA-Q1				190	ns
		ISO722xC-Q1				10	
$t_{sk(o)}$	Channel-to-channel output skew ⁽³⁾	ISO7220A-Q1			3	15	ns
t_r	Output signal rise time		See Figure 6-1		2.3		ns
t_f	Output signal fall time		See Figure 6-1		2.3		ns
t_{fs}	Failsafe output delay time from input power loss		See Figure 6-2		3		μ s

- (1) Also referred to as pulse skew.
(2) $t_{sk(pp)}$ is the magnitude of the difference in propagation delay times between any specified terminals of two devices when both devices operate with the same supply voltages, at the same temperature, and have identical packages and test circuits.

- (3) $t_{sk(o)}$ is the skew between specified outputs of a single device with all driving inputs connected together and the outputs switching in the same direction while driving identical specified loads.

5.12 Switching Characteristics

$V_{CC1} = 5\text{ V} \pm 10\%$, $V_{CC2} = 3.3\text{ V} \pm 10\%$, over recommended operating conditions (unless otherwise noted)

PARAMETER			TEST CONDITIONS	MIN	TYP	MAX	UNIT
t_{pLH} , t_{pHL}	Propagation delay	ISO722xA-Q1	See Figure 6-1	253	410	585	ns
PWD	Pulse-width distortion $ t_{pHL} - t_{pLH} ^{(1)}$						
t_{pLH} , t_{pHL}	Propagation delay	ISO722xC-Q1	See Figure 6-1	21	36	48	ns
PWD	Pulse-width distortion $ t_{pHL} - t_{pLH} ^{(1)}$						
$t_{sk(pp)}$	Part-to-part skew ⁽²⁾	ISO722xA-Q1				180	ns
		ISO722xC-Q1				10	
$t_{sk(o)}$	Channel-to-channel output skew ⁽³⁾	ISO7220A-Q1			3	15	ns
t_r	Output signal rise time		See Figure 6-1		2.3		ns
t_f	Output signal fall time		See Figure 6-1		2.3		
t_{fs}	Failsafe output delay time from input power loss		See Figure 6-2		3		μs

(1) Also referred to as pulse skew.

(2) $t_{sk(pp)}$ is the magnitude of the difference in propagation delay times between any specified terminals of two devices when both devices operate with the same supply voltages, at the same temperature, and have identical packages and test circuits.

(3) $t_{sk(o)}$ is the skew between specified outputs of a single device with all driving inputs connected together and the outputs switching in the same direction while driving identical specified loads.

5.13 Switching Characteristics

$V_{CC1} = V_{CC2} = 3.3\text{ V} \pm 10\%$, over recommended operating conditions (unless otherwise noted)

PARAMETER			TEST CONDITIONS	MIN	TYP	MAX	UNIT
t_{pLH} , t_{pHL}	Propagation delay	ISO722xA-Q1	See Figure 6-1	267	400	610	ns
PWD	Pulse-width distortion $ t_{pHL} - t_{pLH} ^{(1)}$						
t_{pLH} , t_{pHL}	Propagation delay	ISO722xC-Q1	See Figure 6-1	23	40	52	ns
PWD	Pulse-width distortion $ t_{pHL} - t_{pLH} ^{(1)}$						
$t_{sk(pp)}$	Part-to-part skew ⁽²⁾	ISO722xA-Q1				190	ns
		ISO722xC-Q1				10	
$t_{sk(o)}$	Channel-to-channel output skew ⁽³⁾	ISO7220A-Q1			3	15	ns
t_r	Output signal rise time		See Figure 6-1		2.3		ns
t_f	Output signal fall time		See Figure 6-1		2.3		ns
t_{fs}	Failsafe output delay time from input power loss		See Figure 6-2		3		μs

(1) Also referred to as pulse skew.

(2) $t_{sk(pp)}$ is the magnitude of the difference in propagation delay times between any specified terminals of two devices when both devices operate with the same supply voltages, at the same temperature, and have identical packages and test circuits.

(3) $t_{sk(o)}$ is the skew between specified outputs of a single device with all driving inputs connected together and the outputs switching in the same direction while driving identical specified loads.

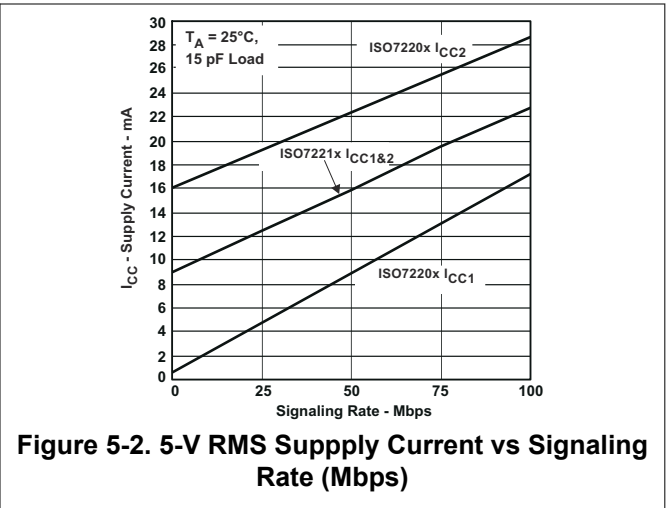
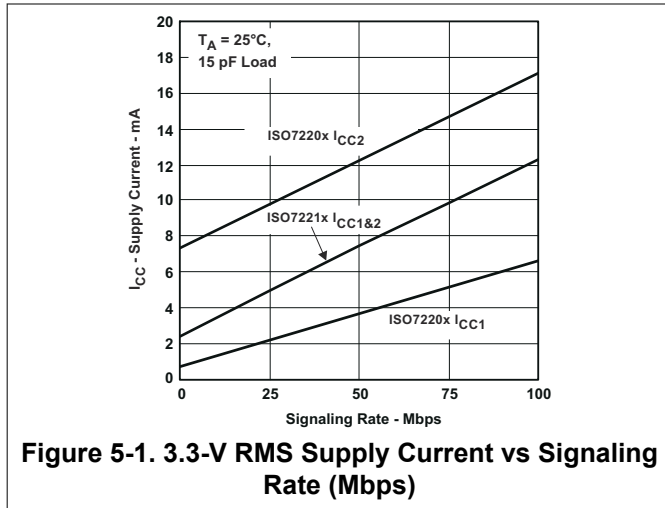
5.14 Switching Characteristics

$V_{CC1} = V_{CC2} = 5\text{ V} \pm 10\%$, over recommended operating conditions (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT	
t_{pLH}, t_{pHL}	Propagation delay	ISO722xA-Q1 See Figure 6-1	252	405	600	ns	
PWD	Pulse-width distortion $ t_{pHL} - t_{pLH} ^{(1)}$						
t_{pLH}, t_{pHL}	Propagation delay	ISO722xC-Q1 See Figure 6-1	21	32	42	ns	
PWD	Pulse-width distortion $ t_{pHL} - t_{pLH} ^{(1)}$						
$t_{sk(pp)}$	Part-to-part skew ⁽²⁾	ISO722xA-Q1			180	ns	
		ISO722xC-Q1			10		
$t_{sk(o)}$	Channel-to-channel output skew ⁽³⁾	ISO7220A-Q1			3	15	ns
t_r	Output signal rise time	See Figure 6-1			2.3	ns	
t_f	Output signal fall time	See Figure 6-1			2.3	ns	
t_{fs}	Failsafe output delay time from input power loss	See Figure 6-2			3	μs	

- (1) Also referred to as pulse skew.
- (2) $t_{sk(pp)}$ is the magnitude of the difference in propagation delay times between any specified terminals of two devices when both devices operate with the same supply voltages, at the same temperature, and have identical packages and test circuits.
- (3) $t_{sk(o)}$ is the skew between specified outputs of a single device with all driving inputs connected together and the outputs switching in the same direction while driving identical specified loads.

5.15 Typical Characteristics



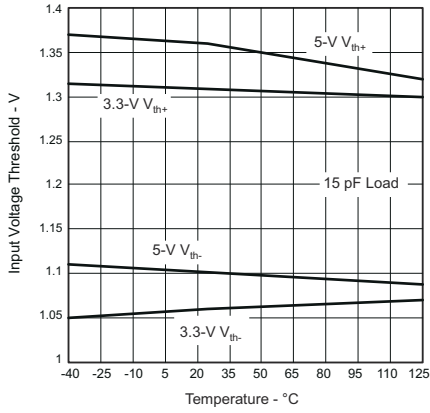


Figure 5-3. ISO722xA-Q1 and ISO722xC -Q1 Input Voltage Low-to-High Switching Threshold vs Free-Air Temperature

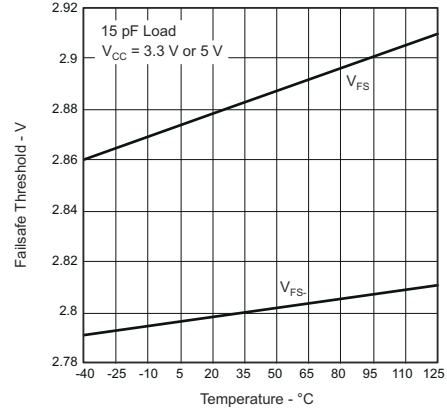


Figure 5-4. V_{CC} Failsafe Threshold vs Free-Air Temperature

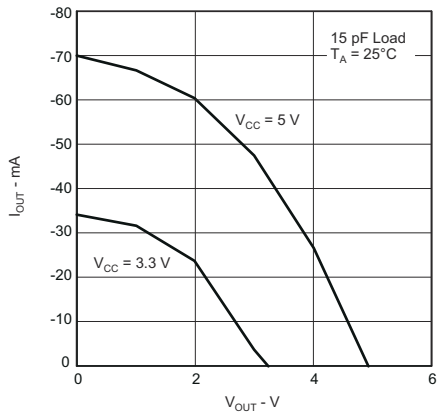


Figure 5-5. High-Level Output Current vs High-Level Output Voltage

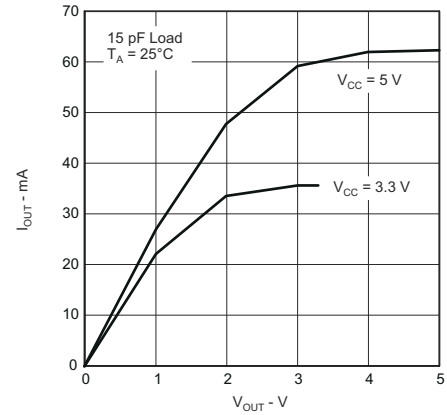
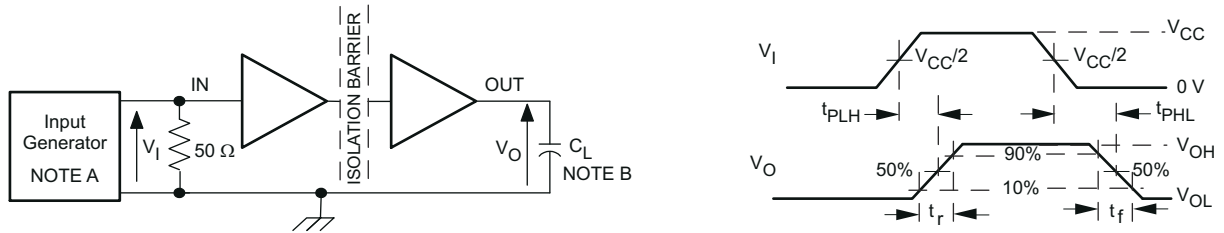


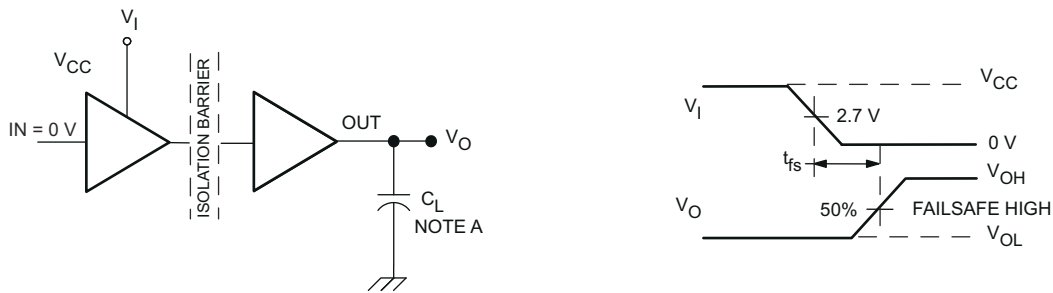
Figure 5-6. Low-Level Output Current vs Low-Level Output Voltage

6 Parameter Measurement Information



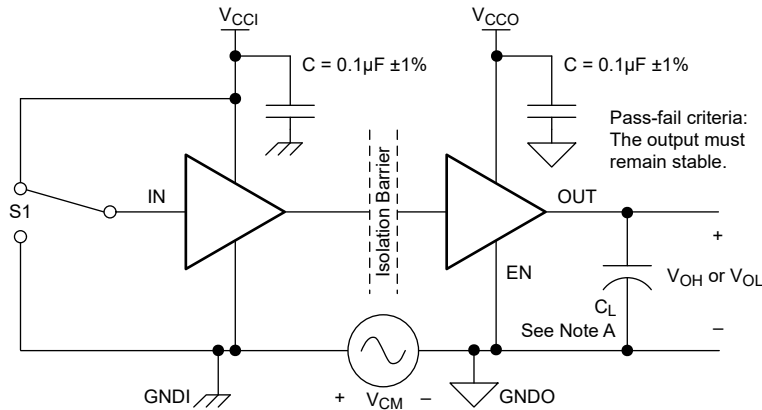
- A. The input pulse is supplied by a generator having the following characteristics: PRR \leq 50 kHz, 50% duty cycle, $t_r \leq$ 3 ns, $t_f \leq$ 3 ns, $Z_O = 50 \Omega$.
- B. $C_L = 15$ pF and includes instrumentation and fixture capacitance within $\pm 20\%$.

Figure 6-1. Switching Characteristic Test Circuit and Voltage Waveforms



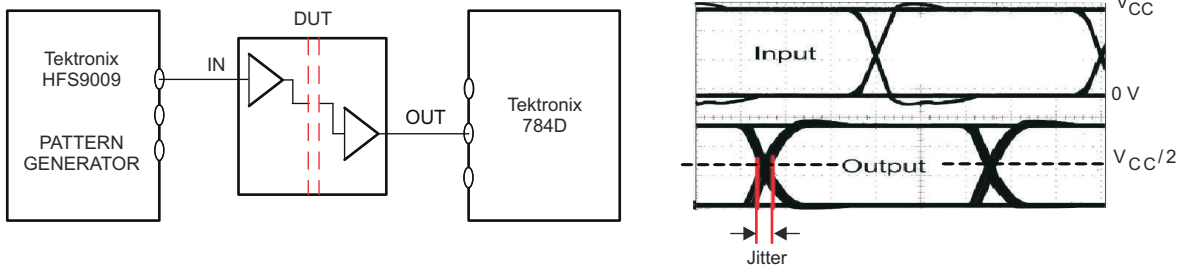
- A. $C_L = 15$ pF and includes instrumentation and fixture capacitance within $\pm 20\%$.

Figure 6-2. Failsafe Delay Time Test Circuit and Voltage Waveforms



- A. $C_L = 15$ pF and includes instrumentation and fixture capacitance within $\pm 20\%$.

Figure 6-3. Common-Mode Transient Immunity Test Circuit



PRBS bit pattern run length is $2^{16} - 1$. Transition time is 800 ps.

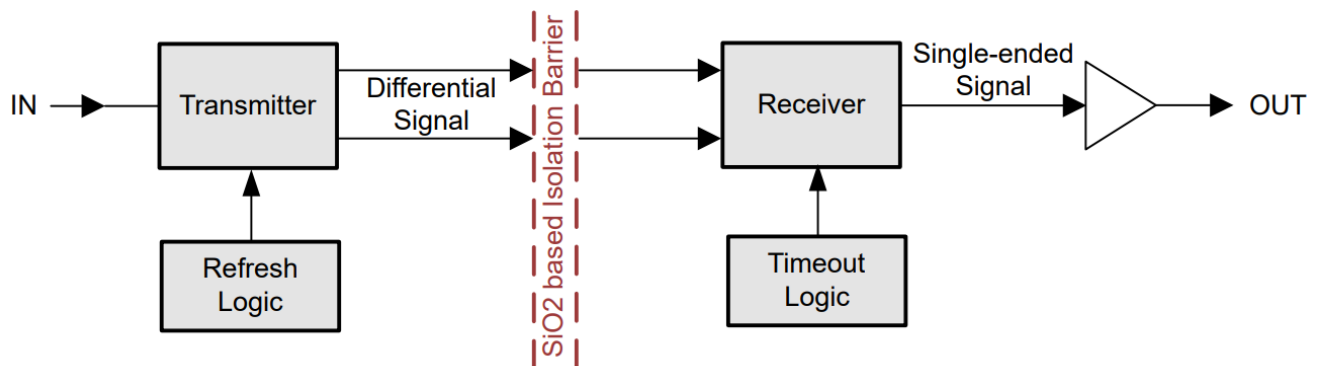
Figure 6-4. Peak-to-Peak Eye-Pattern Jitter Test Circuit and Voltage Waveform

7 Detailed Description

7.1 Overview

The ISO722xx-Q1 family of devices transmit digital data across a silicon dioxide based isolation barrier. The digital input signal (IN) of the device is sampled by a transmitter and at every data edge the transmitter sends a corresponding differential signal across the isolation barrier. When the input signal is static, the refresh logic periodically sends the necessary differential signal from the transmitter. On the other side of the isolation barrier, the receiver converts the differential signal into a single-ended signal which is output on the OUT pin through a buffer. If the receiver does not receive a data or refresh signal, the timeout logic detects the loss of signal or power from the input side and drives the output to the default level.

7.2 Functional Block Diagram



7.3 Feature Description

Table 7-1 provides an overview of the device features.

Table 7-1. Device Features

PART NUMBER	MAXIMUM SIGNALING RATE	INPUT THRESHOLD	CHANNEL DIRECTION
ISO7220A-Q1	1 Mbps	≅ 1.5 V (TTL) (CMOS compatible)	2/0
ISO7221A-Q1	1 Mbps	≅ 1.5 V (TTL) (CMOS compatible)	1/1
ISO7221C-Q1	25 Mbps	≅ 1.5 V (TTL) (CMOS compatible)	

7.4 Device Functional Modes

The ISO7220x-Q1 and ISO7221x-Q1 family of devices functional modes are listed in Table 7-2.

Table 7-2. ISO7220x-Q1 or ISO7221x-Q1 Function Table

INPUT SIDE $V_{CC}^{(1)}$	OUTPUT SIDE V_{CC}	INPUT (IN)	OUTPUT (OUT)
PU	PU	H	H
		L	L
		Open	H
PD	PU	X	H
X	PD	X	Undetermined

- (1) PU = Powered Up ($V_{CC} \geq 3.0$ V), PD = Powered Down ($V_{CC} \leq 2.5$ V), X = Irrelevant, H = High Level, L = Low Level

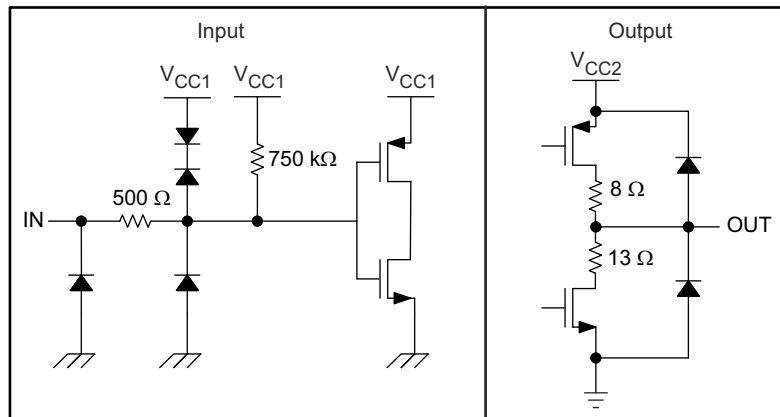


Figure 7-1. Device I/O Schematics

8 Application and Implementation

Note

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes, as well as validating and testing their design implementation to confirm system functionality.

8.1 Application Information

The ISO7220x and ISO7221x family devices use single-ended TTL or CMOS-logic switching technology. The supply voltage range is from 3 V (2.8 V for C-grade) to 5.5 V for both supplies, V_{CC1} and V_{CC2} . When designing with digital isolators, because of the single-ended design structure, digital isolators do not conform to any specific interface standard and are only intended for isolating single-ended CMOS or TTL digital signal lines. The isolator is typically placed between the data controller (that is, μC or UART), and a data converter or a line transceiver, regardless of the interface type or standard.

8.2 Typical Application

The ISO7221x-Q1 family of devices can be used with Texas Instruments' mixed signal micro-controller, digital-to-analog converter, transformer driver, and voltage regulator to create an isolated 4- to 20-mA current loop.

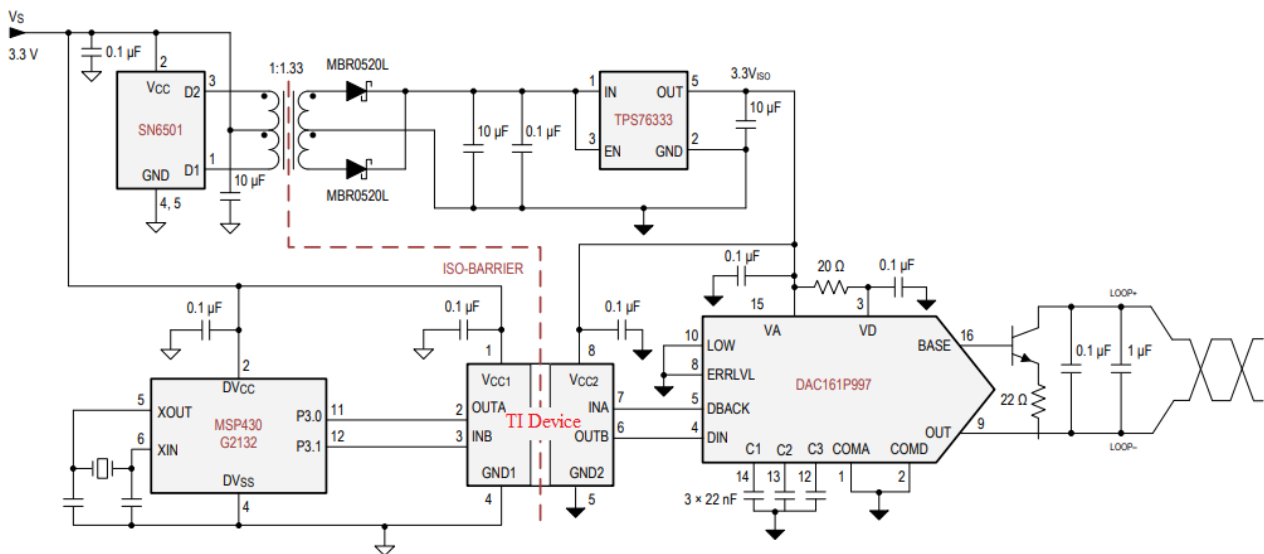


Figure 8-1. Isolated 4- to 20-mA Current Loop

8.2.1 Design Requirements

Unlike optocouplers, which require external components to improve performance, provide bias (or limit current), the ISO7220x-Q1 and ISO7221x-Q1 devices require only two external bypass capacitors to operate.

8.2.2 Detailed Design Procedure

Figure 8-2 and Figure 8-3 show the hookup of a typical ISO7220x-Q1 and ISO7221x-Q1 circuit. The only external components are two bypass capacitors.

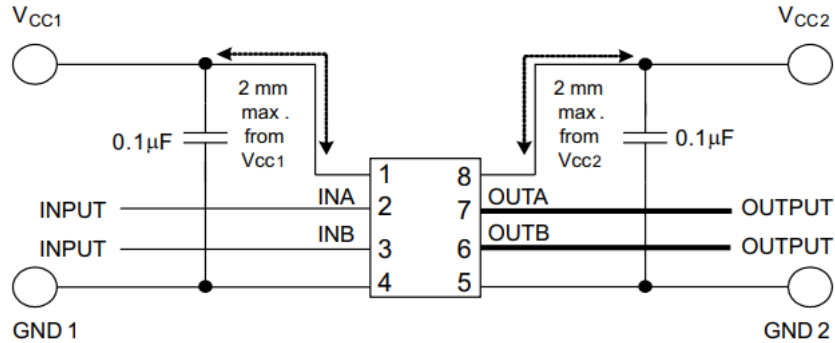


Figure 8-2. Typical ISO7220x-Q1 Circuit Hook-Up

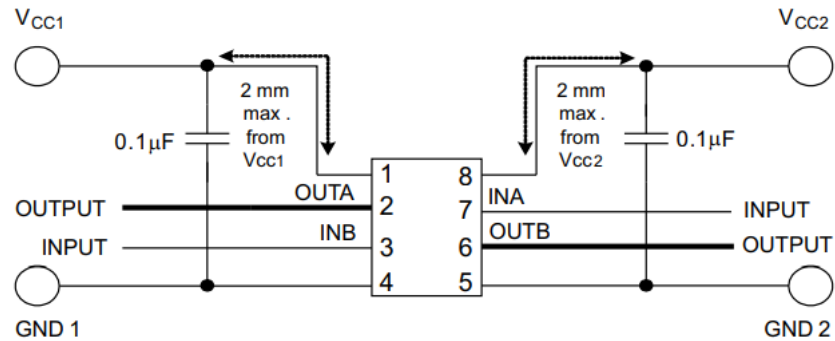


Figure 8-3. Typical ISO7221x-Q1 Circuit Hook-Up

8.2.3 Insulation Lifetime

At maximum working voltage, the isolation barrier of the ISO72x and ISO72xM family of devices has more than 28 years of life.

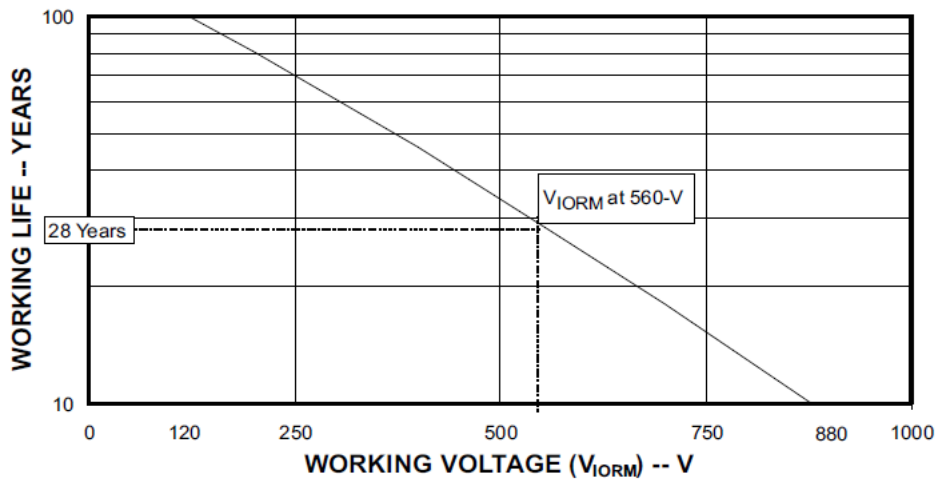


Figure 8-4. Insulation Lifetime Projection

8.3 Power Supply Recommendations

To help provide reliable operation at all data rates and supply voltages, a 0.1- μ F bypass capacitor is recommended at input and output supply pins (V_{CC1} and V_{CC2}). The capacitors must be placed as close to the supply pins as possible. If only a single primary-side power supply is available in an application, isolated power can be generated for the secondary-side with the help of a transformer driver such as Texas Instruments SN6501 device. For such applications, detailed power supply design and transformer selection recommendations are available in [SN6501 Transformer Driver for Isolated Power Supplies](#).

8.4 Layout

8.4.1 Layout Guidelines

A minimum of four layers are required to accomplish a low EMI PCB design (see [Figure 8-5](#)). Layer stacking must be in the following order (top-to-bottom): high-speed signal layer, ground plane, power plane and low-frequency signal layer.

- Route the high-speed traces on the top layer to avoid the use of vias (and the introduction of the inductances) and allow for clean interconnects between the isolator and the transmitter and receiver circuits of the data link.
- Place a solid ground plane next to the high-speed signal layer to establish controlled impedance for transmission line interconnects and provide an excellent low-inductance path for the return current flow.
- Place the power plane next to the ground plane to create additional high-frequency bypass capacitance of approximately 100 pF/in².
- Route the slower speed control signals on the bottom layer to allow for greater flexibility as these signal links typically have margin to tolerate discontinuities such as vias.

If an additional supply voltage plane or signal layer is needed, add a second power or ground plane system to the stack to keep the planes symmetrical. Adding a second plane system to the stack makes the stack mechanically stable and prevents warping. The power and ground plane of each power system can be placed closer together, thus increasing the high-frequency bypass capacitance significantly.

For detailed layout recommendations, refer to the [Digital Isolator Design Guide](#).

8.4.1.1 PCB Material

For digital circuit boards operating at less than 150 Mbps, (or rise and fall times greater than 1 ns), and trace lengths of up to 10 inches, use standard FR-4 UL94V-0 printed circuit board. This PCB is preferred over cheaper alternatives because of lower dielectric losses at high frequencies, less moisture absorption, greater strength and stiffness, and the self-extinguishing flammability-characteristics.

8.4.2 Layout Example

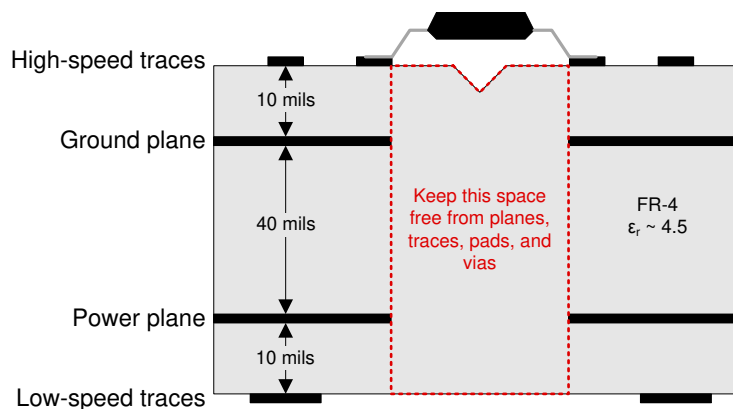


Figure 8-5. Recommended Layer Stack

9 Device and Documentation Support

9.1 Device Support

9.1.1 Development Support

For development support, refer to:

- [AC-mains LED Lighting with DALI DMX512 & Power Line Communications Reference Design](#)
- [Industrial Servo Drive and AC Inverter Drive Reference Design](#)
- [Low-Cost Single/Dual-Phase Isolated Electricity Measurement Reference Design](#)
- [Noise Tolerant Capacitive Touch HMI Reference Design](#)
- [Type 2 PoE PSE, 6kV Lightning Surge Reference Design](#)

9.2 Documentation Support

9.2.1 Related Documentation

For related documentation, see the following:

- Texas Instruments, [DAC161P997 Single-Wire 16-bit DAC for 4- to 20-mA Loops data sheet](#)
- Texas Instruments, [Digital Isolator Design Guide](#)
- Texas Instruments, [High-Voltage Lifetime of the ISO72x Family of Digital Isolators application report](#)
- Texas Instruments, [Isolation Glossary](#)
- Texas Instruments, [MSP430G2x32 Mixed Signal Microcontroller data sheet](#)
- Texas Instruments, [SN6501 Transformer Driver for Isolated Power Supplies data sheet](#)
- Texas Instruments, [TPS763xx Low-Power 150-mA Low-Dropout Linear Regulators data sheet](#)

9.3 Receiving Notification of Documentation Updates

To receive notification of documentation updates, navigate to the device product folder on [ti.com](#). Click on *Notifications* to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

9.4 Support Resources

[TI E2E™ support forums](#) are an engineer's go-to source for fast, verified answers and design help — straight from the experts. Search existing answers or ask your own question to get the quick design help you need.

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9.6 Electrostatic Discharge Caution



This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

9.7 Glossary

[TI Glossary](#) This glossary lists and explains terms, acronyms, and definitions.

10 Revision History

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

Changes from Revision D (April 2020) to Revision E (November 2024)	Page
• Updated the content throughout the document to better align with the commercial version of the device	1
• Updated reference from capacitive isolation to isolation barrier throughout the document.....	1
• Updated VDE V 0884-11 to DIN VDE 0884-17 throughout the document.....	1
• Updated the numbering format for tables, figures, and cross-references throughout the document.....	1
• Updated Thermal Characteristics, Safety Limiting Values, and Thermal Derating Curves to provide more accurate system-level thermal calculations.....	4
• Updated electrical and switching characteristics to match device performance.....	6
• Added the <i>Detailed Description</i> , <i>Overview</i> , <i>Feature Description</i> , <i>Functional Block Diagram</i> , and <i>Device Functional Modes</i> sections.....	16
• Added the <i>Typical Application</i> , <i>Power Supply Recommendations</i> , and <i>Layout</i> sections.....	18

Changes from Revision C (May 2012) to Revision D (April 2020)	Page
• Change standard names From: 'IEC 60747-5-2 (VDE 0884, Rev 2), IEC 61010-1' To: 'DIN VDE V 0884-11:2017-01, DIN EN 61010-1' and add 'IEC 62368-1' in FEATURES	1
• Made editorial and cosmetic changes throughout the document	1
• Deleted typical values (TYP) for 'Input pulse width' and 'Signaling rate' specifications in Recommended Operating Conditions table.....	4
• Added 'Ambient temperature' specification in Recommended Operating Conditions table.....	4
• Changed 'Propagation delay' maximum (MAX) limit for ISO722xA From: 480 ns To: 605 ns in Switching Characteristics at $V_{CC1} = 3.3\text{ V} \pm 10\%$, $V_{CC2} = 5\text{ V} \pm 10\%$	9
• Changed 'Pulse-width distortion' maximum (MAX) limit for ISO722xA From: 18 ns To: 22 ns in Switching Characteristics at $V_{CC1} = 3.3\text{ V} \pm 10\%$, $V_{CC2} = 5\text{ V} \pm 10\%$	9
• Changed 'ISO722xA' to 'ISO7220A' and deleted 'ISO722xC' row from 'Channel-to-channel output skew' specification in Switching Characteristics at $V_{CC1} = 3.3\text{ V} \pm 10\%$, $V_{CC2} = 5\text{ V} \pm 10\%$	9
• Changed 'Propagation delay' maximum (MAX) limit for ISO722xA From: 480 ns To: 585 ns in Switching Characteristics at $V_{CC1} = 5\text{ V} \pm 10\%$, $V_{CC2} = 3.3\text{ V} \pm 10\%$	11
• Changed 'ISO722xA' to 'ISO7220A' and deleted 'ISO722xC' row from 'Channel-to-channel output skew' specification in Switching Characteristics at $V_{CC1} = 5\text{ V} \pm 10\%$, $V_{CC2} = 3.3\text{ V} \pm 10\%$	11
• Changed 'Pulse-width distortion' maximum (MAX) limit for ISO722xA From: 14 ns To: 18 ns in Switching Characteristics at $V_{CC1} = 5\text{ V} \pm 10\%$, $V_{CC2} = 3.3\text{ V} \pm 10\%$	11
• Changed 'Propagation delay' maximum (MAX) limit for ISO722xA From: 485 ns To: 610 ns in Switching Characteristics at $V_{CC1} = V_{CC2} = 3.3\text{ V} \pm 10\%$	11
• Changed 'ISO722xA' to 'ISO7220A' and deleted 'ISO722xC' row from 'Channel-to-channel output skew' specification in Switching Characteristics at $V_{CC1} = V_{CC2} = 3.3\text{ V} \pm 10\%$	11
• Changed 'Pulse-width distortion' maximum (MAX) limit for ISO722xA From: 18 ns To: 22 ns in Switching Characteristics at $V_{CC1} = V_{CC2} = 3.3\text{ V} \pm 10\%$	11
• Changed 'ISO722xA' to 'ISO7220A' and deleted 'ISO722xC' row from 'Channel-to-channel output skew' specification in Switching Characteristics at $V_{CC1} = V_{CC2} = 5\text{ V} \pm 10\%$	12
• Changed 'Pulse-width distortion' maximum (MAX) limit for ISO722xA From: 14 ns To: 18 ns in Switching Characteristics at $V_{CC1} = V_{CC2} = 5\text{ V} \pm 10\%$	12
• Changed 'Propagation delay' maximum (MAX) limit for ISO722xA From: 480 ns To: 600 ns in Switching Characteristics at $V_{CC1} = V_{CC2} = 5\text{ V} \pm 10\%$	12

11 Mechanical, Packaging, and Orderable Information

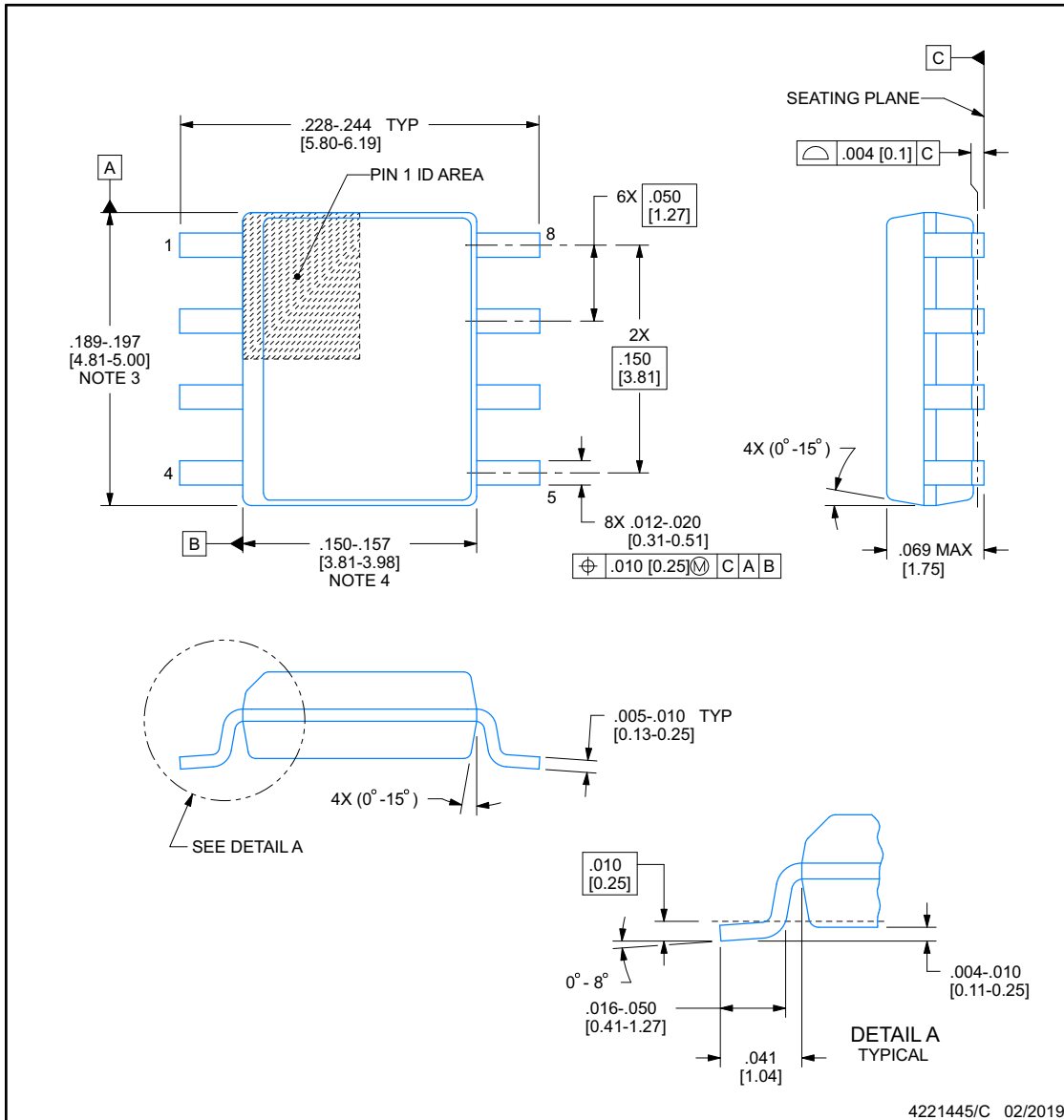
The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.



D0008B

PACKAGE OUTLINE
SOIC - 1.75 mm max height

SMALL OUTLINE INTEGRATED CIRCUIT



NOTES:

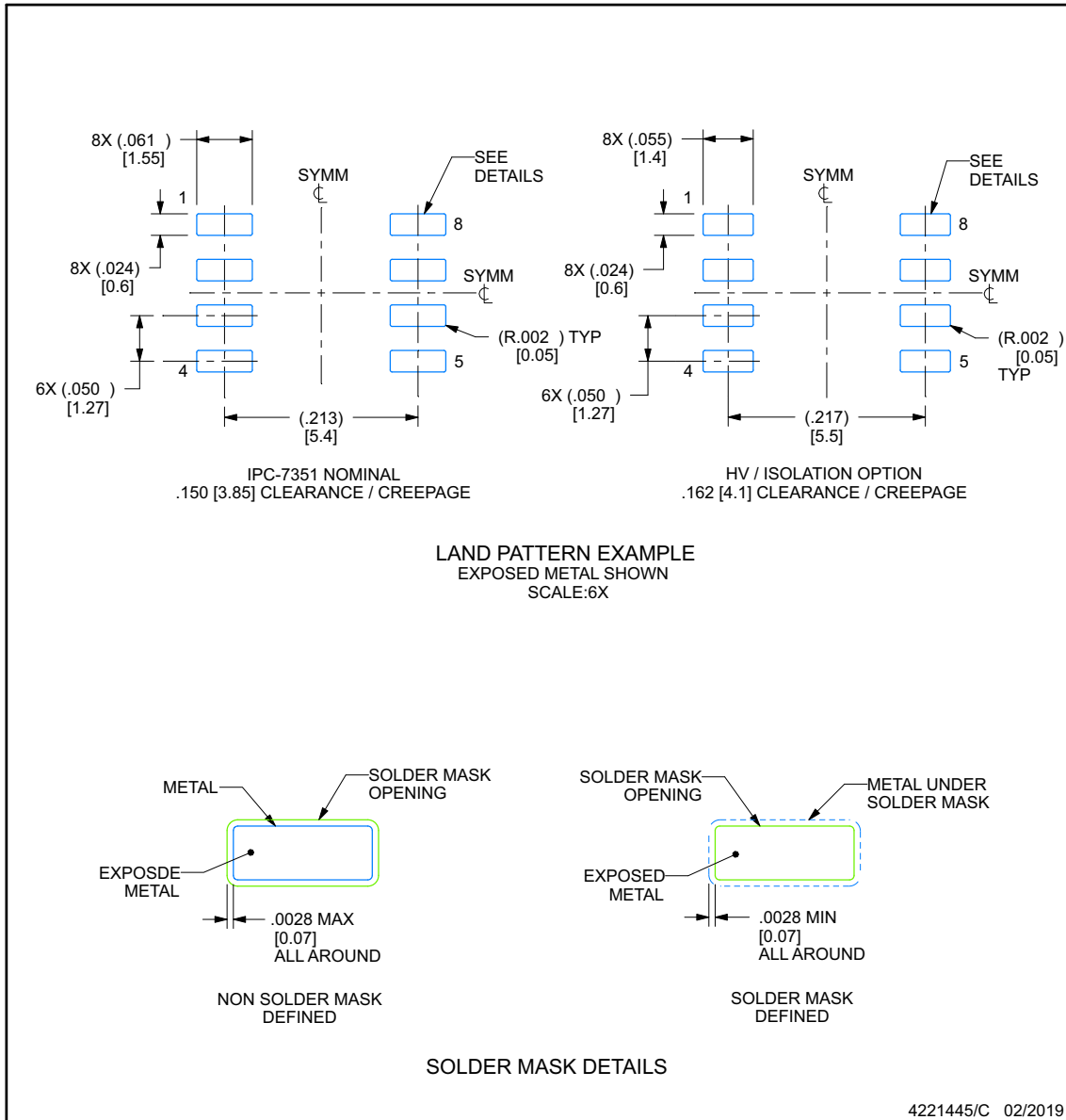
1. Linear dimensions are in inches [millimeters]. Dimensions in parenthesis are for reference only. Controlling dimensions are in inches. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed $.006 [0.15]$, per side.
4. This dimension does not include interlead flash.
5. Reference JEDEC registration MS-012, variation AA.

EXAMPLE BOARD LAYOUT

D0008B

SOIC - 1.75 mm max height

SMALL OUTLINE INTEGRATED CIRCUIT



NOTES: (continued)

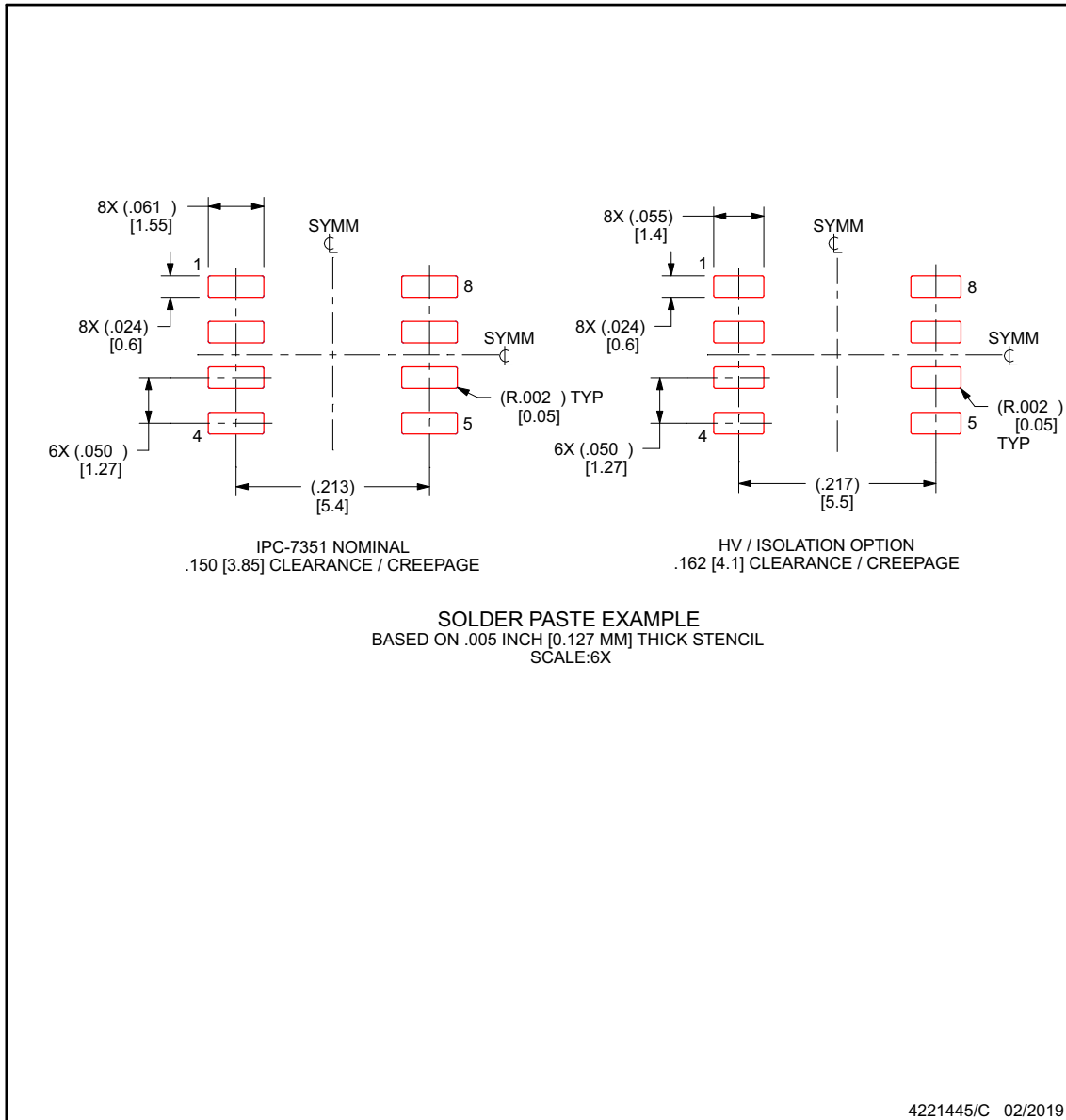
- 6. Publication IPC-7351 may have alternate designs.
- 7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.

EXAMPLE STENCIL DESIGN

D0008B

SOIC - 1.75 mm max height

SMALL OUTLINE INTEGRATED CIRCUIT



NOTES: (continued)

- 8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
- 9. Board assembly site may have different recommendations for stencil design.

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