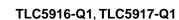


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TLC591x-Q1 8-Bit Constant-Current LED Sink Drivers

Technical

Documents

1 Features

- Qualified for Automotive Applications
- AEC-Q100 Qualified With the Following Results:
 - Device Temperature Grade 1: -40°C to 125°C Ambient Operating Temperature Range
 - Device HBM ESD Classification Level 1C
 - **Device CDM ESD Classification Level C4**
- **Eight Constant-Current Output Channels**
- **Output Current Adjusted Through External** Resistor
- Constant Output Current Range: 5 mA to 120 mA
- Constant Output Current Invariant to Load Voltage Change
- Open Load, Short Load, and Overtemperature Detection
- 256-Step Programmable Global Current Gain
- Excellent Output Current Accuracy: •
 - Between Channels: < ±3% (Maximum)
- Between ICs: $< \pm 6\%$ (Maximum)
- Fast Response of Output Current
- 30-MHz Clock Frequency
- Schmitt Trigger Input
- 3.3-V or 5-V Supply Voltage
- Thermal Shutdown for Overtemperature Protection

2 Applications

Tools &

- General LED Lighting Applications
- LED Display Systems
- LED Signage ٠
- Automotive LED Lighting
- White Goods .
- Gaming Machines and Entertainment

3 Description

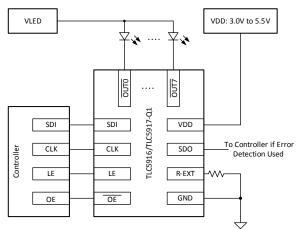
The TLC591x-Q1 Constant-Current LED Sink Drivers is designed to work alone or cascaded. Because each output is independently controlled, they can be programmed to be on or off by the user. The high LED voltage (VLED) allows for the use of one LED per output or multiple LEDs on a single string. With independently controlled outputs supplied with constant current, the LEDs can be combined in parallel to create higher currents on a single string. The constant sink current for all channels is set through a single external resistor. This allows different LED drivers in the same application to sink various currents which provides optional implementation of multicolor LEDs. An additional advantage of the independent outputs is the ability to leave unused channels floating. The flexibility of the TLC591x-Q1 LED driver is ideal for applications such as (but not limited to): automotive LED lighting, 7segment displays, scrolling single-color displays, gaming machines, white goods, video billboards, and video panels.

Device Information⁽¹⁾

PART NUMBER	PACKAGE	BODY SIZE (NOM)
TLC591x-Q1	SOIC (16)	9.90 mm × 3.91 mm

(1) For all available packages, see the orderable addendum at the end of the data sheet.

Single Implementation of TLC591x-Q1 Device



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Added ESD Ratings table, Feature Description section, Device Functional Modes, Application and Implementation section, Power Supply Recommendations section, Layout section, Device and Documentation Support section, and Mechanical, Packaging, and Orderable Information section1

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5 Device Comparison Table

DEVICE ⁽¹⁾	OVERTEMPERATURE DETECTION	OPEN-LOAD DETECTION	SHORT TO GND DETECTION	SHORT TO V _{LED} DETECTION
TLC5916-Q1	Х	Х	Х	
TLC5917-Q1	Х	Х	Х	Х

(1) The device has one error register for all these conditions (1 error bit per channel).

6 Pin Configuration and Functions

	1 6- I	Packa Pin S op Vi	OIC	;
GND [SDI [CLK [LE(ED1) [OUT0 [OUT1 [OUT2 [1 2 3 4 5 6 7	σ	16 15 14 13 12 11 10] VDD] R-EXT] SDO] OE(ED2)] OUT7] OUT6] OUT5
<u>оитз</u> [8		9] <u>OUT4</u>

Pin Functions

Р	IN	1/0	DESCRIPTION
NAME	NO.	1/0	DESCRIPTION
CLK	3	I	Clock input for data shift on rising edge
GND	1	—	Ground for control logic and current sink
LE(ED1)	4	I	Data strobe input. Serial data is transferred to the respective latch when LE(ED1) is high. The data is latched when LE(ED1) goes low. Also, LE(ED1) is a control signal input for an Error Detection mode and Current Adjust mode (See Timing Diagram). LE(ED1) has an internal pulldown.
OE(ED2)	13	I	Output enable. When $\overline{OE}(ED2)$ is active (low), the output drivers are enabled; when $\overline{OE}(ED2)$ is high, all output drivers are turned OFF (blanked). Also, $\overline{OE}(ED2)$ is a control signal input for an Error Detection mode and Current Adjust mode (See Timing Diagram). $\overline{OE}(ED2)$ has an internal pullup.
OUT0 to OUT7	5 to 12	0	Constant-current outputs
R-EXT	15	I	Input used to connect an external resistor for setting up all output currents
SDI	2	I	Serial-data input to the Shift register
SDO	14	0	Serial-data output to the following SDI of next driver IC or to the microcontroller
VDD	16	I	Supply voltage

7 Specifications

7.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted) ⁽¹⁾

		MIN	MAX	UNIT
$V_{DD}^{(2)}$	Supply voltage	0	7	V
VI ⁽³⁾	Input voltage	-0.4	V _{DD} + 0.4	V
$V_{0}^{(4)}$	Output voltage	-0.5	20	V
f _{clk}	Clock frequency		25	MHz
I _{OUT}	Output current		120	mA
I _{GND}	GND terminal current		960	mA
T _A	Operating free-air temperature	-40	125	°C
TJ	Operating junction temperature	-40	150	°C
T _{stg}	Storage temperature	-55	150	°C

 Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under Recommended Operating Conditions is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

(2) All voltages are with respect to GND.

(3) Absolute negative voltage on these terminals must not go below 0 V.

(4) Absolute maximum voltage is 7 V for 200 ms.

7.2 ESD Ratings

				VALUE	UNIT
		Human body model (HBM), per AEC Q100-002 ⁽¹⁾		±1500	
		Charged device model (CDM) per	Other pins	±1000	
V _(ESD)	Electrostatic discharge	Charged device model (CDM), per AEC Q100-011	Corn <u>er pins</u> (GND, OUT3, VDD, and OUT4)	±1000	V
		Machine Model		±150	

(1) AEC Q100-002 indicates HBM stressing is done in accordance with the ANSI/ESDA/JEDEC JS-001 specification.

7.3 Recommended Operating Conditions

				MIN	NOM MAX	UNIT
V_{DD}	Supply voltage			3	5.5	5 V
Vo	Supply voltage to output pins	OUT0-OUT7			17	V V
		DC test circuit	V _O ≥ 0.6 V	5		~^^
IO	Output current	DC test circuit	V _O ≥1 V		120	mA
I _{OH}	High-level output current source	SDO shorted to GN	ND	-1		mA
I _{OL}	Low-level output current sink	SDO shorted to V _C	C	1		mA
VIH	High-level input voltage	CLK, OE(ED2), LE	(ED1), and SDI	$0.7 \times V_{DD}$	V _{DI}	, V
V _{IL}	Low-level input voltage	CLK, OE(ED2), LE	(ED1), and SDI	0	0.3 × V _{DI}	V



7.4 Thermal Information

		TLC591x-Q1	
	THERMAL METRIC ⁽¹⁾	D (SOIC)	UNIT
		16 PINS	
$R_{\theta JA}$	Junction-to-ambient thermal resistance	86.9	°C/W
R _{0JC(top)}	Junction-to-case (top) thermal resistance	47.7	°C/W
$R_{\theta JB}$	Junction-to-board thermal resistance	43.9	°C/W
ΨJT	Junction-to-top characterization parameter	11.9	°C/W
Ψ _{JB}	Junction-to-board characterization parameter	43.7	°C/W

(1) For more information about traditional and new thermal metrics, see the Semiconductor and IC Package Thermal Metrics application report, SPRA953.

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7.5 Electrical Characteristics: VDD = 3 V

 $V_{DD} = 3 \text{ V}, \text{ T}_{J} = -40^{\circ}\text{C} \text{ to } 125^{\circ}\text{C} \text{ (unless otherwise noted)}$

	PARAMETER	TEST CONDITIONS	MIN	TYP ⁽¹⁾	MAX	UNIT
V _{DD}	Input voltage		3		5.5	V
Vo	Supply voltage to the output pins				17	V
1		$V_{O} \ge 0.6 V$	5			m۸
lo	Output current	V _O ≥1V			120	mA
I _{ОН}	High-level output current, source		-1			mA
l _{OL}	Low-level output current, sink		1			mA
V _{IH}	High-level input voltage		$0.7 \times V_{DD}$		V_{DD}	V
V _{IL}	Low-level input voltage		GND		$0.3 \times V_{DD}$	V
l _{leak}	Output leakage current	$V_{OH} = 17 \text{ V} \qquad \frac{T_J = 25^{\circ}\text{C}}{T_J = 125^{\circ}\text{C}}$			0.5	μA
V _{OH}	High-level output voltage	SDO, $I_{OL} = -1 \text{ mA}$	V _{DD} – 0.4			V
V _{OL}	Low-level output voltage	SDO, $I_{OH} = 1 \text{ mA}$			0.4	V
	Output current 1	$V_{OUT} = 0.6 \text{ V}, \text{ R}_{ext} = 720 \Omega,$ CG = 0.992		26		mA
I _{O(1)}	Output current error, die-die	I_{OL} = 26 mA, V_O = 0.6 V, R_{ext} = 720 Ω , T_J = 25°C		±3%	±6%	
	Output current skew, channel-to- channel	$ I_{OL} = 26 \text{ mA}, \text{ V}_{O} = 0.6 \text{ V}, \text{ R}_{ext} = 720 \Omega, \\ \text{T}_{J} = 25^{\circ}\text{C} $		±1.5%	±3%	
	Output current 2	V _O = 0.8 V, R _{ext} = 360 Ω, CG = 0.992		52		mA
I _{O(2)}	Output current error, die-die	$\begin{array}{l} I_{OL} = 52 \text{ mA}, V_{O} = 0.8 \text{V}, \text{R}_{ext} = 360 \Omega, \\ \text{T}_{J} = 25^{\circ}\text{C} \end{array}$		±2%	±6%	
	Output current skew, channel-to- channel	$\begin{array}{l} I_{OL}=52 \text{ mA}, V_{O}=0.8 \text{V}, \text{R}_{ext}=360 \Omega, \\ \text{T}_{J}=25^{\circ}\text{C} \end{array}$		±1.5%	±3%	
	Output ourront vo	$V_0 = 1 V \text{ to } 3 V, I_0 = 26 \text{ mA}$		±0.1		
I _{OUT} vs V _{OUT}	Output current vs output voltage regulation	$V_{DD} = 3 V \text{ to } 5.5 V,$ $I_{O} = 26 \text{ mA}/120 \text{ mA}$		±1		%/V
	Pullup resistance	OE(ED2)		500		kΩ
	Pulldown resistance	LE(ED1)		500		kΩ
T _{sd}	Overtemperature shutdown ⁽²⁾		150	175	200	°C
T _{hys}	Restart temperature hysteresis			15		°C
I _{OUT,Th1}	Threshold current for open error detection	I _{OUT,target} = 26 mA		0.5 × I _{target} %		
I _{OUT,Th2}	Threshold current for open error detection	I _{OUT,target} = 52 mA		0.5 × I _{target} %		
I _{OUT,Th3}	Threshold current for open error detection	I _{OUT,target} = 104 mA		0.5 × I _{target} %		
I _{OUT,Th}	Threshold current for open error detection	I _{OUT,target} = 5 mA to 120 mA		0.5 × I _{target} %		
V _{OUT,TTh}	Trigger threshold voltage for short-error detection (TLC5917- Q1 only)	I _{OUT,target} = 5 mA to 120 mA	2.44	2.7	3.1	V
V _{OUT,RTh}	Return threshold voltage for short-error detection (TLC5917- Q1 only)	I _{OUT,target} = 5 mA to 120 mA	2.2			V
		R _{ext} = Open		5	10	
	Supply ourrent	R _{ext} = 720 Ω		8	14	~ ^
DD	Supply current	R _{ext} = 360 Ω		11	18	mA
		R _{ext} = 180 Ω		16	22	

(1) Typical values represent the likely parametric nominal values determined at the time of characterization. Typical values depend on the application and configuration and may vary over time. Typical values are not ensured on production material.

(2) Specified by design



7.6 Electrical Characteristics: VDD = 5.5 V

 $V_{DD} = 5.5 \text{ V}, \text{ T}_{J} = -40^{\circ}\text{C}$ to 125°C (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP ⁽¹⁾	MAX	UNIT
V _{DD}	Input Voltage		3		5.5	V
Vo	Supply voltage to the output pins				17	V
		V _O ≥ 0.6 V	5			
lo	Output current	V _O ≥1 V			120	mA
I _{OH}	High-level output current, source		-1			mA
I _{OL}	Low-level output current, sink		1			mA
V _{IH}	High-level input voltage		$0.7 \times V_{DD}$		V_{DD}	V
V _{IL}	Low-level input voltage		GND		$0.3 \times V_{DD}$	V
l _{leak}	Output leakage current	$V_{OH} = 17 \text{ V} \qquad \frac{T_J = 25^{\circ}\text{C}}{T_J = 125^{\circ}\text{C}}$			0.5	μA
V _{OH}	High-level output voltage	SDO, $I_{OL} = -1 \text{ mA}$	V _{DD} – 0.4			V
V _{OL}	Low-level output voltage	SDO, I _{OH} = 1 mA			0.4	V
	Output current 1	$V_{OUT} = 0.6 \text{ V}, \text{ R}_{ext} = 720 \Omega,$ CG = 0.992		26		mA
I _{O(1)}	Output current error, die-die	$I_{OL} = 26 \text{ mA}, V_O = 0.6 \text{ V}, R_{ext} = 720 \Omega$ $T_J = 25^{\circ}\text{C}$,	±3%	±6%	
	Output current skew, channel-to- channel	$I_{OL} = 26 \text{ mA}, V_O = 0.6 \text{ V}, R_{ext} = 720 \Omega$ $T_J = 25^{\circ}\text{C}$,	±1.5%	±3%	
	Output current 2	$V_{O} = 0.8 V, R_{ext} = 360 \Omega, CG = 0.992$		52		mA
I _{O(2)}	Output current error, die-die	I_{OL} = 52 mA, V _O = 0.8 V, R _{ext} = 360 Ω T _J = 25°C	,	±2%	±6%	
-(-)	Output current skew, channel-to- channel	I_{OL} = 52 mA, V_O = 0.8 V, R_{ext} = 360 Ω T_J = 25°C	,	±1.5%	±3%	
		$V_{\rm O}$ = 1 V to 3 V , $I_{\rm O}$ = 26 mA		±0.1		
I _{OUT} vs V _{OUT}	Output current vs output voltage regulation	$V_{DD} = 3 V \text{ to } 5.5 V,$ $I_{O} = 26 \text{ mA}/120 \text{ mA}$		±1		%/V
	Pullup resistance	OE(ED2),		500		kΩ
	Pulldown resistance	LE(ED1),		500		kΩ
T _{sd}	Overtemperature shutdown ⁽²⁾		150	175	200	°C
T _{hys}	Restart temperature hysteresis			15		°C
I _{OUT,Th1}	Threshold current for open error detection	I _{OUT,target} = 26 mA		0.5 × I _{target} %		
I _{OUT,Th2}	Threshold current for open error detection	I _{OUT,target} = 52 mA		0.5 × I _{target} %		
I _{OUT,Th3}	Threshold current for open error detection	I _{OUT,target} = 104 mA		0.5 × I _{target} %		
I _{OUT,Th}	Threshold current for open error detection	I _{OUT,target} = 5 mA to 120 mA		0.5 × I _{target} %		
V _{OUT,TTh}	Trigger threshold voltage for short-error detection (TLC5917- Q1 only)	I _{OUT,target} = 5 mA to 120 mA	2.44	2.7	3.1	V
V _{OUT,RTh}	Return threshold voltage for short-error detection (TLC5917- Q1 only)	I _{OUT,target} = 5 mA to 120 mA	2.2			V
		R _{ext} = Open		6	10	
	Complex compact	R _{ext} = 720 Ω		11	14	
DD	Supply current	R _{ext} = 360 Ω		13	18	mA
		R _{ext} = 180 Ω		19	24	

(1) Typical values represent the likely parametric nominal values determined at the time of characterization. Typical values depend on the application and configuration and may vary over time. Typical values are not ensured on production material.

(2) Specified by design

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TLC5916-Q1, TLC5917-Q1

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7.7 Timing Requirements

 V_{DD} = 3 V to 5.5 V (unless otherwise noted)

			MIN	NOM MA	
t _{w(L)}	LE(ED1) pulse duration	Normal mode	20		ns
t _{w(CLK)}	CLK pulse duration	Normal mode	20		ns
	OF(FD2) pulse duration	Normal mode, I _{OUT} < 60 mA	675		
t _{w(OE)}	OE(ED2) pulse duration	Normal mode, I _{OUT} > 60 mA	800		ns
t _{su(D)}	Setup time for SDI	Normal mode	3		ns
t _{h(D)}	Hold time for SDI	Normal mode	2		ns
t _{su(L)}	Setup time for LE(ED1)	Normal mode	15		ns
t _{h(L)}	Hold time for LE(ED1)	Normal mode	15		ns
t _{w(CLK)}	CLK pulse duration	Error Detection mode	20		ns
t _{w(ED2)}	OE(ED2) pulse duration	Error Detection mode	2000		ns
t _{su(ED1)}	Setup time for LE(ED1)	Error Detection mode	4		ns
t _{h(ED1)}	Hold time for LE(ED1)	Error Detection mode	10		ns
t _{su(ED2)}	Setup time for $\overline{OE}(ED2)$	Error Detection mode	8.5		ns
t _{h(ED2)}	Hold time for $\overline{OE}(ED2)$	Error Detection mode	10		ns
f _{CLK}	Clock frequency	Cascade operation		3) MHz

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7.8 Switching Characteristics: VDD = 3 V

 V_{DD} = 3 V, T_{J} = -40°C to 125°C (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP ⁽¹⁾	MAX	UNIT
t _{PLH1}	Low-to-high propagation delay time, CLK to OUTn		40	65	95	ns
t _{PLH2}	Low-to-high propagation delay time, LE(ED1) to OUTn		40	65	95	ns
t _{PLH3}	Low-to-high propagation delay time, OE(ED2) to OUTn		40	65	95	ns
t _{PLH4}	Low-to-high propagation delay time, CLK to SDO		12	20	30	ns
t _{PHL1}	High-to-low propagation delay time, CLK to OUTn			300	365	ns
t _{PHL2}	High-to-low propagation delay time, LE(ED1) to OUTn			300	365	ns
t _{PHL3}	High-to-low propagation delay time, $\overline{OE}(ED2)$ to \overline{OUTn}			300	365	ns
t _{PHL4}	High-to-low propagation delay time, CLK to SDO		12	20	30	ns
t _{w(CLK)}	Pulse duration, CLK		20			ns
t _{w(L)}	Pulse duration, LE(ED1)		20			ns
t _{w(OE)}	Pulse duration, OE(ED2)		500			ns
t _{w(ED2)}	Pulse duration, $\overline{OE}(ED2)$ in Error Detection mode	$V_{IH} = V_{DD}, V_{IL} = GND,$ $R_{ext} = 360 \ \Omega, V_L = 4 \ V,$	2			μs
t _{h(ED1,ED2)}	Hold time, LE(ED1) and $\overline{OE}(ED2)$	$R_{L} = 44 \Omega, C_{L} = 10 \text{ pF},$ CG = 0.992	10			ns
t _{h(D)}	Hold time, SDI	CG = 0.992	2			ns
t _{su(D,ED1)}	Setup time, SDI, LE(ED1)		4			ns
t _{su(ED2)}	Setup time, OE(ED2)		8.5			ns
t _{h(L)}	Hold time, LE(ED1), Normal mode		15			ns
t _{su(L)}	Setup time, LE(ED1), Normal mode		15			ns
t _r	Rise time, CLK ⁽²⁾				500	ns
t _f	Fall time, CLK ⁽²⁾				500	ns
t _{or}	Rise time, outputs (off)		40	85	105	ns
t _{or}	Rise time, outputs (off), $T_J = 25^{\circ}C$			83	100	ns
t _{of}	Rise time, outputs (on)		100	280	370	ns
t _{of}	Rise time, outputs (on), $T_J = 25^{\circ}C$			170	225	ns
f _{CLK}	Clock frequency	Cascade operation			30	MHz

(1) Typical values represent the likely parametric nominal values determined at the time of characterization. Typical values depend on the application and configuration and may vary over time. Typical values are not ensured on production material.

(2) If the devices are connected in cascade and t_r or t_f is large, it may be critical to achieve the timing required for data transfer between two cascaded devices.

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7.9 Switching Characteristics: VDD = 5.5 V

 $V_{DD} = 5.5 \text{ V}, \text{ T}_{J} = -40^{\circ}\text{C}$ to 125°C (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP ⁽¹⁾	MAX	UNIT
t _{PLH1}	Low-to-high propagation delay time, CLK to OUTn		40	65	95	ns
t _{PLH2}	Low-to-high propagation delay time, LE(ED1) to OUTn		40	65	95	ns
t _{PLH3}	Low-to-high propagation delay time, $\overline{OE}(ED2)$ to \overline{OUTn}		40	65	95	ns
t _{PLH4}	Low-to-high propagation delay time, CLK to SDO		8	20	30	ns
t _{PHL1}	High-to-low propagation delay time, CLK to OUTn			300	365	ns
t _{PHL2}	High-to-low propagation delay time, LE(ED1) to OUTn			300	365	ns
t _{PHL3}	High-to-low propagation delay time, $\overline{\text{OE}}(\text{ED2})$ to $\overline{\text{OUTn}}$			300	365	ns
t _{PHL4}	High-to-low propagation delay time, CLK to SDO		8	20	30	ns
t _{w(CLK)}	Pulse duration, CLK		20			ns
t _{w(L)}	Pulse duration, LE(ED1)		20			ns
t _{w(OE)}	Pulse duration, $\overline{OE}(ED2)$		500			ns
t _{w(ED2)}	Pulse duration, $\overline{OE}(ED2)$ in Error Detection mode	$V_{IH} = V_{DD}, V_{IL} = GND,$ $R_{ext} = 360 \Omega, V_L = 4 V,$	2			μs
t _{h(D,ED1,ED2)}	Hold time, SDI, LE(ED1), and $\overline{OE}(ED2)$	$R_L = 44 \Omega, C_L = 10 pF,$ CG = 0.992	10			ns
t _{h(D)}	Hold time, SDI	CG = 0.992	2			ns
t _{su(D,ED1)}	Setup time, SDI, LE(ED1)		4			ns
t _{su(ED2)}	Setup time, OE(ED2)		8.5			ns
t _{h(L)}	Hold time, LE(ED1), Normal mode		15			ns
t _{su(L)}	Setup time, LE(ED1), Normal mode		15			ns
t _r	Rise time, CLK ⁽²⁾				500	ns
t _f	Fall time, CLK ⁽²⁾				500	ns
t _{or}	Rise time, outputs (off)		40	85	105	ns
t _{or}	Rise time, outputs (off), $T_J = 25^{\circ}C$			83	100	ns
t _{of}	Rise time, outputs (on)		100	280	370	ns
t _{of}	Rise time, outputs (on), $T_J = 25^{\circ}C$			170	225	ns
f _{CLK}	Clock frequency	Cascade operation			30	MHz

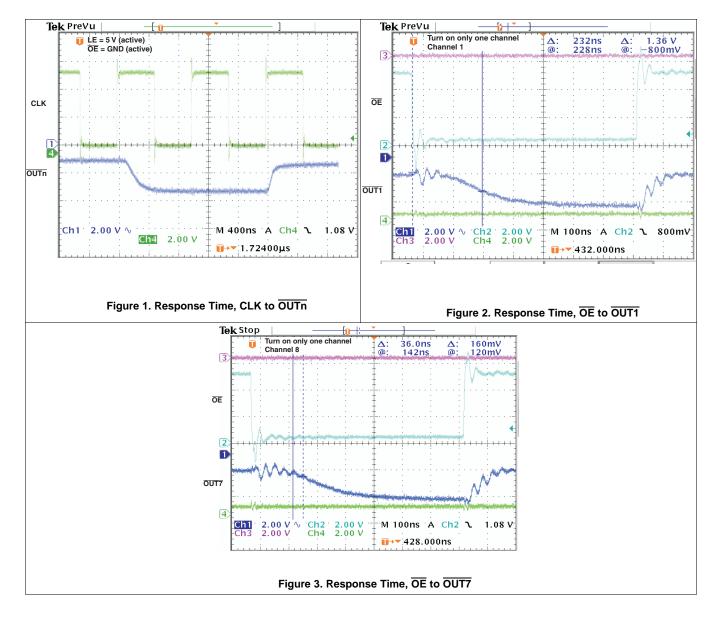
(1) Typical values represent the likely parametric nominal values determined at the time of characterization. Typical values depend on the application and configuration and may vary over time. Typical values are not ensured on production material.

(2) If the devices are connected in cascade and t_r or t_f is large, it may be critical to achieve the timing required for data transfer between two cascaded devices.

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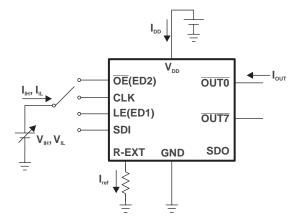


7.10 Typical Characteristics





8 Parameter Measurement Information





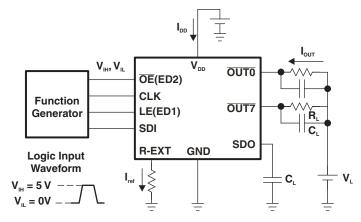


Figure 5. Test Circuit for Switching Characteristics



$\mathbf{t}_{w(CLK)}$ 50% CLK 50% 50% 50% $\mathbf{t}_{su(D)}$ **t**_{h(D)} SDI 50% 50% $\mathbf{t}_{_{\mathrm{PLH4}}}, \mathbf{t}_{_{\mathrm{PHL4}}}$ 50% SDO $\boldsymbol{t}_{w(L)}$ 50% LE(ED1) t_{su(L)} $\mathbf{t}_{h(L)}$ OE(ED2) LOW $\mathbf{t}_{\text{PLH2}}, \mathbf{t}_{\text{PHL2}}$ Output off OUTn 50% Output on $\mathbf{t}_{\text{PLH1}}, \mathbf{t}_{\text{PHL1}}$ t_{w(OE)} HIGH OE(ED2) 50% 50% \mathbf{t}_{PHL3} t_{PLH3} Output off 80% 80% OUTn 50% 50% 20% 20%

Parameter Measurement Information (continued)

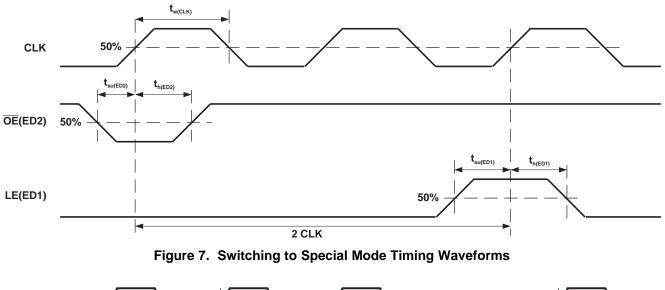
Figure 6. Normal Mode Timing Waveforms

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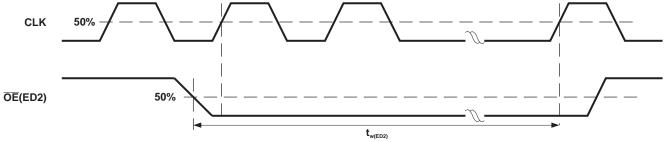


Figure 8. Reading Error Status Code Timing Waveforms

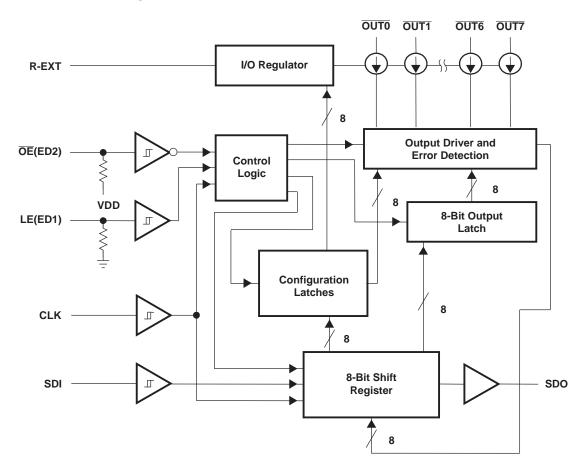


9 Detailed Description

9.1 Overview

The TLC591x-Q1 is designed for LED displays and LED lighting applications with constant-current control and open-load, shorted-load, and overtemperature detection. The TLC591x-Q1 contains an 8-bit shift register and data latches, which convert serial input data into parallel output format. At the output stage, eight regulated current ports are designed to provide uniform and constant current for driving LEDs within a wide range of VF variations. Used in system design for LED display applications, for example, LED panels, the TLC591x-Q1 device provides great flexibility and device performance. Users can adjust the output current from 5 mA to 120 mA through an external resistor, R-EXT, which gives flexibility in controlling the light intensity of LEDs. The devices are designed for up to 17 V at the output port. The high clock frequency, 30 MHz, also satisfies the system requirements of high-volume data transmission.

9.2 Functional Block Diagram



9.3 Feature Description

9.3.1 Open-Circuit Detection Principle

The LED Open-Circuit Detection compares the effective current level I_{out} with the open load detection threshold current $I_{OUT,Th}$. If I_{OUT} is below the $I_{OUT,Th}$ threshold, the TLC591x-Q1 detects an open-load condition. This error status can be read as an error status code in the Special mode. For open-circuit error detection, a channel must be on.

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Feature Description (continued)

STATE OF OUTPUT PORT	STATE OF OUTPUT PORT CONDITION OF OUTPUT CURRENT		- OF OUTPUT PORT FRROR STATUS CODE		MEANING
Off	I _{OUT} = 0 mA	0	Detection not possible		
07	I _{OUT} < I _{OUT,Th} ⁽¹⁾	0	Open circuit		
On	I _{OUT} ≥ I _{OUT,Th} ⁽¹⁾	Channel n error status bit 1	Normal		

Table 1. Open-Circuit Detection

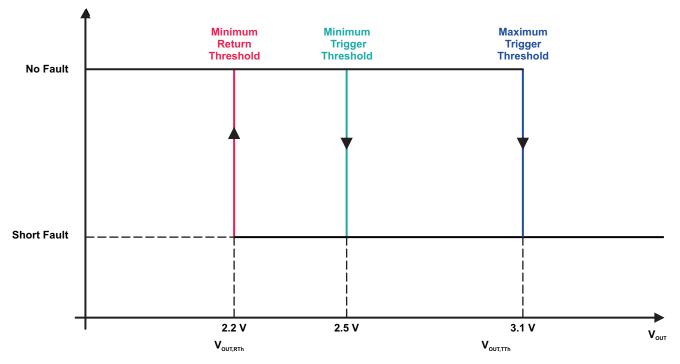
(1) $I_{OUT,Th} = 0.5 \times I_{OUT,target}$ (typical)

9.3.2 Short-Circuit Detection Principle (TLC5917-Q1 Only)

The LED short-circuit detection compares the effective voltage level (V_{OUT}) with the shorted-load detection threshold voltages $V_{OUT,TTh}$ and $V_{OUT,RTh}$. If V_{OUT} is above the $V_{OUT,TTh}$ threshold, the TLC5917-Q1 detects an shorted-load condition. If V_{OUT} is below the $V_{OUT,RTh}$ threshold, no error is detected and the error bit is reset. This error status can be read as an error status code in the Special mode. For short-circuit error detection, a channel must be on.

Table 2. Shorted-Load Detection

STATE OF OUTPUT PORT	CONDITION OF OUTPUT VOLTAGE	ERROR STATUS CODE	MEANING
Off	I _{OUT} = 0 mA	0	Detection not possible
0-	V _{OUT} ≥ V _{OUT,TTh}	0	Short circuit
On	V _{OUT} < V _{OUT,RTh}	1	Normal





9.3.3 Overtemperature Detection and Shutdown

TLC591x-Q1 is equipped with a global overtemperature sensor and eight individual, channel-specific, overtemperature sensors.

• When the global sensor reaches the trip temperature, all output channels are shutdown, and the error status is stored in the internal Error Status register of every channel. After shutdown, the channels automatically restart after cooling down, if the control signal (output latch) remains on. The stored error status is not reset



after cooling down and can be read out as the error status code in the Special mode.

When one of the channel-specific sensors reaches trip temperature, only the affected output channel is shut down, and the error status is stored only in the internal Error Status register of the affected channel. After shutdown, the channel automatically restarts after cooling down, if the control signal (output latch) remains on. The stored error status is not reset after cooling down and can be read out as error status code in the Special mode.

For channel-specific overtemperature error detection, a channel must be on.

The error status code is reset when TLC591x-Q1 returns to Normal mode.

STATE OF OUTPUT PORT	CONDITION	ERROR STATUS CODE	MEANING
Off	I _{OUT} = 0 mA	0	
On	T _j < T _{j,trip} global	1	Normal
$On \rightarrow all channels$ Off	$T_j > T_{j,trip}$ global	All error status bits = 0	Global overtemperature
On	T _j < T _{j,trip} channel n	1	Normal
$On \rightarrow Off$	$T_j > T_{j,trip}$ channel n	Channel n error status bit = 0	Channel n overtemperature

Table 3. Overtemperature Detection⁽¹⁾

(1) The global shutdown threshold temperature is approximately 170°C.

9.4 Device Functional Modes

The TLC591x-Q1 provides a Special Mode in which two functions are included: Error Detection and Current Gain Control. There are two operation modes and three phases: Normal Mode phase, Mode Switching transition phase, and Special Mode phase. The signal on the multiple function pin OE(ED2) is monitored to determine the mode. When an one-clock-wide pulse appears on OE(ED2), the device enters the Mode Switching phase. At this time, the voltage level on LE(ED1) determines the mode to which the TLC591x-Q1 switches. In the Normal Mode phase, the serial data can be transferred into TLC591x-Q1 via the pin SDI, shifted in the shift register, and transferred out via the pin SDO. LE(ED1) can latch the serial data in the shift register to the output latch. OE(ED2) enables the output drivers to sink current.

In the Special Mode phase, the low-voltage-level signal OE(ED2) can enable output channels and detect the status of the output current, to determine if the driving current level is sufficient. The detected Error Status is loaded into the 8-bit shift register and shifted out via the pin SDO, synchronous to the CLK signal. The system controller can read the error status and determine whether or not the LEDs are properly lit.

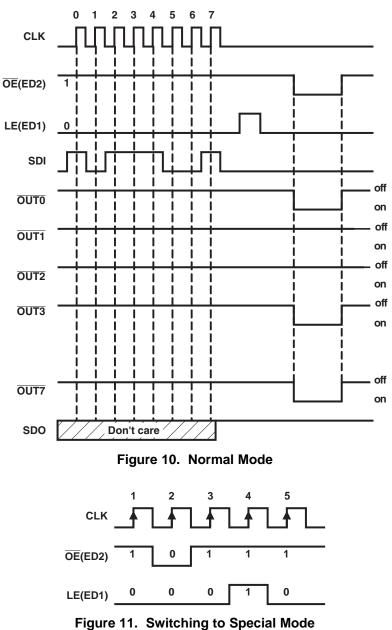
In the Special Mode phase, the TLC591x-Q1 allows users to adjust the output current level by setting a runtimeprogrammable Configuration Code. The code is sent into the TLC591x-Q1 via SDI. The positive pulse of LE(ED1) latches the code in the shift register into a built-in 8-bit configuration latch, instead of the output latch. The code affects the voltage at the terminal R-EXT and controls the output-current regulator. The output current can be finely adjusted by a gain ranging from 1/12 to 127/128 in 256 steps. Therefore, the current skew between ICs can be compensated within less than 1%. This feature is suitable for white balancing in LED color display panels.

CLK	LE(ED1)	OE(ED2)	SDI	OUT0OUT7	SDO
Ť	Н	L	Dn	DnDn – 7	Dn – 7
Ť	L	L	Dn + 1	No change	Dn – 6
Ť	Н	L	Dn + 2	Dn + 2Dn – 5	Dn – 5
\downarrow	Х	L	Dn + 3	Dn + 2Dn – 5	Dn – 5
Ļ	Х	Н	Dn + 3	Off	Dn – 5

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The signal sequence shown in Figure 11 makes the TLC591x-Q1 enter Current Adjust and Error Detection mode.





In the Current Adjust mode, sending the positive pulse of LE(ED1), the content of the Shift register (a current adjust code) is written to the 8-bit configuration latch (see Figure 12).

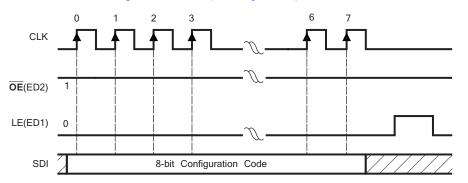


Figure 12. Writing Configuration Code

When the TLC591x-Q1 is in the Error Detection mode, the signal sequence shown in Figure 13 enables a system controller to read error status codes through SDO.

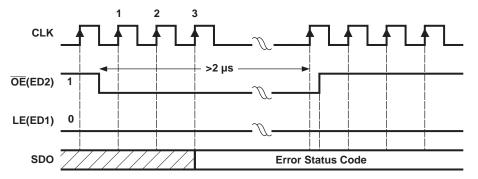


Figure 13. Reading Error Status Code

The signal sequence shown in Figure 14 makes TLC591x-Q1 resume the Normal mode. Switching to Normal mode resets all internal Error Status registers. $\overline{OE}(ED2)$ always enables the output port, whether the TLC591x-Q1 enters Current Adjust mode or not.

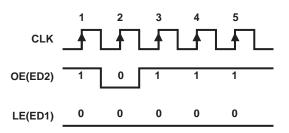
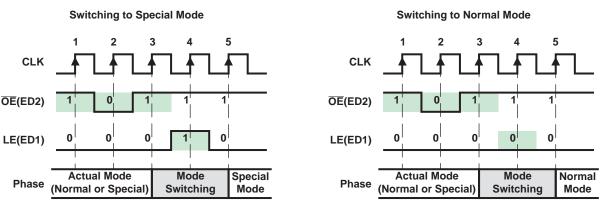


Figure 14. Switching to Normal Mode

9.4.1 Operation Mode Switching

In order to switch between its two modes, TLC591x-Q1 monitors the signal $\overline{OE}(ED2)$. When an one-clock-wide pulse of $\overline{OE}(ED2)$ appears, TLC591x-Q1 enters the two-clock-period transition phase, the Mode Switching phase. After power on, the default operation mode is the Normal Mode (see Figure 15).





As shown in Figure 15, once a one-clock-wide short pulse (101) of $\overline{OE}(ED2)$ appears, TTLC591x-Q1 enters the Mode Switching phase. At the fourth rising edge of CLK, if LE(ED1) is sampled as voltage high, TLC591x-Q1 switches to Special mode; otherwise, it switches to Normal mode. The signal LE(ED1) between the third and the fifth rising edges of CLK cannot latch any data. Its level is used only to determine into which mode to switch. However, the short pulse of $\overline{OE}(ED2)$ can still enable the output ports. During mode switching, the serial data can still be transferred through SDI and shifted out from SDO.

NOTE

- 1. The signal sequence for the mode switching may be used frequently to ensure that TLC591x-Q1 is in the proper mode.
- 2. The 1 and 0 on the LE(ED1) signal are sampled at the rising edge of CLK. The X means its level does not affect the result of mode switching mechanism.
- 3. After power on, the default operation mode is Normal mode.

9.4.1.1 Normal Mode Phase

Serial data is transferred into TLC591x-Q1 via SDI, shifted in the Shift register, and output via SDO. LE(ED1) can latch the serial data in the Shift register to the Output Latch. $\overline{OE}(ED2)$ enables the output drivers to sink current. These functions differ only as described in Operation Mode Switching, in which case, a short pulse triggers TLC591x-Q1 to switch the operation mode. However, as long as LE(ED1) is high in the Mode Switching phase, TLC591x-Q1 remains in the Normal mode, as if no mode switching occurred.

9.4.1.2 Special Mode Phase

In the Special mode, as long as $\overline{OE}(ED2)$ is not low, the serial data is shifted to the Shift register via SDI and shifted out via SDO, as in the Normal mode. However, there are two differences between Special mode and Normal mode, as shown in the following sections.

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9.5 Programming

9.5.1 Reading Error Status Code in Special Mode

When $\overline{OE}(ED2)$ is pulled low while in Special mode, error detection and load error status codes are loaded into the Shift register, in addition to enabling output ports to sink current. Figure 16 shows the timing sequence for error detection. The 0 and 1 signal levels are sampled at the rising edge of each CLK. At least three zeros must be sampled at the voltage low signal $\overline{OE}(ED2)$. Immediately after the second zero is sampled, the data input source of the Shift register changes to the 8-bit parallel Error Status Code register, instead of from the serial data on SDI. Normally, the error status codes are generated at least 2 µs after the falling edge of $\overline{OE}(ED2)$. The occurrence of the third or later zero saves the detected error status codes into the Shift register. Therefore, when $\overline{OE}(ED2)$ is low, the serial data cannot be shifted into TLC591x-Q1 via SDI. When $\overline{OE}(ED2)$ is pulled high, the data input source of the Shift register is changed back to SDI. At the same time, the output ports are disabled and the error detection is completed. Then, the error status codes saved in the Shift register can be shifted out via SDO bit by bit along with CLK, as well as the new serial data can be shifted into TLC591x-Q1 via SDI.

While in Special mode, the TLC591x-Q1 cannot simultaneously transfer serial data and detect LED load error status.

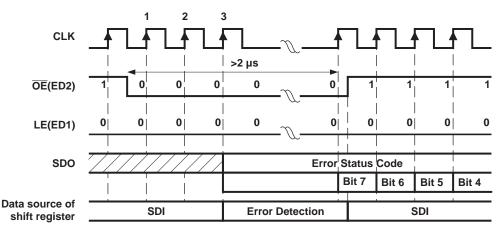


Figure 16. Reading Error Status Code

9.5.2 Writing Configuration Code in Special Mode

When in Special mode, the active high signal LE(ED1) latches the serial data in the Shift register to the Configuration Latch, instead of the Output Latch. The latched serial data is used as the Configuration Code.

The code is stored until power off or the Configuration Latch is rewritten. As shown in Figure 17, the timing for writing the Configuration Code is the same as the timing in the Normal Mode to latching output channel data. Both the Configuration Code and Error Status Code are transferred in the common 8-bit Shift register. Users must pay attention to the sequence of error detection and current adjustment to avoid the Configuration Code being overwritten by Error Status Code.

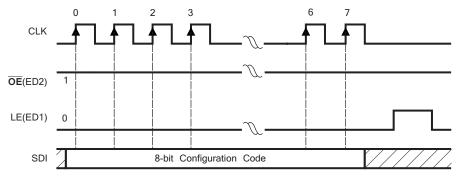


Figure 17. Writing Configuration Code

Programming (continued)

9.5.3 8-Bit Configuration Code and Current Gain

Bit definition of the Configuration Code in the Configuration Latch is shown in Table 5.

	BIT 0	BIT 1	BIT 2	BIT 3	BIT 4	BIT 5	BIT 6	BIT 7
Meaning	СМ	HC	CC0	CC1	CC2	CC3	CC4	CC5
Default	1	1	1	1	1	1	1	1

Table 5. Bit Definition of 8-Bit Configuration Code

Bit 7 is first sent into TLC591x-Q1 via SDI. Bits 1 to 7 {HC, CC[0:5]} determine the voltage gain (VG) that affects the voltage at R-EXT and indirectly affects the reference current, Iref, flowing through the external resistor at R-EXT. Bit 0 is the Current Multiplier (CM) that determines the ratio IOUT, target/Iref. Each combination of VG and CM gives a specific Current Gain (CG).

VG: the relationship between {HC,CC[0:5]} and the voltage gain is calculated as shown below:

$$VG = (1 + HC) \times (1 + D/64) / 4$$

 $D = CC0 \times 2^{5} + CC1 \times 2^{4} + CC2 \times 2^{3} + CC3 \times 2^{2} + CC4 \times 2^{1} + CC5 \times 2^{0}$

Where HC is 1 or 0, and D is the binary value of CC[0:5]. So, the VG could be regarded as a floating-point number with 1-bit exponent HC and 6-bit mantissa CC[0:5]. {HC,CC[0:5]} divides the programmable voltage gain VG into 128 steps and two sub-bands:

Low-voltage subband (HC = 0): VG = $1/4 \sim 127/256$, linearly divided into 64 steps

High-voltage subband (HC = 1): $VG = 1/2 \sim 127/128$, linearly divided into 64 steps

- CM: In addition to determining the ratio I_{OUT,target}/I_{ref}, CM limits the output current range. High Current Multiplier (CM = 1): $I_{OUT,target}/I_{ref}$ = 15, suitable for output current range I_{OUT} = 10 mA to 120 mA. Low Current Multiplier (CM = 0): I_{OUT,target}/I_{ref} = 5, suitable for output current range I_{OUT} = 5 mA to 40 mA
- CG: The total Current Gain is defined as the following.

 $V_{R-FXT} = 1.26 V \times VG$

$$\begin{split} I_{\text{ref}} &= V_{\text{R-EXT}}/R_{\text{ext}}, \text{ if the external resistor, } R_{\text{ext}}, \text{ is connected to ground.} \\ I_{\text{OUT,target}} &= I_{\text{ref}} \times 15 \times 3^{\text{CM} - 1} = 1.26 \text{ V/R}_{\text{ext}} \times \text{VG} \times 15 \times 3^{\text{CM} - 1} = (1.26 \text{ V/R}_{\text{ext}} \times 15) \times \text{CG} \end{split}$$
 $CG = VG \times 3^{CM - 1}$

Therefore, CG = 1/12 to 127/128, and it is divided into 256 steps.

Examples

- Configuration Code {CM, HC, CC[0:5]} = {1,1,11111} VG = 127/128 = 0.992 and $CG = VG \times 3^0 = VG = 0.992$
- Configuration Code = $\{1, 1, 00000\}$ $VG = (1 + 1) \times (1 + 0/64)/4 = 1/2 = 0.5$, and CG = 0.5
- Configuration Code = $\{0, 0, 000000\}$ $VG = (1 + 0) \times (1 + 0/64)/4 = 1/4$, and $CG = (1/4) \times 3^{-1} = 1/12$

After power on, the default value of the Configuration Code {CM, HC, CC[0:5]} is {1,1,11111}. Therefore, VG = CG = 0.992. The relationship between the Configuration Code and the Current Gain is shown in Figure 18.



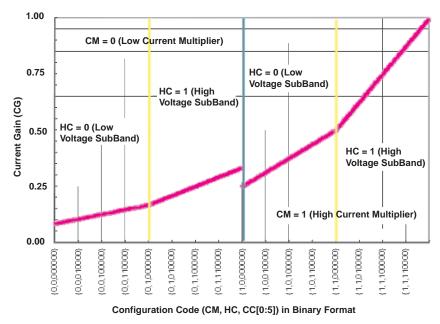


Figure 18. Current Gain vs Configuration Code

TRUMENTS

10 Application and Implementation

NOTE

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

10.1 Application Information

10.1.1 Constant Current

In LED display applications, the TLC591x-Q1 provides nearly no current variations from channel to channel and from IC to IC. While $I_{OUT} \le 100$ mA, the maximum current skew between channels is less than ±3% and between ICs is less than ±6%.

10.1.2 Adjusting Output Current

The TLC591x-Q1 scales up the reference current, I_{ref} , set by the external resistor R_{ext} to sink a current, I_{out} , at each output port. Use the following formulas to calculate the target output current

I_{OUT,target} in the saturation region:

 $I_{ref} = V_{R-EXT}/R_{ext}, \text{ if another end of the external resistor } R_{ext} \text{ is connected to ground} V_{R-EXT} = 1.26 \text{ V} \times \text{VG}$ (2) $I_{OUT,target} = I_{ref} \times 15 \times 3^{CM-1}$

where

- R_{ext} is the resistance of the external resistor connected to the R-EXT terminal.
- V_{R-EXT} is the voltage of R-EXT, which is controlled by the programmable voltage gain (VG), which is defined by the Configuration Code.
 (3)

The Current Multiplier (CM) determines that the ratio $I_{OUT,target}/I_{ref}$ is 15 or 5. After power on, the default value of VG is 127/128 = 0.992, and the default value of CM is 1, so that the ratio $I_{OUT,target}/I_{ref}$ = 15. Based on the default VG and CM.

 $V_{R-EXT} = 1.26 V \times 127/128 = 1.25 V$ $I_{OUT,target} = (1.25 V/R_{ext}) \times 15$ (4) (5)

(1)

Therefore, the default current is approximately 52 mA at 360 Ω and 26 mA at 720 Ω . The default relationship after power on between I_{OUT,target} and R_{ext} is shown in Figure 19.

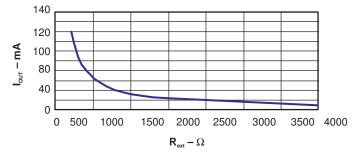


Figure 19. Default Relationship Curve Between I_{OUT,target} and R_{ext} After Power Up



10.2 Typical Applications

10.2.1 Single Implementation of TLC591x-Q1 Device

The TLC5917/TLC5917-Q1 Constant-Current LED Sink Drivers is designed to work alone or cascaded. Figure 20 shows implementation of a single TLC591x-Q1 device.

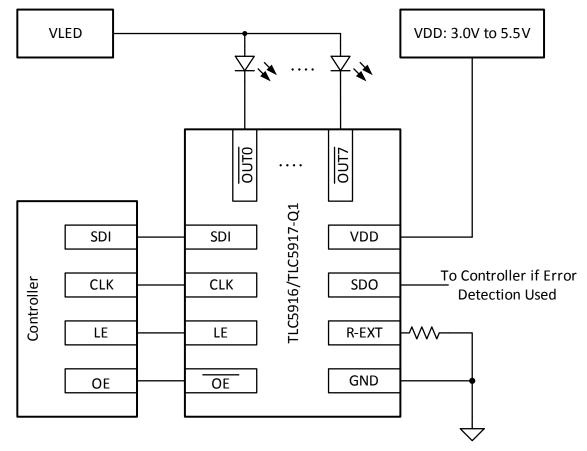


Figure 20. Single Implementation of TLC591x-Q1 Device

10.2.1.1 Design Requirements

For this design example, use the parameters listed in Table 6. The purpose of this design procedure is to calculate the power dissipation in the device and the operating junction temperature.

-	
DESIGN PARAMETER	EXAMPLE VALUE
Number of LED strings	8
Number of LEDs per string	3
LED Current (mA)	20
Forward voltage of each LED (V)	3.5
Junction-to-ambient thermal resistance (°C/W)	86.9
Ambient temperature of application (°C)	115
VDD (V)	5
IDD (mA)	10
Max operating junction temperature (°C)	150

Table 6. Design Parameters

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TLC5916-Q1, TLC5917-Q1

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10.2.1.2 Detailed Design Procedure

Lico the following	oquationa	to c	lotormino	tha	docian	naramatara
Use the following	equations	10 0	Jelennine	uie	uesign	parameters.

where

where	
• T _J is the junction temperature	
T _A is the ambient temperature	
 R_{eJA} is the junction-to-ambient thermal resistance 	
 P_{D_TOT} is the total power dissipation in the IC. 	(6)
$P_{D_{T}TT} = P_{D_{C}CS} + I_{DD} \times V_{DD}$	
where	
 P_{D_CS} is the power dissipation in the LED current sinks. 	
• I _{DD} is the IC supply current.	
 V_{DD} is the IC supply voltage. 	(7)
$P_{D_{C}CS} = I_{O} \times V_{O} \times n_{CH}$	
where	
• I _o is the LED current	
 V_o is the voltage at the output pin 	
 n_{CH} is the number of LED strings. 	(8)
$V_{O} = V_{LED} - (n_{LED} \times V_{F})$	
where	
 V_{LED} is the voltage applied to the LED string 	
 n_{LED} is the number of LEDs in the string 	
 V_F is the forward voltage of each LED. 	(9)
V_O must not be too high as this causes excess power dissipation inside the current sink. How not be too low as this does not allow the full LED current Figure 21.	ever, V _O also must

With $V_{LED} = 12$ V:	
$V_0 = 12 V - (3 \times 3.5 V) = 1.5 V$	(10)
$P_{D_{CS}} = 20 \text{ mA} \times 1.5 \text{ V} \times 8 = 0.24 \text{ W}$	(11)
Using P _{D_CS} , calculate:	
$P_{D_{-}TOT} = P_{D_{-}CS} + I_{DD} \times V_{DD} = 0.24 \text{ W} + 0.01 \text{ A} \times 5 \text{ V} = 0.29 \text{ W}$	(12)

Using $P_{D TOT}$, calculate:

 $T_J = T_A + R_{\theta JA} \times P_{D_TOT} = 115^{\circ}C + 86.9^{\circ}C/W \times 0.29 W = 140^{\circ}C$

This design example demonstrates how to calculate power dissipation in the IC and ensure that the junction temperature is kept below 150°C.

NOTE

This design example assumes that all channels have the same electrical parameters (n_{LED} , I_O , V_F , V_{LED}). If the parameters are unique for each channel, then the power dissipation must be calculated for each current sink separately. Then, add each result together to calculate the total power dissipation in the current sinks.



10.2.1.3 Application Curve

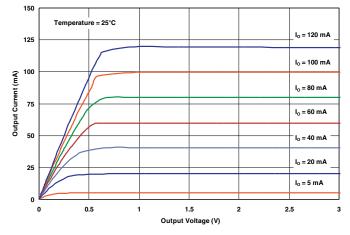
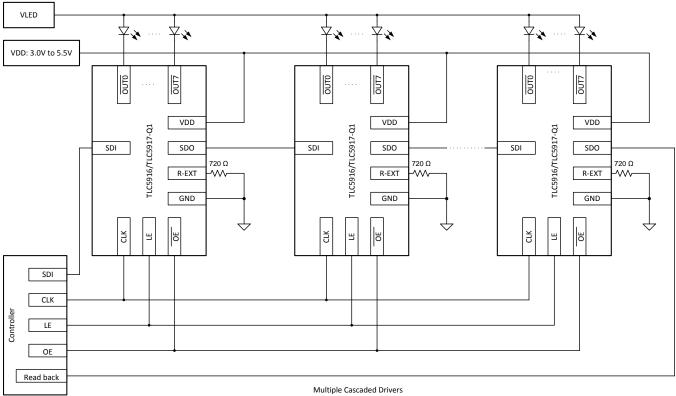


Figure 21. Output Current vs Output Voltage

10.2.2 Cascading Implementation of TLC591x-Q1 Device

The TLC5917/TLC5917-Q1 Constant-Current LED Sink Drivers is designed to work alone or cascaded. Figure 22 shows a cascaded driver implementation.



26mA Application

Figure 22. Cascading Implementation of TLC591x-Q1 Device



The device is designed to operate from a VDD supply between 3 V and 5.5 V. The LED supply voltage is determined by the number of LEDs in each string and the forward voltage of the LEDs (should be \leq 17 V in total).

12 Layout

12.1 Layout Guidelines

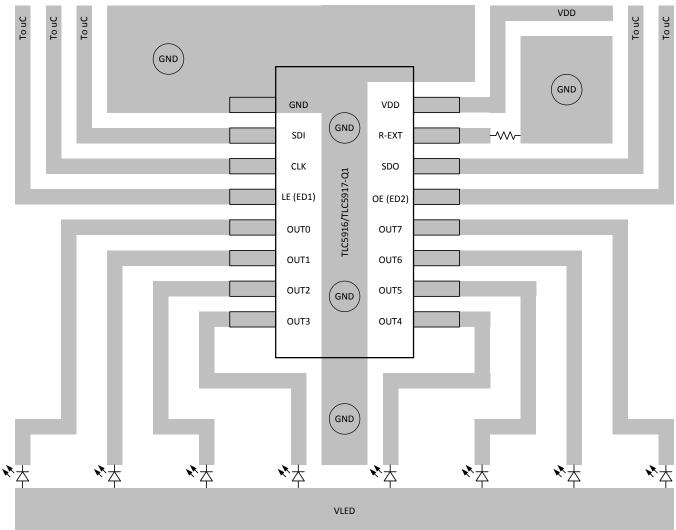
TLC5916-Q1, TLC5917-Q1

The traces that carry current from the LED cathodes to the OUTx pins must be wide enough to support the default current (up to 120 mA).

The SDI, CLK, LE (ED1), OE (ED2), and SDO pins are to be connected to the microcontroller.

There are several ways to achieve this, including the following methods:

- Traces may be routed underneath the package on the top layer.
- The signal may travel through a via to another layer.



12.2 Layout Example





13 Device and Documentation Support

13.1 Related Links

The table below lists quick access links. Categories include technical documents, support and community resources, tools and software, and quick access to sample or buy.

PARTS	PRODUCT FOLDER	SAMPLE & BUY	TECHNICAL DOCUMENTS	TOOLS & SOFTWARE	SUPPORT & COMMUNITY
TLC5916-Q1	Click here	Click here	Click here	Click here	Click here
TLC5917-Q1	Click here	Click here	Click here	Click here	Click here

Table 7. Related Links

13.2 Trademarks

All trademarks are the property of their respective owners.

13.3 Electrostatic Discharge Caution



These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

13.4 Glossary

SLYZ022 — TI Glossary.

This glossary lists and explains terms, acronyms, and definitions.

14 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.



10-Dec-2020

PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead finish/ Ball material (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
TLC5916QDRQ1	ACTIVE	SOIC	D	16	2500	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	TLC5916Q	Samples
TLC5917QDRQ1	ACTIVE	SOIC	D	16	2500	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	TLC5917Q	Samples

⁽¹⁾ The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

⁽²⁾ RoHS: TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

Green: TI defines "Green" to mean the content of Chlorine (CI) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <= 1000ppm threshold. Antimony trioxide based flame retardants must also meet the <= 1000ppm threshold requirement.

⁽³⁾ MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

⁽⁴⁾ There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

⁽⁵⁾ Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

⁽⁶⁾ Lead finish/Ball material - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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PACKAGE OPTION ADDENDUM

10-Dec-2020

OTHER QUALIFIED VERSIONS OF TLC5916-Q1, TLC5917-Q1 :

• Catalog: TLC5916, TLC5917

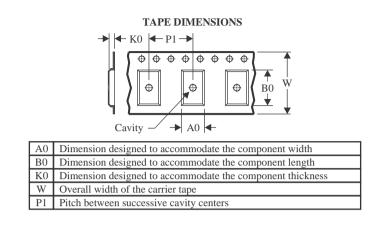
NOTE: Qualified Version Definitions:

• Catalog - TI's standard catalog product



TAPE AND REEL INFORMATION





QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



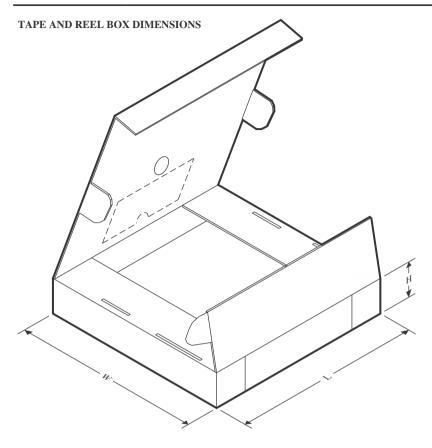
*All dimensions are nominal	

Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TLC5917QDRQ1	SOIC	D	16	2500	330.0	16.4	6.5	10.3	2.1	8.0	16.0	Q1



PACKAGE MATERIALS INFORMATION

25-Sep-2024



*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TLC5917QDRQ1	SOIC	D	16	2500	353.0	353.0	32.0

D (R-PDSO-G16)

PLASTIC SMALL OUTLINE



NOTES: A. All linear dimensions are in inches (millimeters).

- B. This drawing is subject to change without notice.
- Body length does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.006 (0,15) each side.
- Body width does not include interlead flash. Interlead flash shall not exceed 0.017 (0,43) each side.
- E. Reference JEDEC MS-012 variation AC.



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