Analytic Expressions for currents in the CCM PFC stage

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Agenda

- CCM Boost PFC stage with current waveforms
	- RMS calculation
	- RMS current expressions, Diode, MOSFET, C_{OUT}, C_{IN}
- Two Phase Interleaved CCM Boost PFC
	- RMS current expressions, Diode, MOSFET
	- RMS current calculation, \textsf{C}_OUT
	- RMS current expression, $\mathtt{C_{OUT}}$
- Results
- Conclusions
- References

Note: currents are rms unless otherwise stated

CCM Boost PFC Waveforms

Currents are functions of –

Parameters: L_{PFC} , F_{sw} , V_{OUT} , I_{OUT} Assumed to be constant:

Variables: Vac, Duty Cycle (depends on Line phase angle)

CCM Boost PFC Waveforms

RMS Calculation – PFC Diode Current

I $I²(t)$ is first averaged over a switching cycle then averaged over the AC line period

RMS Currents, MOSFET and Diode

RMS Currents, COUT

The total current in $\mathtt{C_{OUT}}$ has two components.

A LF component at twice line frequency Ref (3)

$$
I_{\text{Cout_LF}} = \frac{P_{\text{in}}}{\eta \cdot \sqrt{2} \cdot V_{\text{out}}}
$$

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A HF component at the switching frequency and its harmonics. This current is I_{Diode} – AVG(I_{Diode}) – the calculation will be outlined later (slide 16)

$$
I_{\text{Cout_HF_1Ph}} = \sqrt{\frac{16}{3 \cdot \pi} \cdot \frac{P_{\text{in}}^2}{v_{\text{in_pk}} \cdot v_{\text{out}}} - \frac{3}{8} \cdot \frac{4 \cdot P_{\text{in}}^2}{v_{\text{out}}^2} + \frac{4}{3 \cdot \pi} \cdot \frac{T_s^2 \cdot v_{\text{in_pk}}^3}{12 \cdot L_{\text{pfc}}^2 \cdot v_{\text{out}}} - \frac{3}{8} \cdot \frac{T_s^2 \cdot v_{\text{in_pk}}^4}{6 \cdot L_{\text{pfc}}^2 \cdot v_{\text{out}}^2} + \frac{16}{15 \cdot \pi} \cdot \frac{T_s^2 \cdot v_{\text{in_pk}}^5}{12 \cdot L_{\text{pfc}}^2 \cdot v_{\text{out}}^3}
$$
\n
$$
I_{\text{Cout}} = \sqrt{I_{\text{Cout_HF}}^2 + I_{\text{Cout_LF}}^2}
$$

RMS Currents, C_{IN}

The total ripple current in ${\sf C}_{\sf IN}$ is

$$
I_{\text{Cin}} = \sqrt{I_{\text{LPFC}}^2 - I_{\text{ac}}^2}
$$

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 ${\sf C}_{\sf IN}$ is usually small and there is effectively no line frequency current in this capacitor The RMS current in this capacitor is high frequency only.

2 Phase Interleaved boost PFC

I $_{\rm{Cout}}$ Ripple at twice line frequency is unaffected by interleaving HF ripple is at twice the switching frequency Switching frequency ripple reduction in ${\sf C}_{\sf IN}$ and ${\sf C}_{\sf OUT}$ is a function of the duty cycle

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RMS Currents, 2 Phase, MOSFET and Diode

Calculation for MOSFET and diode currents is the same as in the 1 Phase case

The input power is shared between the two phases so P_in is replaced by $\mathsf{P}_\mathsf{in}/2$ There is no ripple current cancellation in the MOSFET or Diodes

$$
I_{Q} = \frac{\sqrt{2} \cdot \frac{P_{in}}{2}}{V_{in_pk}} \cdot \sqrt{1 - \frac{8}{3\pi} \cdot \frac{V_{in_pk}}{V_{out}}} + \frac{T_s^2 \cdot V_{in_pk}}{48L_{pfc}^2 \cdot P_{in}^2} \cdot \left(1 - \frac{8}{\pi} \cdot \frac{V_{in_pk}}{V_{out}} + \frac{9}{4} \frac{V_{in_pk}}{V_{out}}^2 - \frac{32}{15 \cdot \pi} \cdot \frac{V_{in_pk}}{V_{out}}^3\right)}{V_{out}}
$$

$$
I_{D} = \sqrt{\frac{16}{3\pi} \cdot \left(\frac{P_{in}}{2}\right)^2 + \frac{1}{9 \cdot \pi} \cdot \frac{T_s^2 \cdot V_{pk}^3}{L_{pfc}^2 \cdot V_{out}} - \frac{1}{16} \cdot \frac{T_s^2 \cdot V_{pk}^4}{L_{pfc}^2 \cdot V_{out}} + \frac{4}{45 \cdot \pi} \cdot \frac{T_s^2 \cdot V_{pk}^5}{L_{pfc}^2 \cdot V_{out}}^3}
$$

2 Phase, I_{Cout rms} Calculation Outline

 $\bullet~$ The LF current ripple in $\mathtt{C_{OUT}}$ is unchanged by interleaving

- $\bullet~$ Calculate the 1 Ph HF current in C_{OUT}
- Determine a correction factor $(K_{MS}(t))$
- Apply the correction factor to the 1 Ph HF solution to get the 2 Ph HF solution
- Add the 2 Ph HF current to the LF current to get the total

RMS Calculation – Calculate the HF ripple in C_{out}

I $I²(t)$ is first averaged over a switching cycle then averaged over the AC line period

Correction Factor

1 Ph and 2 Ph I_{cout} currents simulated (PSIM) $\begin{array}{llll} \text{Cancellation Factor (K_MS)}\\ \text{R} & \text{R} & \text{R} \\ \text{R} & \text{R} & \text{R} \\ \text{R} & \text{R} & \text{R} \end{array}$ Mean Square values taken across d_{on} range I_{2Ph} Plot $\frac{27\,n}{I}$ results against d_{on} I_{1Ph} Linearise $K_{MS}(d_{on})=md\ d_{on}+cd$ 0.1 $md = 1.2$ $cd = -0.6$ If $d_{on} > 50\%$ $\mathbf 0$ 0% 20% 40% $md = -1.2$ $cd = 0.6$ If d_{on} < 50% On time duty cycle (d_{on}) \rm{V}_{pk}

Duty cycle d_{on} is a function of time Restate in terms of t

 $K_{MS}(t) = mt$ V_{in_pk} $\frac{ln_2 p}{v_{out}}$ sin(ωt) + ct

If $\rm{d_{on}}>50\%$ If $\rm{d_{on}}$ < 50% $\,$ $mt = -1.2$ $ct = 0.6$ $mt = 1.2$ $ct = -0.6$

 $d_{\text{on}}(t) = 1$

 0.6

Analytic Expressions for currents in the CCM PFC stage

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 $-\frac{1}{V_{\text{out}}} \cdot \sin(\omega \cdot t)$

60%

80%

100%

C_{OUT} – Switching Cycle HF Mean Square Current

Diode and MOSFET currents are trapezoidal $-i_{avg}(t)$ $I_{\text{MS_Trap}} = d \cdot I_{\text{mid}}^2 + \frac{d}{3} \cdot \Delta I(t)^2$ During Ton current is During Toff current is $i_{\rm ac}^{}(\mathrm{t})-i_{\rm avg}^{}(\mathrm{t})$ ි
ර $\int \Delta I$ I_{mid} $\left[i(t)^2 \right]_{\text{Ts}} = i_{\text{avg}}(t)^2 \cdot \left(1 - d_{\text{off}} \right) + d_{\text{off}} \cdot i_{\text{mid}}(t)^2 + \frac{d_{\text{off}}}{3} \cdot \Delta i(t)^2$ So Multiply by $K_{MS}(t)$ for the interleaved case t off t on T_s $\left[i\text{Cout}^2(t)\right]_{\text{Ts}} = \text{K}_{\text{MS}}(t) \cdot \left[\left(i\text{avg}(t)\right)^2 \cdot \left(1 - d_{\text{off}}(t)\right) + d_{\text{off}}(t) \cdot \left(i\text{mid}(t)\right)^2 + \frac{d_{\text{off}}(t)}{3} \cdot \Delta i(t)^2\right]$ Now we need expressions for the functions of t in this equation 14

Substitutions

• We have the following

 $d_{\text{off}}(t)$ V_{in_pk} V_{out} $\cdot \sin(\omega \cdot t)$ Δi (t) $T_S \cdot V_{in_pk}$ 2.L $_{\rm pfc}$ 1 Vin_pk $-\frac{1}{\text{V}_{\text{out}}} \cdot \sin(\omega \cdot t)$ $\cdot \left(1 - \frac{V_{in_pk}}{V_{out}} \cdot \sin(\omega \cdot t)\right) \cdot \sin(\omega \cdot t)$ $I_{\text{mid}}(t) = I_{\text{ac}}(t) - I_{\text{avg}}(t)$ $K_{\text{MS}}(t) = m t$ V_{pk} $\frac{1}{v_{\text{out}}}$ sin(ω ·t) + ct $i_{\text{ac}}(t)$ $2 \cdot P_{in}$ Vin_pk $\cdot \sin(\omega \cdot t)$ 15

C_{OUT} – Switching Cycle HF Mean Square Current

$$
\left[i_{\text{Cout}}^{2}(t)\right]_{Ts} = K_{\text{MS}}(t) \cdot \left[\left(i_{\text{avg}}(t)\right)^{2} \cdot \left(1 - d_{\text{off}}(t)\right) + d_{\text{off}}(t) \cdot \left(i_{\text{mid}}(t)\right)^{2} + \frac{d_{\text{off}}(t)}{3} \cdot \Delta i\left(t\right)^{2}\right]
$$

Set $K_{MS}(t)$ to 1 for the 1Ph solution

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• Making the substitutions and simplifying gives

$$
\left[i_{\text{Cout}}^{2}(t)\right] = m t \frac{4 P_{in}^{2}}{V_{out}^{2}} \cdot \left(\sin(\omega \cdot t)^{4} - \frac{V_{pk} \cdot \sin(\omega \cdot t)^{5}}{V_{out}^{2}}\right) + ct \cdot \frac{4 P_{in}^{2}}{V_{out}} \cdot \left(\frac{\sin(\omega \cdot t)^{3}}{V_{pk}} - \frac{\sin(\omega \cdot t)^{4}}{V_{out}}\right)
$$
\n• Integrating gives the rms value\n
$$
I_{RMS} = \frac{2}{T_{AC}} \sqrt{\int_{0}^{T_{AC}} [i^{2}(t)]_{T_{SW}} dt}
$$
\n
$$
I_{\text{Cout_HF_2Ph}} = \sqrt{\frac{m \left(\frac{3}{2} \cdot \frac{P_{in}^{2}}{V_{out}} - \frac{64}{15 \cdot \pi} \cdot \frac{P_{in}^{2} \cdot V_{in_pk}}{V_{out}}\right) + ct \cdot \left(\frac{16}{3 \cdot \pi} \cdot \frac{P_{in}^{2}}{V_{in_pk}} \cdot V_{out} - \frac{3}{2} \cdot \frac{P_{in}^{2}}{V_{out}}\right)}
$$

COUT – RMS Current

We have

$$
I_{\text{Cout_HF_2Ph}} = \sqrt{\text{mt} \left(\frac{3}{2} \cdot \frac{P_{\text{in}}^2}{V_{\text{out}}} - \frac{64}{15 \cdot \pi} \cdot \frac{P_{\text{in}}^2 \cdot V_{\text{in_pk}}}{V_{\text{out}}} \right) + \text{ct} \cdot \left(\frac{16}{3 \cdot \pi} \cdot \frac{P_{\text{in}}^2}{V_{\text{in_pk}} \cdot V_{\text{out}}} - \frac{3}{2} \cdot \frac{P_{\text{in}}^2}{V_{\text{out}}^2} \right)}
$$

And $I_{\text{Cout_LF}} = \frac{P_{\text{in}}}{T}$

$$
Ind \tI_{Cout_LF} = \frac{1}{\eta \cdot \sqrt{2} \cdot V_{out}}
$$

So

$$
I_{\text{Cout_2Ph}} = \sqrt{I_{\text{Cout_HF}}^2 + I_{\text{Cout_LF}}^2}
$$

Analytic Expressions for currents in the CCM PFC stage

Valid for V_{in_pk} < Vout/2

If V_{inpk} > Vout/2 then the MS calculation and the integral have to be split. One for 0 < Vin < Vout/2 One for Vout/2 < Vin < Vout/2And the results added $d \overline{17}$

Results*

Taken on a Texas Instruments UCC28070 EVM - 2Ph interleaved CCM Boost PFC

C_{OUT} , Ripple reduction example

* With thanks to Sonal Singh for taking the measurements **18** 18

Conclusions

• Equations for the RMS currents in the CCM boost PFC developed and presented

Diode

MOSFET

Input Capacitor

Output Capacitor

- Significant reduction in RMS currents in the interleaved CCM Boost PFC
- Output capacitor High Frequency ripple current significantly reduced by interleaving
- Output capacitor Low Frequency ripple current unaffected by interleaving
- Output capacitor total ripple current reduced by interleaving
- Good agreement with experimental results

References

Ref (1): 'Fundamentals of Power Electronics, Erickson and Maksimovic; Springer 2001, Table 18.3, summary of rectifier current stresses.

Ref (2): SLUP279 An Interleaving PFC Pre-Regulator for High-Power Converters. Mike O'Loughlin

Ref (3): Capacitor Ripple current in an interleaved PFC converter, Pratt and Jinsong,

IEEE transactions on Power Electronics, Vol 24, No 6 June 2009.

UCC28070: http://www.ti.com/product/UCC28070

UCC28180: http://www.ti.com/product/UCC28180

Useful Integrals

$$
\frac{1}{\pi} \cdot \int_0^{\pi} \sin(\theta) d\theta = \frac{2}{\pi}
$$

$$
\frac{1}{\pi} \cdot \int_0^{\pi} (\sin(\theta))^2 d\theta = \frac{1}{2}
$$

$$
\frac{1}{\pi} \cdot \int_0^{\pi} (\sin(\theta))^3 d\theta = \frac{4}{3\pi}
$$

$$
\frac{1}{\pi} \cdot \int_0^{\pi} (\sin(\theta))^5 d\theta = \frac{16}{15\pi}
$$

$$
\frac{1}{\pi} \cdot \int_0^{\pi} (\sin(\theta))^6 d\theta = \frac{15}{48}
$$

Thank You

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