## Analytic Expressions for currents in the CCM PFC stage

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**SLYY131** 

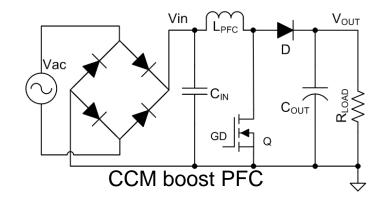


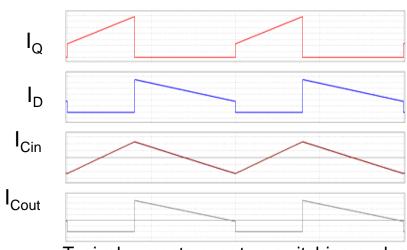
### **Agenda**

- CCM Boost PFC stage with current waveforms
  - RMS calculation
  - RMS current expressions, Diode, MOSFET,  $C_{OUT}$ ,  $C_{IN}$
- Two Phase Interleaved CCM Boost PFC
  - RMS current expressions, Diode, MOSFET
  - RMS current calculation, C<sub>OUT</sub>
  - RMS current expression, C<sub>OUT</sub>
- Results
- Conclusions
- References

Note: currents are rms unless otherwise stated

### **CCM Boost PFC Waveforms**





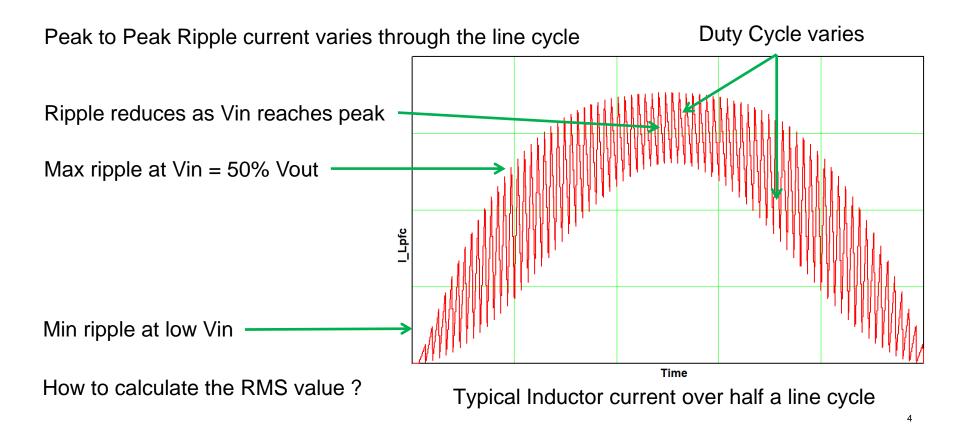
Typical currents over two switching cycles

Currents are functions of -

Parameters: L<sub>PFC</sub>, F<sub>sw</sub>, V<sub>OUT</sub>, I<sub>OUT</sub> Assumed to be constant:

Variables: Vac, Duty Cycle (depends on Line phase angle)

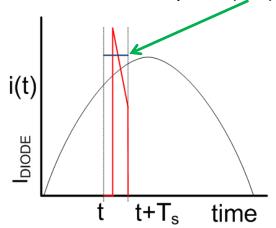
#### **CCM Boost PFC Waveforms**



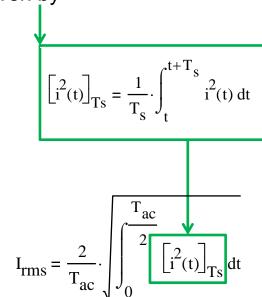
#### **RMS Calculation – PFC Diode Current**

I<sup>2</sup>(t) is first averaged over a switching cycle then averaged over the AC line period

1/ The Mean Square (MS) current in each switching cycle is given by



2/ Integrate the MS current over a half line cycle, — average it and take the square root



### **RMS Currents, MOSFET and Diode**

$$I_{Q} = \frac{\sqrt{2} \cdot P_{in}}{V_{in\_pk}} \cdot \sqrt{1 - \frac{8}{3\pi} \cdot \frac{V_{in\_pk}}{V_{out}}} + \frac{T_{s}^{2} \cdot V_{in\_pk}^{4}}{48L_{pfc}^{2} \cdot P_{in}^{2}} \cdot \left(1 - \frac{8}{\pi} \cdot \frac{V_{in\_pk}}{V_{out}} + \frac{9}{4} \cdot \frac{V_{in\_pk}^{2}}{V_{out}^{2}} - \frac{32}{15 \cdot \pi} \cdot \frac{V_{in\_pk}^{3}}{V_{out}^{3}}\right)$$

$$I_{D} = \sqrt{\frac{\frac{16}{3\pi} \cdot \frac{P_{in}^{2}}{V_{pk} \cdot V_{out}}}{V_{pk} \cdot V_{out}}} + \frac{1}{9 \cdot \pi} \cdot \frac{T_{s}^{2} \cdot V_{pk}^{3}}{L_{pfc}^{2} \cdot V_{out}} - \frac{1}{16} \cdot \frac{T_{s}^{2} \cdot V_{pk}^{4}}{L_{pfc}^{2} \cdot V_{out}^{2}} + \frac{4}{45 \cdot \pi} \cdot \frac{T_{s}^{2} \cdot V_{pk}^{5}}{L_{pfc}^{2} \cdot V_{out}^{3}}$$

and then:

Erickson and Maksimovic give the part in red (Ref 1)

 $I_{LPFC} = \sqrt{I_Q^2 + I_D^2}$   $I_{Cin} = \sqrt{I_{LPFC}^2 - I_{ac}^2}$   $I_{Cout} = \sqrt{I_D^2 - I_{Out}^2}$ 

## RMS Currents, Cout

The total current in  $C_{\text{OUT}}$  has two components.

A LF component at twice line frequency Ref (3)

$$I_{\text{Cout\_LF}} = \frac{P_{\text{in}}}{\eta \cdot \sqrt{2} \cdot V_{\text{out}}}$$

A HF component at the switching frequency and its harmonics. This current is  $I_{Diode} - AVG(I_{Diode})$  – the calculation will be outlined later (slide 16)

$$I_{Cout\_HF\_1Ph} := \sqrt{\frac{\frac{16}{3 \cdot \pi} \cdot \frac{P_{in}^{2}}{V_{in\_pk} \cdot V_{out}} - \frac{3}{8} \cdot \frac{4 \cdot P_{in}^{2}}{V_{out}^{2}} + \frac{4}{3 \cdot \pi} \cdot \frac{T_{s}^{2} \cdot V_{in\_pk}^{3}}{12 \cdot L_{pfc}^{2} \cdot V_{out}} - \frac{3}{8} \cdot \frac{T_{s}^{2} \cdot V_{in\_pk}^{4}}{6 \cdot L_{pfc}^{2} \cdot V_{out}^{2}} + \frac{\frac{16}{15 \cdot \pi} \cdot \frac{T_{s}^{2} \cdot V_{in\_pk}^{5}}{12 \cdot L_{pfc}^{2} \cdot V_{out}^{3}}}{12 \cdot L_{pfc}^{2} \cdot V_{out}^{3}}}$$

The total current in  $C_{\text{OUT}}$  is then

$$I_{\text{Cout}} = \sqrt{I_{\text{Cout\_HF}}^2 + I_{\text{Cout\_LF}}^2}$$

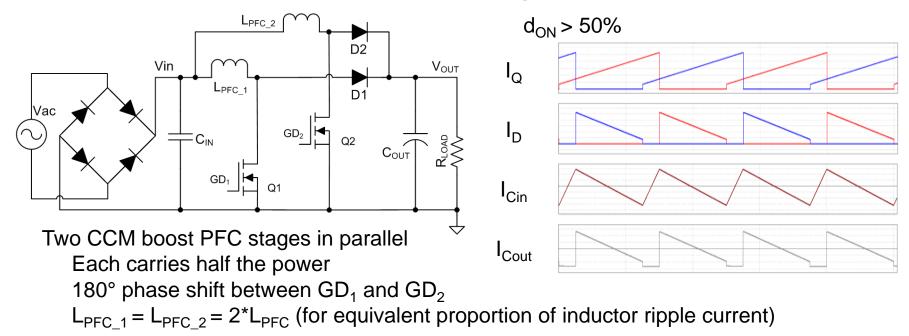
## RMS Currents, C<sub>IN</sub>

The total ripple current in  $C_{\text{IN}}$  is

$$I_{Cin} = \sqrt{I_{LPFC}^2 - I_{ac}^2}$$

C<sub>IN</sub> is usually small and there is effectively no line frequency current in this capacitor. The RMS current in this capacitor is high frequency only.

#### 2 Phase Interleaved boost PFC



 $I_{Cout}$  Ripple at twice line frequency is unaffected by interleaving HF ripple is at twice the switching frequency Switching frequency ripple reduction in  $C_{IN}$  and  $C_{OUT}$  is a function of the duty cycle

### RMS Currents, 2 Phase, MOSFET and Diode

Calculation for MOSFET and diode currents is the same as in the 1 Phase case

The input power is shared between the two phases so  $P_{in}$  is replaced by  $P_{in}/2$ . There is no ripple current cancellation in the MOSFET or Diodes

$$I_{Q} = \frac{\sqrt{2} \cdot \frac{P_{in}}{2}}{V_{in\_pk}} \cdot \sqrt{1 - \frac{8}{3\pi} \cdot \frac{V_{in\_pk}}{V_{out}} + \frac{T_{s}^{2} \cdot V_{in\_pk}}{48L_{pfc}^{2} \cdot P_{in}^{2}}} \cdot \left(1 - \frac{8}{\pi} \cdot \frac{V_{in\_pk}}{V_{out}} + \frac{9}{4} \cdot \frac{V_{in\_pk}}{V_{out}^{2}} - \frac{32}{15 \cdot \pi} \cdot \frac{V_{in\_pk}}{V_{out}^{3}}\right)$$

$$I_{D} = \sqrt{\frac{\frac{16}{3\pi} \cdot \frac{\left(\frac{P_{in}}{2}\right)^{2}}{V_{pk} \cdot V_{out}} + \frac{1}{9 \cdot \pi} \cdot \frac{T_{s}^{2} \cdot V_{pk}^{3}}{L_{pfc}^{2} \cdot V_{out}} - \frac{1}{16} \cdot \frac{T_{s}^{2} \cdot V_{pk}^{4}}{L_{pfc}^{2} \cdot V_{out}^{2}} + \frac{4}{45 \cdot \pi} \cdot \frac{T_{s}^{2} \cdot V_{pk}^{5}}{L_{pfc}^{2} \cdot V_{out}^{3}}}$$

# 2 Phase, I<sub>Cout rms</sub> Calculation Outline

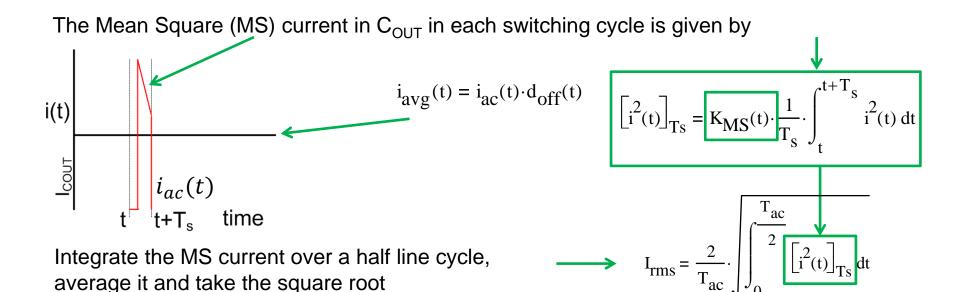
• The LF current ripple in C<sub>OUT</sub> is unchanged by interleaving

$$I_{\text{Cout\_LF}} = \frac{P_{\text{in}}}{\eta \cdot \sqrt{2} \cdot V_{\text{out}}}$$

- Calculate the 1 Ph HF current in C<sub>OUT</sub>
- Determine a correction factor (K<sub>MS</sub>(t))
- Apply the correction factor to the 1 Ph HF solution to get the 2 Ph HF solution
- Add the 2 Ph HF current to the LF current to get the total

# RMS Calculation – Calculate the HF ripple in Cout

I<sup>2</sup>(t) is first averaged over a switching cycle then averaged over the AC line period



#### **Correction Factor**

1 Ph and 2 Ph I<sub>cout</sub> currents simulated (PSIM) Mean Square values taken across don range

Plot 
$$\frac{I_{2Ph}}{I_{1Ph}}$$
 results against  $d_{on}$  Linearise  $K_{MS}(d_{on}) = md \ d_{on} + cd$  ---

$$m_{NS}(\alpha_{ON})$$

md = 1.2 cd = -0.6 If  $d_{on} > 50\%$ 

$$cd = -0.6$$

If 
$$d_{on} > 50\%$$

$$md = -1.2$$

$$cd = 0.6$$

$$md = -1.2$$
  $cd = 0.6$  If  $d_{on} < 50\%$ 

Duty cycle d<sub>on</sub> is a function of time Restate in terms of t

$$K_{MS}(t) = mt \frac{V_{in\_pk}}{V_{out}} \sin(\omega t) + ct$$

$$mt = -1.2$$

$$mt = 1.2$$

$$ct = 0$$

Cancellation Factor (K\_MS)

$$ct = -0.6$$

$$mt = -1.2$$
  $ct = 0.6$  If  $d_{on} > 50\%$   $mt = 1.2$   $ct = -0.6$  If  $d_{on} < 50\%$ 

6 If 
$$d_{on} < 50$$

On time duty cycle (don)

100%

# **C<sub>OUT</sub> – Switching Cycle HF Mean Square Current**

Diode and MOSFET currents are trapezoidal

During Ton current is  $-i_{avg}(t)$ 

$$I_{MS\_Trap} = d \cdot I_{mid}^{2} + \frac{d}{3} \cdot \Delta I(t)^{2}$$

t on

Ts

During Toff current is  $i_{ac}(t) - i_{avg}(t)$ 

So 
$$\left[i(t)^{2}\right]_{Ts} = i_{avg}(t)^{2} \cdot \left(1 - d_{off}\right) + d_{off} \cdot i_{mid}(t)^{2} + \frac{d_{off}}{3} \cdot \Delta i(t)^{2}$$

Multiply by K<sub>MS</sub>(t) for the interleaved case

$$\left[i_{Cout}^{2}(t)\right]_{Ts} = K_{MS}(t) \cdot \left[\left(i_{avg}(t)\right)^{2} \cdot \left(1 - d_{off}(t)\right) + d_{off}(t) \cdot \left(i_{mid}(t)\right)^{2} + \frac{d_{off}(t)}{3} \cdot \Delta i(t)^{2}\right]$$

Now we need expressions for the functions of t in this equation

t off

#### **Substitutions**

We have the following

$$d_{\text{off}}(t) = \frac{V_{\text{in}} pk}{V_{\text{out}}} \cdot \sin(\omega \cdot t)$$

$$\Delta i(t) = \frac{T_{s} \cdot V_{in\_pk}}{2 \cdot L_{pfc}} \cdot \left(1 - \frac{V_{in\_pk}}{V_{out}} \cdot \sin(\omega \cdot t)\right) \cdot \sin(\omega \cdot t)$$

$$i_{ac}(t) = \frac{\sqrt{2} \cdot P_{in}}{V_{in\_pk}} \cdot \sin(\omega \cdot t)$$

$$I_{mid}(t) = I_{ac}(t) - I_{avg}(t)$$

$$K_{MS}(t) = mt \cdot \frac{V_{pk}}{V_{out}} \cdot \sin(\omega \cdot t) + ct$$

# **C<sub>OUT</sub> – Switching Cycle HF Mean Square Current**

$$\left[i_{Cout}^{2}(t)\right]_{Ts} = K_{MS}(t) \cdot \left[\left(i_{avg}(t)\right)^{2} \cdot \left(1 - d_{off}(t)\right) + d_{off}(t) \cdot \left(i_{mid}(t)\right)^{2} + \frac{d_{off}(t)}{3} \cdot \Delta i(t)^{2}\right]$$
 Set  $K_{MS}(t)$  to 1 for the 1Ph solution

Making the substitutions and simplifying gives

$$\left[i_{Cout}^{2}(t)\right] = mt \cdot \frac{4P_{in}^{2}}{V_{out}^{2}} \cdot \left(\sin(\omega \cdot t)^{4} - \frac{V_{pk} \cdot \sin(\omega \cdot t)^{5}}{V_{out}^{2}}\right) + ct \cdot \frac{4 \cdot P_{in}^{2}}{V_{out}} \cdot \left(\frac{\sin(\omega \cdot t)^{3}}{V_{pk}} - \frac{\sin(\omega \cdot t)^{4}}{V_{out}}\right)$$

• Integrating gives the rms value

$$I_{RMS} = \frac{2}{T_{AC}} \int_{0}^{\frac{T_{AC}}{2}} [i^{2}(t)]_{T_{SW}} dt$$

$$I_{Cout\_HF\_2Ph} = \sqrt{mt \left(\frac{3}{2} \cdot \frac{P_{in}^{2}}{V_{out}^{2}} - \frac{64}{15 \cdot \pi} \cdot \frac{P_{in}^{2} \cdot V_{in\_pk}}{V_{out}^{3}}\right) + ct \cdot \left(\frac{16}{3 \cdot \pi} \cdot \frac{P_{in}^{2}}{V_{in\_pk} \cdot V_{out}} - \frac{3}{2} \cdot \frac{P_{in}^{2}}{V_{out}^{2}}\right)}$$

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### **C**<sub>OUT</sub> – **RMS** Current

We have

$$I_{Cout\_HF\_2Ph} = \sqrt{mt \left(\frac{3}{2} \cdot \frac{P_{in}^{2}}{V_{out}^{2}} - \frac{64}{15 \cdot \pi} \cdot \frac{P_{in}^{2} \cdot V_{in\_pk}}{V_{out}^{3}}\right) + ct \cdot \left(\frac{16}{3 \cdot \pi} \cdot \frac{P_{in}^{2}}{V_{in\_pk} \cdot V_{out}} - \frac{3}{2} \cdot \frac{P_{in}^{2}}{V_{out}^{2}}\right)}$$

And 
$$I_{Cout\_LF} = \frac{P_{in}}{\eta \cdot \sqrt{2} \cdot V_{out}}$$

So

$$I_{\text{Cout\_2Ph}} = \sqrt{I_{\text{Cout\_HF}}^2 + I_{\text{Cout\_LF}}^2}$$

Valid for  $V_{in\_pk}$  < Vout/2

If  $V_{in\_pk}$  > Vout/2 then the MS calculation and the integral have to be split.

One for 0 < Vin < Vout/2

One for Vout/2 < Vin < Vout/2

And the results added

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### Results\*

#### Taken on a Texas Instruments UCC28070 EVM - 2Ph interleaved CCM Boost PFC

ICout_2Ph				
Vin	Pin	Meas, mA	Calc, mA	Error
120V	300W	817	732	10.4%
120V	200W	702	610	13.1%
120V	100W	595	488	18.0%
90V	300W	976	946	3.1%
90V	200W	821	788	4.0%
90V	100W	684	631	7.7%

$C_{OUT}$ , F	Ripple	reduction	example
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Icout (Calc, mA)			
	1Ph	2Ph	Ratio
300W, 120V	1407	732	52%
300W, 90V	1633	946	58%

IQ_2Ph				
Vin	Pin	Meas, mA	Calc, mA	Error
120V	300W	1185	1170	1.3%
120V	200W	1029	1035	0.6%
120V	100W	880	910	3.4%
90V	300W	1581	1525	3.5%
90V	200W	1355	1309	3.4%
90V	100W	1128	1102	2.3%

ID_2Ph				
V_in	Pin	Meas, mA	Calc, mA	Error
120V	300W	788	895	13.6%
120V	200W	670	788	17.6%
120V	100W	555	690	24.3%
90V	300W	880	949	7.8%
90V	200W	741	814	9.9%
90V	100W	602	684	13.6%

<sup>\*</sup> With thanks to Sonal Singh for taking the measurements

#### **Conclusions**

Equations for the RMS currents in the CCM boost PFC developed and presented

Diode

**MOSFET** 

**Input Capacitor** 

**Output Capacitor** 

- Significant reduction in RMS currents in the interleaved CCM Boost PFC
- Output capacitor High Frequency ripple current significantly reduced by interleaving
- Output capacitor Low Frequency ripple current unaffected by interleaving
- Output capacitor total ripple current reduced by interleaving
- Good agreement with experimental results

#### References

Ref (1): 'Fundamentals of Power Electronics, Erickson and Maksimovic; Springer 2001, Table 18.3, summary of rectifier current stresses.

Ref (2): SLUP279 An Interleaving PFC Pre-Regulator for High-Power Converters. Mike O'Loughlin

Ref (3): Capacitor Ripple current in an interleaved PFC converter, Pratt and Jinsong,

IEEE transactions on Power Electronics, Vol 24, No 6 June 2009.

UCC28070: http://www.ti.com/product/UCC28070

UCC28180: http://www.ti.com/product/UCC28180

$$\frac{1}{\pi} \cdot \int_0^{\pi} \sin(\theta) \, d\theta = \frac{2}{\pi}$$

$$\frac{1}{\pi} \cdot \int_0^{\pi} \left( \sin(\theta) \right)^2 d\theta = \frac{1}{2}$$

$$\frac{1}{\pi} \cdot \int_0^{\pi} \sin(\theta) \, d\theta = \frac{2}{\pi} \qquad \qquad \frac{1}{\pi} \cdot \int_0^{\pi} \left(\sin(\theta)\right)^2 d\theta = \frac{1}{2} \qquad \qquad \frac{1}{\pi} \cdot \int_0^{\pi} \left(\sin(\theta)\right)^3 d\theta = \frac{4}{3 \cdot \pi}$$

$$\frac{1}{\pi} \cdot \int_0^{\pi} \left( \sin(\theta) \right)^4 d\theta = \frac{3}{8}$$

$$\frac{1}{\pi} \cdot \int_0^{\pi} \left( \sin(\theta) \right)^5 d\theta = \frac{16}{15 \cdot \pi}$$

$$\frac{1}{\pi} \cdot \int_{0}^{\pi} (\sin(\theta))^{4} d\theta = \frac{3}{8} \qquad \frac{1}{\pi} \cdot \int_{0}^{\pi} (\sin(\theta))^{5} d\theta = \frac{16}{15 \cdot \pi} \qquad \frac{1}{\pi} \cdot \int_{0}^{\pi} (\sin(\theta))^{6} d\theta = \frac{15}{48}$$



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